

Comparison Between Fully and Partially Filled Dielectric Materials on the Waveguide of Circularly Polarised Radial Line Slot Array Antennas

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Abstract—This paper presents an investigation on the waveguide of circularly polarised radial line slot array (RLSA) antennas to improve gain and radiation bandwidth. Two circularly polarised (CP) RLSA antennas were designed with two different waveguide configurations. In the first configuration the waveguide is fully filled with dielectric materials and in the second configuration the waveguide is partially filled with dielectric materials and rest of the waveguide is filled with air. Numerical results of these two CP-RLSA antennas with two different waveguide configurations are presented and compared. Significant improvements have been made in the 3-dB directivity bandwidth and aperture efficiency of the antenna having waveguide partially filled with dielectric material. The 3-dB directivity bandwidth was measured 6.2% and aperture efficiency increased to 55.5%. The CP-RLSA antenna has also achieved a peak directivity of 31.7 dBic and a gain of 31.2 dBic as compared to the directivity 30.1 dBic and gain 29.5 dBic, respectively achieved with the CP-RLSA antenna having waveguide fully filled with dielectric material.

Index Terms—low-profile, radial line slot array, improved bandwidth, wireless communication, high gain, high aperture efficiency

I. INTRODUCTION

Recent years have witnessed a fast growing interest for satellite communication and wireless applications. In order to attract the competitive commercial market, highly efficient, low-cost, light weight antennas with wider radiation bandwidth are extremely desirable for satellite applications on mobile platform such as long distance trains, buses and aircrafts. Among the diversity of microwave and millimeter wave antennas, parabolic reflectors have been utilized for several years in satellite and terrestrial communication system. Parabolic reflectors provide high gain and good coverage but their bulky and large volume make them undesirable for mobile satellite application such as SATCOM, on-the-move connectivity. Also the non-planar configuration makes it difficult for transportation. The patch array antennas provide low-profile and planar configuration but the gain drops at higher frequency. The radial line slot array (RLSA) antenna was first introduced in Japan in 1980s as an alternative to parabolic dish used for direct broadcast satellite TV reception [1], [2]. Radial line slot arrays are slotted waveguide planar antennas with several advantages including high gain, high efficiency, low profile, and compact size. The RLSA

antennas can be designed for linear, circular or elliptical polarization based on the slot layout in the top radiating surface. The antenna can be installed easily in roof-top on mobile vehicles and not subject to leaf or water build-up due to its planar surface. Many investigations were done to improve the performances of RLSA antennas [3]–[10]. In the conventional RLSAs, the waveguide was fully filled with dielectric materials [2]–[4]. The dielectric materials increase the weight and cost of the antenna. Also the thickness of the dielectric materials at higher frequency affects the antenna’s radiation performance.

In this research, we have designed two circularly polarised radial line slot array (CP-RLSA) antennas with two different waveguide configuration. The motivation of this work is to design and compare the two different antennas’ radiation performances. The design examples of these two antennas are explained in Section II. Predicted results such as near-field phase distributions and far-field patterns are given in Section III.

II. CONFIGURATIONS OF CP-RLSA ANTENNAS

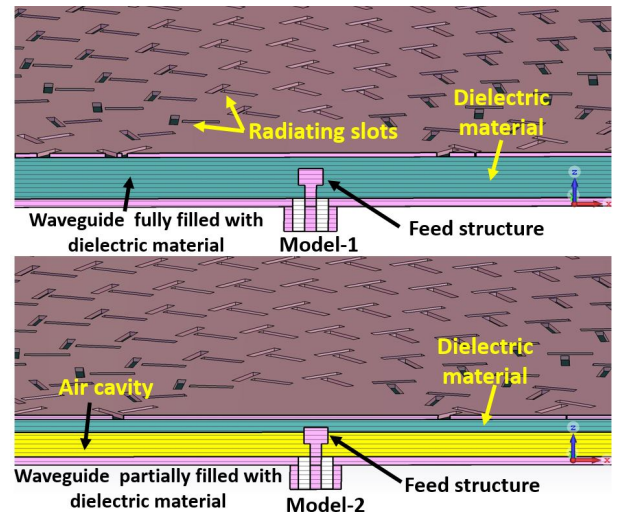


Fig. 1: Cross-section view of the two CP-RLSA antennas with two different waveguide configurations.

The CP-RLSA antenna is made of two parallel conducting metal plates, forming a waveguide cavity inside these two plates. The upper plate is consisted of radiating slots and the bottom plate works as a ground plane. The waveguide is filled with dielectric materials of relative permittivity $\epsilon_r > 1$ to create slow wave structure in the cavity with the guided wavelength, λ_g to avoid grating lobes in the radiation patterns. The power is fed to the CP-RLSA antenna at the center of the ground plane with a 50Ω coaxial cable. The feed probe has a head-disk which resides inside the waveguide cavity. The radiating slots on the top plate radiate circular polarized electric fields.

Figure 1 shows the two CP-RLSA antennas with two different waveguide models operating at 20 GHz. In the first model (model-1) antenna, the waveguide is fully filled with dielectric materials. In the second model (model-2) antenna, the waveguide is partially filled with dielectric materials and rest of the waveguide is filled with air. The height of the waveguides were optimised to less than one-half of the guided wavelength ($\lambda_g/2$) to ensure the propagation of the transverse electromagnetic (TEM) wave. Table I shows the design parameters for the waveguide of the two models. In model-2, the average permittivity value was calculated from the dielectric material and air. All other design parameters were kept constant for these two antennas. Both models have an aperture size of $16 \lambda_0$ (240 mm).

TABLE I: Design parameters of the two CP-RLSA antennas.

Specification	Model-1	Model-2
Height of the Waveguide	4.5 mm	4.5 mm
Dielectric material	Rogers RT5880(lossy)	Rogers RT5880(lossy)
Thickness of the dielectric material	4.5 mm	1.575 mm
Thickness of the air cavity	0	3 mm
Permittivity	2.2	1.41
Guided wavelength	10.1 mm	12.63 mm

III. SIMULATION RESULTS

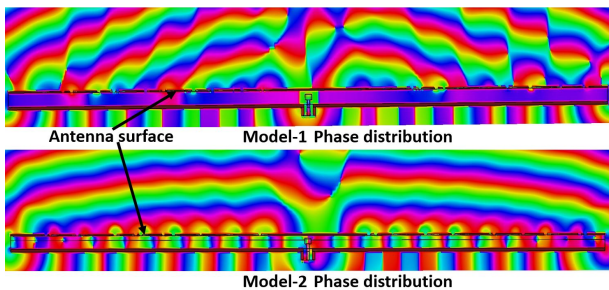


Fig. 2: Near-field phase distributions of the two antennas.

The two models were simulated in the transient solver of CST Microwave Studio from 18 GHz to 20.5 GHz. Both near and far-field results were studied and compared. Both antennas have acceptable matching in the operating frequency range. Figure 2 shows the near-field phase distributions of the two CP-RLSA antennas at 20 GHz. The electric

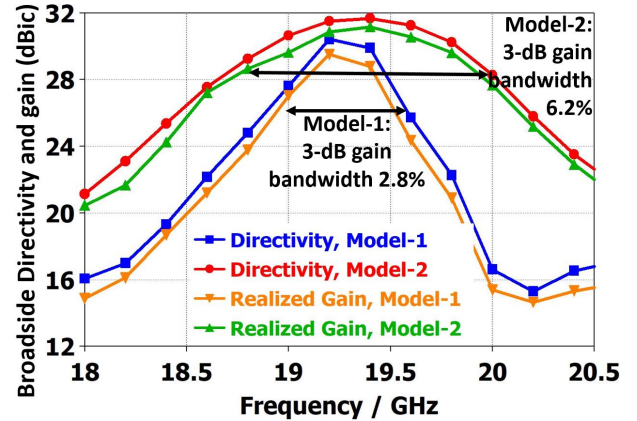


Fig. 3: Broadside far-field directivity and realized gain of the two CP-RLSA antennas.

near-field patterns (E_x) were taken in the XZ plane at 20 GHz, extending from the top plate to a distance of $2\lambda_0$ (30 mm above the top plate). As it can be seen from the figure, Model-1 has a non-uniform phase distribution throughout the aperture whereas Model-2 has shown more uniform phase distribution as compared to Model-1. The antennas have also successfully fulfilled the 3-dB axial ratio condition of circular polarisation.

Figure 3 shows the broadside far-field directivity and realized gain of the two CP-RLSA antennas. Model-1 antenna has achieved a peak directivity of 30.4 dBic with a realized gain of 29.5 dBic at 19.2 GHz. Model-2 antenna has shown a peak broadside directivity 31.7 dBic with a gain of 31.2 dBi at 19.4 GHz. Model-2 antenna has shown 1.3 dB more directivity and 1.7 dB more gain compared to Model-1 antenna. The 3-dB directivity bandwidth and 3-dB gain bandwidth of Model-1 antenna is only 2.8%. Model-2 antenna has provided a 3-dB directivity and 3-dB gain bandwidth of 6.2% which is 3.4% more than the Model-1 antenna.

Figure 4 shows the 3D CST view of the radiation patterns of the model-1 and Model-2 antennas at 19.2 GHz and 19.4 GHz, respectively. The model-2 antenna has demonstrated a relatively good antenna efficiency. The aperture efficiency achieved with Model-1 antenna is 38%, whereas Model-2 antenna has achieved an aperture efficiency of 55.5% which is 17.5% more than Model-1 antenna. Figure 5 shows the far-field radiation cut patterns of the two models at 19.2 GHz

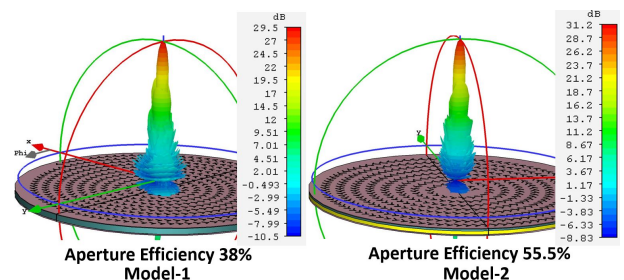


Fig. 4: 3D CST view of the radiation patterns of the two models.

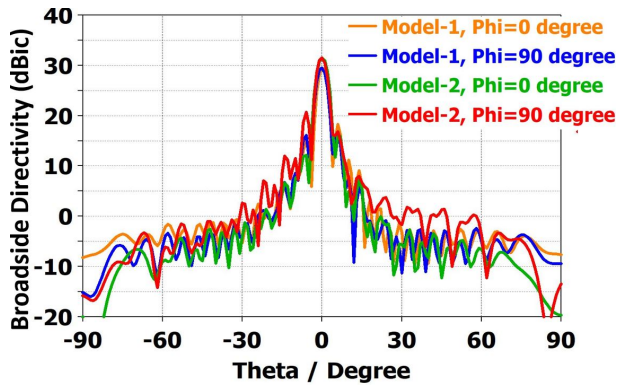


Fig. 5: Far-field radiation cut patterns of the two models at 19.2 GHz.

in both $\phi = 0^\circ$ plane and $\phi = 90^\circ$ plane. Model-1 antenna has provided a side lobe level (SLL) of -11.2 dB in $\phi = 0^\circ$ plane, whereas Model-2 has provided a side lobe level (SLL) of -15.2 dB in $\phi = 0^\circ$ plane. In $\phi = 90^\circ$ plane, Model-1 and Model-2 antennas have shown the SLLs of -12.9 dB and -11 dB, respectively.

IV. CONCLUSION

Two CP-RLSA antennas were designed with two different waveguide configurations. The numerical results predict that the CP-RLSA antenna (Model-1) having the waveguide partially filled with dielectric material provides more uniform field distribution and good antenna performance compared to the antenna (Model-2) having the waveguide fully filled with dielectric material. The Model-2 CP-RLSA antenna has provided 1.3 dB more directivity and 1.7 dB more gain than the Model-1 antenna, which also increases the total aperture efficiency of Model-2 antenna (17.5% more than Model-1 antenna). The 3-dB gain bandwidth of model-2 CP-RLSA antenna is 6.2% which is 3.4% more than Model-1 CP-RLSA antenna. The thickness of the dielectric material of the Model-2 antenna is less compared to Model-1, which also reduces the total antenna weight.

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