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A SINGLE/DOUBLE PRECISION FLOATING-POINT MULTIPLIER DESIGN FOR MULTIMEDIA APPLICATIONS

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ABSTRACT

Modern graphic processors, multimedia processors, and general-purpose processors with multimedia extensions provide SIMD floating-point instructions. SIMD floating-point multiplication is commonly used in 2D and 3D applications, which mostly use single precision floating-point operands. Consequently, efficient single precision multiplier units are crucial for high performance systems. This paper introduces a novel floating-point multiplications. The proposed design uses approximately 10% more hardware and has 33% more delay compared to a conventional double precision floating-point multiplier.

Keywords: Multimedia, floating-point numbers, multiply

1. INTRODUCTION

Modern general purpose processors provide special instructions for multimedia applications [1]. New generations of the general purpose processors use larger sets of multimedia instructions than the ones offered by the previous generations, since the variety of the operations and the performance requirement for multimedia applications increase [1-6]. Consequently, providing efficient multimedia hardware has become an important design task. In the past,

Received Date: 12.08.2008 Accepted Date: 05.01.2009 modifying only integer data path were enough to implement most multimedia instructions; however, today, the floating-point data path is modified as well, since many applications use floating-point operands [2-3].

Floating-point multiplication is one of the most supported operation in multimedia extensions offered by popular chip producers. For example, Intel's Pentium 3 has multimedia extensions SSE and SSE2 that support the multiplication of two double precision or four single precision floating-point numbers [5], similarly AMD's 3DNow multimedia extension contains an instruction that can perform two simultaneous single precision floating-point multiplications [6]. The AltiVec extension of PowerPC also contains an instruction that executes four singleprecision floating-point multiplications [7].

IEEE 754 describes single and double-precision floating-point number formats. A doubleprecision floating-point multiplier can be used to perform a single precision multiplication, however, this requires additional processing, since the operands should be translated to double precision, and after the multiplication, the result should be translated back to single precision format [2]. Furthermore, a double precision floating-point multiplier can perform a single multiplication per cycle. The Dual Mode IEEE multiplier design presented in [2] performs two single precision multiplications using a double precision multiplier. However, the proposed multiplier in [2] needs a delay cycle before starting a new multiplication.

This paper presents a new floating-point multiplier which can perform a double-precision floating-point multiplication or two simultaneous single precision floating-point multiplications. Since in single precision floating-point multiplication two results are generated in parallel, the multiplier's performance is almost doubled compared to a conventional floatingpoint multiplier.

2. PROPOSED FLOATING-POINT MULTIPLIER DESIGN

This section presents proposed design, which performs the following floating-point multiplications described in Figure 1.

Figure 1a shows the alignments of two double precision floating point numbers X, Y and their product Z that are placed in three 64-bit registers. Figure 2b shows the alignments of four single precision floating-point numbers A,B,C and D and the product of A and B, H, and the product of C and D, J that are placed in three 64-bit registers.

The multiplication of X and Y is performed as

$$E_z = E_x + E_y \tag{1}$$

$$M_{z} = M_{x} \times M_{y} \tag{2}$$

$$S_{z} = S_{y} \otimes S_{y} \tag{3}$$

The multiplication of A and B, and the multiplication of C and D are performed as

$$E_h = E_a + E_h, E_i = E_c + E_d \tag{4}$$

$$M_h = M_a \times M_b, \ M_j = M_c \times M_d \tag{5}$$

$$S_{ab} = S_a \otimes S_b, \ S_{ab} = S_a \otimes S_b \tag{6}$$

The proposed design performs these two floating-point multiplications in parallel. In [9], a design method for the multiplication of two unsigned integer operands is presented. Figure 2 presents the adaptation of the technique in [9] to implement the proposed method. In this figure, the matrices generated for the two single precision floating-point multiplications are placed in the matrix generated for a double precision floating-point multiplication. All the bits are generated in double precision mode, the shaded areas Z1, Z2, Z3 are not generated when single precision multiplication is performed, the non-shaded regions designate the generated bits.

	63	6252	510
Х	Sx	Ex	M _x
Y	Sy	Ey	My
Ζ	Sz	Ez	Mz

(a) The alignment of two double precision floating numbers and their product.

÷								
		63	6255	5432		31	3023	220
	А	S_a	Ea	M_{a}	С	S_{c}	Ec	M_{c}
	В	S_b	Eb	M_{b}	D	S_d	E_{d}	\mathbf{M}_{d}
	Η	S_h	E _h	M_{h}	J	Si	E	Mi

(b) The alignment of four single prezision floating point numbers and their products.

Figure 1. The alignments for double and single precision numbers.

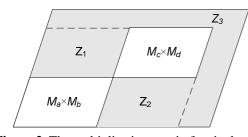


Figure 2. The multiplication matrix for single and double precision mantissas.

The partial products within the regions Z1, Z2 and Z3 are generated using following equations

$$\hat{b}_j = s \cdot b_j \text{ and } p_{ij} = a_i \cdot \hat{b}_j$$
 (7)

,the rest of the partial products are produced with

$$p_{ii} = a_i \cdot b_i \tag{8}$$

's' is used as control signal. When s = 0', only the bits in no shaded regions are generated. When s = 1', all bits are generated.

High-speed multipliers reduce the partial product matrix to two vectors using a reduction method. Then, these two vectors are added to produce the result with carry-propagate adder. The reduction method and the type of the carry-propagate adder are not important for the proposed design, since it only modifies the generation of the partial products. also means that the reduction algorithm and the carry-propagate adder is not modified for the implementation of the proposed method.

Figure 3 presents a block diagram for a standard floating-point multiplier. This multiplier directly implements Eqns. (1)-(3). Figure 4 presents the proposed single/dual floating-point multiplier which is designed by slightly modifying the standard floating-point multiplier. The modifications can be used on every type of double precision floating-point multiplier. The data flow and the functionality of each unit in the proposed design are explained as follows:

- The control signal determines the mode of execution; when s='0' double precision floating-point multiplication is performed, otherwise two single precision multiplications are performed.
- The 11-bit adder is used for double precision exponent addition and two 8-bit adders are used for single precision exponent additions.
- The exponent updaters remove extra bias values form the exponent additions.
- The mantissa modifier implements the logic equations given in Eqn 7.
- The mantissa multiplier generates carry-save vectors.
- The add, normalize and round unit generates normalized and rounded result.
- The signs of the products are obtain by XOR gates.

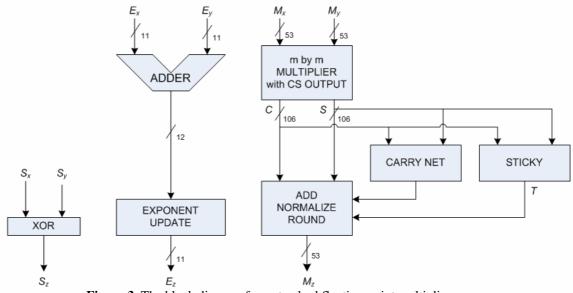


Figure 3. The block diagram for a standard floating-point multiplier.

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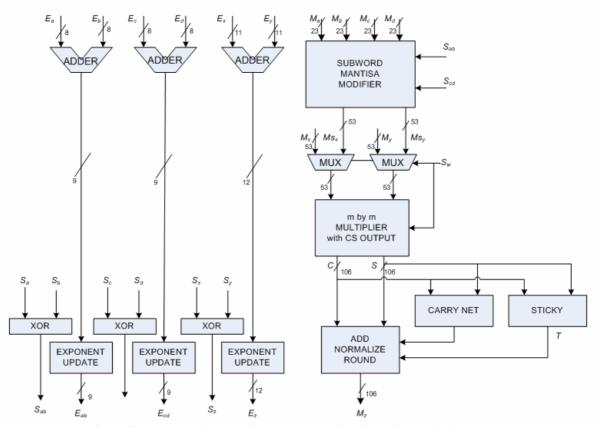


Figure 4. The block diagram for the proposed floating-point multiplier.

3. SYNTHESIS RESULTS

In this section we present the synthesis results for the proposed single/double precision floating point multiplier and the standard dual precision floating-point multiplier. Both circuits are modeled using structural VHDL code. Synthesis of the circuit is done using TSMC 0.18 micron ASIC library and Leonarda Spectrum program. Both circuits are optimized for delay. The values in Table 1 are in nanoseconds for time and in number of gate for area.

Table 1. The comparison of the standard double precision and proposed floating-point multiplier designs.

Design	# of Gates	Latency
Double Precision	25175	4.10 Ns
Single/Double Precision	27566	5.49 Ns

The single/double precision multiplier has approximately 9.49% more area and has about 34% more critical delay. The floating-point multipliers used in modern processors are usually pipelined designs. If the proposed method is applied to a pipelined multiplier the area increase is expected to fall down below 5% and also critical delay increase will be dissolved in pipeline stages.

4. CONCLUSION

In this study a floating point multiplier design which is capable of executing a double precision floating-point multiplication or two parallel single precision multiplications is presented. One of the important aspects of the presented design method is that it can be applicable to all kinds of floating-point multipliers. The presented design is compared with a standard floating point multiplier via synthesis. The synthesis results showed that proposed design is %10 larger than

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conventional multiplier and critical path increment is only one or two gate delay. Since modern floating-point multiplier designs have significantly larger area than the standard floating-point multiplier, the percentage of the extra hardware will be less for those units. The methods presented in this study will be used on the design of floating-point multiplier-adder circuits. Also, the future work will enhance the proposed designs to support all IEEE754 rounding modes.

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