An Improved Speed Sensorless Control of Self Excited Induction Generators Considering Magnetic Nonlinearity

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Abstract—**Self excited induction generators (SEIG) are widely used for wind power generation in remote areas. The challenge is to maintain constant voltage and frequency under variable load and variable wind speeds. In this paper a reactive power control technique of the generator partially through fixed capacitor bank connected at the stator terminals and switched dc power source is proposed for constant voltage operation. The dc power source can be a battery charged from a solar module and from the generator during light loaded condition. The capacitor provides the bulk excitation current for the induction generator (IG) while the inverter adds the reactive current needed to regulate the IG output voltage under variable wind speeds and loads. The method is simple and can also be applicable for a wider speed range of operation. Different simulations and experiments are performed on a real machine to validate the proposed concept.**

Index Terms—**self excited induction generators, speed-sensorless control, reactive power management, magnetization curve**

I. Introduction

Three phase self excited induction machines with a squirrel-cage rotor are widely used for small scale power generation. The excitation capacitor bank is connected to the stator terminals for such schemes [1]-[6]. The main disadvantage of using capacitor for the supply of reactive power for self-excitation of an induction generator is poor voltage regulation. So it is not suitable for wider range of operating speed and load. To overcome this problem of voltage regulation some schemes proposed switched capacitor bank. But this type of discrete switching introduces sudden voltage fluctuations. As an alternate scheme, thyristorised control of the capacitors can be employed but it introduces voltage harmonics in the generated voltage. Various control strategies have been proposed to overcome these limitations. Some of these proposals use sensor less field-orientation algorithms to excite and control the induction generator [2]-[9] which introduce serious voltage harmonics problem. Of which there are some schemes which proposed shunt–connected PWM voltage source inverter supplying constant frequency voltage [5] and some schemes exist on

supplying reactive current to the SEIG by using a static reactive power compensator [9]. Another strategy is also developed based on the instantaneous reactive power theory [10]. All these proposed algorithms assume a constant magnetizing inductance of the machine during its operation. Due to this insensitivity of the controller towards the variation of the mutual inductance at different operating points, the system becomes unstable or uncontrollable at some conditions.

In the proposed work, a strategy of a combination of inverter fed from a battery and fixed capacitances are considered for voltage stabilization of the system over a wide speed range. However, all the existing control schemes do not consider the variations in magnetizing inductances during the generator operation. Although this assumption simplifies the control schemes, this can introduce errors in various estimated quantities during the control of the generator or can add some unstable operating points or zones. The proposed technique assumes the nonlinear magnetization curve within its operating zone in order to improve the accuracy of inverter switching for voltage stabilization. The method is useful to widen the speed range of the generator without much variation of the generated voltage. The magnetization curve of the machine can be plotted during commissioning of the system. The inverter acts as a source of reactive power during the change in load or operating speed. The battery can be charged during light loaded conditions, which helps to stabilize the system voltage during such condition. A PWM strategy of switching for the inverter has been used to reduce the harmonics in the generated voltage. Various simulations and experiments are performed to justify proposed control strategy.

II. Proposed Induction Generator Model

The induction generator is modelled in synchronously rotating frame. The stator and rotor voltage equations in d-q synchronous reference frame can be written as

$$V_{ds} = r_s i_{ds} + \frac{d\lambda_{ds}}{dt} - \omega_s \lambda_{qs} \qquad (1)$$

$$V_{qs} = r_s i_{qs} + \frac{d\lambda_{qs}}{dt} + \omega_s \lambda_{ds} \qquad (2)$$

$$0 = R_r i_{dr} + \frac{d\lambda_{dr}}{dt} - (\omega_s - \omega_r)\lambda_{qr} \qquad (3)$$

$$0 = R_r i_{qr} + \frac{d\lambda_{qr}}{dt} + (\omega_s - \omega_r)\lambda_{dr} \qquad (4)$$

The flux linkage equations are,

$$\lambda_{ds} = L_s i_{ds} + L_m i_{dr} \qquad (5)$$

$$\lambda_{qs} = L_s i_{qs} + L_m i_{qr} \qquad (6)$$

$$\lambda_{dr} = L_r i_{dr} + L_m i_{ds} \qquad (7)$$

$$\lambda_{qr} = L_r i_{qr} + L_m i_{qs} \qquad (8)$$

III. PROPOSED TECHNIQUE FOR CAPACITOR CALCULATION

The steady state model of the machine is considered for capacitor calculation where the derivative terms of (1)-(4) are zero. If leakage inductances are neglected for simplicity in equations (5)-(8), $\lambda_{qs}, \lambda_{qr} = \lambda_{qm}$ and $\lambda_{ds}, \lambda_{dr} = \lambda_{dm}$.

where, λ_{dm} and λ_{qm} are d-q axes air gap flux linkages.

Therefore, from equations (1)-(4),

$$V_{ds} = r_s i_{ds} - \omega_s \lambda_{qm} \qquad (9)$$

$$V_{qs} = r_s i_{qs} + \omega_s \lambda_{dm} \qquad (10)$$

$$0 = R_r i_{dr} - (\omega_s - \omega_r)\lambda_{qm} \qquad (11)$$

$$0 = R_r i_{qr} + (\omega_s - \omega_r)\lambda_{dm} \qquad (12)$$

Also,

$$\lambda_{dm} = L_m i_{dm} \qquad (13)$$

$$\lambda_{qm} = L_m i_{qm} \qquad (14)$$

where, $i_{dm} = i_{dr} + i_{ds}$ and, $i_{qm} = i_{qr} + i_{qs}$

Neglecting stator resistance in equation (9) and (10),

$$V_{ds} = -\omega_s \lambda_{qm} \qquad (15)$$

$$V_{qs} = \omega_s \lambda_{dm} \qquad (16)$$

Assuming air-gap flux orientation of d-axis, $\lambda_{dm} = \lambda_m$ and $\lambda_{qm} = 0$. The latter gives $i_{qm} = 0$ from equation (14).

Also substitution of the same condition in equation (11) yields,

$$R_r i_{dr} = 0 \qquad (17)$$

As the rotor resistance is a non-zero quantity, $i_{dr} = 0$ from equation (17).

Therefore, $i_{ds} = i_{dm}$.

Thus the air-gap flux can be expressed as,

$$\lambda_m = L_m i_{ds} \qquad (18)$$

where, i_{ds} is the stator magnetizing current of the machine.

The corresponding phasor diagram for the machine is shown in Fig. 1. The capacitor bank is used as a bulk uncontrolled source of reactive current, whereas the inverter provides the controlled source of reactive power to regulate the stator voltage current. The design criterion for the capacitor is considered is such that it has to supply the whole reactive-current needed to generate the nominal voltage under nominal speed and no load conditions. Thus the current i_{ds} will be supplied by the capacitor only. Based on equations (15)-(18), the d-axis equivalent circuit for the generator can be developed, which is shown in Fig. 2.
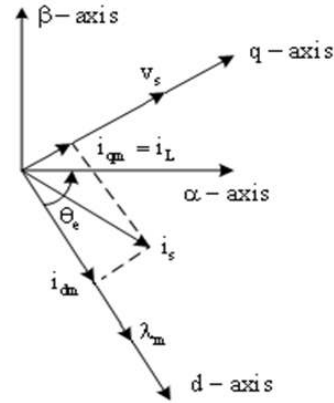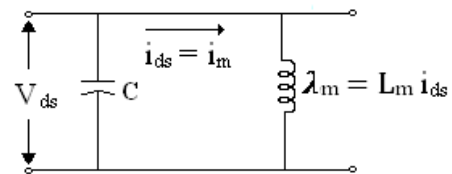

Figure 1.   Phasor diagram of SEIG

Figure 2.   D-axis equivalent circuit of the proposed generator

Therefore, from Fig. 2 it can be shown as,

$$i_{ds} = jV_{ds}\omega C$$

Or,

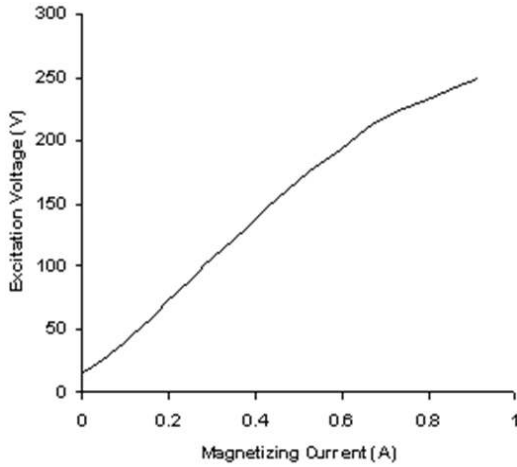$$i_{ds} = V_{qs}\omega C \qquad (19)$$

Figure 3. Open circuit characteristics (OCC)

To calculate the capacitance for the purpose to provide total reactive power at no load condition under nominal speed generating nominal voltage, the open circuit characteristics (OCC) of the machine can be considered. The magnetization curve obtained from the experiments for the given machine at nominal speed is shown in Fig. 3. The excitation voltage and magnetizing current of Fig. 3 are the q-axis stator voltage V_{qs} and d-axis stator current i_{ds} respectively. Fig. 3 shows high degree of nonlinearity between V_{qs} and i_{ds}. Therefore assumption of a linear magnetizing curve will provide a calculated value of capacitor that can result in wider variation of generated voltage with rotor speed. This can result in increased amount of reactive power to be supplied by the compensating inverter resulting in its higher size design. In the suggested method this nonlinearity has been taken into account for calculation of the required excitation capacitance. The q-axis stator voltage V_{qs} is a linear function of λ_m as observed from (16). Thus λ_m can be obtained as a non-linear function of i_{ds} similar to OCC. The mixed model equations [11] for λ_m takes the form of,

$$\lambda_m = k + k_1 i_{ds} - k_2 i_{ds}^2 \qquad (20)$$

where k, k_1 and k_2 are constants which can be obtained from the magnetization curve for the machine. Thus solving equations (16), (19) and (20),

$$C = \frac{k_1 - \sqrt{k_1^2 - 4k_2(\dfrac{v_{qs} - \omega_r k}{\omega_r})}}{2k_2 \omega_r v_{qs}} \text{ Farads} \qquad (21)$$

Equation (21) can be used to calculate the initial capacitance for the generator at no load rated voltage.

IV. PROPOSED CONTROL TECHNIQUE

A scheme of the proposed system is shown in Fig. 4. The self excited induction generator system consists of a conventional three phase squirrel cage IM, driven by a prime mover (a DC motor is used as the prime mover for experimental purpose). The system does not require any position or speed sensor. The IM stator is connected to an ac load, a fixed capacitor bank and a current controlled inverter. The capacitor bank and the inverter supplied from a battery provide the reactive current needed to excite the IM. The battery is charged from a solar module to keep its voltage constant. This scheme can operate at a remote place where no utility supply is available. The capacitor bank will act as an uncontrolled frequency-dependent reactive current source that provides the bulk excitation and the inverter will act as a controlled reactive current source that is used to regulate the stator voltage and frequency. Here the stator reactive current will be the sum of the uncontrolled capacitor-bank reactive current and the inverter reactive current. The inverter reactive q-axis current reference is kept at a minimum value near zero and the d-axis current reference is generated from the voltage loop as shown in Fig. 4.
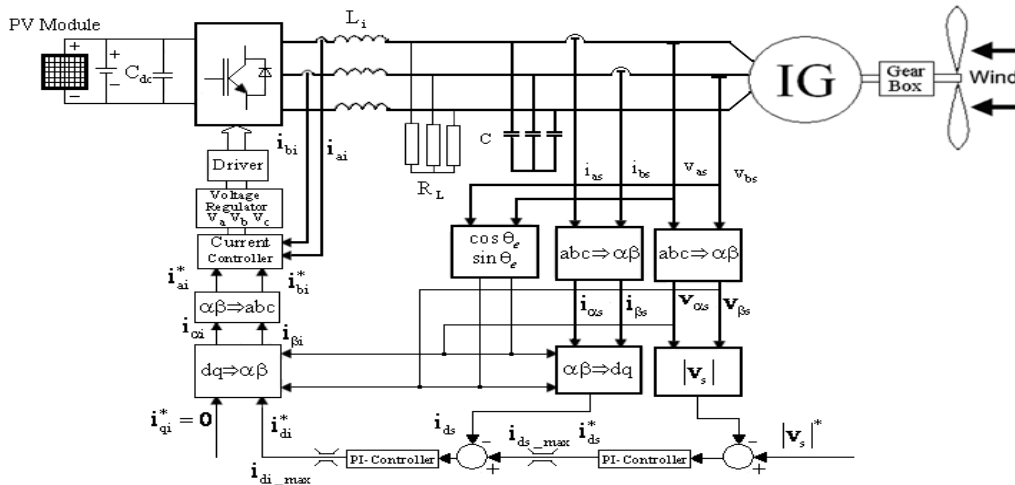

Figure 4. Experimental block diagram for SEIG.

The generation of unit vectors for air gap flux orientation can be achieved through stationary $\alpha - \beta$ reference frame stator voltage equations as,

$$V_{\alpha s} = r_s i_{\alpha s} + \frac{d\lambda_{\alpha s}}{dt} \qquad (22)$$

$$V_{\beta s} = r_s i_{\beta s} + \frac{d\lambda_{\beta s}}{dt} \qquad (23)$$

Neglecting the stator resistance drops and stator leakage inductances in equations (22) and (23),

$$\lambda_{\alpha m} = \int v_{\alpha s} dt \qquad (24)$$

$$\lambda_{\beta m} = \int v_{\beta s} dt \qquad (25)$$

Therefore the unit vectors are,

$$\cos\theta_e = \frac{\lambda_{\alpha m}}{\sqrt{(\lambda_{\alpha m})^2 + (\lambda_{\beta m})^2}} \qquad (26)$$

$$\sin\theta_e = \frac{\lambda_{\beta m}}{\sqrt{(\lambda_{\alpha m})^2 + (\lambda_{\beta m})^2}} \qquad (27)$$

V. SIMULATION AND EXPERIMENTAL RESULTS

The proposed algorithm was tested experimentally with the scheme shown in Fig. 4. The inverter q-axis reference current was kept at a minimum value so that no active power is drawn from it. The machine specification is given in Table I. The calculated capacitor considering magnetic nonlinearity to generate nominal voltage at nominal speed is 13uF was initially connected with the stator. The speed was varied by varying the DC motor speed and corresponding generated voltage both at no load and full load was recorded. The theoretical and experimental variation of the generated voltage is shown in Fig. 5 which shows a good conformity between the two. Then the same experiment was repeated with a capacitance of 10uF, which is the value considering a linear magnetic circuit of the machine. The comparative experimental voltage variation against the speed of the generator for both the capacitances is shown in Fig. 6 which shows a clear advantage of the proposed method. Then the proposed control loop was initiated and the same speed variation at full load and no load was tried. The generated voltage was recorded against the rotor speed. The comparative voltage variation with and without control at full load with 13uF capacitor is shown in Fig. 7. It is evident that the proposed technique maintains a good voltage regulation over a speed range of 0.65p.u. to 1.2p.u., which can be of good practical value. The corresponding line voltage and line current at 1.0p.u speed and 1.0p.u. load is shown in Fig. 8 and Fig. 9 respectively.

TABLE I. MACHINE PARAMETERS

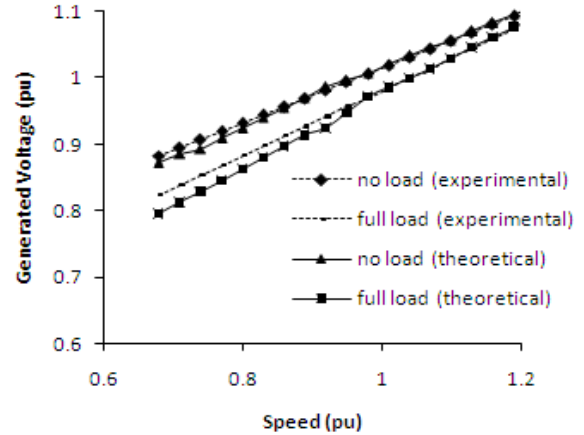| Rated output	No. Of Poles	Rated speed	Stator voltage	Stator winding	Rotor
1hp	4	1400	400V, 50Hz	Star connected	Squirrel cage

Figure 5. Theoretical and experimental variation of the generated voltage (at no-load and full-load)
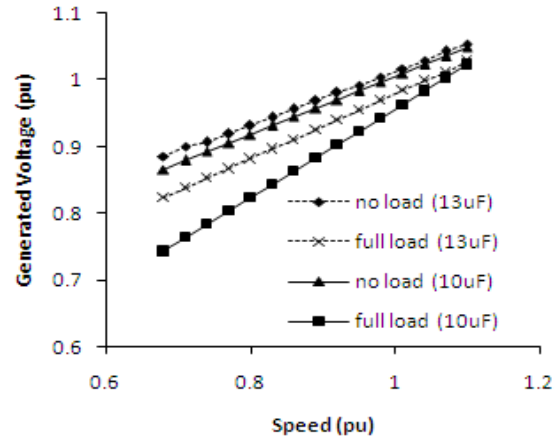

Figure 6. Comparative experimental voltage variation against the speed of the generator for both the capacitances
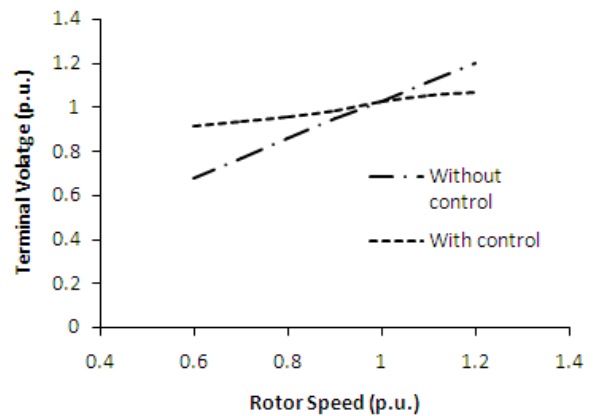

Figure 7. Terminal voltage variation with and without control at full load with 13uF capacitor
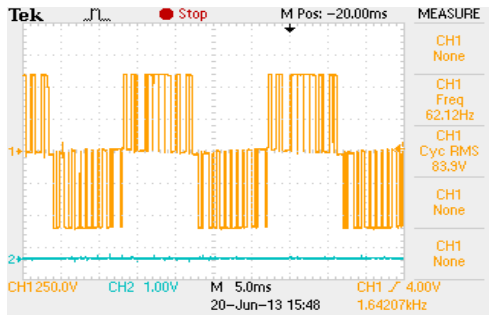
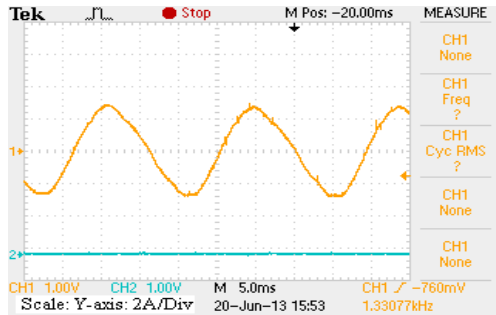Figure 8.   Line to line voltage at 1p.u. load and 1p.u. speed

Figure 9.   Line to line current at 1p.u. load and 1p.u. speed

VI.   CONCLUSION

A technique for calculation of capacitance considering magnetic nonlinearity for three phase SEIG is shown. The proposed method widens the operating speed range of the generator with stabilized terminal voltage. The method requires an additional inverter for supplying the reactive power only during the speed or load variation. The required inverter power can be shown to be much less than the capacity of the SEIG and for the proposed scheme a solar panel is used to supply inverter dc power. The proposed method is simple to implement and will require no high-end hardware. Various experimental results are shown to validate the proposed concept.

ACKNOWLEDGMENT

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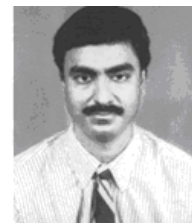
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