

1986

AN EXACT ANALYSIS FOR EFFICIENT COMPUTATION OF RANDOM-PATTERN TESTABILITY IN COMBINATIONAL CIRCUITS

Sharad C. Seth

University of Nebraska-Lincoln, seth@cse.unl.edu


Bhargab B. Bhattacharya

Indian Statistical Institute, Kolkata, India, bhargab@isical.ac.in

Vishwani Agrawal

AT&T Bell Laboratories, Murray Hill, NJ

Follow this and additional works at: <http://digitalcommons.unl.edu/cseconfwork>

 Part of the [Computer Engineering Commons](#), [Electrical and Computer Engineering Commons](#), and the [Other Computer Sciences Commons](#)

Seth, Sharad C.; Bhattacharya, Bhargab B.; and Agrawal, Vishwani, "AN EXACT ANALYSIS FOR EFFICIENT COMPUTATION OF RANDOM-PATTERN TESTABILITY IN COMBINATIONAL CIRCUITS" (1986). *CSE Conference and Workshop Papers*. 302. <http://digitalcommons.unl.edu/cseconfwork/302>

This Article is brought to you for free and open access by the Computer Science and Engineering, Department of at DigitalCommons@University of Nebraska - Lincoln. It has been accepted for inclusion in CSE Conference and Workshop Papers by an authorized administrator of DigitalCommons@University of Nebraska - Lincoln.

**AN EXACT ANALYSIS FOR EFFICIENT COMPUTATION
OF RANDOM-PATTERN TESTABILITY IN
COMBINATIONAL CIRCUITS**

**Sharad C. Seth, Bhargab B. Bhattacharya,
and Vishwani D. Agrawal**

Reprinted from the 16th ANNUAL INTERNATIONAL
SYMPOSIUM ON FAULT-TOLERANT COMPUTING SYSTEMS
July 1-4, 1986 Vienna, Austria

• S. C. Seth, B. B. Bhattacharya, and V. D. Agrawal, "An exact analysis for efficient computation of random-pattern testability in combinational circuits", Proc. 16th Int. Symposium on Fault Tolerant Computing (FTCS), 318-323, 1986.

AN EXACT ANALYSIS FOR EFFICIENT COMPUTATION OF RANDOM-PATTERN TESTABILITY IN COMBINATIONAL CIRCUITS

*Sharad C. Seth
Bhargab B. Bhattacharya*

University of Nebraska
Lincoln, Nebraska 68588

Vishwani D. Agrawal

AT&T Bell Laboratories
Murray Hill, New Jersey 07974

ABSTRACT - Experimental evidence shows that low testability in a typical circuit is much more likely due to poor observability than poor controllability. Thus, from theoretical and practical standpoints, it is important to develop an accurate model for observability computation. One such model, in terms of supergates, is proposed in the first part of this paper thus complementing our earlier work. It is now possible to obtain exact random-pattern testability for each line in a circuit.

The second part of the paper analyzes the supergate structure of a circuit from a graph theoretic viewpoint. Finding a supergate is related to determining the dominator tree in a modified circuit graph which provides an exact bound on the complexity of computation. The uniqueness of the cover extends to multiple output circuits and the complexity of finding it is shown to be quadratic in the number of nodes in the circuit graph. By prior determination of the maximal supergate cover, unnecessary duplication in computation of testability values can be avoided.

1. INTRODUCTION

Design for testability has assumed increasing urgency due to higher circuit densities. Concomitantly, the need for developing good measures of testability and for computing them efficiently with sufficient accuracy also becomes greater. Increased understanding of the limitations of the traditional deterministic measures [AGR82, GOLD79] has prompted recent efforts on improved measures based on applying random patterns to the circuit [SAV84, SET85]. Random-pattern testability is shown to be effective not only in identifying areas of poor testability but also for simulation-free fault analysis and automatic test-pattern generation [AGR85a, AGR85b, BRG84, WUN85].

Both deterministic and probabilistic methods calculate, for each circuit node, values related to its controllability and observability. The deterministic measures, such as SCOAP [GOL79], are non-normalized and only intuitively justified. Probabilistic measures, on the other hand, are signal probabilities which always lie in the [0.0, 1.0] range. For a line in a combinational circuit, these measures are related to the boolean function realized by the line. Thus, if all input patterns are equiprobable, a line's one-controllability is identical to its syndrome [SAV80]; its observability at an output, can similarly be related to the boolean difference [SEL68].

This paper extends the supergate analysis of PREDICT [SET85] for exact computation of observabilities. Also, sections 4 and 5 analyze the supergate structure of a circuit from a graph-theoretic viewpoint. Finding a supergate is related to determining the dominator tree in a modified circuit graph thus providing an exact bound on the complexity of this computation. Determination of the maximal supergate cover avoids unnecessary duplication in the computation of testability.

2. BACKGROUND AND NOTATION

Supergates: In PREDICT, computation of controllability for a node (a primary input or a gate output) is carried out on a circuit subgraph called the "supergate" of that node. A graph-theoretic definition of supergate appears in Section 4. To determine the supergate of a node X in the circuit, one starts tracing the circuit back from X. If X is a primary input we are done. Otherwise, the signals on the inputs of gate X are compared pairwise. If the signals are mutually independent, we are again done. Otherwise, such signals as are not pairwise independent define a new "frontier" for further backtracing. One of the frontier nodes is arbitrarily chosen for backtracing and a new "frontier" is defined after tracing signals at its input. The process stops as soon as the frontier becomes null. As an example, Fig. 1 shows a NAND circuit with the supergate of primary output shown within the shaded triangle.

The inputs of a supergate are partitioned into two classes: fanout and nonfanout. The former are characterized by more than one path to the supergate output. In the example, lines c, d, and e define supergate inputs with d and e as the fanout inputs. It may be noted that a fanout input of a supergate might not fanout itself (e.g., line d) but may have multiple paths to the supergate output through an internal fanout.

Conditional Computation within a Supergate: The computation within a supergate is carried out in the context of a fixed pattern of binary inputs applied to the fanout inputs of the supergate. Assume that all possible patterns on the fanout inputs are indexed in some arbitrary way and let A_i represent the i-th pattern. Then $C0_i(k)$ is the *conditional zero-controllability* of line k in the supergate when A_i is applied to the fanout inputs. This is the probability of setting line k to zero when a randomly selected input to the network impresses the pattern A_i to the fanout inputs of the supergate. *Conditional one-controllability* is similarly defined and is denoted by $C1_i(k)$. In [SET85] it was shown how these conditional controllability values can be propagated forward in a supergate essentially by assuming them to be independent. For example, for a two-input AND gate with lines m and n as inputs and p as the output, we have:

$$C0_i(p) = C0_i(m) + C0_i(n) - C0_i(m)C0_i(n), \text{ and}$$

$$C1_i(p) = C1_i(m)C1_i(n)$$

The *zero-controllability (one-controllability)* of line k is the probability of setting line k to zero (one) when a randomly selected input pattern is applied to the network. These are denoted, respectively, as $C0(k)$ and $C1(k)$ and can be computed from the conditional controllabilities [SET85].

$$C1(k) = \sum_{\text{all } A_i} C1_i(k)P(A_i)$$

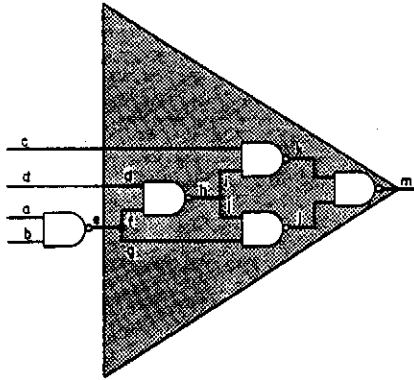


Fig. 1 Supergate example.

where, $P(A_i)$ is the probability of A_i appearing at the fanout inputs when a random pattern is applied to the network.

Example: For the circuit in Fig. 2(a), assume each input can be independently set to one with probability $1/2$ and we want to compute the one-controllability of line g . The supergate of g is the whole circuit with the primary input B as a fanout input. Let A_0 and A_1 represent the assignments $B=0$ and $B=1$, respectively. Then, as shown in Figs. 2(b) and 2(c), $C1_0(g) = 1$ and $C1_1(g) = 1/2$. Thus, the above sum yields $C1(g) = (1 + 1/2)(1/2) = 3/4$.

The *conditional zero-observability (conditional one-observability)* of line k is the probability of observing line k at the supergate output under the condition that A_i is applied to the fanout inputs and line k is set to zero (one). These are denoted, respectively, as $B0_i(k)$ and $B1_i(k)$.

The *conditional zero-detectability (conditional one-detectability)* of line k is the probability of simultaneously setting line k to zero (one) and observing it at the supergate output when a random input is applied to the network. Denoting these by $D0_i(k)$ and $D1_i(k)$, respectively, we write:

$$D0_i(k) = C0_i(k)B0_i(k)$$

$$D1_i(k) = C1_i(k)B1_i(k)$$

$D0_i(k)$ is the probability of detecting the fault "line k stuck-at-one" at the supergate output when a random input to the network applies A_i to the fanout inputs of the supergate. Since the observabilities, as defined above, are conditioned on setting the line being observed to a value, we avoid the anomaly, noted in [SAV83] that high values of a line's controllability and observability may not imply high testability.

3. EXACT CALCULATION OF DETECTABILITY

We consider the problem of determining the detectability of a line in a supergate at its output. All the proposed solutions to this problem (based on the circuit structure) involve approximations and thus are inexact [JAI85, SET85, BRG84]. The errors arising in such computations can be attributed to two distinct classes of simplifications: (1) the observability of line through a chain of gates can be found by considering the gates one at a time, and (2) the observability of a reconvergent fanout stem is a fixed function of its branch observabilities. The first simplification is, indeed, valid in the context of a pattern applied to fanout inputs. This leads to an efficient procedure for computing detectabilities for all lines which are not reconvergent.

Observabilities of Non-reconvergent Lines: Consider a two-input AND gate with input lines a and b and output line c . Observing a zero value on line a requires that line b should be set to 1 and the 0 on line c should be observable. That is,

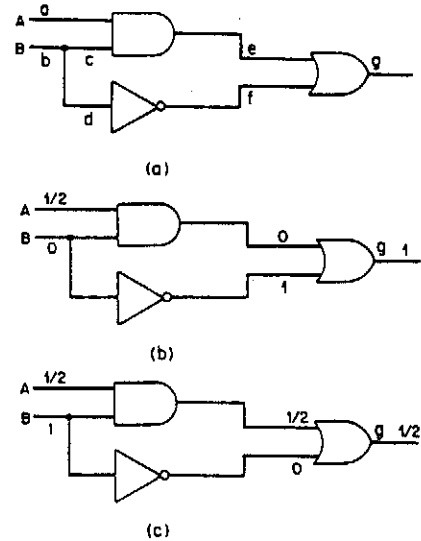


Fig. 2 Conditional controllability and detectability computation for nonreconvergent lines. (a) Circuit, (b) One-controllabilities when $B=0$, and (c) One-controllabilities when $B=1$.

$$\begin{aligned} B0(a) &= B0(c)\text{Prob}(b=1 | a=0) \\ &= B0(c)[\text{Prob}(b=1, a=0)/\text{Prob}(a=0)] \\ &= B0(c)[S(a) - C1(c)]/C0(a) \end{aligned}$$

where, $S(a)$ is the probability of sensitizing a path from a to c . For the two-input AND gate $S(a) = C1(b)$. Similarly,

$$B1(a) = B1(c)\text{Prob}(b=1 | a=1) = B1(c)C1(c)/C1(a)$$

Expressions for observabilities for other types of gates are derived in a similar manner [JAI85]. Notice that in the above $B1(a)$ is obtained by multiplying two probabilities, $B1(c)$ and $\text{Prob}(b=1 | a=1)$. This assumes that these two probabilities correspond to independent events. To examine the validity of this assumption, consider again the circuit shown in Fig. 2(a). Line g is a primary output with observabilities set to unity. If the inputs A and B have equal probabilities of 0 and 1, then the expressions of the above type give the following result:

	a	b	e	f	g
C0	1/2	1/2	3/4	1/2	1/4
C1	1/2	1/2	1/4	1/2	3/4
B0	1/6		1/3	1/2	1/1
B1	1/2		1/1	1/1	1/1

Line b , a reconvergent stem, will be considered in the next section. Line a is observable at g whenever b is 1. Therefore, $B0(a) = B1(a) = 1/2$. However, our computation, in the above table, incorrectly gives $B0(a) = 1/6$. This is because the observability of e depends on f which is derived from b . Such situations are common in the presence of reconvergent fanouts. The Stafan formulas take account of the correlation between the lines at the input of the same gate, e.g., e and f are correlated due to the common influence of b on them; their observabilities are correctly computed. However, correlations extending beyond the inputs of a single gate are ignored by Stafan.

Consider a two-input gate, with inputs a and b , and output c , embedded in a supergate. Assume a pattern A_i on the fanout inputs of the supergate. Under these conditions, the signals on the two inputs of the AND gate indeed become independent, therefore, Stafan's

assumption are true for conditional computation. In particular, to detect a 0 on line a (when A_i is applied to fanout inputs of the supergate), a 1 must be applied to line b and the resulting 0 on line c must be observed. Similar arguments will apply to detection of 1 on line a. Thus,

$$D0_i(a) = C1_i(b) \overline{p}_i(c) \quad (1a)$$

$$D1_i(a) = \overline{C1_i(b)} D1_i(c) \quad C0_i(c) \quad (1b)$$

Similarly, for a two-input OR gate:

$$D0_i(a) = C0_i(b) \overline{p}_i(c) \quad (2a)$$

$$D1_i(a) = \overline{C0_i(b)} D1_i(c) \quad C1_i(c) \quad (2b)$$

These and similar equations for other elementary gates may be used to obtain conditional detectabilities of all lines up to (but not including) fanout stems.

Theorem 1: [SET86] Let k be a non-reconvergent line in a supergate and let $D0(k)$ and $D1(k)$ represent, respectively, its zero and one detectabilities. Then

$$(a) D0(k) = \sum_{\text{all } A_i} D0_i(k) P(A_i)$$

$$(b) D1(k) = \sum_{\text{all } A_i} D1_i(k) P(A_i)$$

where, $P(A_i)$ is the probability of applying pattern A_i to the fanout inputs of the supergate.

Returning to our example of Fig. 2, the conditional detectabilities can be computed from Figs. 2(b) and 2(c). Assuming again that line g is a primary output, using Eqns. (1) and (2), the results are tabulated below.

	a	b	c	f	g
$C1_{b=0}$	1/2	0	0	1	1
$C1_{b=1}$	1/2	1	1/2	0	1/2
$D0_{b=1}$	0		0	0	0
$D0_{b=0}$	1/2		1/2	1/4	1/2
$D1_{b=0}$	0		0	0	0
$D1_{b=1}$	1/2		1/2	1/4	1/2

From Theorem 1, therefore, $D0(a) = D1(a) = (1/2 + 0)(1/2) = 1/4$, which is verified by noting that only one of the four input patterns detects each of the faults: "a stuck-at-0" and "a stuck-at-1". Note that $B0(a) = D0(a)/C0(a) = 1/2$. Thus non-stem line observabilities can also be correctly computed.

Detectabilities of Reconvergent Stems: It is attractive to think of extending the above method to determine the observabilities of reconvergent stems. Then, all line detectabilities could be found in a single backward trace of the circuit. Unfortunately, a stem's detectability is not just a function of its branch detectabilities but also depends on the inversion parity of the branches along reconvergent paths. The solution proposed in this paper gives up the goal of determining all detectabilities in a single backward trace. Instead, reconvergent stem detectabilities are found in a forward trace combined with the computation of line controllabilities.

Let s be a reconvergent fanout stem in a supergate and let A_i be applied to the fanout inputs of the supergate. Note that, from the definition of fanout inputs to supergate, it follows that the value on stem s is uniquely determined by A_i . For an arbitrary line k in the supergate, we consider the probabilities of disjoint events of sensitizing k to 0 and 1 values from s . These will be called, respectively, *conditional zero-*

sensitization and *conditional one-sensitization* of k from s , symbolically written as $S_i(s, k_0)$ and $S_i(s, k_1)$. If k has only even-parity or only odd-parity inversion paths from s , one of these probabilities will be zero but, in general, both probabilities must be considered.

The conditional sensitizations of line k from stem s can be stated in functional terms. Let the vector $X = (x_1, x_2, \dots, x_n)$ be the inputs to the supergate and let $k(X)$ be the function realized on line k (we will abbreviate $k(X)$ as k whenever there is no ambiguity). Denote:

- (a) the restriction of $k(X)$ when A_i is assigned to the fanout inputs as the function $\{k(X)\}_i$,
- (b) the syndrome [SAV80] of $k(X)$ (defined as the number of one's in the truth table of $k(X)$ divided by 2^n) as $\{k(X)\}_i$,
- (c) $k(X)$ considered as a function of the stem s and supergate inputs X as $k(s, X)$, and
- (d) the boolean difference [SEL68] of k with respect to s as k' .

Lemma 1: Under the assumption that all input patterns to a supergate are equiprobable,

$$(a) S_i(s, k_1) = |\{kk'\}_i|$$

$$(b) S_i(s, k_0) = |\{\bar{k}k'\}_i|$$

Proof: The boolean difference k' represents the condition of sensitizing s to line k . Thus, the boolean function $\{kk'\}_i$ represents the condition of sensitizing a 1 to line k from s when A_i is applied to fanout inputs. The syndrome of this function is then equal to $S_i(s, k_1)$ under the equiprobable assumption as stated above. A similar proof can be given for (b).

Theorem 2: Let $S_i(s, k_b)$ denote the conditional b -sensitization of a line k in the supergate from the stem s , where b is either 0 or 1. (a)

$$S_i(s, p_0) = S_i(s, m_0)[C1_i(n) - S_i(s, n_1)] + S_i(s, n_0)[C1_i(m) - S_i(s, m_1)] + S_i(s, m_0)S_i(s, n_0)$$

(b) For a two-input OR gate with inputs m and n and output p :

$$S_i(s, p_0) = S_i(s, m_0)[C0_i(n) - S_i(s, n_0)] + S_i(s, n_0)[C0_i(m) - S_i(s, m_0)] + S_i(s, m_0)S_i(s, n_0)$$

(c) For a NOT gate with input m and output p :

$$S_i(s, p_0) = S_i(s, m_0)$$

Proof: (c) is trivial. The proofs of (a) and (b) are very similar, hence we will only prove (a). Consider the case when $b=1$. From Lemma 1

$$S_i(s, p_1) = |\{pp'\}_i|$$

where, p' is the boolean difference of p with respect to s . Now $p = m \cdot n$ for an AND gate, therefore [BRE76, Section 2.1],

$$p' = (mn)' = mn' \oplus nm' \oplus m'n' \text{ and}$$

$$pp' = mn(mn' \oplus nm' \oplus m'n')$$

$$= mnmn' \oplus mnam' \oplus mnm'n'$$

$$= m(m' \oplus \bar{m})nn' \oplus mm'n(n' \oplus \bar{n}) \oplus mnm'n'$$

$$= m\bar{m}nn' \oplus mm'n\bar{n}' \oplus mnm'n'$$

Therefore, from Lemma 1,

$$S_i(s, p_1) = |\{pp'\}_i| = |\{m\bar{m}nn' \oplus mm'n\bar{n}' \oplus mnm'n'\}_i|$$

Since all the three terms are mutually disjoint on the right hand side, the syndrome is the sum of syndromes of each term, hence

$$\begin{aligned}
S_i(s, p_i) &= |[mm\bar{n}n']_i| + |[mm'n\bar{n}']_i + |[mm'nn']_i \\
&= |[m\bar{m}]_i| \cdot |[nn']_i + |[mm']_i| \cdot |[n\bar{n}']_i \\
&\quad + |[mm']_i| \cdot |[nn']_i
\end{aligned} \tag{3}$$

Since $[mm']_i$ and $[nn']_i$ do not involve any variables in common.

$$\begin{aligned}
C1_i(m) &= |[m]_i| - |[m(m' \oplus \bar{m}')]_i| = |[mm']_i| + |[m\bar{m}']_i \\
&= S_i(s, m_i) + |[m\bar{m}']_i
\end{aligned}$$

Therefore,

$$|[m\bar{m}']_i| = C1_i(m) - S_i(s, m_i)$$

When this substitution is made on the right hand side of (3) we get the desired result. A similar proof can be given for the case when $b=0$.

Next, suppose the output line of a supergate is x . For a stem s , it is possible to determine its conditional sensitizations to line x by Theorem 2. The stem's observabilities at x are obtained by appropriately combining these conditional sensitizations as indicated in the following theorem:

Theorem 3: [SET86] Let s be a reconvergent stem in a supergate and let Z_s (N_s) be the subset of patterns on the fanout inputs of the supergate which cause a zero (one) to appear on s . Then

$$(a) D0(s) = \sum_{A \in Z_s} [S_i(s, x_0) + S_i(s, x_1)]P(A_i)$$

$$(b) D1(s) = \sum_{A \in N_s} [S_i(s, x_0) + S_i(s, x_1)]P(A_i)$$

Example: Consider the carry-logic circuit of Fig. 3(a). It is easily verified that the supergate of the carry output is the whole circuit with A and B as the fanout inputs. For each line k in the network, we show the sensitizations $S_i(s, k_0)$ and $S_i(s, k_1)$ as an ordered pair within square brackets. Also shown for line k is $C1_i(k)$. Stem s is 1 for three combinations of values on A and B : 00, 01, and 10. Figures 3(a)-3(c) show the conditional sensitizations (obtained from Theorem 2) and one-controllabilities (as derived in PREDICT) for the three cases. Each of these combinations occurs with probability, $P(A_i) = 1/4$, therefore

$$D1(s) = [(1+0) + (1/2+0) + (1/2+0)](1/4) = 1/2$$

4. SUPERGATE STRUCTURE AND ACCELERATION OF TESTABILITY ANALYSIS

As we saw in Section 2, informally, the supergate of a line X in the circuit is the smallest predecessor subnetwork feeding X whose inputs are logically independent, i.e., have no signal correlation. We now introduce a formal definition.

For simplicity, consider first a single-output combinational network N . The structure of N has an equivalent representation in the form of a directed graph $G(V, E)$, called the *circuit graph*, whose nodes are the primary inputs, the primary outputs, and the gates in N , and whose edges represent the connections in N , oriented in the direction of signal flow (see Figs. 4(a) and 4(b)). A fanout in N is represented by a node with outdegree greater than one in G ; a primary input (output) becomes a node whose indegree (outdegree) is zero. Two distinct directed paths $P1$ and $P2$ in G are said to be *reconvergent* if they emanate from a common vertex (say A) and terminate at another common vertex (say B). Node B is called a *reconvergent node*. Obviously, A is a fanout node.

Let $R(X)$ denote the set of all nodes in G , from which node X is reachable by a directed path. Then, the *cone of influence* $C(X)$ of node X is a proper subgraph $G'(V', E')$ of G , such that $V' = R(X)$.

The *supergate* of a node X in a circuit graph G , denoted by $SG(X)$, is a proper subgraph (V'', E'') of the cone of influence $C(X)$, such that the following conditions hold:

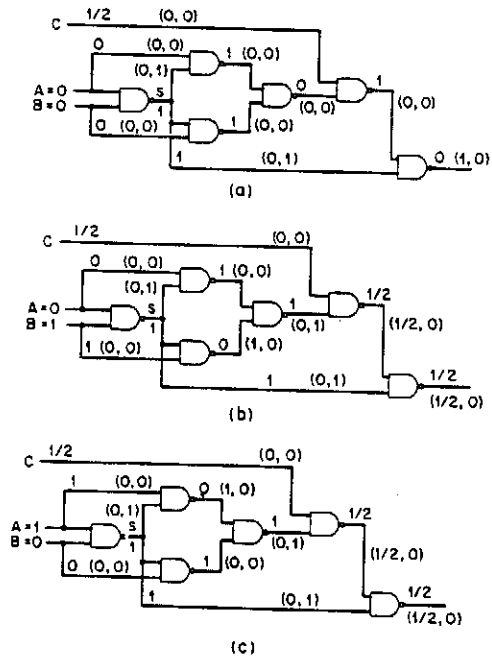


Fig. 3 Detectability computation for reconvergent stems.

- (i) X and all its predecessor nodes (i.e., those with edges directed towards X) are in $SG(X)$.
- (ii) Let v be a node in $SG(X)$ such that a predecessor of v is also in $SG(X)$. Then, all predecessors of v are in $SG(X)$.
- (iii) For every pair of nodes v_j, v_k in $SG(X)$, each with indegree zero, $R(v_j)$ is disjoint from $R(v_k)$.
- (iv) V'' is a minimum vertex set satisfying the above properties.

From the definition it is obvious that the supergate of every node in G is unique. As an example, the vertex set of the supergate of node 16 in Fig. 4(b) is given by $\{16, 13, 12, 11\}$.

The set of *input nodes* $I(X)$ of a supergate $SG(X)$ is the set of nodes in $SG(X)$ with in degree zero. These are further partitioned into two groups:

- (i) $INF(X)$, the *non-fanout* inputs, are those with a unique directed path to X , and
- (ii) $IF(X)$, the *fanout* inputs, are those inputs each of which has at least two distinct direct paths to X (this does *not* imply that these nodes have an outdegree greater than one, since two distinct paths may have a common initial segment).

The *output node* of a supergate $SG(X)$ is the node X itself.

Remark 1: For every node Y in $SG(X)$ which is not an input node, $SG(Y)$ is a proper subgraph of $SG(X)$, i.e., supergates of interior nodes in $SG(X)$ are contained within $SG(X)$.

A supergate is said to be *maximal* if it is not properly contained in a larger supergate. As an example, for the network of Fig. 4(a), the circuit graph and the maximal supergates (shown within dotted lines) are given in Fig. 4(b).

Remark 2: From the previous remark and the definition of maximal supergates, it follows immediately that the partitioning of the circuit graph affected by maximal supergates is unique.

For a single-output combinational circuit, there exists an interesting topological relationship amongst the maximal supergates. We define a *reduced circuit graph* (RCG) as a directed graph (V_1, E_1) , where V_1 denotes the set of supergates and a directed edge (i, j) exists if the output node of $SG(i)$ is an input node of $SG(j)$.

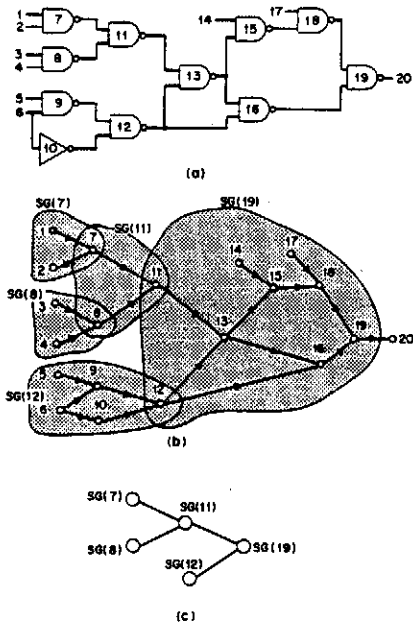


Fig. 4 (a) Example Circuit, (b) Circuit graph and maximal supergates, and (c) Reduced circuit graph.

Remark 3: The RCG is a tree, i.e., for every pair of nodes in RCG there is no more than one directed path.

For the circuit in Fig. 4(a), RCG is shown in Fig. 4(c).

Identification of Maximal Supergates: Intuitively, each supergate represents a minimum subcircuit with logically independent inputs. A more precise statement of the same idea is that for every interior node i in a supergate $SG(X)$, there exists another interior node j in $SG(X)$, such that $R(i)$ and $R(j)$ both include at least one common (fanout) input of $SG(X)$. Clearly, there is an one-to-one correspondence between the notions of supergate and dominance in flow graph [TAR74]. We will use the concept of flow dominance for identifying the set of maximal supergates in a combinational circuit.

Given a directed graph $G(V, E)$ with a distinguished source vertex s of indegree zero, we say that a node v_i *dominates* a node v_j if all directed paths from s to v_j also pass through v_i [TAR74]. It is known that the dominance relation induces a partial order and that the set of nodes which dominate a given vertex is linearly ordered [AHO74]. Because of this linear ordering, the *immediate dominator* of a node n can be uniquely defined as the dominator that is closest to n on any path from the source s to n . This allows depiction of dominance relationship amongst vertices using a tree called the *dominator tree* [TAR74] in which the source vertex is the root and the predecessor of every other node is its immediate dominator.

Algorithm: To find all maximal supergates in a circuit graph of a single-output combinational network, proceed as follows:

Step 1: Given the circuit graph G , construct a directed flow graph FG as follows: (i) Delete the primary-output node and its incident edge. (ii) Reverse the directions on the edges which are incident on the node corresponding to the output gate and make all other edges bidirectional

Step 2: Construct the dominator tree of FG .

Step 3: Start from the root node of the tree and collect all the nodes (including the root) which are children of the root. Include also all single successors of these children, if any. (This set gives the maximal supergate corresponding to the output node.) Remove all the edges so far covered and iterate the process until the tree is empty.

Example: Consider the circuit graph in Fig. 4(b). The dominator tree

corresponding to its flow graph FG is shown in Fig. 5. The maximal supergates are $SG(19)$, $SG(11)$, $SG(12)$, $SG(7)$, and $SG(8)$, the vertex sets of which are given by:

$$V''[SG(19)] = \{19, 18, 17, 16, 15, 14, 13, 11, 12\}$$

$$V''[SG(11)] = \{11, 7, 8\}$$

$$V''[SG(12)] = \{12, 9, 10, 6, 5\}$$

$$V''[SG(7)] = \{7, 1, 2\} \text{ and}$$

$$V''[SG(8)] = \{8, 3, 4\}$$

Remark 4: Clearly, the complexity of the above algorithm is dominated by that of Step 2, which can be implemented with a space complexity of $O(|V| + |E|)$ and with time complexity $O(|V|\log|V| + |E|)$, by using an efficient depth-first search technique [TAR74], where V and E denote the vertex and edge sets of the circuit graph.

5. SUPERGATE STRUCTURE OF MULTIOUTPUTS

For a multioutput circuit all maximal supergates can be easily found by using the earlier dominance algorithm for each output individually. Obviously, the computational effort required to evaluate the detectability of the lines in the circuit depends strongly on the complexity of maximal supergates, determined by their size and the number of fanout inputs. For a multioutput circuit, a maximal supergate of a primary output might partially or fully overlap a maximal supergate of another primary output. To avoid computation of testability parameters for overlapping nodes/edges more than once we require a minimum covering of all the circuit nodes with maximal supergates. We will show that because of certain structural relations amongst the supergates, such a minimum cover is unique and can be found in polynomial time.

Example: Fig. 6 shows a NAND realization of a full-adder circuit and its circuit graph. The vertex sets of the maximal supergates corresponding to the sum output are:

$$V''[SG(11)] = \{11, 10, 9, 8, 7, 1\} \text{ and}$$

$$V''[SG(7)] = \{7, 6, 5, 4, 3, 2\}$$

and for the carry output

$$V''[SG(13)] = \{13, 8, 1, 7, 5, 6, 4, 3, 2\}.$$

Note that $SG(13)$ properly includes $SG(7)$. Clearly, the computation of controllability and detectability for the nodes in $SG(13)$ obviates the need to recompute these parameters for the nodes in $SG(7)$. In a multioutput circuit, such a redundant computation can be eliminated by finding a minimum set of maximal supergates which cover all the nodes in the circuit graph.

Lemma 2: [SET86] In the graph of a multioutput circuit, every maximal supergate $SG(X)$ corresponding to a primary output is either properly contained within another maximal supergate $SG(Y)$ (corresponding to another primary output), or has at least one node which is not covered by any other maximal supergate.

Corollary: In a multioutput circuit graph, no maximal supergate can be completely covered by the union of all other supergates, unless it is totally covered by one maximal supergate alone.

Theorem 4: The minimum cover of all nodes in a multioutput circuit graph with maximal supergates is unique.

Example: In Fig. 6, the minimum cover consists of $\{SG(11), SG(13)\}$. [Nodes 12 and 14 can be ignored without loss of information.]

Remark 5: From the uniqueness of the cover, it follows that once maximal supergates for individual primary outputs are determined by the algorithm in the last section, finding the minimum cover in a circuit graph having $|V|$ nodes is at most $O(|V|^2)$. Since this dominates the complexity of finding maximal supergates (Remark 4) for

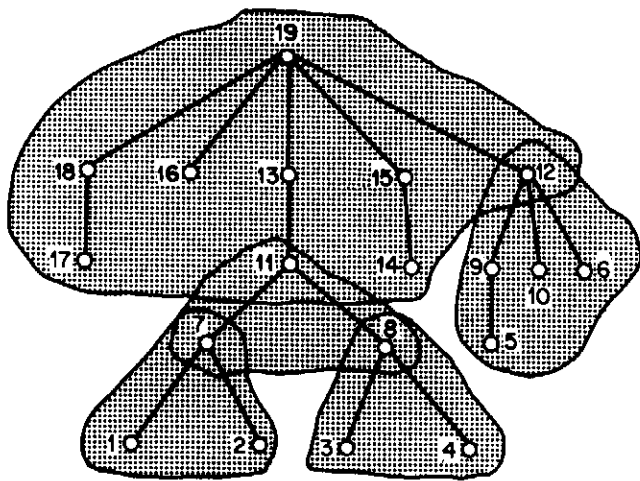


Fig. 5 Dominator tree and maximal supergate partitioning.

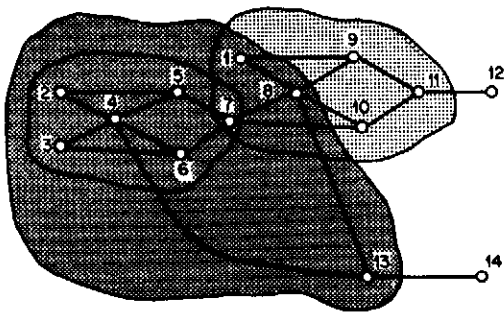
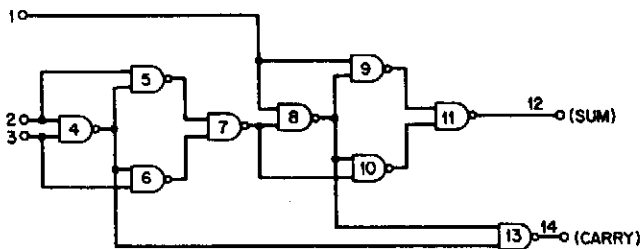


Fig. 6 The full-adder circuit and its circuit graph with the maximal supergates.

individual primary outputs, the unique cover can be identified in at most $O(|V|^2)$.

If so identified in a preprocessing step, the minimum cover of a multioutput circuit graph will help accelerate computation.

6. CONCLUSION

We have introduced a model for exact computation of line observabilities but, in order to do so it is necessary to handle non-reconvergent lines in a supergate differently from reconvergent stems. The observabilities of the latter are computed in a forward trace through the circuit along with all the line controllabilities. Then,

the primary output observabilities are initialized and a backward trace determines the observabilities of non-reconvergent lines. Whenever a reconvergent stem is encountered, the backward trace restarts with its observability value computed earlier. PREDICT-like approximations are easily extended to include this more accurate method for observability computation. For multiple output circuits, the order in which the supergates are considered, affects the efficiency of computation. We have analyzed the supergate structure of single and multiple output circuits and shown that the covering of the circuit in terms of maximal supergates is unique in both cases. Further, finding such cover has the worst-case time complexity that is quadratic in the number of nodes in the circuit graph.

Acknowledgment — Authors thank F. Brglez for technical discussions and M. Pashan for comments on manuscript.

REFERENCES

- [AGR82] V. D. Agrawal and M. R. Mercer, "Testability measures — What do they tell us?" Int. Test Conf., Philadelphia, PA, 1982, *Digest of Papers*, pp. 391-396.
- [AGR85a] V. D. Agrawal, S. C. Seth, and C. C. Chuang, "Probabilistically guided test generation," *Proc. ISCAS-85*, Kyoto, Japan, June 6-8 1985, pp. 687-690.
- [AGR85b] V. D. Agrawal and S. C. Seth, "Probabilistic testability," *Proc. ICCD '85*, New York, Oct. 7-10, 1985, pp. 562-565.
- [AHO74] A. V. Aho, J. E. Hopcroft, & J. D. Ullman, *The Design and Analysis of Computer Algorithms*, Addison Wesley, 1974.
- [BRE76] M. A. Breuer and A. D. Friedman, *Diagnosis and Reliable Design of Digital Systems*, Computer Science Press, 1976.
- [BRG84] F. Brglez, P. Pownall, and R. Hung, "Applications of testability analysis: From ATPG to critical delay path tracing," Int. Test Conf., Philadelphia, PA, Oct. 16-18, 1984, *Digest of Papers*, pp. 705-712.
- [GOL79] L. H. Goldstein, "Controllability/Observability analysis of digital circuits," *IEEE Trans. Circuits and Systems*, Vol. CAS-26, Sept. 1979, pp. 685-693.
- [JAI85] S. K. Jain & V. D. Agrawal, "Statistical fault analysis," *IEEE Design and Test*, Vol. 2, Feb. 1985, pp. 38-44.
- [SAV80] J. Savir, "Syndrome testable design of combinational circuits," *IEEE Trans. Computers*, Vol. C-29, June 1980, pp. 442-451; see also *IEEE Trans. Computers*, Vol. C-29, Nov. 1980, pp. 1012-1013 for corrections.
- [SAV84] J. Savir, G. S. Ditlow, and P. H. Bardell, "Random pattern testability," *IEEE Trans. Computers*, Vol. C-33, Jan 1984, pp. 79-90.
- [SAV83] J. Savir, "Good controllability and good observability do not guarantee good testability," *IEEE Trans. Computers*, Vol. C-32, Dec. 1983, pp. 1198-1200.
- [SEL68] E. F. Sellers, M. Y. Hsiao, and L. W. Bearnson, "Analyzing errors with the boolean difference," *IEEE Trans. Computers*, Vol. C-17, July 1968, pp. 676-683.
- [TAR74] R. E. Tarjan, "Finding dominators in a directed graph," *SIAM Jour. of Computing*, Vol. 3, 1974, pp. 62-89.
- [SET85] S. C. Seth, L. Pan, and V. D. Agrawal, "PREDICT — Probabilistic Estimation of Digital Circuit Testability," *Proc. FTCS-15*, Ann Arbor, Mich., June 19-21, 1985, pp. 220-225.
- [SET86] S. C. Seth, B. B. Bhattacharya, and V. D. Agrawal, "An Exact Analysis for Efficient Computation of Random-Pattern Testability in Combinational Circuits," Technical Report, 1986, Dept. of Computer Science, University of Nebraska, Lincoln, Nebraska.
- [WUN85] H. J. Wunderlich, "PROTEST: A tool for probabilistic testability analysis," *Proc 22nd Des. Auto. Conf.* Las Vegas, Nev. June 23-25, 1985, pp. 204-211.