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# A Snapback-Free and Low Turn-Off Loss Reverse-Conducting SOI-LIGBT With Embedded Diode and MOSFET

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**ABSTRACT** A novel snapback-free and low turn-off loss reverse-conducting (RC) SOI-LIGBT is proposed and investigated by numerical simulations. An n-MOSFET ( $M_{N2}$ ) is embedded in the anode side of the LIGBT to short the P-anode/N-buffer junction during the turn-off transient, thus allowing the LIGBT to be turned off rapidly without excessive tail current. In addition,  $M_{N2}$  enable the LIGBT to conduct the reverse conducting current like the freewheeling diode. In the forward-conducting state,  $M_{N2}$  is turned off, then the proposed LIGBT operates like a conventional one and the snap-back is avoided. The gate electrode of  $M_{N2}$  can be controlled synchronously by the gate signal of the LIGBT which is level-shifted by a p-i-n diode ( $D_1$ ) and processed by an anode-controlling circuit, and therefore, the proposed RC-LIGBT still maintains a three-terminal configuration.  $D_1$  and  $M_{N2}$  are embedded in the drift region and anode-side of the LIGBT, respectively, and they can be isolated by deep-oxide trenches. The numerical simulation results reveal that the turn-off loss ( $E_{off}$ ) and reverse recovery charge of the proposed LIGBT is reduced by 58.3% and 38.9%, respectively, compared with the conventional LIGBT combining with antiparallel freewheeling diode.

**INDEX TERMS** RC-LIGBT, snapback-free, turn-off loss, reverse recovery charge.

#### I. INTRODUCTION

The lateral insulated gate bipolar transistor (LIGBT) on silicon-on-insulator (SOI) has been widely used in the medium voltage range of smart power integrated circuits [1], [2]. As a unidirectional device, the conventional LIGBT needs an antiparallel diode in various power applications to conduct reverse current, which induces extra cost and parasitic effects [3]. The shorted-anode LIGBT implements reverse-conducting by introducing an  $N^+$  region in the anode side of the LIGBT, which undesirably leads to the snap-back effect in the forward-conducting state [4], [5]. Some methods have been reported to suppress the snap-back problem [6]–[9] by increasing the resistor for the electron current from the  $N^+$  anode to the  $P^+$  anode, which causes a higher on-state voltage drop ( $V_{\rm on}$ ) compared with that of

the conventional LIGBT. Dual-gate devices [10]–[13] also have been reported to be able to eliminate the snap-back effect and improve the turn-off speed, nevertheless, these devices require extra control circuits or high voltage devices to switch the anode-gate.

In this paper, we propose a novel RC-LIGBT with embedded p-i-n diode  $(D_1)$  and n-MOS transistor  $(M_{N2})$ . The gate signal of the LIGBT which is level-shifted by  $D_1$  and processed by an anode-controlling circuit can synchronously control  $M_{N2}$  to short the anode junction of the LIGBT during the turn-off transient and the reverse-conducting state. Thus, the LIGBT is a three-terminal device without extra controlling signal. The snap-back phenomenon is eliminated in forward-conducting state. The turn-off loss and reverse recovery charge are reduced. The electrical characteristics

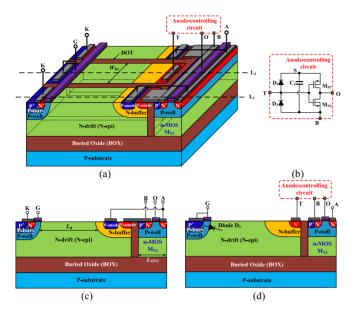


FIGURE 1. (a) 3-D cell view of the proposed LIGBT with (b) the block diagram of the anode-controlling circuit. Cross sections of the proposed structure along the line L<sub>1</sub> (c) and L<sub>2</sub> (d).

of the proposed RC-LIGBT are investigated by Sentaurus TCAD tools.

#### II. DEVICE STRUCTURE AND MECHANISM

Fig. 1(a) shows the 3-D cell view of the proposed RC-LIGBT. The inside and outside part of the proposed structure are all active LIGBT region, and the middle part isolated by deep-oxide trench is D<sub>1</sub>, which has the same breakdown voltage (BV) as the LIGBT region. Its width  $(W_{D1})$ is 10  $\mu$ m. On the right side of the cell,  $M_{N2}$  is formed and also isolated by the deep-oxide trench. The drain electrode and source electrode of M<sub>N2</sub> are connected to the P-anode and N-anode of the LIGBT, named as electrodes A and B, respectively. The gate electrode of  $M_{N2}$  is connected to the output terminal (O) of the anode-controlling circuit. The input terminal (T) of the anode-controlling circuit is connected to the  $N^+$  cathode of  $D_1$ . The  $P^+$  anode of  $D_1$  is connected to the gate electrode (G) of the LIGBT. The anodecontrolling circuit, as depicted in Fig. 1(b), consists of two low-voltage diodes (D2 and D3), a power supply capacitor  $(C_1)$  and a CMOS inverter (consists of  $M_{P1}$  and  $M_{N1}$ ), which can be integrated on the right side of M<sub>N2</sub>. The contact B serves as the reference ground of the anode-controlling circuit. Table 1 lists the key design parameters for the proposed structure and the anode-controlling circuit. A conventional LIGBT and a freewheeling diode with the same parameters are also designed for comparison. The operation principle of the proposed RC-LIGBT with the anode-controlling circuit will be described later.

## A. POWER-ON PROCESS OF THE PROPOSED RC-LIGBT

As shown in Fig. 2(a), during the power-on process,  $V_{\rm GK} = 0$  V,  $V_{\rm AK}$  and  $V_{\rm BK}$  rise to the bus voltage. The diodes D<sub>2</sub>

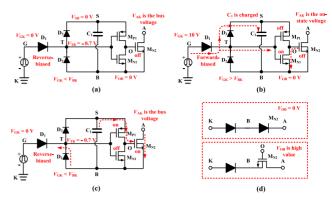


FIGURE 2. Operation of the proposed RC-LIGBT during (a) the power-on process, (b) the forward-conducting state, (c) the turn-off process, and (d) the reverse-conducting state.

is forward biased.  $D_1$  and  $D_3$  are reverse biased to make  $V_{TB} = -0.7$  V. Since the LIGBT has not been turned on before,  $C_1$  is not charged and  $V_{SB}$  (the voltage across  $C_1$ ) is 0 V. The output of the CMOS inverter ( $V_{OB}$ ) is 0 V and  $M_{N2}$  is in OFF-state.

## B. FORWARD-CONDUCTING STATE OF THE PROPOSED RC-LIGBT

As depicted in Fig. 2(b), during the forward-conducting state,  $V_{\rm GK}=10$  V, and  $V_{\rm BK}$  is close to the ON-state voltage drop of the LIGBT ( $V_{\rm BK}< V_{\rm GK}=10$  V). Thus, D<sub>1</sub> and D<sub>3</sub> are forward biased, and D<sub>2</sub> is reverse biased to make  $V_{\rm TB}$  rise from -0.7 V in the blocking state to a high value. C<sub>1</sub> is charged, M<sub>P1</sub> is turned off and M<sub>N1</sub> is turned on.  $V_{\rm OB}$  keeps to 0 V, and then, M<sub>N2</sub> is still off. Since the current path from the N-anode to the P-anode is blocked, the electron current can only flow through the P-anode, thus the proposed LIGBT can inject holes into the drift region like the conventional one without snap-back phenomenon.

## C. TURN-OFF PROCESS OF THE PROPOSED LIGBT

When  $V_{\rm GK}$  drops from 10 V to 0 V, both  $V_{\rm AK}$  and  $V_{\rm BK}$  will rise as the LIGBT is turned off ( $V_{\rm BK} > V_{\rm GK}$ ). Hence, D<sub>2</sub> is forward biased, D<sub>1</sub> and D<sub>3</sub> are reverse biased to make  $V_{\rm TB}$  drops from high value to -0.7 V.  $V_{\rm SB}$  maintains high value attributed to the fact that the voltage drop of the capacitor C<sub>1</sub> cannot be abruptly changed, and accordingly, M<sub>P1</sub> is turned on and M<sub>N1</sub> is turned off,  $V_{\rm OB}$  increases from 0 V to the value of  $V_{\rm SB}$ . Consequently, M<sub>N2</sub> is turned on, and the P-anode/N-buffer junction is shorted to prevent the injection of holes and thereby the long tail current is eliminated. After the LIGBT is turned off and in the blocking state, M<sub>N2</sub> will remain in the ON-state because  $V_{\rm TB} = -0.7$  V and C<sub>1</sub> has been charged until the LIGBT is turned on.

## D. REVERSE-CONDUCTING STATE OF THE PROPOSED LIGHT

Fig. 2(d) reveals the equivalent circuit diagram of the proposed LIGBT when it is in the reverse-conduction state.

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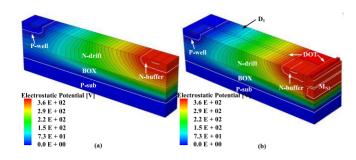


FIGURE 3. Potential contours of (a) the conventional LIGBT and (b) the proposed LIGBT at BV of 363 V and 362 V, respectively (10V/contour).

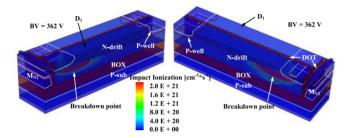


FIGURE 4. The total generation rate due to impact ionization of the proposed structure.

If the LIGBT is not turned on before, it is known from the previous discussion that  $M_{N2}$  is in the OFF-state. Hence, the reverse-conducting current through  $M_{N2}$  is mainly the body diode current. Then,the whole current path from electrode K to A consists of two diode in series. If the LIGBT has been turned on before,  $M_{N2}$  is in the ON-state, and the reverse-conducting current through  $M_{N2}$  is mainly the channel current. Consequently, the whole current path from electrode K to A consists of one diode and one n-MOSFET in series. In both cases, the reverse-conduction of the LIGBT can be implemented.

#### **III. SIMULATION RESULTS AND ANALYSIS**

The behaviors of the new device is investigated by 3-D numerical simulations. Since the proposed structure is symmetrical, half of its cell are used for simulation to reduce the number of grid points. Fig. 3 depicts the potential contours of the proposed LIGBT and the conventional one at BV of 362 V and 363 V, respectively (10 V/contour). The BV of the proposed structure is not significantly reduced in comparison with that of the conventional one. Fig. 4 exhibits the total generation rate of the proposed structure due to impact ionization. It can be seen that the breakdown point of the device locates at the bottom of the SOI layer, and there is no significant difference from that of other SOI lateral devices.

Fig. 5 depicts the conduction characteristics of the LIGBTs. The SSA LIGBT [6] is designed with the same parameters in Table 1 for comparison and the length between  $N^+$  anode to the N-buffer region ( $L_S$ ) is set as 20  $\mu$ m to suppress the snap-back effect. The width (z-dimensions) ratio of the conventional LIGBT to the conventional antiparallel

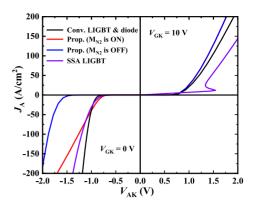


FIGURE 5. Conduction characteristics of the LIGBTs with  $\emph{V}_{GK}=$  10 V and 0 V.

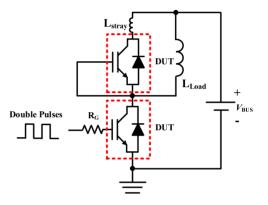


FIGURE 6. Mixed-mode inductive switching circuit with  $V_{\rm BUS}=$  150 V,  $L_{\rm Load}=$  1.5 mH,  $L_{\rm stray}=$  10 nH, and  $R_{\rm G}=$  10  $\Omega.$ 

freewheeling diode is 4:1, and the total area is considered for calculating the current density. For the proposed structure, the width ratio of the active LIGBT region to the  $M_{\rm N2}$  is also 4:1 and the total area is considered for calculating the current density. As shown in Fig. 5, in the forward-conducting state, the proposed structure is snapback-free since  $M_{\rm N2}$  is in OFF-state. Meanwhile, the current density of the proposed structure is higher than other structures when the total area is considered for calculating the current density. In the reverse-conducting state, the knee voltages of the proposed structure with  $M_{\rm N2}$  in OFF-state and in ON-state are 1.4 V and 0.7 V, respectively.

The mixed-mode inductive switching circuit shown in Fig. 6 is utilized to simulate the turn-off and reverse recovery characteristics. All of the three devices are turned off under the current of 1 A ( $J = 100 \text{ A/cm}^2$ ). The area of the proposed structure and the SSA LIGBT are all set as 1 mm<sup>2</sup>. And the total area of the conventional LIGBT combining with the conventional antiparallel freewheeling diode is set as 1 mm<sup>2</sup> for comparison. The lifetimes of the electrons and the holes are both set as 1  $\mu$ s.

Fig. 7 reveals the transient waveforms of  $V_{\rm GK}$  and  $V_{\rm OB}$  (the gate voltage of  $M_{\rm N2}$ ) during the switching process of the proposed LIGBT. When the LIGBT is in the turn-off process,  $V_{\rm GK}$  jumps from high level to low level and  $V_{\rm OB}$ 

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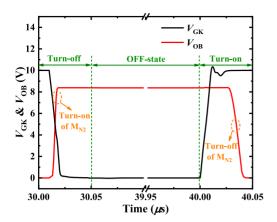


FIGURE 7. Transient waveforms of  $V_{GK}$  and  $V_{OB}$  during the switching process.

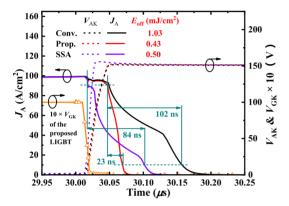


FIGURE 8. Turn-off characteristics of the three LIGBTs.

increases to high level almost simultaneously. Then,  $M_{N2}$  is turned on and the P-anode/N-buffer junction of the LIGBT is shorted. When the LIGBT is in the turn-on process,  $V_{GK}$  jumps from low level to high level and  $V_{OB}$  decreases to 0 V. Then,  $M_{N2}$  is turned off and the holes can be injected to the N-drift region from the P-anode with no snap-back effect.

The turn-off characteristics of the three LIGBTs are illustrated in Fig. 8. The proposed device exhibits the shorter fall time of tail-current and the lower turn-off loss ( $E_{\rm off}$ ) than the other two LIGBTs. The  $E_{\rm off}$  of the proposed device decreases by 58.3% at  $V_{\rm on}=1.38$  V, compared with the conventional LIGBT.

The reverse recovery characteristics of the proposed LIGBT, the SSA LIGBT and the antiparallel freewheeling diode are demonstrated in Fig. 9. The reverse recovery charge  $(Q_{rr})$  of the proposed LIGBT with  $M_{N2}$  in ON-state are less than that of the freewheeling diode by 38.9%. The reverse recovery charge and time of the proposed device with  $M_{N2}$  in OFF-state are larger and longer than that of the proposed device with  $M_{N2}$  in ON-state. This is because that during the reverse recovery process of the device with  $M_{N2}$  in OFF-state, more electrons are injected into P-anode which causes more holes to be injected into the N-drift region.

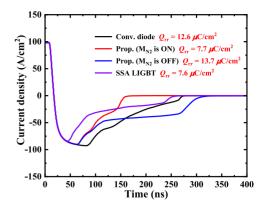


FIGURE 9. Reverse recovery characteristics of the proposed LIGBT, the SSA LIGBT and the conventional diode.

**TABLE 1.** Key design parameters.

Structure Parameters	Values	
BOX thickness (μm)	3	
N-drift thickness (μm)	6	
N-drift length, $L_{\rm d}$ ( $\mu$ m)	27	
N-drift concentration (cm <sup>-3</sup> )	$2 \times 10^{15}$	
P-well concentration (cm <sup>-3</sup> )	$2 \times 10^{17}$	
P-bury concentration (cm <sup>-3</sup> )	$3 \times 10^{18}$	
N-buffer concentration (cm <sup>-3</sup> )	$2 \times 10^{17}$	
P-anode concentration (cm <sup>-3</sup> )	$1 \times 10^{18}$	
N-anode concentration (cm <sup>-3</sup> )	$1 \times 10^{20}$	
Gate oxide thickness (nm)	50	
Width of deep-oxide trenches (μm)	1	
Width of the diode $D_1$ , $W_{D1}$ ( $\mu m$ )	10	
Length of the $M_{N2}$ , $L_{MN2}$ ( $\mu m$ )	3	
The value of capacitor C <sub>1</sub> (pF)	100	
The $W/L$ of $M_{P1}$ ( $\mu m$ )	200 / 1	
The $W/L$ of $M_{\rm Nl}$ ( $\mu m$ )	100 / 1	
The breakdown voltage of D <sub>2</sub> and D <sub>3</sub>	15 V	

TABLE 2. Comparison of the characteristics among these devices.

@ J = 100 A/cm <sup>2</sup>		Conv. LIGBT & diode	SSA LIGBT	The proposed LIGBT	
				M <sub>N2</sub> is ON	M <sub>N2</sub> is OFF
FC-State	$V_{\text{on-FC}}\left(\mathbf{V}\right)$	1.47	1.75	-	1.38
	$E_{\rm off}$ (mJ/cm <sup>2</sup> )	1.03	0.50	0.43	ı
	$FOM1 = V_{on} \times E_{off}$	1.514	0.875	0.593	
RC-State	$V_{\text{on-RC}}\left(\mathbf{V}\right)$	1.09	1.20	1.26	1.88
	$Q_{\rm rr} (\mu {\rm C/cm^2})$	12.6	7.6	7.7	13.7
	$FOM2 = V_{on} \times Q_{rr}$	13.734	9.12	9.702	25.756

Table 2 shows the comparison of the characteristics between the conventional LIGBT combining with freewheeling diode, the SSA LIGBT and the proposed LIGBT. In the forward-conducting state, the proposed structure exhibits superior tradeoff relationship between  $V_{\rm on}$  and  $E_{\rm off}$  and its figure of merit (FOM1 =  $V_{\rm on-FC} \times E_{\rm off}$ ) is reduced

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by 60.8% compared with the conventional LIGBT. In the reverse-conducting state, the FOM2 (=  $V_{\text{on-RC}} \times Q_{\text{rr}}$ ) of the proposed structure with  $M_{\text{N2}}$  in ON-state reduced by 33.9% compared with the conventional antiparallel free-wheeling diode. Although the FOM2 of the proposed structure with  $M_{\text{N2}}$  in OFF-state is relatively higher than other devices, in practical applications, once the LIGBT has been turned on and  $C_1$  has been charged,  $M_{\text{N2}}$  will always maintain ON-state when the proposed structure is in the reverse-conducting state.

### **IV. CONCLUSION**

3-D numerical simulation results demonstrate that the proposed RC-LIGBT has promising performances on the output characteristics, the turn-off loss, and the reverse recovery charge. And the fabrication process of the proposed structure is compatible with the existing SOI process [9], [14]. The proposed structure decreases the turn-off loss and reverse recovery charge by 58.3% and 38.9%, respectively, compared with the conventional LIGBT combining with antiparallel freewheeling diode.

### **REFERENCES**

- W. Sun et al., "A novel silicon-on-insulator lateral insulated-gate bipolar transistor with dual trenches for three-phase single chip inverter ICs," *IEEE Electron Device Lett.*, vol. 36, no. 7, pp. 693–695, Jul. 2015. doi: 10.1109/LED.2015.2434611.
- [2] Q. Qian, S. Liu, W. Sun, H. Wang, and C. Zhang, "A robust W-shape-buffer LIGBT device with large current capability," *IEEE Trans. Power Electron.*, vol. 29, no. 9, pp. 4466–4469, Sep. 2014. doi: 10.1109/TPEL.2014.2299822.

- [3] K. Hara et al., "600V single chip inverter IC with new SOI technology," in Proc. ISPSD, 2014, pp. 418–421. doi: 10.1109/ISPSD.2014.6856065.
- [4] P. A. Gough, M. R. Simpson, and V. Rumennik, "Fast switching lateral insulated gate transistor," in *Proc. IEDM*, 1986, pp. 218–221. doi: 10.1109/IEDM.1986.191153.
- [5] M. R. Simpson, "Analysis of negative differential resistance in the I–V characteristics of shorted-anode LIGBT's," *IEEE Trans. Electron Devices*, vol. 38, no. 7, pp. 1633–1640, Jul. 1991. doi: 10.1109/16.85160.
- [6] J.-H. Chul, D.-S. Byeon, J.-K. Oh, M.-K. Han, and Y.-I. Choi, "A fast-switching SOI SA-LIGBT without NDR region," in *Proc. ISPSD*, 2000, pp. 149–152. doi: 10.1109/ISPSD.2000.856793.
- [7] J. Oh, D. H. Chun, R. Oh, and H. S. Kim, "A snap-back suppressed shorted-anode lateral trench insulated gate bipolar transistor (LTIGBT) with insulated trench collector," in *Proc. IEEE Int. Symp. Ind. Electron.*, Gdansk, Poland, 2011, pp. 1367–1370. doi: 10.1109/ISIE.2011.5984358.
- [8] L. Huang et al., "A snapback-free fast-switching SOI LIGBT with polysilicon regulative resistance and trench cathode," IEEE Trans. Electron Devices, vol. 64, no. 9, pp. 3961–3966, Sep. 2017. doi: 10.1109/TED.2017.2726080.
- [9] L. Zhang et al., "A high current density SOI-LIGBT with segmented trenches in the anode region for suppressing negative differential resistance regime," in *Proc. ISPSD*, Hong Kong, 2015, pp. 49–52. doi: 10.1109/ISPSD.2015.7123386.
- [10] F. Udrea, U. N. K. Udugampola, T. Trajkovic, and G. A. J. Amaratunga, "Reverse conducting double gate lateral insulated gate bipolar transistor in SOI based technology," in *Proc. 19th Int. Symp. Power Semicond. Devices IC's*, Jeju, South Korea, 2007, pp. 221–224. doi: 10.1109/ISPSD.2007.4294972.
- [11] B.-H. Lee et al., "A new gradual hole injection dual-gate LIGBT," IEEE Electron Device Lett., vol. 19, no. 12, pp. 490–492, Dec. 1998. doi: 10.1109/55.735756.
- [12] K. Sheng et al., "Dual gate lateral inversion layer emitter transistor," in Proc. Int. Conf. Power Electron. Mach. Drives, Jun. 2002, pp. 37–40.
- [13] J. Wei et al., "Ultrafast and low-turn-off loss lateral IEGT with a MOS-controlled shorted anode," IEEE Trans. Electron Devices, vol. 66, no. 1, pp. 533–538, Jan. 2019. doi: 10.1109/TED.2018.2873766.
- [14] X-FAB SOI Process Technologies. Accessed: Feb. 15, 2019. [Online]. Available: https://www.xfab.com/technology/soi/

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