

CMOS VDIBAs-Based Single-Resistance-Controlled Voltage-Mode Sinusoidal Oscillator

Kanhaiya Lal Pushkar¹, Ghanshyam Singh², Rajendra Kumar Goel¹

¹Department of Electronics and Communication Engineering, Maharaja Agrasen Institute of Technology, New Delhi, India

²Department of Electronics and Communication Engineering, HMR Institute of Technology and Management (HMRIT), Delhi, India

Email: klpushkar17@gmail.com, ghanshyamsingh_09@rediffmail.com, kumarrajendra2003@gmail.com

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Abstract

In this communication, a new single-resistance controlled sinusoidal oscillator (SRCO) has been presented. The presented SRCO uses two voltage differencing inverting buffered amplifiers (VDIBAs), one resistor and two capacitors in which one is grounded (GC) and the other one is floating (FC). The proposed structure offers the following advantageous features: 1) independent control of oscillation condition (OC) and oscillation frequency (OF); 2) low passive and active sensitivities and 3) very good frequency stability. The non-ideal effects of the VDIBA on the proposed oscillator have also been investigated. The proposed SRCO has been tested for its robustness using Monte-Carlo simulations. The check of the validity of the presented SRCO has been established by SPICE simulations using 0.18 μm TSMC technology.

Keywords

Sinusoidal Oscillator, Voltage Differencing Inverting Buffered Amplifier, Voltage-Mode Circuits, Analog Circuit Design

1. Introduction

In analog signal processing and circuit design, realization of active filters and oscillators has become the important research areas. In reference [1] Biolek, Senani, Biolkova, and Kolka have introduced a number of modern analog active building blocks and VDIBA is one of them which is emerging very flexible and versatile active building block for analog signal processing and signal generation. The role played by SRCOs in control systems, signal processing, instrumentation

and measurement and communication systems is well established in the open literature (see [2] [3] [4] and the references cited therein). Considerable attention has been given by the various researchers in the realization of SRCOs using various active building blocks because of their several merits over conventional op-amp-based SRCOs (see [5]-[16] and the references cited therein). The applications, advantages and usefulness of VDIBA have now been recognized in the realization of the first-order all-pass filter and oscillator [17] [18], and universal biquadratic filters [19] [20]. However, to the best knowledge and belief of the authors, none of the SRCOs using VDIBAs has yet been presented in the literature with independent control of oscillation condition (OC) and oscillation frequency (OF) so far. Therefore, the purpose of this communication is to present a new SRCO using two VDIBAs along with a bare minimum number of three passive components. The proposed structure offers: 1) independent electronic control of oscillation condition; 2) independent control of oscillation frequency through a resistor; 3) low passive and active sensitivities and 4) very good frequency stability. The workability of the proposed SRCO has been confirmed by SPICE simulations using 0.18 μm TSMC technology.

2. New Oscillator Circuit

The symbolic notation and equivalent model of the VDIBA are given in **Figure 1(a)** and **Figure 1(b)** respectively [17]. The structure of VDIBA has two voltage inputs of high impedance, a voltage input terminal of low impedance and a current output terminal of high impedance. The ideal terminal equations between port voltages and currents can be expressed as: $I_+ = 0 = I_-$, $I_z = g_m (V_+ - V_-)$ and $V_{w-} = -V_z$, where g_m , represents the transconductance of VDIBA.

The presented single-resistance-controlled sinusoidal oscillator circuit is shown in **Figure 2**.

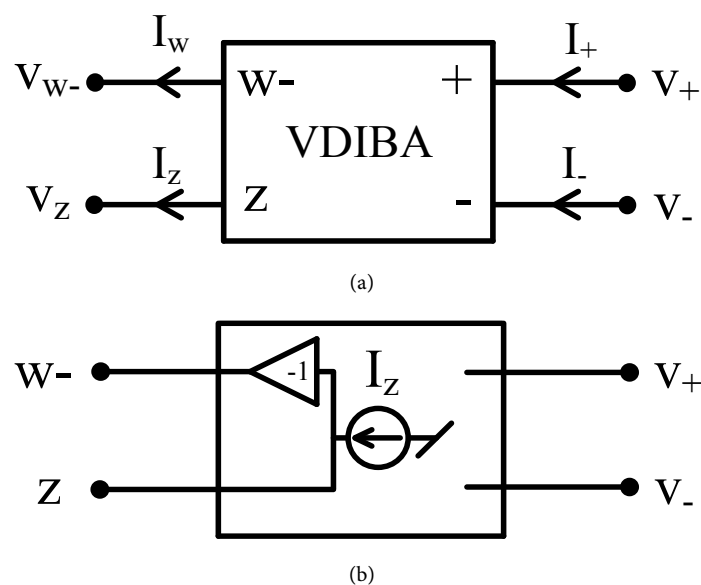


Figure 1. (a) Symbolic notation; (b) equivalent model of VDIBA.

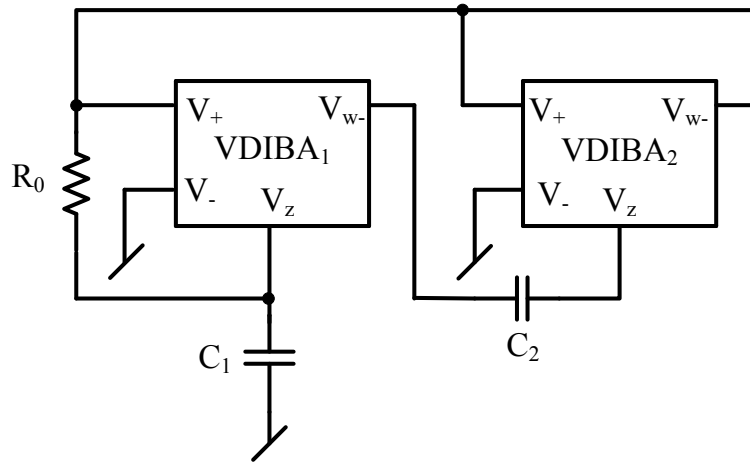


Figure 2. The new SRCO structure.

The characteristic equation (CE) of the proposed SRCO of **Figure 2**, using a routine circuit analysis can easily be obtained as:

CE:

$$s^2 C_1 C_2 + s \{ C_1 g_{m2} - C_2 g_{m1} \} + \frac{g_{m2}}{R_0} = 0 \tag{1}$$

From Equation (1), the oscillation condition (OC) and oscillation frequency (OF) can be determined as:

OC:

$$\{ C_1 g_{m2} - C_2 g_{m1} \} \leq 0 \tag{2}$$

and

OF:

$$\omega_0 = \sqrt{\frac{g_{m2}}{C_1 C_2 R_0}} \tag{3}$$

From Equations (2) and (3), it is obvious that OF is independently controllable by resistor R_0 and OC is independently controllable electronically by transconductance g_{m1} .

3. Frequency Stability Analysis of the Presented SRCO

Frequency stability may be considered to be an important figure of merit of an oscillator. The frequency stability factor is defined as $S^F = d\varphi(u)/du$ [4], where $u = \omega/\omega_0$ is the normalized frequency, and $\varphi(u)$ represents the phase function of the open loop transfer function of the oscillator circuit, with $C_1 = C/2$, $C_2 = C$, $R_0 = R/n$ and $g_{m1} = g_{m2} = 1/R$, S^F for the proposed SRCO is found to be

$$S^F = 2\sqrt{2n} \tag{4}$$

Thus for larger values of n , the presented oscillator circuit enjoys a very good frequency stability.

4. Non-Ideal Analysis and Sensitivity Performance

Let R_z and C_z denote the parasitic resistance and parasitic capacitance of the

Z-terminal of VDIBA. Taking the non-idealities into account, namely, the voltage of W-terminal $V_{W-} = (-\beta^+ V_z)$ where $\beta^+ = 1 - \varepsilon_p$ ($\varepsilon_p \ll 1$) denotes the voltage tracking error of Z-terminal of VDIBA, the expressions for characteristic equation, CO and FO respectively become:

$$s^2 \{C_1 C_2 + (C_1 + C_2 + C_z) C_z\} + s \left\{ (C_1 + C_z) \left(\frac{1}{R_z} + \beta^+ g_{m2} \right) + (C_2 + C_z) \left(\frac{1}{R_0} + \frac{1}{R_z} \right) \right\} - \beta^{+2} C_2 \left(g_{m1} + \frac{1}{R_0} \right) + \left(\frac{1}{R_0} + \frac{1}{R_z} \right) \left(\frac{1}{R_z} + \beta^+ g_{m2} \right) = 0 \tag{5}$$

Therefore the expressions for OC and OF are given as:

OC:

$$\left\{ (C_1 + C_z) (1 + \beta^+ g_{m2} R_z) R_0 + (C_2 + C_z) (R_0 + R_z) - \beta^{+2} R_z C_2 (1 + g_{m1} R_0) \right\} \leq 0 \tag{6}$$

OF:

$$\omega_0 = \sqrt{\frac{R_0 + R_z + \beta^+ g_{m2} R_z (R_0 + R_z)}{R_0 R_z^2 \{C_z (C_1 + C_2 + C_z) + C_1 C_2\}}} \tag{7}$$

Therefore the active and passive sensitivities can be obtained as:

$$S_{C_1}^{e_0} = -\frac{1}{2} \frac{1}{1 + \frac{C_z^2 + C_2 C_z}{C_1 (C_2 + C_z)}}, S_{C_2}^{e_0} = -\frac{1}{2} \frac{1}{1 + \frac{C_z^2 + C_1 C_z}{C_2 (C_1 + C_z)}}, S_{R_0}^{e_0} = -\frac{1}{2} \left\{ \frac{R_z}{R_0 + R_z} \right\} \tag{8}$$

$$S_{C_z}^{e_0} = -\frac{1}{2} \frac{1}{1 + \frac{C_1 C_2 - C_z^2}{C_z (C_1 + C_2 + 2C_z)}}, S_{R_z}^{e_0} = -\frac{1}{2} \left\{ \frac{2R_0 + R_z (1 + \beta^+ g_{m2} R_0)}{R_0 + R_z + \beta^+ g_{m2} R_z (R_0 + R_z)} \right\} \tag{9}$$

$$S_{\beta^+}^{e_0} = \frac{1}{2} \frac{1}{1 + \frac{1}{\beta^+ g_{m2} R_z}} = S_{g_{m2}}^{e_0} \tag{10}$$

Ideally, the various sensitivities of OF with respect to passive elements C_z , R_z , C_1 , and C_2 are found to be

$$S_{C_z}^{e_0} = S_{R_z}^{e_0} = 0, S_{C_1}^{e_0} = S_{C_2}^{e_0} = -\frac{1}{2} \tag{11}$$

For the typical values of $C_z = 0.81$ pF, $R_z = 53$ k Ω , $\beta^+ = 1$ along with $C_1 = 0.5$ nF, $C_2 = 1.0$ nF, $R_0 = 950$ Ω , the various sensitivities are found to be $S_{C_1}^{e_0} = -0.391$, $S_{C_2}^{e_0} = -0.276$, $S_{C_z}^{e_0} = -0.533$, $S_{R_0}^{e_0} = -0.491$, $S_{\beta^+}^{e_0} = 0.477 = S_{g_m}^{e_0}$, $S_{R_z}^{e_0} = -0.0241$ which are all low.

Figure 3 shows the CMOS implementation of the VDIBA used, which was biased with $V_{DD} = 0.9$ V D.C. = $-V_{SS}$ and I_b was taken 100 μ A.

5. Simulation Results

To confirm theoretical analysis, the proposed SRCO was simulated using CMOS

VDIBA (as shown in **Figure 3**). The passive components were used as $C_1 = 0.5$ nF, $C_2 = 1.0$ nF, $R_0 = 950 \Omega$. The transconductance of VDIBA was controlled by bias current I_b . SPICE generated output waveforms indicating transient and steady state responses are shown in **Figure 4(a)** and **Figure 4(b)** respectively.

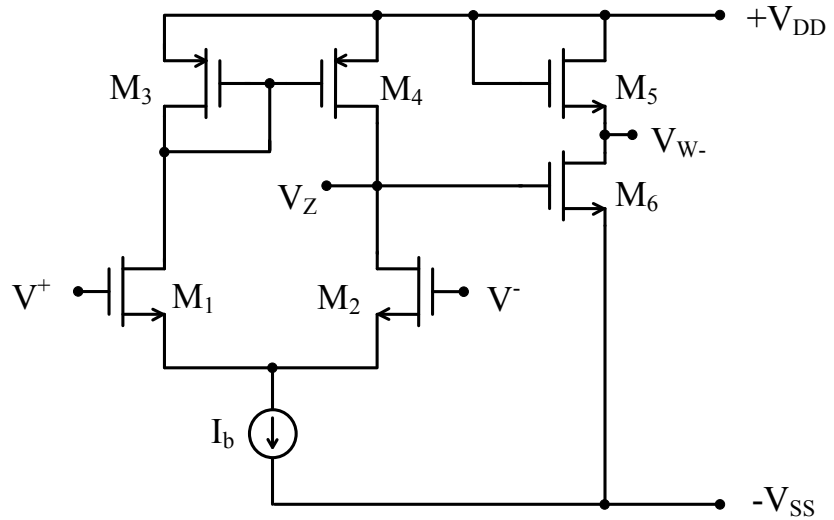
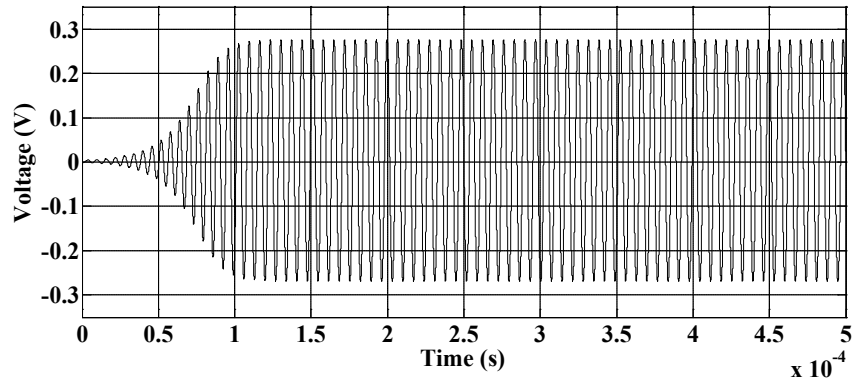
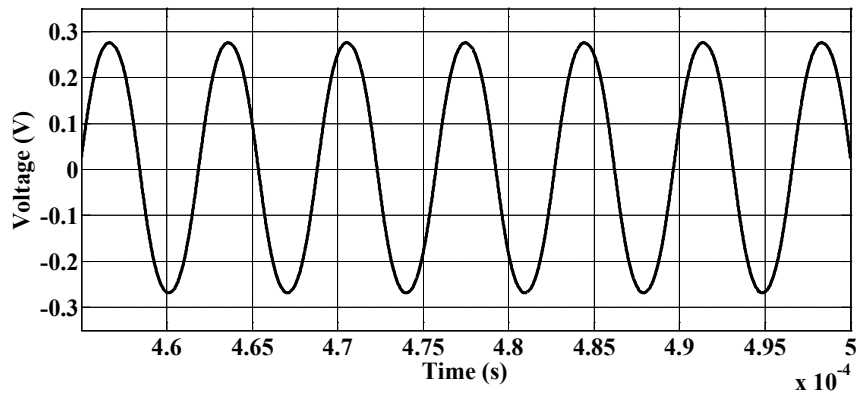


Figure 3. Implementation of CMOS VDIBA [17].



(a)



(b)

Figure 4. (a) Transient output waveform; (b) Steady state response of the output.

These results, thus, confirm the validity of the proposed configuration. **Figure 5** shows the output spectrum, where the total harmonic distortion (THD) is found to be 1.996%. **Figure 6** shows the variation of frequency with resistance R_0 . A comparison with other previously known SRCOs using different active building blocks has been given in **Table 1**.

The implementation of CMOS VDIBA employing 0.18 μm TSMC technology was used from [17] and the device parameters were taken from [21]. The aspect ratios of various MOSFETs used in CMOS VDIBA of **Figure 7** were taken from reference [18].

From Equations (8) - (10), this is obvious that the values of various sensitivities of passive and active components are less than half.

6. Conclusion

This work presents VDIBAs-based SRCO which employs minimum number of passive elements (namely, one resistor, two capacitors) and offers independent control of OF through the resistor R_0 and OC through the transconductance g_{m1} (thus the circuit enjoys the electronic control of OC), low passive and active sen-

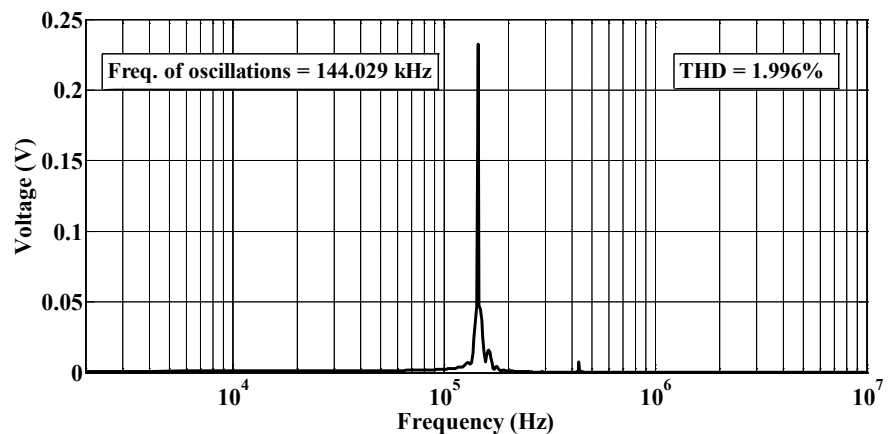


Figure 5. Simulation result of the output spectrum.

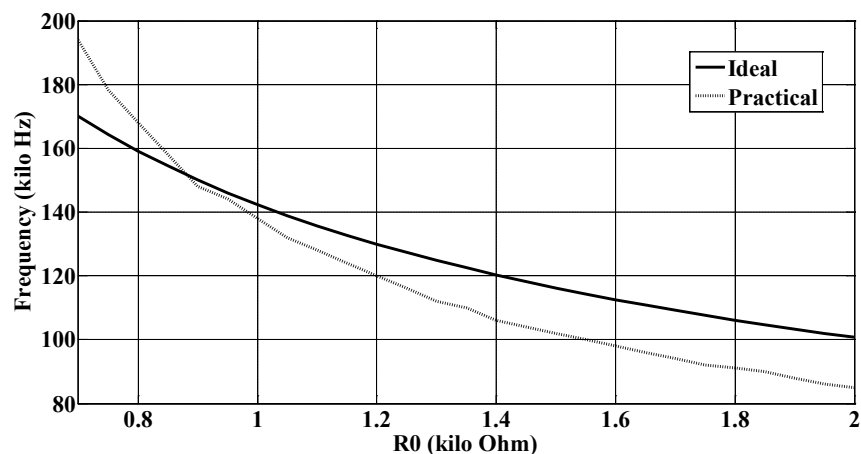


Figure 6. Variation of frequency with R_0 of the proposed SRCO.

Table 1. A comparison with other previously known SRCOs using different active building blocks.

Reference	Active Element	Number of Active Element(s)	Number of GC	Number of FC	Number. of Resistors	Whether OC and OF are Independently Controllable?
[5]	CFOA	1	1	1	3	YES
[6]	CC-II (+)	1	1	1	3	YES
[7]	CC-II (-) + Buffer	2	2	0	3	YES
[8]	PFTFN	1	1	1	3	YES
[9]	PNFTN	1	2	0	4	NO
[10]	NFTFN + Buffer	2	2	0	3	YES
[11]	DVCCC	1	2	0	3	YES
[12]	DVCCC	1	2	0	3/2	YES
[13]	CDBA	1	1 virtually grounded)	1	3	YES(only in second topology of Table 1)
[14]	OTRA	1	1 virtually grounded)	1	3	NO
[15]	CDTA	1	1	1	2	YES
[16]	VD-DIBA	2	2	0	1	YES
[17]	VDIBA	2	1	1	1	NO
[22]	VD-DIBA	1	2	0	2	YES
[23]	VD-DIBA	1	1	1	2	YES
Proposed	VDIBA	2	1	1	1	YES

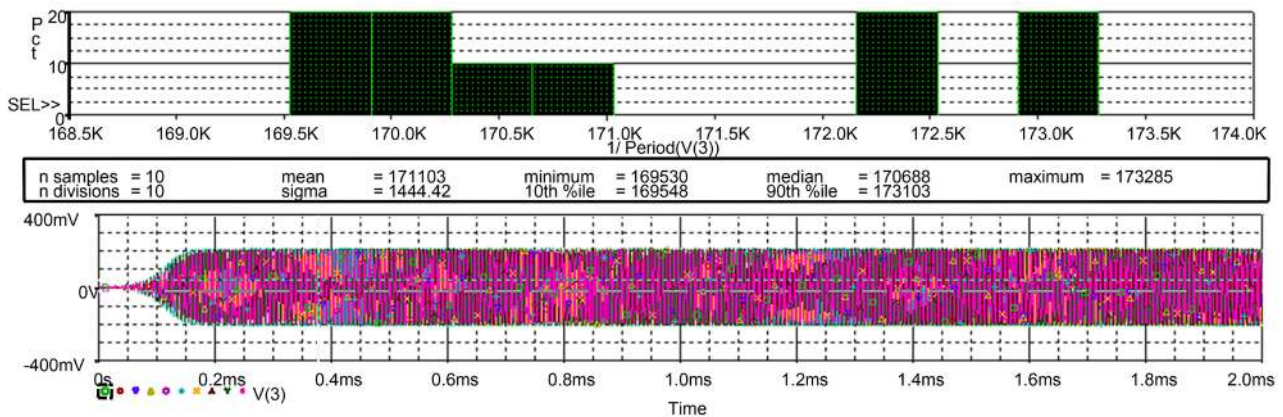


Figure 7. Monte-Carlo analysis of the SRCO.

sivities and a very good frequency stability. This communication, therefore, added a new application circuit to the existing repertoire of VDIBAs-based application circuits.

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