

A Constrained Optimization Approach to Dynamic State Estimation for Power Systems including PMU Measurements

Liang Hu, Zidong Wang, Izaz Rahman and Xiaohui Liu

Abstract—In this paper, a hybrid filter algorithm is developed to deal with the state estimation problem for power systems by taking into account the impact from the phasor measurement units (PMU). Our aim is to include PMU measurements when designing the dynamic state estimators for power systems with traditional measurements. Also, as data dropouts inevitably occur in the transmission channels of traditional measurements from the meters to the control centre, the missing measurement phenomenon is also tackled in the state estimator design. In the framework of extended Kalman filter (EKF) algorithm, the PMU measurements are treated as inequality constraints on the states with the aid of the statistical criterion, and then the addressed state estimation problem becomes a constrained optimization one based on the probability-maximization method. The resulting constrained optimization problem is then solved by using the particle swarm optimization (PSO) algorithm together with the penalty function approach. The proposed algorithm is applied to estimate the states of the power systems with both traditional and PMU measurements in the presence of probabilistic data missing phenomenon. Extensive simulations are carried out on the IEEE 14-bus test system and it is shown that the proposed algorithm gives much improved estimation performances over the traditional EKF method.

Index Terms—Power systems, state estimation, extended Kalman filter, missing measurements, particle swarm optimization, constrained optimization

NOMENCLATURE

k	Time index.
$k k-1$	Time index for prediction from instant $k-1$ to k .
$k k$	Time index for update at instant k .
tj	The line connecting nodes t and j .
$tj0$	t side (to the ground) of the line connecting nodes t and j .
r and i	Real and imaginary component.
N	Total number of bus nodes of interest.
M	Total number of PMUs.
n_v	Total number of voltage meters.
n_p	Total number of power meters at the nodes.
n_f	Total number of power meters at the lines.
\mathbb{N}_s	The set of bus numbers directly connected to node s .
t_l^j	The l th bus directly connected to bus j .

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I. INTRODUCTION

State estimation (SE) has long been one of the fundamental problems in the research on power systems. Traditional SE approach is typically static where the single-scan weighted least-squares estimators are adopted [1]. Static SE method exhibits the features of fast convergence and easy implementation, but it suffers from the accuracy problems since the dynamics of the power system is ignored.

With rapid development of the sensing techniques, online monitoring has recently become popular which gives rise to the renewed research interests on the design of the dynamic state estimator (DSE). Comparing with the static state estimation scheme, the DSE is capable of achieving better estimation accuracy since more information about the state evolution is utilized. Another advantage of the DSE is its potential ability to provide prediction database that could be adopted as a set of pseudo-measurements in case of missing data or meter outages in the power grids.

Note that the missing data phenomenon constitutes one of the major concerns in state estimation for power systems since data dropouts inevitably occur in the transmission channels of traditional measurements from the meters to the control centre. As discussed in [7], [8], [11], [18], [21], the communication constraints (e.g. limited bandwidth) have inevitably led to network-induced phenomena such as random communication delays and missing measurements. As for missing measurements, a conventional way is to treat them as normal bad data *without* in-depth characterization of the dropouts. Very recently, the missing measurement problem has been tackled in [18], [19] where a certain stochastic variable is involved in the estimator, and this renders the difficulties in the implementation. In our paper, a recursive algorithm is developed to mitigate the effect of missing measurements through modifying the traditional DSE approaches.

On the other hand, advanced techniques for synchronized phasor measurements have recently been applied in power systems. Different from traditional SCADA systems, where the magnitude of the nodal voltage can be measured directly, PMUs are capable to measure both the magnitude and the phase of nodal voltages. And due to their intrinsic advantages, the phasor measurement units (PMU) have made it possible to measure the system states in a more accurate and timely way as compared with the traditional measurements. Unfortunately, for economic reasons, it is not affordable to replace all the RTUs with PMUs in the foreseeable future [14], [22]. In other

words, only states of partial buses could be measured directly by PMUs and the rest would have to be estimated by using the conventional RTUs. As such, an emerging research issue is how to integrate PMU measurements into traditional SE algorithms, and this issue has started to gain some initial research attention, see [3], [14]–[16], [22], [24]. It should be noted that all the corresponding results available in the literature have been concerned with static SE problems, and the DSE problem in the presence of partial PMU involvements remains as a challenging topic of research [12]. This situation motivates our current investigation.

The main purpose of the present research is to design dynamic state estimators for power systems by making one of the first attempts to solve the aforementioned two challenging problems, i.e., 1) how to account for the probabilistic missing data phenomenon? 2) how to include the PMU measurements in the state estimator design? In this paper, the phenomenon of missing measurements is assumed to occur randomly and the missing probability for each channel is governed by an individual random variable satisfying a certain probability distribution over the interval $[0, 1]$. The impact of missing measurements on the overall estimation performance is considered when designing the estimator. On the other hand, to incorporate the PMU measurements into the widely used extended Kalman filter (EKF) algorithm, the PMU measurements are characterized via a set of inequality constraints based on the well-known 3-sigma rule of the Gaussian distribution, and then the EKF problem with state constraints becomes a constrained optimization problem that can be effectively solved by the particle swarming optimization (PSO) algorithm. As PSO has been developed primarily as an unconstrained optimization method, the penalty function approach is utilized to convert the constrained optimization problem into an unconstrained one.

In this paper, a hybrid EKF and PSO algorithm is developed to estimate the states of power system. The main contribution of this paper is threefold. 1) *A new dynamic state estimation scheme is first proposed to improve the estimation performance of power system including PMU measurements. Such a scheme has the advantages of being scalable to the numbers of the installed PMUs and of being compatible with existing DSE software.* 2) *Practical issues of missing measurements in communication network are investigated thoroughly and a modified EKF algorithm is developed which is insensitive to the measurement unreliability in terms of acceptable probability.* 3) *Extensive comparative experiments have been implemented based on different missing rates of the RTU measurements and it is confirmed that our proposed estimation algorithm provides better performance than the traditional EKF in the presence of the missing measurements.*

Notation The notation used here is fairly standard except where otherwise stated. $I_{m,1}$ denotes the m -dimensional vector with all elements equal to 1. For given matrices A and B with the same dimension, \circ is the Hadamard product defined as $[A \circ B]_{ij} = [A]_{ij} \times [B]_{ij}$. $\mathbb{E}\{x\}$ stands for the expectation of the stochastic variable x . $|C|$ describes the determinants of a square matrix C . $\text{diag}\{\dots\}$ and $\text{diag}_n\{*\}$ stand for a

block-diagonal matrix and $\text{diag}\{\overset{n}{*}, \dots, *\}$, respectively.

II. PROBLEM FORMULATION AND PRELIMINARIES

A. System Model with Missing RTU Measurements

In this paper, the power network is assumed to operate among quasi-steady states and such kind of steady-state dynamics is typically different from the transient ones generated by the electro-mechanical power systems. Let us consider the following model that represents the slow system dynamics of N buses ([4]–[6], [19], [20]):

$$x(k+1) - u = A(x(k) - u) + \omega(k) \quad (1)$$

where the state $x(k) \in \mathbb{R}^{2N}$ is the vector of the real parts and the imaginary parts of the voltages at all buses in the rectangular form, that is, $x(k) = [x_{r,1}(k) \ x_{r,2}(k) \ \dots \ x_{r,N}(k) \ x_{i,1}(k) \ x_{i,2}(k) \ \dots \ x_{i,N}(k)]^T$, and $u \in \mathbb{R}^{2N}$ is the trend behavior of the state trajectory. $\omega(k)$ is a Gaussian sequence with zero mean and covariance matrix $W(k)$. A represents how fast the transitions between states are. The initial value of state $x(0)$ is a white Gaussian noise with mean value $\bar{x}(0)$ and covariance matrix $\Sigma(0|0)$. For computational convenience, the state transition matrix A has been traditionally assumed to be diagonal in the dynamic state estimation algorithms of power systems [6].

For the purpose of simplicity, define $B := I - A$, then (1) can be rewritten in the following compact form:

$$x(k+1) = Ax(k) + Bu + \omega(k). \quad (2)$$

The ideal measurement (without missing phenomena) $z^{(r)}(k) \in \mathbb{R}^m$ collected by RTUs is given as follows

$$z^{(r)}(k) = [V^T(k) \ P^T(k) \ Q^T(k) \ P^f T(k) \ Q^f T(k)]^T.$$

Assuming the general two-port π -model for the network branches, the explicit element for each aforementioned measurement is given as follows (the symbol of time instant, k , is omitted for brevity):

$$\begin{aligned} V_s &= \sqrt{x_{r,s}^2 + x_{i,s}^2} \\ P_s &= x_{r,s} \sum_{j \in \mathbb{N}_s} (G_{sj} x_{r,j} - B_{sj} x_{i,j}) \\ &\quad + x_{i,s} \sum_{j \in \mathbb{N}_s} (G_{sj} x_{i,j} + B_{sj} x_{r,j}) \\ Q_s &= x_{i,s} \sum_{j \in \mathbb{N}_s} (G_{sj} x_{r,j} - B_{sj} x_{i,j}) \\ &\quad - x_{r,s} \sum_{j \in \mathbb{N}_s} (G_{sj} x_{i,j} + B_{sj} x_{r,j}) \\ P_s^f &:= P_{tj}^f = (x_{r,t}^2 + x_{i,t}^2)(g_{tj} + g_{tj}) - x_{r,t} x_{r,j} g_{tj} \\ &\quad - x_{i,t} x_{i,j} g_{tj} - x_{i,t} x_{r,j} b_{tj} + x_{r,t} x_{i,j} b_{tj} \\ Q_s^f &:= Q_{tj}^f = -(x_{r,t}^2 + x_{i,t}^2)(b_{tj} + b_{tj}) - x_{i,t} x_{r,j} g_{tj} \\ &\quad + x_{r,t} x_{i,j} g_{tj} + x_{r,t} x_{r,j} b_{tj} + x_{i,t} x_{i,j} b_{tj} \end{aligned} \quad (3)$$

where $V(k) = [V_1(k) \ V_2(k) \ \dots \ V_{n_v}(k)]^T$ denotes the bus voltage magnitude measurements, $P(k) = [P_1(k) \ P_2(k) \ \dots \ P_{n_p}(k)]^T$ and $Q(k) = [Q_1(k) \ Q_2(k) \ \dots \ Q_{n_p}(k)]^T$ stand

for the real and reactive bus power injections measurements, and $P^f(k) = [P_1^f(k) \ P_2^f(k) \ \cdots \ P_{n_l}^f(k)]^T$ and $Q^f(k) = [Q_1^f(k) \ Q_2^f(k) \ \cdots \ Q_{n_l}^f(k)]^T$ are the real and reactive transmission line power flows, respectively. $G_{sj} + jB_{sj}$ is the sj th element of the complex bus admittance matrix, $g_{tj} + jb_{tj}$ is the admittance of the series branch connecting bus t and j , $g_{tj0} + jb_{tj0}$ is the half admittance of the shunt branch of the line collecting bus t and j in the π -model circuit, and \mathbb{N}_s is the set of bus numbers which are directly connected to bus s .

Taking the measurement noise into consideration, $z^{(r)}(k)$ can be rewritten as the following compact form

$$z^{(r)}(k) = h(x(k)) + v_1(k) \quad (4)$$

The nonlinear function $h(x)$ is given as follows:

$$h(x(k)) = [V^T(k), P^T(k), Q^T(k), P^{fT}(k), Q^{fT}(k)]^T,$$

where the variables $V(k)$, $P(k)$, $Q(k)$, $P^f(k)$ and $Q^f(k)$ are defined in (3). $v_1(k)$ is the RTU measurement noise which is also a Gaussian noise with zero mean and covariance matrix $R_1(k)$. Assume $\omega(k)$ and $v_1(k)$ are uncorrelated with $x(0)$ and with each other.

Considering missing measurements, the *actual* measurement $z(k)$ is described by

$$z(k) = \Xi(k)h(x(k)) + v_1(k) \quad (5)$$

where $\Xi(k) = \text{diag}\{\gamma_1(k), \gamma_2(k), \dots, \gamma_m(k)\}$ with $\gamma_i(k)$ ($i = 1, 2, \dots, m$) being m unrelated random variables. $\Xi(k)$ is also unrelated with $\omega(k)$, $v_1(k)$ and $x(0)$. Furthermore, it is assumed that the stochastic variable $\gamma_i(k)$ is a Bernoulli-distributed white noise sequence taking values on 0 or 1 with:

$$\text{Prob}\{\gamma_i(k) = 0\} = 1 - \mu_i(k), \quad \text{Prob}\{\gamma_i(k) = 1\} = \mu_i(k)$$

where the value of $\text{Prob}\{\gamma_i(k) = 0\}$ is also called the missing rate of the i th measurement.

In RTU measurements, one bus is usually chosen as the reference bus for all the other buses to obtain the relative phase angles, while in PMU measurements, all PMU measurements provide the direct phase angles with respect to the time reference provided by the GPS system. In this paper, we use both RTU and PMU measurements, and therefore all the bus phase angles are relative to the reference dictated by the GPS [14]. As a result, no reference buses are needed.

B. PMU Measurements and Inequality Constraints

In this paper, both the state variables and measured variables are in the rectangular form, which makes a linear PMU measurement model. Assume that the l th PMU is installed at bus j , and the measurement $z_l^{(p)} \in \mathbb{R}^{2(1+N_j)}$ can be described as follows

$$z_l^{(p)} = \begin{bmatrix} z_{r,j}^{(p)} & z_{i,j}^{(p)} & z_{r,t_1^j}^{(p)} & z_{i,t_1^j}^{(p)} & \cdots & z_{r,t_{N_l}^j}^{(p)} & z_{i,t_{N_l}^j}^{(p)} \end{bmatrix}^T.$$

To be more specific, the voltage measurement in the above vector is given as follows

$$z_{r,j}^{(p)} = x_{r,j}, \quad z_{i,j}^{(p)} = x_{i,j} \quad (6)$$

The current measurement of the line collecting the bus j and t is as follows

$$\begin{aligned} z_{r,jt}^{(p)} &= x_{r,t}g_{jt0} - x_{i,j}b_{jt0} + (x_{r,j} - x_{r,t})g_{jt} - (x_{i,j} - x_{i,t})b_{jt}, \\ z_{i,jt}^{(p)} &= x_{i,t}g_{jt0} + x_{r,j}b_{jt0} + (x_{i,j} - x_{i,t})g_{jt} + (x_{r,j} - x_{r,t})b_{jt}. \end{aligned}$$

Considering the measurement noise, the PMU measurements can be presented in the following compact vector form:

$$z^{(p)}(k) = H^{(p)}x(k) + v_2(k) \quad (7)$$

where $z^{(p)}$ is the PMU measurement vector, and $v_2(k)$ is the PMU measurement noise, which is also a Gaussian noise with zero mean and covariance matrix $R_2(k)$. $H^{(p)}$ can be obtained directly from PMU configurations, and it can be found that the measurement $z^{(p)}(k)$ is linearly related to the state $x(k)$.

A seemingly natural idea is to treat the PMU measurements as an additional set similar to the RTU measurements. Note the fact reported in [3], [14], that the standard deviation of the errors of PMU measurements is one to two order magnitude less than the one of traditional RTU measurements. Unfortunately, since the PMU measurements are much more accurate than the RTU measurements, including these two kinds of measurements in the estimation process often results in ill-conditioned filtering procedure due primarily to the low covariance matrix for the PMU measurement noises.

As $R_2(k)$ is always a real symmetric matrix, we can find a transformation matrix $M(k)$ of appropriate dimension such that the matrix $M(k)R_2(k)M^T(k)$ is diagonal. Accordingly, we can obtain the following equation from (7):

$$M(k)z^{(p)}(k) = M(k)H^{(p)}x^p(k) + M(k)v_2(k) \quad (8)$$

where $M(k)v_2(k)$ is still a Gaussian noise with zero mean and covariance matrix $M(k)R_2(k)M^T(k)$.

Based on the well-known 3-sigma rule of Gaussian distribution, we can conclude that the following inequality sets are satisfied with probability 99.7%:

$$-3\tilde{R}_2(k) \leq M(k)z^{(p)}(k) - M(k)H^{(p)}x^p(k) \leq 3\tilde{R}_2(k) \quad (9)$$

where $\tilde{R}_2(k) := M(k)R_2(k)M^T(k)I_{m,1}$. From the perspective of engineering applications, it is reasonable to assume that the above inequality sets are satisfied all the time. So far, we have characterized the PMU measurements by a set of inequality constraints on the states for the power system.

III. FILTER SCHEMES

A. EKF Design for the System with RTU Measurements

In this subsection, we first introduce the EKF approach to estimating the system state for the system (2) with missing measurements (5). The EKF is of the following form:

$$\hat{x}(k|k-1) = A\hat{x}(k-1|k-1) + Bu$$

$$\hat{x}(k|k) = \hat{x}(k|k-1) + K(k)[z(k) - \Xi(k)h(\hat{x}(k|k-1))]$$

where $\hat{x}(k|k)$ is the estimate of $x(k)$ at time instant k with $\hat{x}(0|0) = \bar{x}(0)$, and $\hat{x}(k|k-1)$ is the one-step prediction of $x(k)$ at time $k-1$. $K(k)$ is the filter gain to be determined at time instant k , and $\Xi(k) := \mathbb{E}\{\Xi(k)\} = \text{diag}\{\mu_1(k), \mu_2(k), \dots, \mu_m(k)\}$. $P(k|k-1)$ and $P(k|k)$ are

the covariance matrices of, respectively, the one-step prediction error and the filtering error defined by

$$\begin{aligned}\tilde{x}(k|k-1) &= x(k) - \hat{x}(k|k-1), \quad \tilde{x}(k|k) = x(k) - \hat{x}(k|k), \\ P(k|k-1) &= \mathbb{E}\{\tilde{x}(k|k-1)\tilde{x}(k|k-1)^T\}, \\ P(k|k) &= \mathbb{E}\{\tilde{x}(k|k)\tilde{x}(k|k)^T\}.\end{aligned}$$

Denoting $H(k) = \left. \frac{\partial h(x(k))}{\partial x(k)} \right|_{x(k)=\hat{x}(k|k-1)}$, the gain $K(k)$ can be obtained using the following recursive algorithm:

$$P(k|k-1) = AP(k-1|k-1)A^T + W(k-1) \quad (10)$$

$$P(k|k) = [I - K(k)\tilde{\Xi}(k)H(k)]P^{-1}(k|k-1) \quad (11)$$

$$\begin{aligned}S(k) &= \tilde{\Xi}(k) \circ (h(\hat{x}(k|k-1))h^T(\hat{x}(k|k-1))) \\ &\quad + \tilde{\Xi}(k) \circ (H(k)P(k|k-1)H^T(k)) + R(k) \\ &\quad + \tilde{\Xi}(k)H(k)P(k|k-1)H^T(k)\tilde{\Xi}(k) \quad (12)\end{aligned}$$

$$K(k) = P(k|k-1)H^T(k)\tilde{\Xi}(k)S^{-1}(k) \quad (13)$$

where $\tilde{\Xi}(k) := \text{diag}\{\tilde{\mu}_1(k), \tilde{\mu}_2(k), \dots, \tilde{\mu}_m(k)\}$ with $\tilde{\mu}_i(k) = \mu_i(k)(1 - \mu_i(k))$ ($i = 1, 2, \dots, m$).

Remark 1: In this paper, the exact occurrence time for the randomly missing measurements is not required to be exactly known, and this reflects the practical situation in power system. Nonetheless, the statistical law (i.e., the first- and second- order of moments) of the random occurrence of missing measurements is needed in the filter design, where the statistical law could be obtained through statistic tests.

Remark 2: There are mainly two kinds of DSE paradigms in power system state estimation. These two paradigms differ from each other in system dynamics model and time scale. In one paradigm (see e.g. [5], [6], [20] and the references therein) called forecasting-aided state estimation, the bus voltages are chosen as state variables and a succession of the quasi steady-states is assumed to evolve in time. Therefore, a dynamic model is adopted to describe the slow time evolution of the quasi steady-state. In the other paradigm (see e.g. [9] and the references therein), rotor angles and rotor speeds of generators are chosen as state variables, and the classic dynamic model of generators is considered. The DSE of such a paradigm is concerned with the low frequency electromechanical dynamics.

B. The Probability-Maximum Method

For the constrained estimation problem, it is difficult to incorporate the inequality/equality constraint of system states into traditional EKF estimator. Fortunately, the probability-maximum method has been successfully exploited in [17] to convert the constrained estimation problem into a constrained optimization one after each step of the EKF algorithm and, therefore, this method is chosen to handle the constrained EKF problem in this paper.

For presentation conciseness, the notation for time instant, k , is omitted in this subsection. It is known from [2] that, based on the Kalman filter theory, the state estimate of x maximizes the conditional probability density

$$\mathbb{P}(x|Z) = (2\pi)^{-\frac{n}{2}} |P|^{-\frac{1}{2}} \exp\left\{-\frac{1}{2}(x-\bar{x})^T P^{-1}(x-\bar{x})\right\} \quad (14)$$

where n is the dimension of x , P is the covariance of the Kalman filter estimate, $Z \triangleq \{z(0), z(1), \dots, z(k)\}$ denotes

the set of measurements available at time instant $0, 1, \dots, k$, and \bar{x} is the conditional mean of x given Z .

The constrained EKF can be derived by finding an estimate \hat{x} such that the conditional probability $\mathbb{P}(\hat{x}|Z)$ is maximized and \hat{x} satisfies the constraint (9). Since maximizing $\mathbb{P}(\hat{x}|Z)$ is equivalent to maximizing its natural logarithm, the problem to be solved can be expressed as

$$\begin{aligned}\max \ln \mathbb{P}(\hat{x}|Y) &\Rightarrow \min(\hat{x} - \bar{x})^T P^{-1}(\hat{x} - \bar{x}) \\ \text{subject to } &-3\tilde{R}_2 \leq Mz^{(p)} - MH^{(p)}C\hat{x} \leq 3\tilde{R}_2.\end{aligned} \quad (15)$$

So far, the constrained state estimation problem has been converted into an equivalent constrained optimization problem that can be solved after each time step of the EKF algorithm. As is impossible to develop a deterministic method for the constrained nonlinear optimization problem (15) in the global optimization category, we adopt the PSO algorithm, which is a popular evolutionary algorithm in solving the nonlinear optimization problem.

IV. PSO FOR CONSTRAINED OPTIMIZATION PROBLEM

Particle Swarm optimization (PSO) is a metaheuristic that optimizes a problem by iteratively searching in a large spaces of candidate solutions [13]. In PSO, a population of candidate solutions (called as particles) moves in the search space according to two simple mathematic formulae over the particle's position and velocity. More specifically, each particle's movement is influenced by its local best known position and also the best known positions, which are updated by other particles, in the search space. By such an iterate approach, the swarm of the particles moves towards the best solutions. The velocity and position of the particle at the next iteration are updated according to the following equations:

$$\begin{cases} v_i(s+1) = \omega v_i(s) + c_1 r_1 (p_i(s) - x_i(s)) \\ \quad + c_2 r_2 (p_g(s) - x_i(s)) \\ x_i(s+1) = x_i(s) + v_i(s+1) \end{cases} \quad (16)$$

where $x_i(s) = [x_{i1}(s), \dots, x_{id}(s)]$, $x_i(s)$ is the position of the i th particle at the s th iteration, and $x_i(s) \in [x_{min,n}, x_{max,n}]$, with $x_{min,n}$ and $x_{max,n}$ being the lower and the upper bounds for all particles' positions. $v_i(s) = [v_{i1}(s), \dots, v_{id}(s)]$, $v_i(s)$ is the velocity of the i th particle at the s th iteration. ω is the inertia weight, c_1 and c_2 are called acceleration coefficients, namely, cognitive and social parameters, respectively. r_1 and r_2 are two uniform random number samples from $[0, 1]$. $p_i(s)$ is the local best position encountered by i th particle at the s th iteration, and $p_g(s)$ is the global best position in the swarm at the s th iteration.

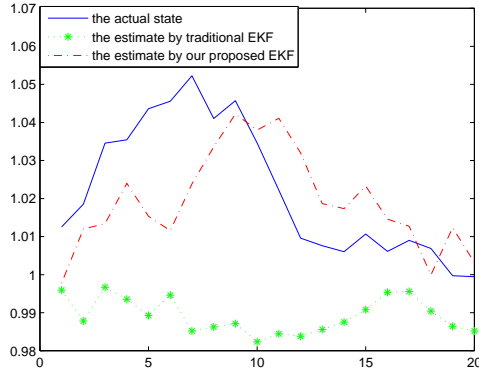
PSO has been successfully applied to various optimization problems. As to constrained optimization problem, PSO is still valid with the aid of the popular constraint-handling technique: the penalty function approach. By using the penalty function approach, a constrained optimization problem can be converted into a corresponding unconstrained optimization one by adding a penalty term to the original objective function [23].

In this paper, the penalty function $F(x)$ is defined as

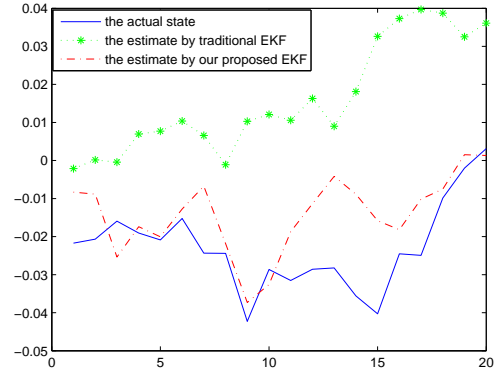
$$F(x) = f(x) + h(s)g(x), x \in \mathbb{R}^n \quad (17)$$

TABLE I
THE TREND VOLTAGE AT NORMAL STATES

Bus	1	2	3	4	5	6	7	8	9	10	11	12	13	14
Real Parts	1.0600	1.0368	0.9609	0.9858	0.9958	1.0016	1.0022	1.0270	0.9827	0.9769	0.9850	0.9806	0.9755	0.9552
Imaginary Parts	0	0.0943	0.2173	0.1821	0.1563	0.2694	0.2512	0.2643	0.2743	0.2744	0.2724	0.2759	0.2748	0.2812

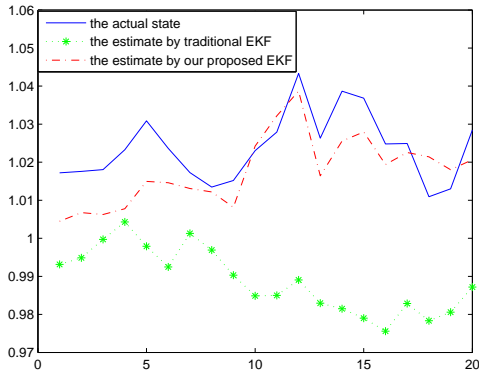


(a) The real part.

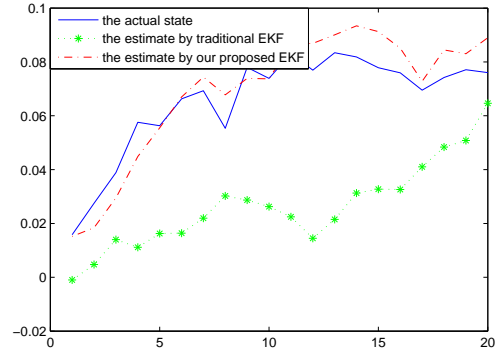


(b) The imaginary part.

Fig. 1. The estimated states of bus 2 from the traditional EKF and our proposed EKF



(a) The real part.



(b) The imaginary part.

Fig. 2. The estimated states of bus 5 from the traditional EKF and our proposed EKF

where $f(x)$ is the original objective function of the constrained optimization problem in (15), $h(s)$ is a dynamic penalty coefficients with the s th iteration steps, $g(x)$ is a penalty factor defined as $g(x) = \theta \sum_{i=1}^{m_1} q_i^2(x)$. Here $q_i(x) = \max\{0, c_i(x)\}$ with $c_i(x) = \frac{|M_i(k)(z^{(p)} - H^{(p)}x^p)|}{3\tilde{R}_{2i}(k)} - 1$, $i = 1, \dots, m_1$, where $M_i(k)$ and $\tilde{R}_{2i}(k)$ are the i th row of $M(k)$ and $\tilde{R}_2(k)$ in inequality (9).

V. SIMULATION RESULTS

In this section, the proposed hybrid algorithm of EKF and PSO is tested in the case study of the IEEE 14-bus test system. The simulation is implemented in Matlab with the Matpower package [25]. First, the IEEE 14-bus test system can be model as (1) with parameters $A = \text{diag}_{28}\{0.98\}$, $B = \text{diag}_{28}\{0.02\}$ and $W(k) = \text{diag}_{28}\{0.01^2\}$. The trend u of the normal state is the base-case voltages given in Table I. Furthermore, assume

that the initial voltages of all buses are at flat start, that is, $x_{r,l}(0) = 1$ p.u., $x_{i,l}(0) = 0$ for all $l = 1, 2, \dots, 14$.

The measurement configuration is the same as the one used in [14], where RTU measurements consist of three categories: the voltage magnitude at bus 1, power injections at bus 3, 5, 13 and 14, and power flows at branches 1-2, 1-5, 2-5, 3-4, 4-7, 4-9, 6-11, 6-12, 6-13, 7-8, 7-9, 9-10, 9-14, 10-11, 12-13 and 13-14. In addition, PMUs are deployed at buses 2, 7 and 9. Furthermore, the covariance matrices of the traditional RTU measurement and PMU measurement noise are $R_1(k) = \text{diag}_{43}\{0.1^2\}$ and $R_2(k) = \text{diag}_{28}\{0.01^2\}$, respectively.

The algorithm is implemented in Matlab R2010a. The simulation is performed on a PC with a Intel(R) Core(TM) CPU i5-2500 @3.30 GHz and 4 GB RAM. The time required by the proposed EKF without PSO algorithm at each step is 0.81 seconds. For the proposed EKF with PSO algorithm, the computation time is related to the population of the swarm (ps) and the iterations (iter). In the simulation, we

have set $ps = 100$ and $iter = 200$, and the time required by the proposed EKF with PSO algorithm at each step is 1.47 seconds. It can be concluded that the proposed EKF with PSO algorithm is quite fast and hence is suitable for online implementations. Moreover, the integration of PSO into EKF slows the computational speed slightly, yet improves the performance of state estimation obviously.

In this test system, three comparative experiments regarding the estimation accuracy are carried out as follows:

- Case 1) Both the proposed EKF considering measurements with certain missing rate and the traditional EKF ignoring the missing measurements are implemented for the system with missing measurements;
- Case 2) When the missing rate of the measurements varies from zero to higher values, the proposed EKF is implemented in all the cases;
- Case 3) The state estimations based on the proposed EKF with/without PSO algorithm are compared.

In order to have more general and significant experimental results, 100 Monte-Carlo simulations are run in Cases 2 and 3. The notion mean square error (MSE) is adopted to evaluate the estimation accuracy, where MSE_i denotes MSE for the estimate of the i th state, i.e. $MSE_i(k) = \frac{1}{100} \sum_{j=1}^{100} (x_i(k) - \hat{x}_i(k))^2$. To evaluate the average estimation performance of all states, average mean square error (AMSE) is defined as $AMSE(k) := \frac{1}{n} \sum_{j=1}^n MSE_j(k)$, where n is the number of the state variables. In all the figures, “R.V” and “I.V” denote the real and imaginary part of voltage, respectively.

A. Traditional EKF vs. the Proposed EKF

In this case, the probability density function for the missing $\Xi(k)$ is $\text{Prob}\{\Xi_i(k) = 0\} = 0.5$, $\text{Prob}\{\Xi_i(k) = 1\} = 0.5$. The expectation can be easily calculated as $\mu_i(k) = 0.5$. The estimated states of the representative buses 2, 5 obtained from traditional EKF without considering the missing measurements and our proposed EKF considering missing measurements are plotted in Figs. 1 and 2, respectively. From the comparison, it can be found that our proposed EKF algorithm performs well in the presence of missing measurements, whereas the state estimate obtained from the traditional EKF cannot track the real states when missing measurements occur randomly.

B. EKF with Individual Missing Measurements

In order to see how different missing rates impact on the estimation accuracy, three missing rates of 0.15, 0.02 and 0 (without missing measurements) are considered. The MSEs of the estimated states of buses 2, 5 for all the three missing rates are compared in Figs. 3 and 4. The $AMSE(k)$ in all three cases are given in the first three rows of Table II, for $k = 1, \dots, 15$. From the comparisons, it can be found the less the missing rate is, the more accurate the state estimation obtained from the proposed EKF algorithm will be.

C. EKF vs. Hybrid EKF and PSO Algorithm

We are now in a position to evaluate the effectiveness of including the PSO scheme in the EKF design. A comparison

is made between the EKF algorithm alone and the hybrid EKF and PSO algorithm. For this purpose, the missing rate is fixed as 0.15. Regarding the penalty function parameters, $\theta = 1000$ and $h(s) = s$ are chosen in all the iteration steps.

For the same test system, one realization of the EKF and one realization of the hybrid algorithm are simulated simultaneously, and the estimated states of bus voltages 2, 5 obtained from the two algorithms are illustrated in Figs. 5 and 6. It is seen that the trajectory by the proposed hybrid approach is much closer to the true state trajectory than the one only by the EKF. The MSE_2 and MSE_5 at all time instants for both algorithms are plotted in Fig. 7. It can be found that for the same state variable, the MSE of EKF-based state estimation is bigger than the MSE of the state estimation obtained from the hybrid algorithm. Especially, when the accumulated error of EKF-based state estimation becomes bigger after several integrations, the subsequent PSO algorithm can refine the state estimation and diminish the error. The AMSEs of EKF and of the proposed hybrid algorithm are given in the last two rows of Table II. It can be found the $AMSE(k)$ of EKF is bigger than the one of the proposed hybrid algorithm at each step.

From the comparative experiments, it can be concluded that our proposed hybrid EKF and PSO algorithm outperforms the traditional EKF algorithm in the presence of probabilistic missing measurements by including PMU measurements.

VI. CONCLUSION

In this paper, we have developed a hybrid EKF and PSO algorithm for power system dynamic state estimation. In consideration of the missing traditional measurements, a novel EKF estimator has been designed for the power system. The PMU measurements have been incorporated in the designed EKF estimator via the characterization of a set of inequality constraints. The state estimation problem has been transformed to a constrained optimization problem. Then, the PSO algorithm together with the penalty function has been employed to solve the constrained optimization problem. Simulations have confirmed the effectiveness of the propose method.

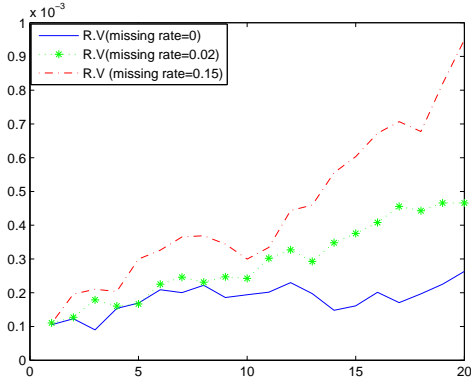
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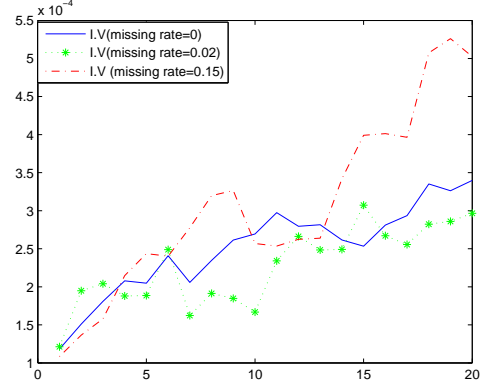
TABLE II
THE AMSE FOR ESTIMATED STATES BY DIFFERENT ALGORITHMS WITH DIFFERENT MISSING RATES, WHERE MR DENOTES THE MISSING RATE

Time instant	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
MR = 0.00 (EKF)	0.1350	0.1609	0.1748	0.2121	0.2342	0.2432	0.2366	0.2564	0.2767	0.2651	0.2707	0.2601	0.2572	0.2503	0.2751
MR = 0.02 (EKF)	0.1344	0.1784	0.2091	0.2134	0.2123	0.2356	0.2550	0.2425	0.2787	0.2884	0.2851	0.3136	0.3202	0.3253	0.3578
MR = 0.15 (EKF)	0.1416	0.1836	0.2152	0.2596	0.2673	0.2913	0.3341	0.3350	0.3590	0.3813	0.3858	0.4410	0.4615	0.4801	0.4612
MR = 0.15 (Hybrid)	0.1342	0.1725	0.2066	0.2047	0.2226	0.2450	0.2682	0.2818	0.3135	0.2985	0.2958	0.2950	0.3050	0.3199	0.3049

Note that the results in the table are magnified 10^3 times for clarity.

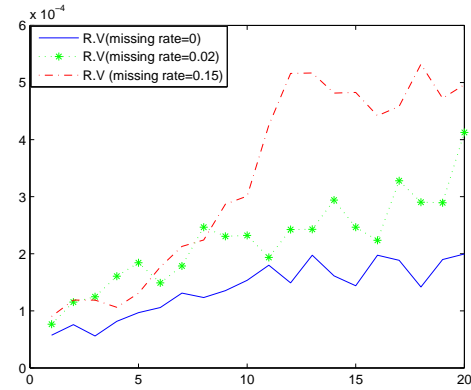


(a) The real part.

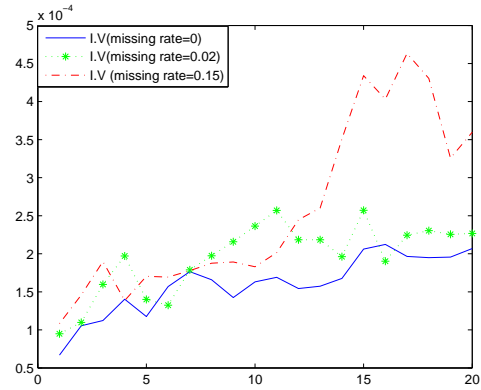


(b) The imaginary part.

Fig. 3. The MSEs of the estimated state at bus 2 under different missing rates

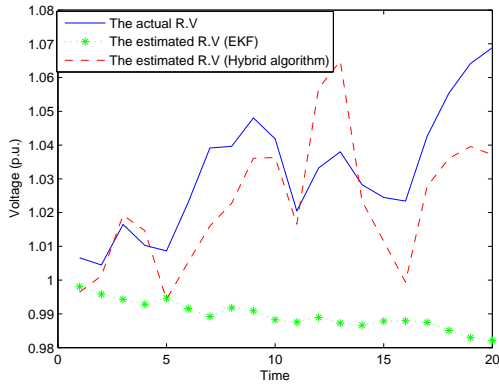


(a) The real part.

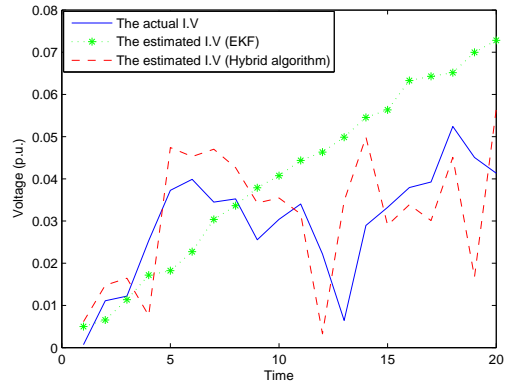


(b) The imaginary part.

Fig. 4. The MSEs of the estimated state at bus 5 under different missing rates

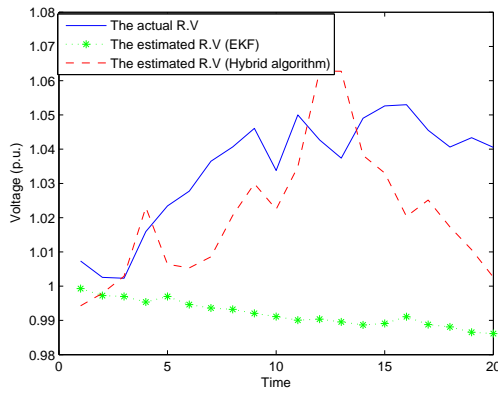


(a) The real part.

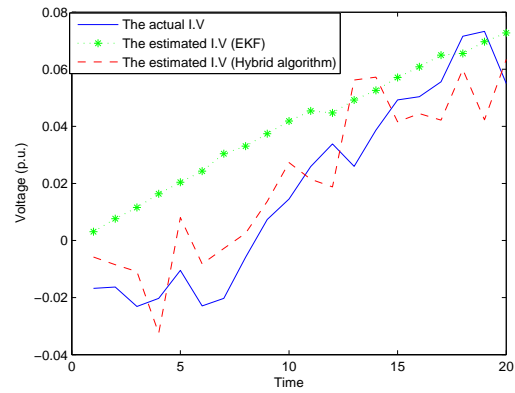


(b) The imaginary part.

Fig. 5. The estimated state at bus 2 by EKF and the proposed hybrid algorithm

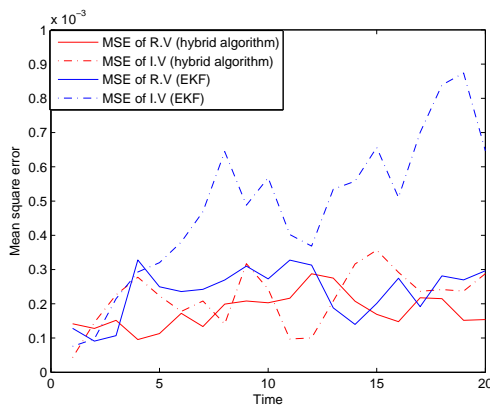


(a) The real part.

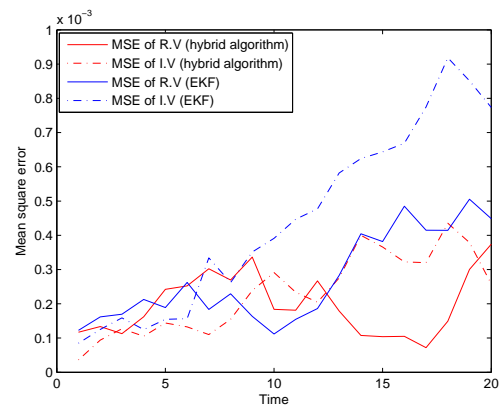


(b) The imaginary part.

Fig. 6. The estimated state at bus 5 by EKF and the proposed hybrid algorithm



(a) Bus 2.



(b) Bus 5.

Fig. 7. The MSEs of estimated states by EKF and the proposed hybrid algorithm

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