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# Design and Simulation of a RF MEMS Shunt Switch for Ka and V Bands and the Impact of Varying Its Geometrical Parameters 

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#### Abstract

RF MEMS plays an important role in microwave switching. The high performance of RF MEMS shunt such as high bandwidth, low insertion loss, and high isolation have made these switches well suitable for high performing microwave and millimeter wave circuits. This paper presents a RF MEMS shunt capacitive switch for $K a$ and $V$ band application. This paper investigates the effect of various geometrical parameters on $R F$ characteristics of the switch. The simulation results are presented and discussed.


## 1. Introduction

RF MEMS is the application of micro electro mechanical systems in radio frequency circuits. The utilization of RF MEMS helps fulfill the increasing demand for more flexible and functional, lightweight, and low-power-consumption wireless systems. RF MEMS switches are one of the applications of RF MEMS. The performance of RF MEMS switches tends to be better than p-i-n or FET switches that have been used extensively in switching networks [1]. Currently, RF switches can be categorized into three main groups: mechanical, semiconductor, and MEMS. Table I summarizes the characteristic of RF switches [2].

| TABLE I: CHARACTERISTIC OF RF SWITCHES [2]. |  |  |
| :--- | :--- | :--- |
| Type Benefit Drawback <br> Mechanical Low insertion loss <br> Large state isolation <br> High power handling Bulky <br> Heavy <br> Slow <br> Semiconductor Small size <br> Low weight <br> High speed DC power <br> consumption <br> MEMS Low insertion loss <br> High isolation <br> Low power consumption <br> Very small size <br> Low height <br> High RF characteristics High voltage <br> actuation |  |  |

There are two main types of forces that can be used for the actuation of RF Switches: electromagnetic and
electrostatic. Table II lists the advantages and disadvantages of these force [3].

TABLE II: ADVANTAGE AND DISADVANTAGES OF ACTUATION FORCES IN RF SWITCHES [3].

| Force | Advantage | Disadvantage |
| :--- | :--- | :--- |
| Electromagnetic | Low actuation voltage | Current consumption |
| Electrostatic | No current <br> consumption | High voltage actuation |

As it can be seen from the table, the electromagnetic force has a low actuation voltage, but a high current consumption. On the other hand, the electrostatic force has no current consumption, but has a high actuation voltage. The electrostatic switches are most common switches that are used in the microwave and mm-wave regions [4].

There are two kinds of electrostatic switches: series and shunt [5]. Series switch is initially disconnected (OFF) and it gets connected (ON) when the switch is actuated. Shunt switch is initially connected (ON) and it gets disconnected (OFF) when a required voltage is applied to the switch [6]. Table III lists the advantages and disadvantages of series and shunt switches:

TABLE III: ADVANTAGES AND DISADVANTAGES OF OHMIC AND SHUNT switch [3].

| Type | Advantage | Disadvantage |
| :--- | :--- | :--- |
| Series | Very low ON insertion loss <br> Very high OFF state isolation | Stiction <br> Microscopic bonding |
| Shunt | Long life time <br> ON state insertion loss <br> Independent from contact force | High low-frequency |

The RF MEMS shunt switch consists of a membrane that anchored from both ends and its middle section is free to move [6] [7].

## 2. Proposed RF MEMS Shunt Switch for Ka and $V$ bands

## A. RF MEMS Shunt Switch

Figure 1 shows the proposed RF MEMS shunt switch for the high frequency applications which is designed and
simulated using Electro Magnetic 3 Dimension Simulation (EM3DS). The switch contains a Co Planner Waveguide (CPW), a membrane, and anchors. The CPW consists of three transmission lines: Ground, Transmission Line and Ground shown in Figure 2.

The membrane is supported by the anchor and it situated on top of the CPW and is allowed to flex in the middle as shown in Figure 5. The Material of the CPW, Anchors and membrane are conductive materials. Figure 5 (a) shows the RF MEMS shunt switch in its upstate position when there is no applied voltage; therefore, the membrane stays on top of the CPW and RF signal propagates through the signal line of the CPW. Figure 5(b) demonstrates the switch in its downstate position. This situation is reached by the apply voltage. The applied voltage creates the electrostatic force which causes the membrane to deflect downwards. While the potential exceeds the threshold voltage, the membrane snaps down to the dielectric layer. The RF signal and the membrane create a high capacitor which couples the RF signals to the ground signal line of the CPW.


Figure 1. Proposed RF MEMS switch.


Figure 2. Co planner waveguide (CPW).

(a)

(b)

Figure 3. Switch in its (a) upstate and (b) downstate positions.

## B. Design Considerations

The RF MEMS shunt switch design involves: (i) designing a coplanar waveguide transmission line with
$Z_{0}=50 \Omega$ characteristic impedance and short circuit gaps, and (ii) designing the switch above the CPW with optimized spring constant, materials, and beam gap height to increase the RF characteristics $\left(\mathrm{S}_{11}, \mathrm{~S}_{21}\right)$ and reduce the activation voltage. The design considerations are discussed briefly in the following.

Coplanar Waveguide: Coplanar waveguide (CPW) is a one-sided three-conductor transmission line. CPW has two grounds and a center conductor in the middle, as shown in Figure 2, which reduces the coupling effects and allows for easy installation of elements. CPW lines are used mostly for circuit elements and interconnecting lines. CPW provides the potential of lower conductor and radiation losses as compared to microstrip lines. The thickness of dielectric in ideal CPW is infinite. However, in practice the thickness should be sufficiently large to remove the electromagnetic field before they get out.
$Z_{0}=\frac{30 \Pi^{2}}{\sqrt{\left(\varepsilon_{\mathrm{r}}+\mathrm{L}\right) / 2}}\left[\operatorname{LN}\left(2 \frac{1+\sqrt{\mathrm{K}}}{1-\sqrt{\mathrm{K}}}\right)\right]$
$K=\frac{w}{w+2 S}$
where w is center strip width and S is slot width and $\varepsilon_{\mathrm{r}}$ is relative dielectric constant of the substrate dielectric.

This design uses the silicon substrate of $\varepsilon_{\mathrm{r}}=11.6$, and CPW center conductor width of $\mathrm{w}=100 \mu \mathrm{~m}$, space line is $60 \mu \mathrm{~m}$, the thickness of AL is $1 \mu \mathrm{~m}$ and the width of the substrate is $400 \mu \mathrm{~m}$. The results at 40 GHz are $Z_{0}=50.09 \Omega$ and $\lambda / 4=0.7492$.

Switch Structure: The key characteristics of the RF MEMS switch are insertion loss, isolation, and actuation voltage. The aim is to increase the RF parameters while reducing the actuation voltage.

## 3. Upstate and Downstate Simulations

## A. Upstate

The switch structure is simulated in the upstate position by the EM3DS. The switch has 6 layers. There is CPW which contains a substrate, transmission line, and grounds. There are two anchors situated on top of the CPW and connects the membrane to the Ground. The height of the anchors is $4 \mu \mathrm{~m}$. The membrane situated on top of the anchors and connects to the ground by the anchors. The wide of the membrane is $80 \times 100 \mu \mathrm{~m}^{2}$. The S parameters are taken from the range of $0-60 \mathrm{GHz}$ using the EM3DS. As can be seen from Figures 4-5, $\mathrm{S}_{21}$ in the upstate position is between 0 dB and -0.4 dB , and $\mathrm{S}_{11}$ varies between -40 dB and -10 dB .


Figure 4. $\mathrm{S}_{21}$ in upstate.


Figure 5. $\mathrm{S}_{11}$ in upstate.

## B. Downstate

The switch in its downstate position does not have anchors because the membrane should be situated on top of the dielectric layer. Figures 6-7 show that for switch in downstate position the $S_{21}$ is less than -10 dB and $S_{11}$ is above -1 dB for the frequency above the 29 GHz .


Figure 6. $\mathrm{S}_{21}$ in downstate.


Figure 7. $\mathrm{S}_{11}$ in downstate.

## 4. Effects of Varying Geometrical Parameters

In this section, the effects of varying anchor height, membrane size, and dielectric height are studied in upstate and downstate positions.

## A. Upstate

When the switch is in upstate position, it creates a capacitor between the transmission line and ground. The capacitor should be small so that it does not affect the signal on the transmission line. The $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ of the scattering parameters illustrate that how this capacitance works on the transmission line.

Tables IV-V show the information about the effects of changing anchor height and membrane size on $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ of the RF MEMS shunt switch for 2 GHz to 50 GHz . These parameters are simulated by EM3DS. The RF characteristics of the upstate switch are improved by decreasing the anchor height and the membrane size. The desirable RF characteristics for the RF MEMS switch, while in the upstate position, is for $\mathrm{S}_{11}$ to be below -10 dB and $S_{21}$ above -1 dB .

Table IV shows that the $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ of the RF MEMS switch are below -10 dB and above -1 dB respectively, while the anchor height is less than $4 \mu \mathrm{~m}$. Decreasing the anchors to 3 and $2 \mu \mathrm{~m}$ causes the $\mathrm{S}_{11}$ to become less than 9 dB and -7 dB respectively.

Table V provides the variation of the size of the membrane, and associated changes in $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$. As it can be seen, for the membrane $40 \times 100 \mu \mathrm{~m}^{2}, S_{11}$ and $S_{21}$ of this membrane are between -44 dB to -14 dB and -0.01 dB to -0.08 dB respectively, for the frequency from 2 to 50 GHz . Increasing the size of the membrane causes the $S_{11}$ and $S_{21}$ of the switch to decline between -40 dB to 10 dB and -0.01 dB to -0.42 dB respectively, for the $80 \times 100 \mu \mathrm{~m}^{2}$. This parameters for the $120 \times 100 \mu \mathrm{~m}^{2}$ is 37 dB to -8 dB and -0.02 dB to -74 dB for $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ respectively.
TABLE IV: EfFECTS OF ANCHOR HEIGHT.

| Height | $\mathbf{S}_{\mathbf{1}}, \mathbf{2} \mathbf{G h z}$ | $\mathbf{S}_{\mathbf{1 1}}, \mathbf{5 0 G h z}$ | $\mathbf{S}_{\mathbf{2 1}}, \mathbf{2} \mathbf{G h z}$ | $\mathbf{S}_{\mathbf{2 1}}, \mathbf{5 0 G h z}$ |
| :---: | :---: | :---: | :---: | :---: |
| $6 \mu \mathrm{~m}$ | -42 | -12 | -0.01 | -0.28 |
| $4 \mu \mathrm{~m}$ | -40 | -10 | -0.01 | -0.42 |
| $3 \mu \mathrm{~m}$ | -39 | -9 | -0.01 | -0.53 |
| $2 \mu \mathrm{~m}$ | -37 | -7 | -0.01 | -0.85 |

TABLE V: EFFECTS OF MEMBRANCE SIZE.

| Size | $\mathbf{S}_{\mathbf{1 1}}, \mathbf{2} \mathbf{G h z}$ | $\mathbf{S}_{\mathbf{1 1}}, \mathbf{5 0 G h z}$ | $\mathbf{S}_{\mathbf{2 1}}, \mathbf{2 G h z}$ | $\mathbf{S}_{\mathbf{2 1}} \mathbf{5 0 G h z}$ |
| :--- | :---: | :---: | :---: | :---: |
| $40 \times 100$ | -44 | -14 | -0.01 | -0.08 |
| $80 \times 100$ | -40 | -10 | -.01 | -0.42 |
| $120 \times 100$ | -37 | -8 | -0.02 | -0.74 |

## B. Downstate

While the switch is actuated in the downstate position, the capacitance to the ground becomes higher and this results in a short circuit and high isolation at microwave frequencies. Tables VI-VII illustrate that how the RF characteristics changed for varying die-electric height and membrane size.

As can be seen from Table VI, decreasing the height of the dielectric makes the bandwidth broader. $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ for the height of $0.1 \mu \mathrm{~m}$ are 20 GHz and 29 GHz , respectively, while $S_{11}$ and $S_{21}$ for $0.15 \mu \mathrm{~m}$ are 24 GHz and 31 GHz . The thinner dielectric performs better but there is limitation for the thickness of dielectric to reduce. This limitation concerns the implementation and the stiffness of dielectric; the minimum height for the semiconductors implementation is around $0.1 \mu \mathrm{~m}$.

Table VII shows the effect of the membrane dimension in downstate position. As can be seen from the table, increasing the membrane dimension improves the RF characteristics of the switch. For example, the membrane with the area of $120 \times 100 \mu \mathrm{~m}^{2}$ has the $\mathrm{S}_{11}$ of 20 GHz and $\mathrm{S}_{21}$ of 22 GHz . Moreover, $\mathrm{S}_{11}$ and $\mathrm{S}_{21}$ for the dimension of $80 \times 100 \mu \mathrm{~m}^{2}$ are 24 GHz and 31 GHz , respectively.

TABLE VI: Effects of dielectric height.

| Height | S11 | S21 |
| :--- | :---: | :---: |
| 0.1 | 20 Ghz | 29 Ghz |
| 0.15 | 24 Ghz | 31 Ghz |
| 0.2 | 26 Ghz | 32 Ghz |
| 0.3 | 25 Ghz | 34 Ghz |
| 0.4 | 28 Ghz | 35 Ghz |

TABLE VII: EfFECTS OF MEMBRANCE SIZE.

| Size | $\mathbf{S 1 1}$ | $\mathbf{S 2 1}$ |
| :--- | :--- | :--- |
| $40 \times 100$ | 38 Ghz | 40 Ghz |
| $80 \times 100$ | 24 Ghz | 31 Ghz |
| $120 \times 100$ | 20 Ghz | 22 Ghz |

The limitation of the size of the membrane relates to the upstate. It was illustrated that decreasing the size of the membrane has a negative effect on $S_{11}$ and $S_{21}$ in upstate position.

## 5. Conclusions

The RF MEMS shunt wide band switch for the Ka and V band has been presented. Both upstate and downstate positions of the switch were considered in high frequency band. The RF behaviors of the switch was simulated and discussed. The simulations reveal the impact of variations in parameters that affect the RF characteristics in both upstate and downstate.

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