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A Theoretical Analysis for Gigabit/Second Pulse Code Modulation of Semiconductor Lasers

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Abstract-Investigation of the rate equations of a semiconductor laser suggests that bit rates of 3-4 Gbit/s can be achieved. Delay, ringing transients, and charge-storage effects can be removed by adjusting the dc-bias current and the peak and width of the current pulse to values prescribed by simple analytical expressions. Also, simple approximate formulas for the light pulse maximum, width, delay, and integrated values are given.

INTRODUCTION

PULSE modulation of a semiconductor laser presents three problems [1]-[5], when high bit rates are required: a time delay between the applied current pulse and the light pulse, ringing transients, and the charge-storage effect of preceding pulses on the light pulse form and magnitude.

This work presents a pulse modulation model which avoids these disadvantages. A dc-bias current near threshold reduces the delay. The first spike of the ringings is suggested to be exploited as a light pulse. This is done by switching off the current pulse immediately after the light pulse appeared. The influence of the charge storage, making the height of the photon pulses dependent on preceding pulses, is overcome by giving the height and duration of the applied current pulses such values that the electron and photon densities return to the start values at the end of the pulse. This makes the start condition identical for each pulse.

Numerical solutions show that the detailed current pulse form is not significant. Only the charge delivered by the current has to fulfill some requirements. Analytical formulas, derived for square current pulses, are therefore at least qualitatively applicable for a wide range of pulse forms.

The following analysis is based on the rate equations including spontaneous emission as reported in [5],

$$dn/dt = J/(ed) - n/\tau_s - gS \tag{1}$$

$$dS/dt = gS - S/\tau_p + \beta n/\tau_s \tag{2}$$

where

n, J, S electron, current, and photon densities, respectively;

d active layer thickness;

- *e* electronic charge;
- τ_s, τ_p electron and photon lifetimes.

The gain of the laser is $g = \alpha(n - N_0)$ [5], [6], where N_0 and α are constants. The factor β is the fraction of spontaneous light that goes into the the lasing modes. In [7] it has been shown

that the spectrum of a narrow pulse is determined by the dc bias, and is almost independent of the current pulse height. Therefore β is taken to depend on the prebias current only and to be constant during a pulse. The value of β is estimated to be of the order of 10^{-4} when one mode is active, and is taken to be proportional to the number of active modes [8].

For simplicity, effects like mode locking, mode coupling, Qswitching, electron diffusion, and circuit parasitics have been disregarded in this analysis.

THE MODULATION FORM

In the present modulation form, the laser is prebiased by a dc current density $J = J_0$ near threshold. This gives the start values $n = n_0$ and $S = S_0 = xS_{th}$, where S_{th} is the static photon density at threshold and x is a parameter used to describe the starting condition.

A current pulse $\Delta J(t)$, sufficiently high to give ringings [5], is superposed on J_0 . The electron density *n* now describes an oscillating curve with a maximum and a minimum (Fig. 1). The photon density reaches a maximum S_{\max} when *n* in its negative slope region has decreased to $n = n_c \simeq 1/(\alpha \tau_P)$ where $\tau'_P = \tau_P/(1 + \alpha N_0 \tau_P)$. If the current has only small relative changes during the output of light, the pulse is nearly symmetric [5], and we have $n_{\max} - n_c \simeq n_c - n_{\min}$. After the minimum, the last part of the current pulse pumps *n* up to n_0 after which the pulse is switched off. From this moment *S* decays from a value only slightly higher than S_0 within 10-100 ps to this value. Another timing of the current would give the laser a longer relaxation time, up to 4 ns, due to the electron life time.

The necessary and sufficient conditions for the use of the present modulation model is expressed by the two following integrals of the current pulse $\Delta J(t) = J(t) - edn_c/\tau_s$:

$$n_{\max} - n_0 = 1/ed \int_{\text{start}}^{t(n=n_{\max})} \Delta J \, dt \tag{3a}$$

$$n_0 - n_{\min} = 1/ed \int_{t(n=n_{\min})}^{end} \Delta J \, dt.$$
 (3b)

These integrals follow directly from (1) and the fact that before and after the light pulse has emerged, we can neglect the term gS. Equation (3a) expresses the pumping of the electrons by the pulse necessary to give the wanted light output pulse, while (3b) expresses the pump necessary to bring the electron density back to its start value. This charge condition shows that the detailed current pulse form is not significant for the applicability of the present model.

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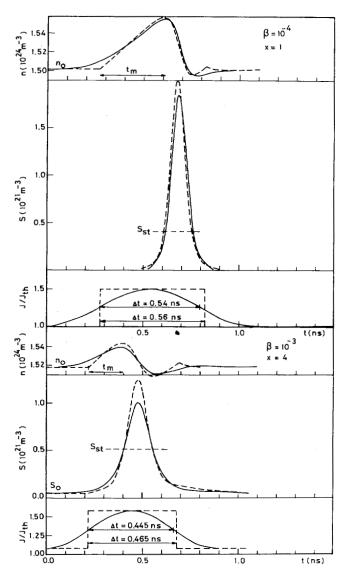


Fig. 1. Numerical results of electron and photon densities as a function of time in a pulse, calculated for a raised cosine (---) and a square (---) current pulse. The applied parameters were $\tau_p = 2.94/(1 + 2.67 \cdot 10^{-25} n(t))$ ps, $\tau_s = 4.0$ ns, $\alpha = 1.125 \cdot 10^{-12}$ s⁻¹ · m³, $N_0 = 0.11 \cdot 10^{25}$ m⁻³, $d = 0.5 \mu$ m, $e = 1.6 \cdot 10^{19}$ As.

NUMERICAL SOLUTION

Fig. 1 shows computer calculations of electron and photon densities for a square and a raised cosine pulse form. Both have the same amplitude. The half-value widths Δt of the current pulses are in both cases fitted to values according to the model. It is seen that Δt , and therefore also the charge in the pulse, is approximate independent of the current pulse form. This has been shown to be true for several sets of parameters β , x. In Fig. 1 two examples are shown where $\beta = 10^{-4}$, x = 1 and $\beta = 10^{-3}$, x = 4.

For low β and x, i.e., low start values S_0 , the peak and width of the light pulse are nearly independent of the current pulse shape. For higher β and x the tendency is that the maximum decreases and the width increases, but in all cases the integrated light output is the same.

Calculations of a long row of pulses showed that no intersymbol interference was present when the correct timing of the current pulses was used, independent of the current pulse spacing down to 0.1 ns. Overtiming of the pulses gave strong intersymbol interference, while undertimed pulses with a deviation up to 10 percent gave stable light output pulses with less than 5 percent change in $S_{\rm max}$.

ANALYTICAL APPROXIMATIONS

As a consequence of the fact that the laser response is only weakly dependent on the detailed current pulse form, and for simplicity, a square current pulse has been analyzed in order to obtain simple approximate formulas to describe the light output and the timing of the current pulse.

The computer calculations show that a square current pulse with the amplitude ΔJ causes *n* to vary approximately as a triangular wave with time. From the start to $n = n_{\text{max}}$ and from $n = n_{\text{min}}$ to the end of the pulse, the slope is $\Delta J' = \Delta J/(ed)$. Between $n = n_{\text{max}}$ and $n = n_{\text{min}}$, the slope was taken to be a constant -n'. Apparently rough, this approximation gives results which are in 5-10-percent agreement with computer calculations for all reasonable values of parameters $\Delta J/J_{\text{th}} < 2$, $10^{-4} < \beta < 5 \cdot 10^{-3}$, and 1 < x < 10. As reference for the start condition for the pulse modulation we use the threshold values. From the definition of the threshold current density in [5] we obtain $J'_{\text{th}} = J_{\text{th}}/(ed) \simeq 1/(\alpha \tau_s \tau'_p)$. The static photon and electron densities with $J' = J'_{\text{th}}$ are given by

$$S_{\rm th} \cong 1/(\alpha \tau_s) \cdot (\beta \tau_p / \tau_p')^{1/2}$$

and

$$\Delta n_{\rm th} = n_{\rm th} - n_c \simeq -1/\alpha \cdot (\beta/\tau_p/\tau_p')^{1/2} \tag{4}$$

where n_c is the electron density for which the gain exceeds the total photon loss in the laser. By introduction of $\Delta n = n - n_c$ in (2) we obtain

$$dS/dt = \alpha \Delta nS + \beta n/\tau_s. \tag{5}$$

The start condition is defined by $J = J_0$ giving $S = S_0 = x \cdot S_{\text{th}}$ and according to (5) $\Delta n = \Delta n_0 = \Delta n_{\text{th}}/x$. The analysis is now performed in three steps.

1) In the time interval $0 < t < t_m$, where t_m is defined by $n(t_m) = n_{max}$, an approximate solution of (5) can be written as

$$S = \{S_1(\operatorname{erf} A(t - t_0) + \operatorname{erf} At_0) + S_0 \exp(-A^2 t_0^2)\}$$

$$\cdot \exp A^2(t - t_0)^2$$
(6)

where $t_0 = -\Delta n_0 / \Delta J'$ is the time it takes for *n* to grow from n_0 to n_c , and

$$S_1 = \beta / (\alpha \tau_s \tau_p') \cdot \sqrt{\pi / (2 \alpha \Delta J')}$$
(7a)

and

$$A = \sqrt{\alpha \Delta J'/2}.$$
 (7b)

At $t = t_m$ we have dn/dt = 0. With $n_{\max} \approx n_c$ in (1) we find that $S \approx S_{st}$, the stationary state value corresponding to $J = J_0 + \Delta J$. This, inserted into (6), gives in practical cases, when $A(t_m - t_0) \ge 1$,

$$t_m = t_0 + \sqrt{2B/(\alpha \Delta J')} \tag{8}$$

where

$$B = \ln \left(S_{\rm st} / S_{\rm 00} \right) \tag{9a}$$

and

$$S_{00} = S_1 (1 + \operatorname{erf} At_0) + S_0 \exp(-A^2 t_0^2).$$
(9b)

For pulses where $\tau_P / \tau_s \cdot x^2 / \beta \cdot \Delta J' / J'_{\text{th}} > 1$, we find that

$$B \simeq \ln \left(\Delta J' / J_{\text{th}} \cdot x^{-1} \cdot (\tau_p / \tau'_p / \beta)^{1/2} \right).$$
(10)

A limiting condition for which the model can be applied is given by $n_{\text{max}} - n_c = n_c - n_{\text{min}}$ and $n_0 > n_{\text{min}}$, which gives $t_m \ge 2t_0$. By insertion of $t = 2t_0 \le t_m$ into the right-hand side and $S = S_{\text{st}}$ into the left-hand side of (6), we obtain, after some calculations,

$$\Delta J/J_{\rm th} \ge (4\tau_s^2 \tau_p' \cdot \tau_p^{-3})^{1/4} \cdot \beta^{3/4} / x^{1/2} + \frac{1}{2}\tau_s / \tau_p \cdot \beta / x^2 \quad (11)$$

which is in 5-percent agreement with computer simulations for all practical cases, x > 1 and $\beta < 5 \cdot 10^{-3}$.

2) In the time interval $t_m < t < t_{\min}$, where t_{\min} is defined by $n(t_{\min}) = n_{\min}$, we use $\Delta n = -n'\Delta t'$. Here $\Delta t' = t - t_m - \Delta n_{\max}/n'$ and $\Delta n_{\max} = (t_m - t_0)\Delta J'$. Equation (5) now gives

$$S = S_{\max} \exp\left(-(\alpha n'/2)\Delta t'^2\right)$$
(12)

$$S_{\max} = \tau_p \Delta J' \exp\left(\alpha \Delta n_{\max}^2 / (2n')\right). \tag{13}$$

The integrated photon output of one pulse is given by [5]

$$S_{\text{out, pulse}} = \frac{1}{\tau_p} \int_{t_m}^{t_{\min}} S \, dt = 2\Delta n_{\max} + \Delta J \cdot \Delta t_{\text{st}}$$
(14)

where $\Delta t_{st} = 2\Delta n_{max}/n'$ is the pulsewidth defined by points where $S = S_{st}$. From (12)-(14) an approximate proportionality between n' and $\Delta J'$ (*B* varies slowly) is found:

$$n'/\Delta J' = 2B/\ln(16B/\pi).$$
 (15)

Finally we can use (15) to obtain solutions for the pulse parameters

$$S_{\max} = \tau_p \,\Delta J' \cdot \sqrt{16B/\pi} \tag{16}$$

$$\Delta t_{\rm st} = \sqrt{2/(\alpha \Delta J')} \ln \left(\frac{16B}{\pi} \right) \sqrt{B} \tag{17}$$

and the half-power width

$$\Delta t_{1/2} = 2(2 \ln 2/\alpha/n')^{1/2}$$

= $\sqrt{2/(\alpha \Delta J')} (2 \ln 2 \cdot \ln (16B/\pi)/B)^{1/2}.$ (18)

3) In the time interval $t_{\min} < t < \Delta t$, where Δt is the current pulse length, we can approximate the slope of n by $\Delta J'$. The value of Δt for which $n(\Delta t) = n_0$ is

$$\Delta t = 2 \Delta n_{\max} (1/\Delta J' + 1/n')$$

= $\sqrt{2/\alpha \Delta J'} \cdot (2 + \ln (16B/\pi)/B)B^{1/2}$ (19)

where we once more have used (15).

The total photon output in the pulse defined by (14) is now found by

$$S_{\text{out, pulse}} = \Delta J' \cdot \Delta t \propto \sqrt{\Delta J'}$$
 (20)

The essence of the last statement is that the linear variation

of *n* leads to the result that no photon is present before and after the moments where $S = S_{st}$. The error introduced in this way is less than 20 percent, which is not serious because almost all of the 20 percent is concentrated as exponential tails at both sides of the pulse.

CONCLUSION

We can conclude that the formula for $S_{\text{out, pulse}}$ agreed within 20 percent, and all other formulas within 10 percent, with computer simulation for the parameters $\Delta J/J_0 < 2 \cdot 10^{-4} < \beta < 5 \cdot 10^{-3}$ and 1 < x < 10. The variation of the times is approximately inversely proportional to $\sqrt{\Delta J'}$, while the photon pulse maximum and the integrated pulse are approximately proportional to $\Delta J'$ and $\sqrt{\Delta J'}$, respectively. These features follow from the fact that the functions of *B* appearing as factors in the pulse parameters vary slowly with $\Delta J'$. The only influence of β and *x* is also through *B* [(9) and (10)], which is why this influence is rather weak. The tendency is that for increasing β and *x*, the values of t_m , Δt , S_{\max} , and $S_{\text{out, pulse}}$ decrease while $\Delta t_{1/2}$ increases. Numerical calculation shows that in all cases $\Delta t_{1/2} < 0.5 \Delta t$. This gives good separation between the light pulses.

Correctly timed and undertimed current pulses with pulse spacings down to 0.1 ns gave no or small intersymbol interference, while overtimed current pulses gave instability and strong intersymbol interference. This suggests that the model is useful for high-speed modulation. It can also be concluded that this modulation form is most useful when the prebias is a little above threshold (x > 1) both because of the dynamic behavior [(11)] and because the spectrum then is narrow according to [7] (β is small). It is expected that with this model, pulse durations of $\Delta t = 0.25 - 0.3$ ns corresponding to bit rates 3-4 Gbit/s can be obtained with $\Delta J/J_{\text{th}} = 1.5$ which is twice the highest experimentally obtained [9]. It is believed that adjustments of the parameters Δt , ΔJ , and x to values which fit with the results of the present model have played a role in obtaining Gbit/s modulation rates in earlier experimental works. The problem of obtaining narrow current pulses ($\Delta t = 300 \text{ ps}$) can, for example, be overcome by using a MESFET driver for the laser [10].

Independently of this work some experimental work has recently been performed which verifies some features of the described modulation method [11].

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Zero-Power-Gain Measurements in CW HF(DF) Laser by Means of a Fast Scan Technique

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Abstract-An improved technique for measuring zero power gain in a CW HF-DF chemical laser has been developed in which a CW HF(DF) single-line, frequency-stabilized TEM₀₀ mode probe laser is used. Through use of a flat rotating mirror and focusing elements, the streamwise distribution of zero-power gain was scanned at a rate of ~1 mm/ μ s with a spatial resolution of ~1 mm. The zero-power-gain profile was observed for two arc-driven chemical-laser nozzles with both HF and DF active species. The nozzles consisted of a 36-slit array with perforated tube H₂ injectors and a 55-slit array with uniform H₂ injection. Results are presented for the variation of zero power gain with axial distance, the magnitude and location of peak gain, and the gain cutoff location for a number of transitions. Peak gain values of 15 percent/cm and 5.5 percent/cm were measured with HF and DF active species, respectively, with the 55-slit nozzle array.

I. INTRODUCTION

SMALL-SIGNAL-GAIN measurements in an arc-driven, supersonic diffusion CW HF chemical laser have been previously made both directly by means of pulsed-probe laser [1] and indirectly by means of a chemiluminescence technique [2]. The measurements made with a pulsed-probe laser had an uncertainty in the magnitude of the gain that resulted from a lack of frequency stability in the multitransverse, multilongitudinal mode probe beam. The gain variation with distance from the nozzle exit plane x was obtained [1] by measuring the gain on the various HF vibrational rotational transitions at fixed values of x during a given arc run. The arc was then shut down, the probe beam was located at a new value of x, and the process repeated.

Clearly, real-time (fast) scans with a CW HF(DF) probe would be desirable in order to reduce run-to-run flow variations and to expedite measurements. The present work incorporates a newly developed single-transverse, singlelongitudinal mode, stable CW probe laser and an optical system that permits scanning of the small-signal CW HF(DF) gain profile from the nozzle exit plane to a distance 8 cm downstream in about 0.3 ms with a 1-mm spatial resolution. Zero-power-gain distributions have been obtained by means of two different nozzle configurations with both HF and DF active species. The results are reported herein.

II. EXPERIMENTAL APPARATUS AND METHOD

A. Scanning Technique

The gain profile scanning technique used in the present experiments is shown in Fig. 1. A technique similar to this was used previously by Falk [3] for zero-power-gain measurements in a DF-CO₂ transfer chemical laser. The rotating mirror was located at the focal points of the focusing elements L_2 (3-m radius concave mirrors), which caused the probe beam to sweep from the nozzle exit plane to a distance 8 cm downstream while remaining parallel to the exit plane. The IR probe beam was aligned with a HeNe reference beam to the center line (vertical) of the nozzle exit. The focusing element L_3 then imaged the split probe beams on diffuse reflectors (sandblasted Al plates) viewed by the signal and reference detectors. The signal detector voltage was observed versus time with H_2/D_2 off and on. The log of the ratio of these observations defines the variation of active region gain versus

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