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Gyftakis, KN.; J. Antonino-Daviu; Garcia-Hernandez, R.; Mcculloch, MD.; Howey, DA.; Marques Cardoso, AJ. (2016). Comparative Experimental Investigation of Broken Bar Fault Detectability in Induction Motors. IEEE Transactions on Industry Applications. 52(2):1452-1459. doi:10.1109/TIA.2015.2505663



The final publication is available at https://doi.org/10.1109/TIA.2015.2505663

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Additional Information

Comparative Experimental Investigation of Broken Bar Fault Detectability in Induction Motors

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Abstract — It has been shown in the past that the zerosequence current spectrum can be reliably used to detect broken bar faults in induction motors. Previous work was carried out with extensive FEM analysis. Although it allows detailed study of spatial and time-dependent electromagnetic characteristics of induction motors, FEM is a heavily timeconsuming tool and this limits full study. So, in this work, extensive experimental testing has been performed to validate the zero sequence current spectrum for detecting rotor asymmetries. Three identical induction motors have been used: one healthy, one with a broken rotor bar, and one with two broken rotor bars. The motors were tested under different voltage supply levels and with different mechanical loads. The zero-sequence current spectrum was calculated after measuring the three phase currents. It is for the first time experimentally shown that this approach offers greater diagnostic potential than traditional MCSA.

Index Terms — Broken rotor bar, Fault diagnosis, Induction motor, Zero-sequence current.

I. INTRODUCTION

ROKEN rotor bar faults account for 5-10% of total induction motor faults [1]-[2]. The mechanisms which lead to this fault are many and are strongly related to the manufacturing process of the motor itself as well as the operating characteristics [3]-[4]. If the induction motor cage is aluminum cast, the broken bar fault is usually the result of air-bubbles inside the bars, created during casting, which lead to hot spots and eventually to local

This work was supported in part by the EPSRC funded FUTURE Vehicles project (EP/I038586/1).

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cracks of the cage [5]. The mechanism of the broken rotor bar fault is usually different in larger industrial induction motors whose cage is fabricated with copper. Thermal expansion of the bars along the shaft direction, vibrations and corrosion are some of the main factors leading to a broken bar fault [6].

Moreover, previous works have shown that, when there is a broken bar fault the neighboring bars are overcharged and are expected in most cases to break subsequently [7]. Despite this, some cases have been reported where the broken bars were not in adjacent positions but in random positions along the rotor circumference influencing negatively the diagnostic reliability of Motor Current Signature Analysis (MCSA) [7]-[10] for this fault.

It is also important to refer to the cases of a false broken bar fault alarm caused by axial cooling air-ducts, a subject that recently has drawn a lot of attention [11]-[13]. Additionally, although the broken bar fault is not usually expected to lead to an abrupt total motor failure, there have been cases where the protrusion of the broken rotor bars caused severe damage to the stator windings leading to significant damage and production shutdowns.

Finite Element Method (FEM) has been used in the past to simulate induction motors under a broken bar fault and offer a deep insight into the magnetic field behavior. In [14] the authors showed that the local saturation has a negative impact on broken bar fault diagnosis. Moreover, in [15] it is shown that a broken bar fault in a 4-pole induction motor caused a parasitic 2-pole magnetic field. Finally, in [16] it was shown that in double cage induction motors the breaking of an upper bar causes the overcharging of the inner bar in the same slot and not the neighboring adjacent bars, clearly showing a different degradation mechanism than in conventional induction motors.

Many methods can be found in the literature dealing with broken rotor bar fault diagnosis [17-18]. Most of the diagnostic techniques used up to now are based on the analysis of the stator current [19]-[21], torque [22]-[23], magnetic flux [24]-[25] and power [26]-[29]. Moreover, different signal processing methods can be found including FFTs [30], time frequency tools [19], [31]-[32], MUSIC [33] etc. It was shown recently [34], with the help of FEM,

that the Zero-Sequence Current (ZSC) spectrum can be used for the detection and identification of a variety of induction motor faults, including broken rotor bar faults. It was also shown in [35] that the ZSC is less sensitive to the rotor slot number compared to other diagnostic means. Finally, it was shown in [36] with the use of FEM that the ZSC broken rotor bar fault signatures had greater amplitudes than those of the stator current signatures in both single and double cage induction motors.

In this work, the aim is to investigate experimentally the broken rotor bar fault diagnostic ability of the ZSC. For this purpose three identical 3-phase cage induction motors have been tested: one healthy, one with a broken rotor bar and one with two broken rotor bars. The motors were set to operate at rated load, half load and at no load for different voltage supply levels. Their stator windings were connected so that the three phase currents were simultaneously monitored for each specific case, which allows the easy calculation of the ZSC.

The experimental results reveal that the ZSC can be exploited to reliably detect broken rotor bar faults and their severity. Moreover, the ZSC contains much stronger broken rotor bar fault related signatures than the line current spectrum. Finally, the ZSC broken rotor bar diagnostic ability is enhanced by the iron core saturation effect. This can be very useful as a complementary diagnostic tool in industrial motors where access to all phase currents may be easily available.

II. THEORETICAL ANALYSIS

Firstly, a theoretical analysis will be presented aiming to properly introduce the reader in the proposed method. It is considered that a 3-phase induction motor (blue) is connected in delta and operates under load (Fig. 1). The yellow sensors are set to monitor the three phase currents namely: I_a , I_b , I_c .

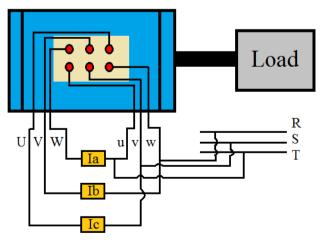


Fig. 1. The delta connected induction motor connected to a load.

The produced first phase-current harmonic constitutes of the following terms:

$$I_a = I_{MMF \ a} + I_{sat \ a} + RSH \tag{1}$$

where:

$$I_{MMF_a} = \sum_{n=2k+1}^{\infty} I_{MMF_n} \cos(\omega_n t)$$
 (2)

$$I_{sat_a} = \sum_{m=2l+1}^{\infty} I_{MMF_m} \cos(\omega_m t + \varphi_{sat})$$
 (3)

RSH: The Rotor Slot Harmonics, if they exist according to [37].

The three phase currents have 120 degrees phase difference. The zero-sequence current is the sum of the three phase currents and occurs as follows:

$$I_{ZSC} = 3\sum_{n=3k}^{\infty} I_{ph_{n}} \cos(\omega_{n}t) +$$

$$+3\sum_{m=3l}^{\infty} I_{sat_{m}} \cos(\omega_{m}t + \varphi_{sat}) + RSH$$
(4)

So, it becomes clear that the fundamental zero-sequence current harmonic is given by:

$$I_{ZSC_1} = 3 \left[I_{ph_3} \cos(3\omega_s t) + I_{sat_3} \cos(3\omega_s t + \varphi_{sat}) \right]$$
 (5)

It is clear that the amplitude of the fundamental zerosequence harmonic is saturation level dependant. This means that at no-load operation or low-load operation, the zero-sequence current can reveal the broken bar fault more reliably than traditional MCSA.

If there is a broken bar fault, then fault-related signatures are expected to appear at frequencies:

$$f_{bb} = f_{zsc_1} - 2ksf_s, \ k \in \square$$
 (6)

III. EXPERIMENTAL TESTING

A. Test Bench Details

The test bench as well as the current sensors configuration is shown in Fig. 1. A DC generator was coupled to the motors shaft playing the role of the load. The variable voltage was regulated by means of an autotransformer that was connected at the output of the supply desk. The autotransformer enabled to vary the supply voltage in a wide range (from 0 to the rated voltage of the motor). The three phase currents were monitored for every operating case using current clamps that were connected to a waveform recorder (YOKOGAWA DL-850). The signals were captured at a sampling rate of 5 kHz and the register length was 100 seconds, which gives a good resolution for the FFT analyses. The current signals were initially stored in the recorder and afterwards transferred to a computer. Moreover, the tested induction motor characteristics are shown in Table I.

TABLE I INDUCTION MOTOR CHARACTERISTICS

Rated Power	1.1 kW	
Rated frequency	50 Hz	
Rated Voltage	230 V	
Rated primary current	4.5 A	
Rated speed	1410 rpm	
Rated slip	0.06	
Stator windings connection	Delta	
Number of pole pairs	2	
Number of rotor bars	28	
Number of stator slots	36	



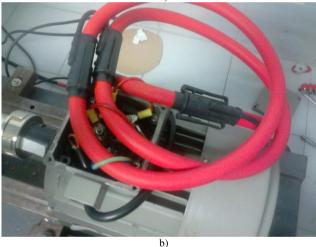


Fig. 1. The test bench where it is shown: a) the coupling between induction motor and load and b) the current sensors configuration.

B. MCSA Results

It is well established that broken bar faults produce a specific signature in the line current spectrum, located at frequency f_s-2sf_s (left sideband component). Due to speed oscillations [17], there will be a right sideband harmonic too, located at f_s+2sf_s . This procedure will end by the production of harmonics located at $f_s\pm 2sf_s$. Similarly, previous contributions have shown that broken

bar fault signatures close to the fifth and seventh current harmonics offer valuable broken bar fault detection potential [30], [38]. Therefore, before analyzing the ZSC results, it is crucial to study the MCSA method's capability in detail.

In Fig. 2 and Fig. 3, the frequency spectra close to the fundamental line current and the fifth harmonic are presented respectively for all tested motors at rated conditions. For a better overview, the amplitudes of the various broken bar fault signatures are illustrated in Table II. The results are consistent with previously published contributions.

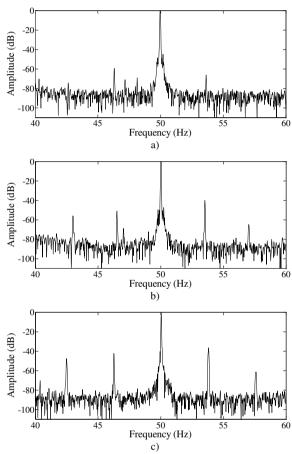
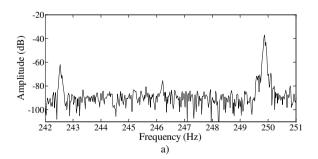


Fig. 2. Frequency spectra around the fundamental line current harmonic for: a) healthy, b) motor with 1 broken bar and c) motor with 2 broken bars.



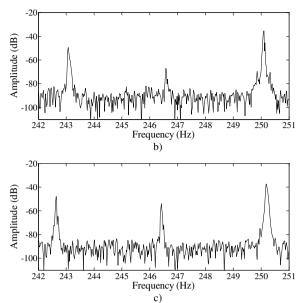


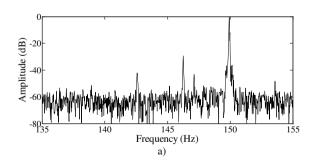
Fig. 3. Frequency spectra around the fifth line current harmonic for: a) healthy, b) motor with 1 broken bar and c) motor with 2 broken bars.

TABLE II Line Current Broken Bar Fault Signatures

	Healthy	1 broken bar	2 broken bars
fs-4sfs	-74.29	-56	-47.58
fs-2sfs	-59.29	-51.05	-42.22
fs+2sfs	-66.19	-40.04	-36.56
fs+4sfs	-88.13	-65.16	-61.45
5fs-4sfs	-61.99	-49.42	-47.8
5fs-2sfs	-75.38	-66.94	-53.94

C. ZSC Results

In this subsection, the results from the application of the ZSC methodology will be shown and its diagnostic ability will be evaluated. In Fig. 4 the ZSC frequency spectrum close to its fundamental frequency $f_{ZSC} = 3f_s$ is presented.



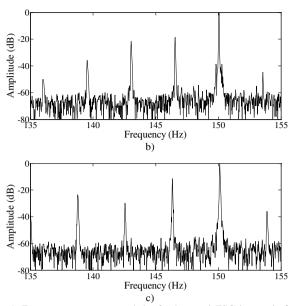


Fig. 4. Frequency spectra around the fundamental ZSC harmonic for: a) healthy, b) motor with 1 broken bar and c) motor with 2 broken bars when the motors operate at rated load.

The amplitudes of the various signatures indicating the broken bar fault are shown in Table III. It is clear that the broken bar fault signatures are characterized by significant amplitudes, which can be more than 20dB greater than those located close to the fundamental component of the line current (Table II).

It can be seen that the healthy motor has a significant signature at $3f_s - 2sf_s$. That signature is well known to be produced by shaft oscillations. The amplitude of this specific harmonic increases by about 9 dB and 18 dB for the motors with 1 and 2 broken rotor bars, respectively. Despite that, it is logical to assume that this harmonic could lead to a false positive alarm by a diagnostics engineer. Alternatively, the $3f_s - 4sf_s$ and $3f_s - 6sf_s$ harmonics increase only under the broken rotor bar fault existence. The 3fs-6sfs seems to be most influenced by the fault and presents the greater amplitude. Moreover, this signature is clearly fault severity sensitive.

TABLE III
ZSC Broken Bar Fault Signatures

	Healthy	1 broken bar	2 broken bars
3fs-6sfs	-51.37	-35.68	-23.57
3fs-4sfs	-42.22	-21.57	-29.82
3fs-2sfs	-29.43	-18.59	-11.4
3fs+2sfs	-48.38	-44.87	-35.94

D. Investigation at No-load Operation

A known disadvantage of traditional MCSA is that it is unreliable at no-load or low load operation of induction motors because a broken rotor bar fault signature frequency is slip dependent, and when slip is close to zero, this harmonic is usually hidden by the stronger fundamental component. Therefore it is important to evaluate and compare the ZSC and line current broken rotor bar fault detection abilities at no-load.

Fig. 5 illustrates the line current frequency spectra of all three motors in the area close to the fundamental component. It is clear that the left sideband harmonic is unreliable for detection because due to load oscillations it is also present in the healthy case. However, the diagnosis is possible using the right sideband signature.

Additionally, it is shown in Fig. 6 that, the frequency band close to $f_{ZSC} = 3f_s$ offers reliable diagnostic potential through the existence of a zone of harmonics related to the fault.

The amplitudes of the broken bar fault signatures are summarized in Table IV for a better overview. Not only do the ZSC signatures have significantly greater amplitudes than the line current but also they offer better severity estimation information. This is evident from the fact that for the MCSA case the broken bar fault signatures in the motors with 1 and 2 broken bars have 2-3 dB amplitude difference. Instead, the $3f_s - 6sf_s$ signature of the ZSC spectrum increases by about 5 dB between the motors with 1 and 2 broken rotor bars respectively.

It is important to note that the motors used for this investigation are small. In larger motors, the no-load speed is much closer to the synchronous speed and thus making the identification of broken bar faults unreliable through the use of the line current spectrum. Instead, it is to be expected that the zone of broken bar fault harmonics offered by the ZSC will be able to reveal the fault, but this is yet to be tested experimentally in large motors.

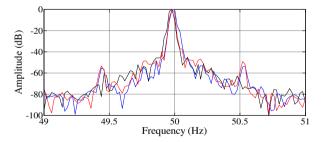


Fig. 5. The frequency spectrum of: healthy (black), motor with one broken bar (blue) and motor with two broken bars (red) at no load.

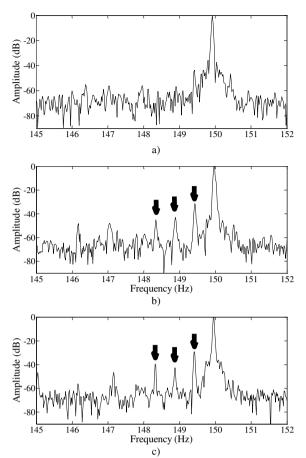


Fig. 6. Frequency spectra around the fundamental ZSC harmonic for: a) healthy, b) motor with 1 broken bar and c) motor with 2 broken bars when the motors operate at no load.

TABLE IV
BROKEN BAR FAULT SIGNATURES AT NO LOAD

	Healthy	1 broken bar	2 broken bars
MCSA fs-2sfs	-61.24	-55.4	-53.27
MCSA fs+2sfs	-62.8	-52	-49.09
ZSC 3fs-6sfs	-60.36	-44.92	-39.51
ZSC 3fs-4sfs	-56.91	-42.71	-42.85
ZSC 3fs-2sfs	-43.59	-31.36	-29.29

E. ZSC Signature Behaviour with Voltage Supply Level

In this subsection, the impact of voltage supply and load level on the broken rotor bar fault signatures of the ZSC spectrum will be shown. For this purpose, measurements at three different load levels: nominal, half nominal and noload were conducted. For each load level, different voltage levels were studied. After analysis, the amplitudes of the $3f_s - 2sf_s$ broken bar fault signatures were extracted from all cases and plotted (Fig. 7).

It can be seen that for nominal load and half nominal load the proposed ZSC signature is fault severity sensitive for every voltage supply level. Similar is the behaviour at no-load operation except some small variation of the signatures amplitude in the healthy motor. This was caused by a random shaft oscillation, which if it happens at no-load, it can significantly influence the current harmonic index due to the rotor irregular movement.

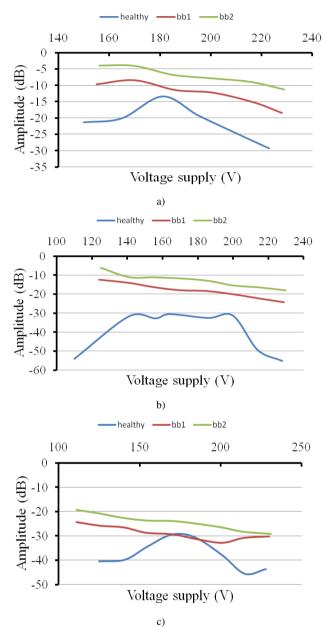


Fig. 7. The amplitude of the 3fs-2sfs broken bar fault signature versus the voltage supply level for: a) nominal load, b) half nominal load and c) no load operation.

IV. MCSA vs ZSC

In this section, the two methods, MCSA and ZSC, will be discussed and compared, taking into consideration the outcomes of this work.

The obvious advantages of MCSA are the following:

- Only one current sensor is needed,
- > Application is independent from the stator winding configuration,
- > The approach is simple,
- It can be applied in already installed induction motors at any time.

Instead, ZSC requires three current sensors and can only be applied in delta connected stator windings or star connected with the neutral connected. For those reasons it is not as simple as MCSA. Moreover, the ZSC is the sum of the three phase currents which means that it is not possible in the case of a delta connected motor to have access to the winding as long as the motor operates. In such a case the monitoring equipment could be installed after scheduled maintenance where a motor would be disconnected from the grid.

On the other hand, the ZSC has a variety of advantages which make it a competitor to MCSA:

- ➤ It offers much stronger broken rotor bar fault signatures compared to those of MCSA. More importantly, the signatures have comparable amplitudes with the ZSC fundamental component. So, the accuracy of the measurement is significantly improved.
- Due to the higher frequency of the ZSC fundamental component (3fs) compared to that of MCSA (fs), the rotor is influenced by a stronger skin effect and thus the flux does not penetrate deep in the rotor core. As a consequence, broken bar fault information is stronger in the ZSC and that is why even the $3f_s 6sf_s$ harmonic offers good diagnostic potential that is independent from shaft oscillations.
- ➤ It has been shown in the past that the saturation effect decreases the diagnostic potential of MCSA [14]. However, the ZSC is saturation related and thus the saturation level increase has a positive impact on broken rotor bar fault identification.
- The ZSC offers better diagnostic capabilities at no-load operation. Moreover, it is obvious that it is less speed sensitive than MCSA and that is because of the zone where broken rotor bar fault harmonics are produced. If the slip is very close to zero, the $f_s \pm 2sf_s$ components of the line current will not be observable. Probably the same will happen with the $3f_s 2sf_s$ ZSC components. But still there will be the 3fs-4sfs and $3f_s 6sf_s$ which can be used for diagnosis.

- As it can be seen from Fig. 7 the ZSC is a diagnostic tool that is sensitive to the severity of the fault. Voltage supply level can influence the diagnosis for no load operation as it can be seen in Fig. 7-c but diagnosis at operation under load is reliable.
- Also, previous works have shown that the ZSC can reveal static eccentricity faults in PSH-induction motors [34], as well as supply imbalances, so it contains much more valuable information for induction motor condition monitoring compared to MCSA.
- Finally, the proposed method has some common advantages with the monitoring of the broken rotor bar fault close to higher frequencies (5th/7th) of the line current (deeper flux penetration, slip independency). However, this paper's proposed method has a distinct advantage over those methods; the increased measurement accuracy. That is because although the amplitudes of the fault signatures 5th and 7th harmonics sidebands are important compared to the amplitudes of the 5th and 7th harmonics themselves, however they are still low compared to the amplitude of the fundamental current harmonic. This does not happen with the ZSC case.

V. CONCLUSIONS

In this paper, the broken rotor bar fault diagnosis through the use of the zero-sequence current spectrum is evaluated. Three motors have been tested: a healthy, one with a broken rotor bar and another with two adjacent broken rotor bars. The results indicate that the ZSC offers much greater diagnostic signatures than the traditional MCSA. Moreover, the diagnosis is more reliable at no-load operation as the ZSC is less speed sensitive than the MCSA. Finally, extensive testing for different voltage levels and for three different load levels has shown that the ZSC broken bar fault signature is fault severity sensitive. Future works should concentrate on testing larger motors with speed very close to the synchronous one, double cage induction motors, as well as the impact of non-adjacent broken bars on the ZSC diagnostics ability.

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