



An Overview of Power Electronics Techniques in Electric Systems for Transport Applications

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Abstract: This paper presents an overview of power electronics techniques used in electric systems for transport applications, including more-electric aircraft (MEA), electric ship (ES), and electric vehicle (EV). As the worldwide economy is continuously increasing, there is more demand for transportation, which is one of the most significant sources of carbon emissions. In order to achieve the target of low greenhouse gas (GHG) emission rate and ultimately realize sustainable development, the concept of using electric systems to replace the other types of systems that utilize conventional energy resources such as fossil fuels. The subsystems utilizing hydraulic, mechanical, and pneumatic power that were previously used in conventional airplanes are replaced by electrical systems in MEA, which reduces the fuel consumption, operating cost, and noise of aircrafts. Besides, in ES applications, by applying electric propulsion systems, the negative impacts caused by using long shaft-lines to the propellers can be eliminated since the flexibility of on-board equipment placement is obtained. Additionally, the concept of EV is becoming popular in the automobile industry due to the stringent constraints on energy resources and environmental concerns. In order to learn about the research interests and trends in power electronics techniques for these more- or all-electric transport applications, a review is carried out in this paper.

Keywords: electric systems, transport applications, more-electric aircraft, electric ship, electric vehicle, power electronics techniques

1. INTRODUCTION

At present, the airplane, ship and vehicle are the three main types of transport tools, and increasing use of these tools can be observed all over the world, since mobility is one of the most important parameters for achieving economic development. However, according to [1], one third of the total greenhouse gas (GHG) emissions are generated from the transportation system. In order to achieve the ultimate aim of sustainable development, more environmentally friendly means for transportation should be developed. In address to the three forms of transportation mentioned above, the concepts of more-electric aircrafts (MEAs)[2], electric ships (ESs) [3] and electric vehicles (EVs) [4] were proposed, and electric propulsion systems are used to in these applications to replace the original ones based on conventional energy resources for carbon emission reduction.

For MEA, the subsystems utilizing hydraulic, mechanical, and pneumatic power that were previously used in conventional airplanes are replaced by electrical systems. The pneumatic power is used to produce bleed air for the Environmental Control System (ECS) to make the passengers feel comfortable as it is used for pressurization and air-conditioning purposes. In addition, it also plays an important role in de-icing for the wings of an airplane. In terms of the mechanical power, it is transferred from the engines to different kinds of mechanically driven hydraulic pumps, generators and subsystems by using the gearboxes [5]. High power density and robustness are obtained for the hydraulic power subsystems, while they are heavy and inflexible [5]. In the relatively new types of airplanes such as the Boeing 787 and Airbus A380, increased use of electrical power systems is realized, which is beneficial for mitigating CO₂ emissions all over the world. With the utilization of MEA concept, all the on-board loads are required to take power from an electrical system [6, 7]. Therefore, the overall weight can be reduced since the mechanical and hydraulic systems are removed. In addition, owing to the elimination of the pneumatic system, the efficiency of the turbine

is greatly improved since there is no need for a bleed air system on the gas turbine. As the technology of MEA develops, minimized mechanical linkages and hydraulic power supply networks are obtained, which reduces the complexity in maintenance [8].

The history of ES dates back to over a century ago [9, 10], where a limited number of such applications was presented. Massive production of ESs was not realized until the 1980s, since at that time the maturity of semiconductor technology was good enough for variable speed control of electric motors. The electric propulsion system utilizes a number of small units for power generation instead of a big prime mover, in which case the generator engines are always loaded close to their optimal operating points [11]. By applying electric propulsion on ships, the placement of equipment on-board is flexible, which eliminates the negative impacts caused by using long shaft-lines to the propellers. There are a number of variants for electric propulsion solutions which are dependent on the type of vessel, operational profiles, and the current available technologies for construction [11]. Three different application areas of electric propulsion systems for ships were introduced in [11], which are 1) ocean going; 2) station keeping; 3) icebreaking [11]. It was also indicated that the future trends and technologies for ES involve DC grid, energy storage, hybrid propulsion and dynamic position (DP) closed bus. Furthermore, future electrical energy sources can be accommodated for ESs, since the relevant necessary infrastructures are available on-board.

Before the millennium, the concept of EV had occurred, but the modern EV is totally different from what was defined for the classical one. In this new EV concept, the electric motor, power converter, and energy sources are involved and distinct characteristics are presented. Moreover, clean and efficient road transportation is to be accomplished by widely applying the modern EVs [12]. Owing to the facts that fossil fuels around the world are limited [13], traffic jams in most of the metropolitan cities are becoming increasingly severe [14], and the procedure of global warming is to be slowed down or even stopped [15], more stringent constraints on the use of traditional internal combustion engine (ICE) driven vehicles are presented. Under this circumstance, it is rather urgent to develop environmentally friendly solutions for transportation. Therefore, the vehicles that utilize partial or full electric propulsion systems emerge, namely electric vehicles (EVs) [12]. The concept of EV can be further divided into three categories in general, which are battery electric vehicle (BEV), hybrid electric vehicle (HEV), and fuel-cell electric vehicle (FCEV) [4, 12].

For the development of these electric systems in applications for transportation, one of the most critical technologies is power electronics (PE) [16-30]. Since the on-board electric power systems for MEA, ES and EV are relatively complex, and there are different types of load demands in these systems, the reasonable control of power flow is important, which is realized by power electronic devices. In this paper a review is carried out on the applications of PE techniques in MEA, ES and EV, respectively, which are illustrated in the following three sections (Section II, III and IV). In Section V, a conclusion will be given.

2. PE TECHNIQUES IN MEA

With the purpose of reducing the fuel consumption, operating cost, and noise of aircrafts, the emerging concept of "More-Electric Aircraft" (MEA) arises and substantial efforts were made in this area. The main idea of MEA is replacing the subsystems utilizing hydraulic, mechanical, and pneumatic power that were previously used in conventional airplanes by electric systems[7]. In a traditional civil airplane, 115V line-to-neutral ac voltage and a stable line frequency of 400Hz are applied in the electric power system [2]. While in MEA, usually high voltage direct current (HVDC) electrical systems are considered. To achieve a lighter weight generator system, 230/400Vac, 400Hz is a commonly adopted option. Besides, the DC voltage value of 270V on the DC-bus can be utilized, and therefore a higher voltage of 540Vdc (± 270 Vdc) can be utilized for the motor controllers [5]. As an alternative, variable bus frequency between 350 and 800Hz according to the changes in the engine speed with the voltage regulated at 115 or 230Vac can be adopted, which increases the reliability of the whole system [31]. Moreover, there are 28Vdc loads on MEA, where the 28V DC voltage is derived by converting the 115Vac, 400Hz with the application of transformer rectifier units (TRUs) [32].

Since there are various types of loads (AC and DC) in MEA, and more complex system configuration is presented, the development of power electronic techniques plays a significant role in the future aerospace applications. According to [33], the hydraulic systems will be replaced by variable-speed

generators along with power electronics, which are lighter and endowed with a simpler structure. A comparative evaluation of three-phase AC-DC converters with high power factor was carried out for future MEA application in [34]. The analysis focused on a passive autotransformer-based 12-pulse rectifier system and an active three-level boost-type pulse width modulation (PWM) rectifier system. The most advantageous solution can be derived in the three-level PWM rectifier, and the three-phase buck plus boost PWM rectifier topology proposed in [35] was demonstrated to have a high system reliability, since the output capacitor is not required to be charged in advance and load current limitation for the short circuit case is avoided for startup. In addition, the black-box modelling of three-phase voltage source inverters (VSIs) was carried out in [36] for system-level analysis in MEA, where the model is simple, the large-signal behaviour of the converter can be derived, and no internal data is required for the converter. On top of that, for the purpose of eliminating the harmonics produced by the on-board nonlinear loads, the design of a fully digitally controlled 400-Hz active power filter (APF) was presented for aircraft applications in [37].

Multi-level converters are used in some aerospace applications since lower output harmonic distortions and lower device voltage rating requirement are presented, and they are good candidates for fault-tolerant ability improvement in MEA systems. The power losses and the harmonic distortion of the output waveforms were compared for 2-level and 3-level power electronic converters for aerospace applications in [38], and the superiority of the 3-level topology was demonstrated. As the increasing use of power electronic subsystems bring several challenges to aircraft power distribution, a five-level voltage source converter (VSC) was employed in [39] to achieve the high performance control of harmonic currents along with a modified PWM strategy. In [40] the compensation of load harmonics in variable-speed constant-frequency (VSCF) aircraft power systems was realized by designing a five-level cascaded H-bridge active shunt filter. In addition, a 50% reduction in the converter switching frequency was accomplished by applying a novel modulation technique, which is based on a phase shift method in [41].

In [42] an experimental tool was introduced for fault-tolerant machine development in aerospace applications, and a multiphase two-level inverter was presented in the research platform. Data acquisition, motor control, and fault monitoring were completed by using a field-programmable gate-array (FPGA)/digital-signal-processor (DSP) controller. In order to make the motor test platform capable of driving different kinds of machines, the following requirements are presented [42]: 1) multiphase output; 2) different number of phases can be used to drive various categories of motors; 3) high-speed controller; PWM generation with high precision; 5) rapid-code-development design cycle. The schematic of the six-phase motor drive system is illustrated in Fig. 1.

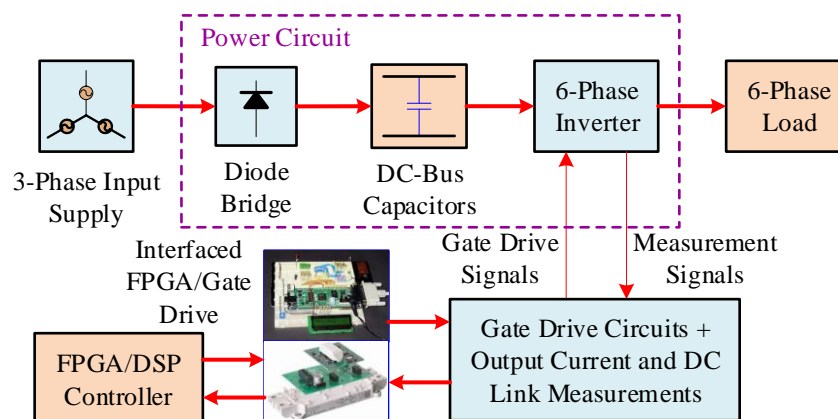


Fig1. Six-phase motor drive system [42]

The design of a unidirectional 5-kW three-phase delta-switch rectifier was proposed in paper [43], which is applicable for the application in aircraft systems with a mains root-mean-square (RMS) voltage of 115V. The three phases can be controlled at the same time with a novel PWM current controller modulation concept, and the optimal switching sequences are generated. Single phase loss can be handled by the proposed three-phase delta-switch topology, and a slightly higher efficiency is derived compared with a 6-switch Vienna-type rectifier, which is the ideal candidate for the increased mains voltages of $230V_{RMS}$ for future airplanes. The equivalent circuit of a two-level three-phase delta-switch rectifier is displayed in Fig. 2.

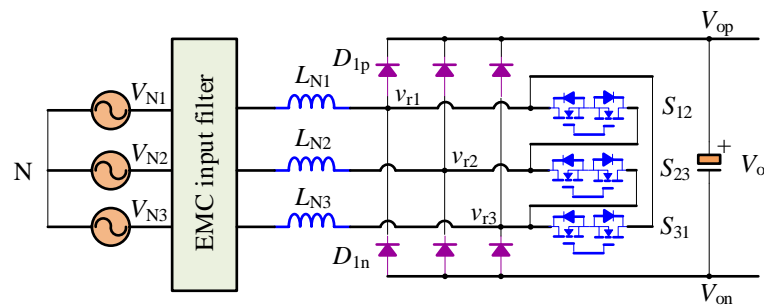


Fig2. Two-level delta-switch rectifier[43]

Good performance of power electronic converters in aircraft applications are based on accurate tracking of the fundamental phase and frequency in real time, which is commonly realized by using phase-locked loop (PLL) based algorithms. Owing to the harmonic contents and measurement noise in aircraft AC power systems, variable fundamental frequency can be observed which fluctuates between 360 and 900Hz. In order to provide higher quality for the phase and frequency estimations under this circumstance, a novel PLL scheme was proposed based on a real-time implementation of the discrete Fourier transform (DFT) in [44]. Compared with a standard PLL algorithm, the proposed DFT-PLL has better dynamic performance. Meanwhile, it is less sensitive to harmonics, noise, and signal amplitude variations. Furthermore, a robust PLL algorithm was raised by using a prediction-correction filter in [45], which was experimentally compared with the aforementioned DFT-PLL proposed in [44]. This PLL solution is based on a third-order linear and time-invariant observation model, which is obtained from a steady-state linear Kalman filter (SSLKF), and the computation burden was mitigated since the calculation of the correction gain vector can be done offline. The frequency estimation can be derived nearly without time delay for the proposed algorithm, which increases the bandwidth of a closed-loop control system.

An attractive type of AC power distribution network in MEAs is a high-voltage direct current (HVDC) one, and the power flows among three HVDC electric networks in MEAs were controlled by a high-frequency transformer, which was included in an isolated multi-port power converter proposed in [46]. Aiming at controlling the DC-bus voltage, a proportional gain DC-bus voltage droop control was implemented in all the converters used in the system. The multi-port power converter was analysed by modelling it as a two input and two output system. Such a type of converter is particularly fitted in multiple voltage electrical systems with storage elements [47]. The topology of the proposed multi-port power converter is displayed in Fig. 3.

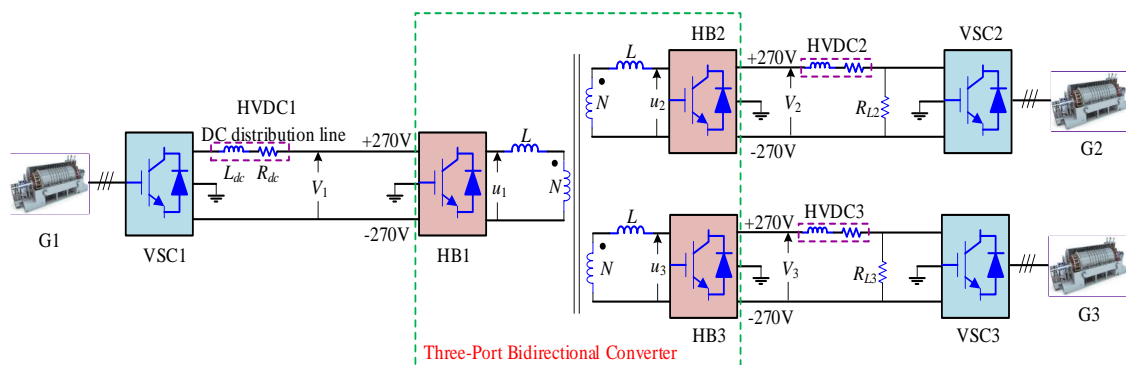


Fig3. Aircraft HVDC electrical network and multi-port converter system for HVDC electrical network power transfer [46]

3. PE TECHNIQUES IN ES

Recently, electric propulsion has been developed for various types of vessels, thanks to the rapid development of semiconductor switching devices, which can be applied in high power drives [11]. Owing to the requirement of saving fuel consumption, the use of electric propulsion in commercial ship applications is widely observed, which substitutes the original mechanical counterparts. In the future electric ship (ES) applications, multiple energy sources, independent operation of individual power producers, and energy storage for different types of applications will be allowed for further optimizing the system efficiency [11].

Since the AC power grid is commonly used in current ES applications, medium-voltage AC (MVAC) was applied in [48], and the power conversion is completed by using solid-state transformers (SSTs) [49, 50]. While the large size of SST is a problem, and synchronous generators have to be applied, and the reactive compensation has to be taken into consideration. On the other hand, the medium-voltage DC (MVDC) power system [51-54] emerges which can make the generators paralleled even if an asynchronous case is encountered. In addition, the elimination of the gearbox between the prime motor and generator is realized thanks to the merit that no frequency limitation is presented for the generators. On top of that, the system losses are reduced due to the fact that no reactive power transmission and skin effect need to be taken into account [55]. The structures of the ES by applying MVAC power system with SSTs and MVDC power system with DC-DC converters are displayed in Fig. 4(a) and (b), respectively.

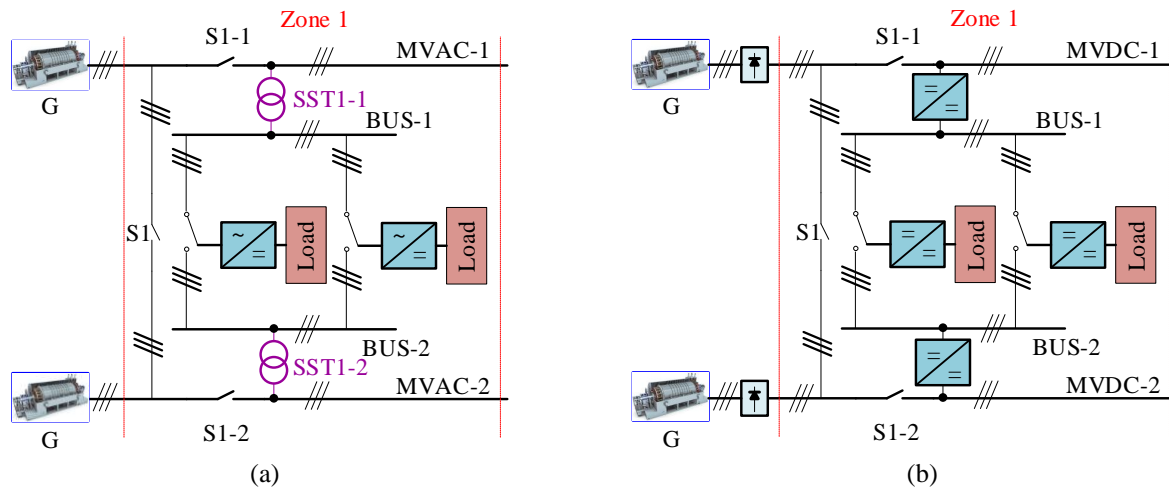


Fig4. Power system for ES (a) MVAC power system with SSTs (b) MVDC power system with DC-DC converters [55]

The integration of power electronic converters in ESs brings about many advantages such as reduction in weight and space, and flexible arrangement of equipment. Therefore, the further development of AESs can be achieved, and the integration of power electronics contributes to the emergence of shipboard hybrid power systems. However, the system structure becomes more complex compared to those on traditional vessels and ships, and more effective analysis tools should be applied to realize the aforementioned benefits. A review of hybrid ES modelling was carried out in [56], and the transitions between two separated models were avoided by proposing a unified model for bidirectional converters. The models were established for various on-board electrical and mechanical elements, including the synchronous generator-rectifier system, inverters, DC-DC converters, diesel engine, propellers, and ship hydrodynamics [56]. The system being focused on is a DC hybrid one, which includes possible integration of clean energy resources such as fuel cells and solar energy, and a single-line diagram of a shipboard DC hybrid distribution system is illustrated in Fig. 5.

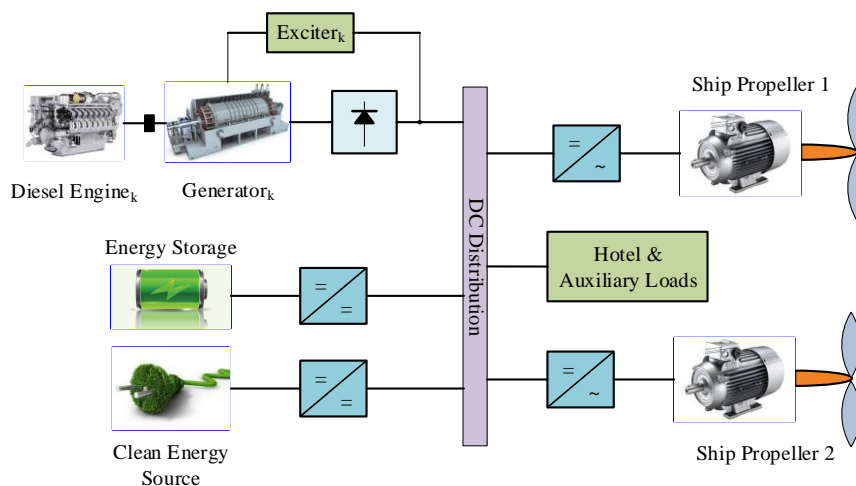


Fig5. Single-line diagram overview of a typical shipboard DC hybrid EPS [56]

The modelling of power electronic converters was based on averaging methods, and the computation burden and time were reduced by neglecting high-frequency switching behaviour. Then a simulation platform was developed for system-level studies with the application of the obtained models. Since significant savings in time and computation intensity, the proposed simulation platform can be useful for studies on AES DC power systems in the long run.

In order to increase the efficiency and compactness of the power electronic converters for AES applications, a novel approach of voltage control in AES power systems was proposed in [57], where the switching losses were to be minimized for these converters. Similar to most other recent studies in AES applications, the notional MVDC ship-board power system (SPS) [58] was investigated in [57]. For a determined MVDC system, the settings of DC voltage reference and the optimal power reference have to be defined in advance for the VSCs operating in the voltage regulator mode and power dispatcher mode, respectively [59]. The equivalent circuit for this configuration is shown in Fig. 6.

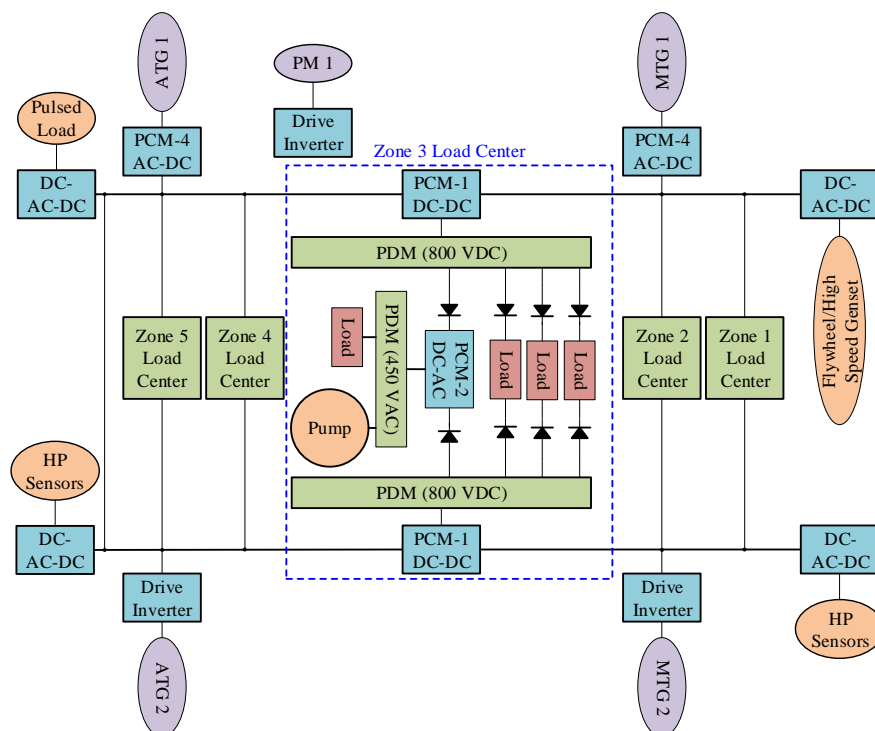


Fig6. MVDC architecture of SPS [58]

In this system, four generators are utilized, including two main generators (MTG1 & 2) and two auxiliary ones (ATG1 & 2). A medium voltage DC ring bus operates at 5kV, which is fed by the generators through transformers and AC-DC converters. In this SPS, the DC power is distributed among five zones including the loads, VSCs, power conversion modules (PCMs), and power distribution modules. The 5000V DC voltage is stepped down to an 800V one by PCM1, while PCM2 is used to perform DC-AC conversion to supply AC zonal loads such as the propulsor motor (PM). Besides, the PCM4s function as AC-DC converters, which are usually connected to the generators. Furthermore, energy storage devices, a pulsed load device, a pump, and high power sensors are involved in this system [57]. By employing the proposed operating principle for the MVDC SPS with the flexible-voltage DC bus, a minimum-voltage operation can be derived which contributes to reduced switching losses, which improves the system efficiency without compromising the converter ratings.

As a three-level converter has the advantage of reducing the voltage stress of the switches to half of the input voltage, it is a promising topology to be applied in ES MVDC application. A three-phase three-level (TPTL) phase-shifted PWM (PSPWM) DC-DC converter was proposed in [60] for ES MVDC direct current application. The semiconductor switches and diodes will experience relatively high current stress when single-phase three-level converters are applied for MVDC MW level application. Therefore, a multiphase DC-DC converter was presented to solve this problem. However, the design and implementation of such a converter in the case of high power is a challenge. With the

purpose of decreasing the total harmonic distortion (THD) of transformer primary side voltage, the utilization of PSPWM was applied with the best operating point. The MW level TPTL DC-DC converter proposed in [60] has the merits of efficiency over 95%, output voltage ripple less than 4%, undershoot and overshoot less than 17% and 11% respectively, and the settling time for the dynamic load is shorter than 0.5s. The circuit structure of the proposed TPTL phase-shifted DC-DC converter is displayed in Fig. 7.

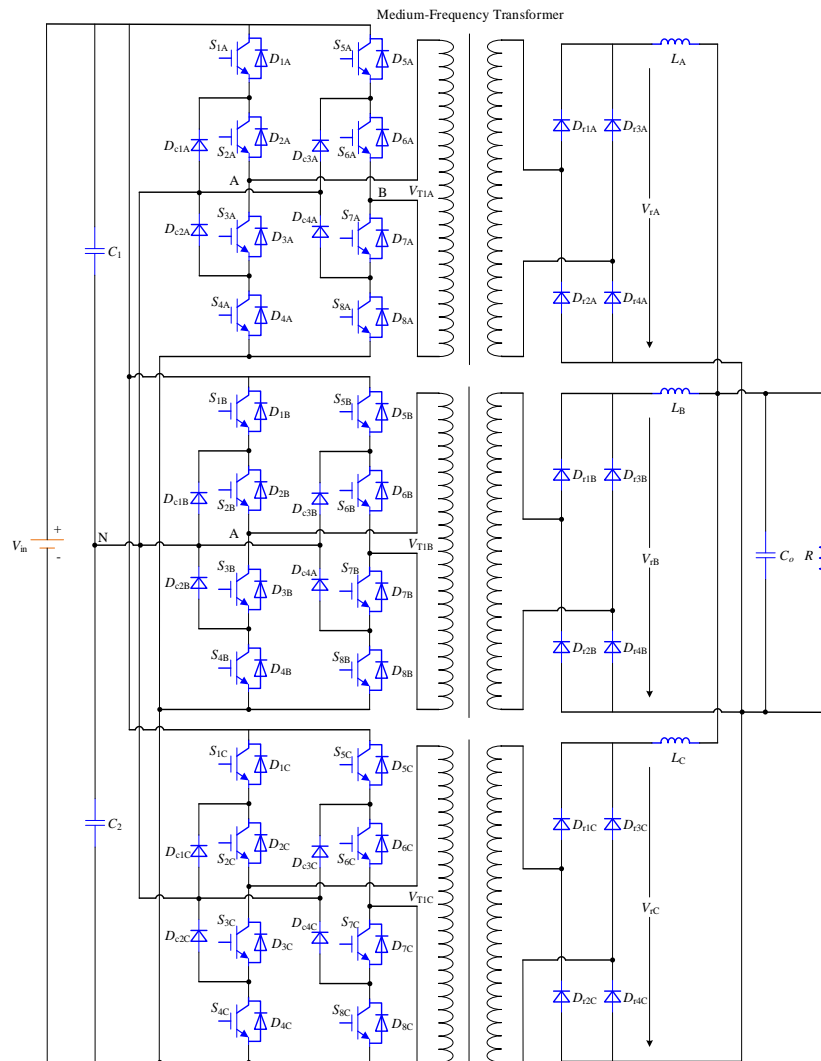


Fig7. Proposed TPTL phase-shifted DC-DC converter [60]

The proposed PSPWM DC-DC converter was proven to be suitable for ES MVDC application, and its performance features can be summarized as [60]: 1) good dynamic performance under low switching frequency; 2) small ripples in the output voltage; 3) good total current sharing; 4) high reliability; 5) plug and play capability; and 6) high efficiency. Therefore, by negotiating the loads which require low output voltage and high currents, a high input voltage can be accepted for the proposed converter.

Apart from TPTL DC-DC converter, the DC-DC modular multilevel converter (MMC) is also a promising candidate for the modern ES MVDC power system applications. In [55] the modelling and control of an isolated DC-DC MMC with a medium-frequency AC-link transformer was investigated for ES MVDC power system. The MMC described was originally based on a nine-order dynamic model. In order to derive the steady-state and small-signal models, the fundamental period averaging (FPA) method was employed owing to the fact that the phase-shifted operation is used for controlling such a DC-DC MMC. In this case, the original dynamic model was transformed into a 15-order one, and the DC and fundamental AC components were regarded as new state variables [55]. In addition, the design of the modulation index control and the simple and efficient competitive dual closed loops was presented, based on the on-ship requirements. The proposed DC-DC MMC system in [55] is displayed in Fig. 8.

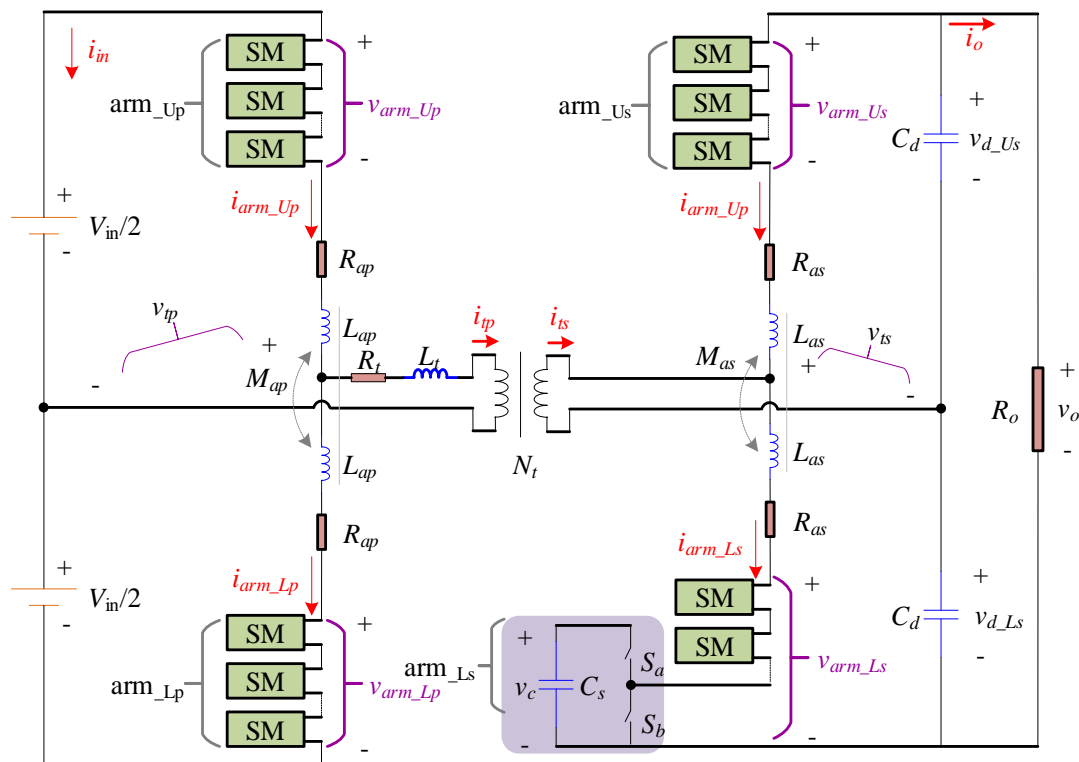


Fig8. Proposed DC-DC MMC system for ES MVDC application[55]

By using an isolated DC-DC MMC, several submodules (SMs) are introduced to form the arms, which are then applied in the process of forming the arrangement of the MMC system. In this case, it is flexible for the configuration to achieve different voltage levels, and the voltage balancing is easy as only float capacitors are involved in the SMs. In addition, since the modular structure and decentralized energy storage are permitted in such a configuration, the fault tolerance is improved. Moreover, the application of the ac-link transformer provides the galvanic isolation and voltage matching, in which case the sizes of the transformer, arm inductors and capacitors in SMs are likely to be reduced by a large extent [55]. In contrast, several challenges are caused such as parasitic capacitances between the heat sink of insulated gate bipolar transistors (IGBTs) and the ground, leakage currents occurred at the instant of switching the SMs, much higher system order hence the complicated steady state and dynamic behaviours, and challenging controller design [55]. After the proposed modelling and control strategies were applied for the specific DC-DC MMC for ES MVDC application, only two arrangements are included, which reduces the number of SMs and their parasitic capacitances. Under this circumstance, the leakage current is alleviated. In addition, the fundamental frequency modulation technique in [61] was applied to minimize the switching times of each SM for further alleviation of the leakage current. On top of that, the useful model information for control purposes was maintained, which enables the steady state and small-signal analysis. Furthermore, the tradeoff between the inner current stresses and the output voltage regulation was achieved by the proposed modulation index control.

4. PE TECHNIQUES IN EV

The concept of EV can be further divided into three categories in general, which are BEV[62], HEV[63-65], and FCEV[66]. The volume of battery storage is the main issue that exists for BEV, and BEV is mainly feasible for the cases when small battery size is needed, for example, the small EVs in short-range low-speed community transportation applications. The customers' requirements can be generally met when HEVs are used, while the cost is a little bit high in this case. Long-term potential is presented by utilizing FCEVs, while the development of FCEVs is still undergoing. In addition, some major issues are induced by using this type of EV [4]: 1) the price is high and the performance is limited by the life cycle of the fuel cells; 2) the energy density has to be increased in order to meet the requirements of the on-board hydrogen storage; and 3) construction of a hydrogen distribution and refueling infrastructure. The comparisons among BEV, HEV and FCEV are displayed in Table 1.

Table1. Characteristics of BEV, HEV, and FCEV [4]

	BEV	HEV	FCEV
Propulsion	* Electric motor drives	* Electric motor drives * Internal combustion engines	* Electric motor drives
Energy storage subsystem	* Battery * Supercapacitor	* Battery * Supercapacitor * Fossil or alternative fuels	* Hydrogen tank * Battery/supercapacitor needed to enhance power density
Energy source & infrastructure	* Electrical grid charging facilities	* Gasoline stations * Electrical grid charging facilities (for plug-in hybrid)	* Hydrogen * Hydrogen production and transportation infrastructure
Characteristics	* Zero local emissions * High energy efficiency * Independent of fossil fuel * Relatively short range * High initial cost * Commercially available	* Low local emissions * High fuel economy * Long driving range * Dependence on fossil fuels * Higher cost than ICE vehicles * Commercially available	* Zero/low local emissions * High energy efficiency * Independent of fossil fuel (if not using gasoline to produce H ₂) * High cost * Under development
Major issues	* Battery sizing and management * Charging facilities * Cost * Battery Lifetime	* Battery sizing and management * Control, optimization and management of multiple energy sources	* Fuel cell cost, life cycle and reliability * Hydrogen production and distribution infrastructure * Cost

The fast development in the power electronics techniques is the basis for modern EV concept, since several technologies for EV applications such as power conversion[67, 68], motor control[69] and fault tolerant operation[70-72] are strongly related to the performance of power electronic devices in the system. There are a number of researches on the topology, modulation and control of DC-DC, AC-DC and DC-AC converters applied in EVs.

A new bidirectional, isolated topology of DC-DC converter was proposed in [73] for FCEV driving system, by considering the difference between the characteristics of the fuel cell and traditional chemical-power battery, along with the consideration of the safety requirements. The comparison between the power circuits of the traditional and the proposed bidirectional DC-DC converters is illustrated in Fig. 9.

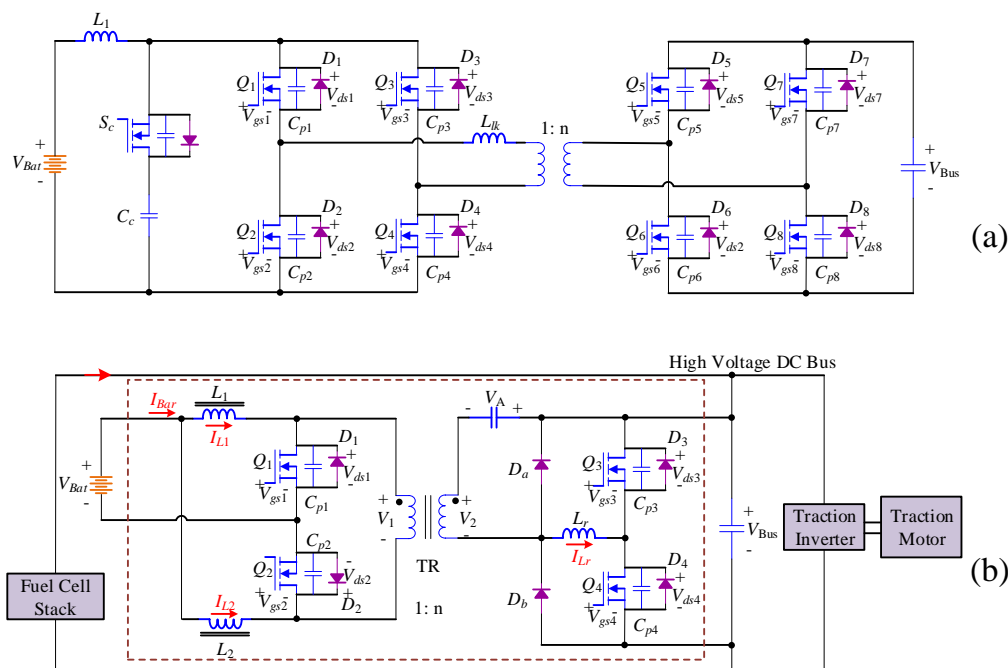


Fig9. Power circuits of (a) a conventional full-bridge bidirectional DC-DC converter; (b) the proposed isolated bidirectional DC-DC converter [73]

The traditional full-bridge bidirectional DC-DC converter is the best choice for EV applications, while the complicated configuration, high cost and large size are the main problems in this configuration. By using the proposed isolated bidirectional DC-DC converter, high efficiency, simple structure and low cost can be achieved, and the safety requirements for the isolated low- and high-voltage sides. There are two operation modes for the proposed converter: 1) the boost mode, where power flows from the low-voltage side (LVS) to the high-voltage side (HVS) and the energy is drawn from the battery; 2) the buck mode, where the battery is recharged from the high-voltage DC bus.

In [74], a load adaptive control approach was proposed for a zero-voltage-switching (ZVS) DC-DC converter for EV application. The converter should have the ability to sustain ZVS from full-load to no-load owing to the demand of a wide range of load variations. The extra conduction losses in the power MOSFET and those in the auxiliary circuit were minimized by determining the optimal value of the reactive current injected by the auxiliary circuit. A complete analysis of the converter behaviour was presented during the switching transitions. Enough current for charging and discharging the snubber capacitors during the deadtime was guaranteed by controlling the peak reactive current through the effective control of the switching frequency. Besides, some practical issues arose in this application were discussed [74]. The diagram of a ZVS full-bridge converter is displayed in Fig. 10.

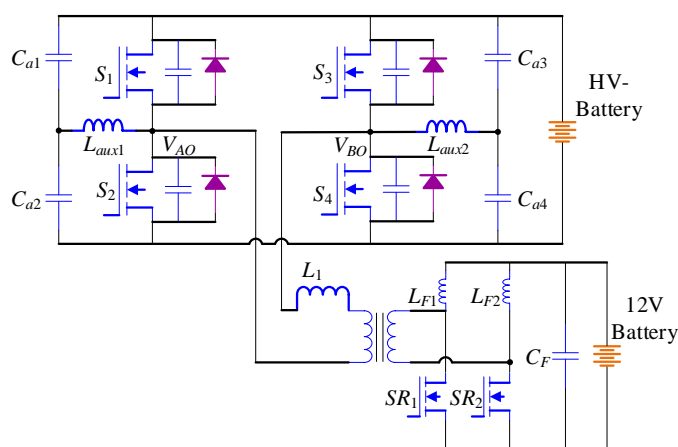


Fig10. ZVS full-bridge converter [74]

Since the processing of battery power is of great importance in EV applications, the selection of DC-DC converter plays a significant role. A three-level DC-DC converter was proposed in [75], where a comparison was made between the proposed three-level converter (TLC) with the conventional two-level bidirectional buck/boost converter (CBC) and interleaved bidirectional buck/boost converter (BIC). The schematics of the three aforementioned converters are shown in Fig. 11.

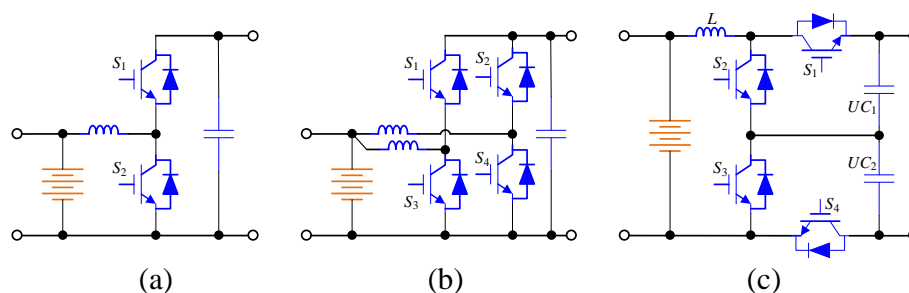


Fig11. Nonisolated bidirectional buck/boost converters (a) two-level (CBC); (b) interleaved (BIC); (c) three-level (TLC) [75]

The magnetic component size/weight and overall efficiency of these three types of converters were taken into consideration in the procedure of comparison. In [75], the power references were derived from a wavelet-transform-based energy management strategy with varying energy source voltages and traction power. The application of TLC in high-power applications is beneficial for mitigate the side effects introduced by using the boost inductor and high-voltage switches. The TLC converter was demonstrated to have the smallest size magnetic component, and considerable efficiency improvement can be obtained under given driving cycle conditions, especially at high switching frequencies [75].

In [76], the performance analysis and comparison of two types of bidirectional DC-DC converters were made in terms of the device requirements, rating of switches and components, control strategy, and performance. The two types of converters are separately the cascaded buck-boost capacitor in the middle (CBB-CIM) and cascaded buck-boost inductor in the middle (CBB-IIM) for EV applications. Both of the two types of converters have the input and output voltage overlap capability, which is an indispensable feature. The circuit structures of these two types of DC-DC converters are displayed as shown in Fig. 12.

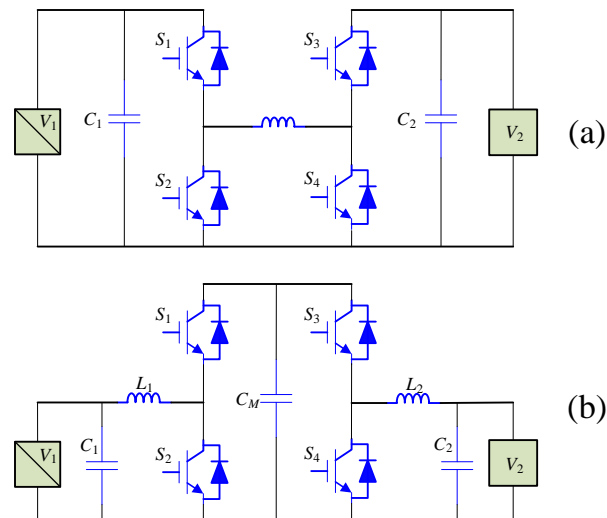


Fig12. The equivalent circuits of (a) CBB-IIM and (b) CBB-CIM [76]

It was concluded in [76] that CBB-CIM is endowed with better performance, system control flexibility and reliability than CBB-IIM. While fewer components are requested for CBB-IMM.

In [77], a two-leg interleaved bidirectional front-end DC-DC buck-boost converter was proposed for the control of an IPMSM drive in EV. The front-end converter can be used to establish adjustable and boostable DC-link voltage, which is important for improving the high-speed driving performance of the motor drive. Apart from that, there are some other benefits that are derived from using such a topology [77]: 1) the battery bank can be charged from the motor during regenerative braking; 2) good current sharing control is achieved between converter modules; 3) the developed robust control scheme preserves the DC-link boosted voltage regulation response; 4) fault disturbance rejection response is obtained against the module fault and recovery can be achieved; 5) more flexible battery voltage selection can be achieved, and the whole battery-powered motor-drive rating utilization is increased. The configuration of the power circuit of the two-leg interleaved front-end DC-DC converter with a battery bank is shown in Fig. 13.

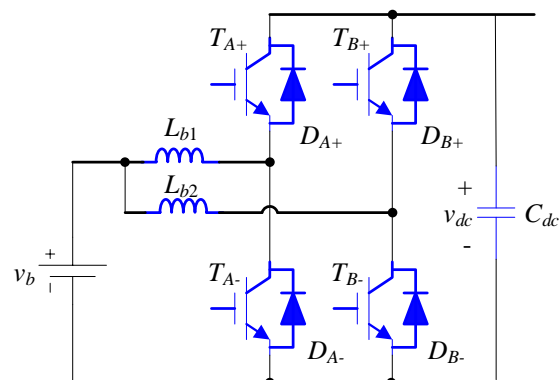


Fig13. A two-leg front-end DC-DC converter with a battery bank [77]

In [78], a bidirectional Z-source nine-switch inverter (BZS-NSI) was proposed for HEV application, which has two AC and one DC bidirectional terminals. In a conventional configuration of a series or series/parallel HEV, a bidirectional DC-DC converter (BDDC) and two DC-AC inverters are contained in the power electronic module. A nine-switch inverter (NSI) was proposed in [79] for the first time, and the concept of NSI was applied in [78] to replace the functions of the two DC-AC

inverters, which saves the space and reduces the cost. Two HEV power electronic circuit configurations were proposed, including the HEV based on BDDC and NSI, and that based on BZS-NSI. The power circuits of the conventional HEV based on BDDC and two six-switch inverters (SSIs), the proposed HEV based on BDDC and NSI, and the proposed HEV based on BZS-NSI are displayed in Fig. 14.

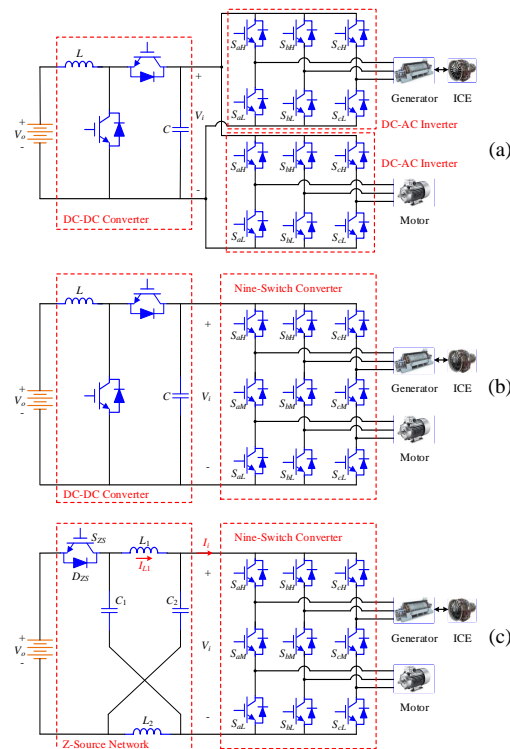


Fig14. The power circuits for (a) the conventional HEV based on BDDC and two SSIs; (b) the proposed HEV based on BDDC and NSI; the proposed HEV based on BZS-NSI

With the employment of NSI in HEV applications, the following advantages can be obtained: 1) reduced cost; 2) smaller volume; 3) efficient thermal management; 4) higher system reliability.

For the AC-DC converters used in EVs, there are two stages to be accomplished [80, 81]. The first stage is the power factor correction (PFC) AC-DC boost converter for the purpose of reducing the input current harmonics to the utility grid and complete the conversion of input AC voltage to the DC-bus voltage. In the second part, high-frequency galvanic isolation is provided by an isolated DC-DC converter. A new control approach based on the differential flatness theory was proposed in [80] for an AC-DC converter used in EVs. Compared to the conventional controller, the dynamic behaviour of the input PFC stage is improved by using the proposed control strategy, and load adaptive regulation of the intermediate DC-bus voltage can be achieved. The DC-DC full-bridge converter can operate optimally from no-load to full-load under the load-adaptive DC-bus voltage regulation. The implementation of the input PFC and AC-DC conversion was accomplished by using two interleaved boost converters in [80]. In addition, the DC-bus voltage is adaptively controlled based on the power demand from the charging curve of the battery in the proposed control approach, and the amount of reactive power is optimized to meet the demand of the power required for the full-bridge converter under different load conditions. Moreover, an optimal nonlinear control approach based on the Control-Lyapunov Function (CFL) was presented in [81], which regulates the input power of the converter instead of the output one. In this control method, the low bandwidth voltage control loop is eliminated and faster dynamic response is derived. Furthermore, superior stability margin and robustness than conventional linear control methods are obtained.

Some investigations in PWM strategies have also been presented for power converters in EV applications. A synthesis of SVPWM control methods was applied for an H-bridge inverter that feeds a three-phase PMSM in EVs in [82]. Six half bridges were applied in the considered configuration, in which case some more degrees of freedom were produced compared with that for the three half bridge based topology. The power structure consists of three H-bridges, where a separate PMSM phase is supplied by each of them, and fault isolation contactors are not required in the proposed architecture.

In addition, there is no neutral point for the machine connected with the converter with three H-bridges, thus the full DC-link voltage to each PMSM stator winding can be applied [83, 84], while the zero-sequence current cannot be structurally rejected. The specific SVPWM method which is most suitable for EV requirement was selected by combining all the following advantages: switching loss minimization, balanced switching rate between the three H-bridges, insensitivity to duty cycle, maximized drive performance, and reduction of the zero-sequence current ripple [82]. In [85], a novel single-reference six-pulse-modulation (SRSPM) technique was proposed for interleaved high-frequency (HF) three-phase inverter for FCVs. Compared with the conventional three-reference three-phase sinusoidal PWM (SPWM), the control complexity is reduced, and the DC-link capacitor is not required in this case. In addition, only one third modulation of the inverter devices is required for generating balanced three-phase voltage waveforms, and two thirds of switching losses are eliminated. Besides, up to 86.6% of switching losses can be reduced compared to a three-phase inverter. An advanced switching sequence for SVPWM-based three-level neutral-point clamped (NPC) inverter was proposed in [86] to keep the voltage difference between the two DC-link capacitors within an acceptable range. The proposed control strategy was based on the nearest three-vector scheme, and the number of switching sequences is reduced, which endows a constant switching frequency for the system. Furthermore, the scheme was modified for reduction of lower-order switching harmonics. In [87], a novel modular multiple-input bidirectional DC-DC power converter (MIPC) was proposed to interface more than two DC sources of different voltage levels. The proposed converter has the following features: 1) more than two DC sources of different voltage levels are likely to be interfaced and they can be extended to any number of sources. 2) Bidirectional power flow capability is available. 3) The power flow between any two of the sources can be controlled independently. 4) Simpler design, implementation and control are achieved.

With the use of photovoltaic (PV) panels on EVs, the reliance on vehicle batteries is decreased, and a tri-port converter was proposed in [88] to control the energy flow between the PV panel, battery and SRM. There are six operating modes in total, where four of them were developed for driving and the other two were applied for standstill on-board charging. The proposed tri-port topology is displayed in Fig. 15.

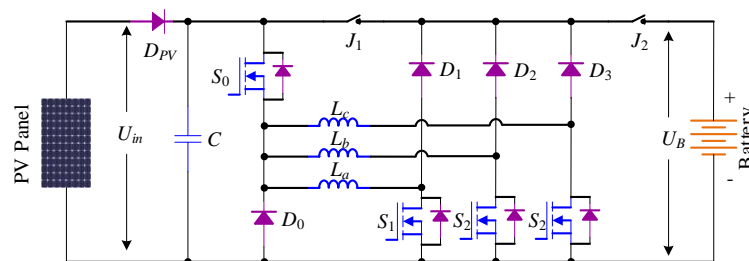


Fig15. The proposed tri-port topology

Furthermore, in [89], an integrated synchronous buck converter using only $(N+1)$ switches to independently generate N -output voltages was proposed for auxiliary power supply system of EVs. Compared with the conventional separate N -output synchronous buck converter, the number of switching components is reduced. The topologies of the conventional and proposed synchronous buck converters are shown in Fig. 16.

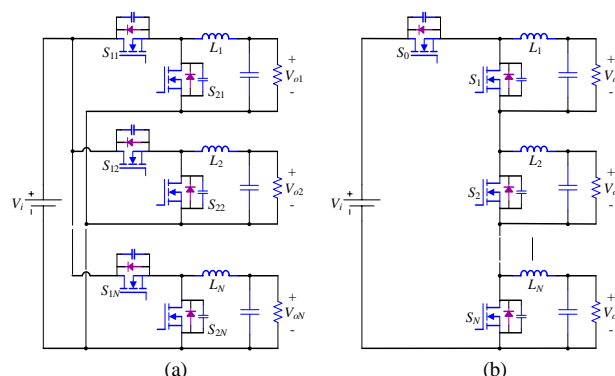


Fig16. The topologies of (a) conventional separate N -output synchronous buck converter; (b) proposed integrated multiple-output synchronous buck converter

By applying the proposed integrated multiple-output synchronous buck converter, high density, high efficiency, and good cross regulation for the auxiliary power supply system of EVs can be obtained [89]. On top of that, ZVS operation and possible lower conduction losses can be achieved. Moreover, a simple control system was also introduced, since the dynamic behaviour of the proposed converter is similar to that of the conventional one, in which case the closed-loop controller of the conventional buck converter can be directly applied.

5. CONCLUSIONS

In this paper, an overview was carried out for power electronics techniques used in MEA, ES and EV applications. As the conventional means of transportation contributes to one thirds of the carbon emissions all over the world, it is essential to develop environmentally friendly alternatives. In recent years, electric propulsion systems are developed and they contribute to the revolution of transportation. The major research interests and trends in MEA, ES and EV in terms of power electronics techniques are listed as shown below.

- 1) For MEAs: Different categories of converters were proposed and investigated, including three-phase AC-DC converters, multi-level converters, and multi-port converters. In addition, novel modulation and PLL schemes were researched in the literatures.
- 2) For ESs: The investigations on MVAC and MVDC power systems were conducted. Besides, researches on TPTL PSPWM DC-DC converter, MMC, hybrid ES modelling and frequency modulation were carried out.
- 3) For EVs: A number of different DC-DC converter configurations were explored, including isolated, ZVS, three-level, CBB-CIM, CBB-IIM, interleaved front-end, and modular multiple-input topology based DC-DC converters. Additionally, some researches on H-bridge inverter, tri-port converter, integrated synchronous buck converter, and BZS-NSI were accomplished. Furthermore, investigations on the load adaptive control, differential flatness theory, optimal nonlinear control and SRSPM for EV applications were conducted.

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