

# Designing a Hardware in the Loop Wireless Digital Channel Emulator for Software Defined Radio

---

Janarбек Matai, Pingfan Meng, Lingjuan Wu, Brad Weals, and Ryan Kastner  
Department of Computer Science and Engineering,  
University of California, San Diego  
Toyon Research Corporation

December 11  
FPT'2012

# Motivation

## ❖ Problem

- ❖ Wireless system testing and verification → Difficult

## ❖ Current wireless system testing methods

- ❖ **Field testing** → Expensive, time consuming and difficult to repeat

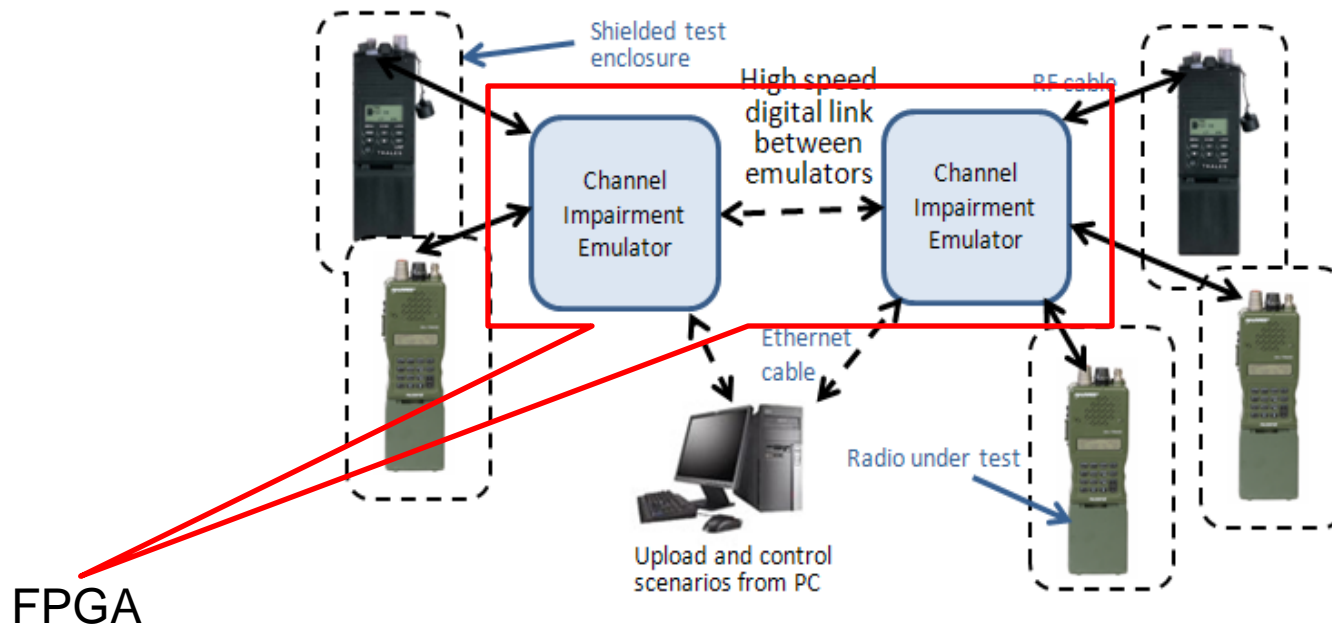
- ❖ **Simulation** → limited by fidelity, excessive run time

## ❖ Wireless Channel Emulator (WCE)

- ❖ Fills the gap left between simulation and field testing
- ❖ Repeatability, high-fidelity, and the opportunity to test complete radio
- ❖ Software implementation is not feasible due to amount of computation for large network

# Hardware in the Loop Wireless Channel Emulator (WCE)

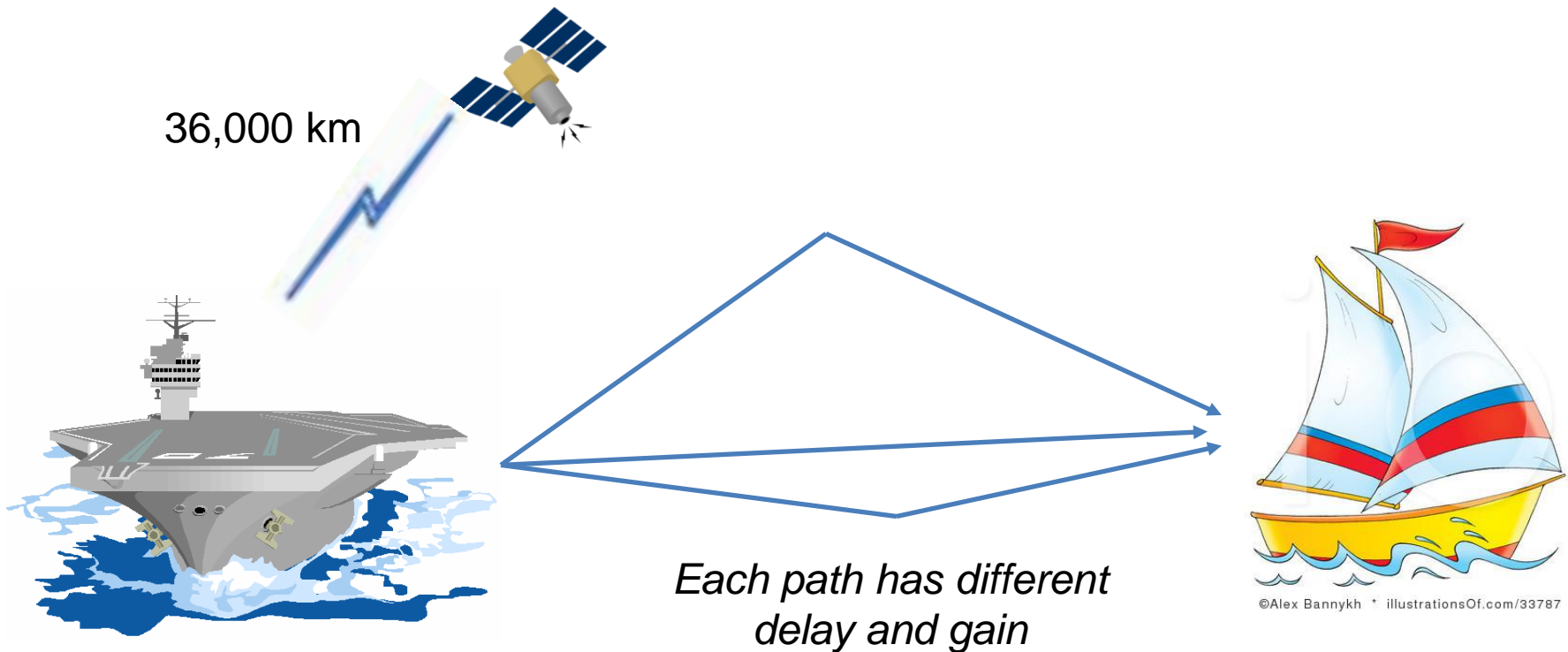
- ❖ Hardware in the loop (HWIL) **Wireless Channel Emulator (WCE)** implementation on an **FPGA** platform is purposed using **High-Level Synthesis Tool**.



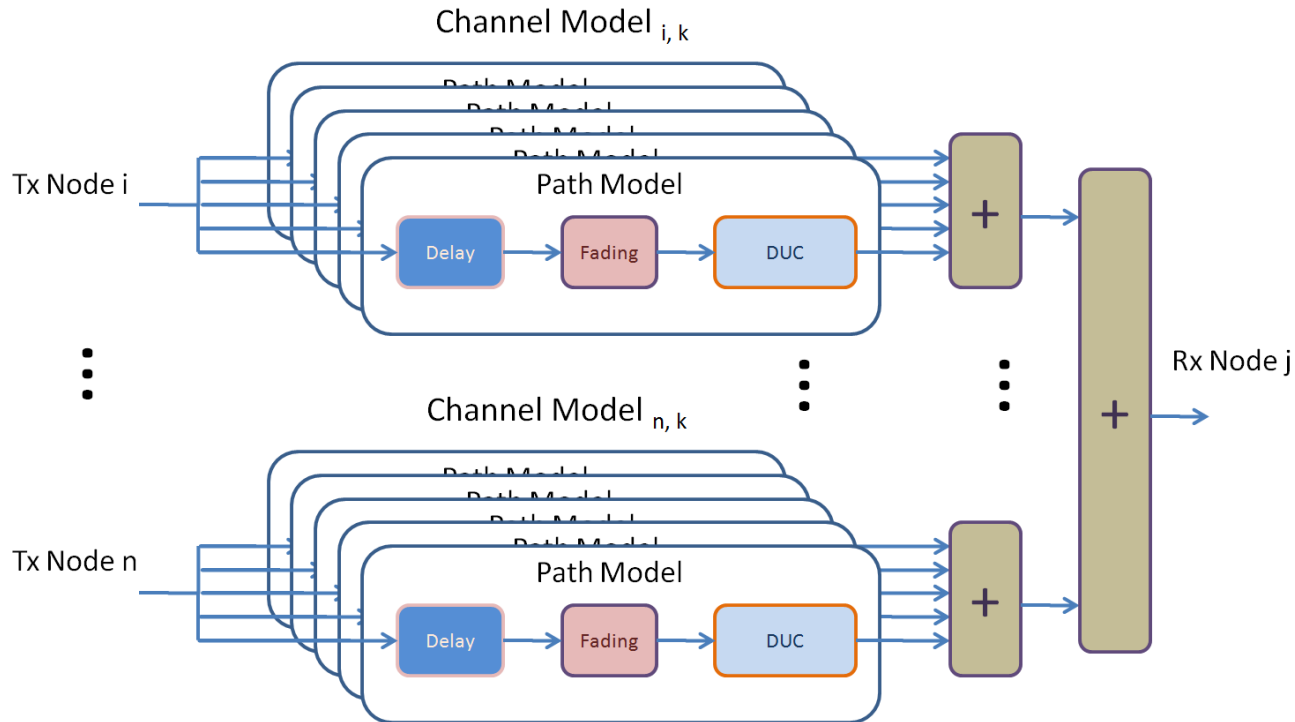
Top level concept of WCE

# Scenario

- ❖ The signal takes multiple paths each with a different **Path delay** and **Path gain**



# Conceptual Multipath Channel Modeling in FPGA



- ❖ Each channel composed of multiple summed paths
- ❖ Delay, path gain, fading, and Doppler modeled in each path
- ❖ **In this work, we focus on implementing a single channel emulator with multi path**

# Single Channel Emulator Model

## ❖ Single channel emulator model

$$so_t = \sum_{i=0}^n si_{t-i\tau} w(t)_i$$

- ❖  $si$ : Previous  $n+1$  input complex samples
- ❖  $so$ : Complex sample output at present time
- ❖  $w(t)$ : Dynamically changing set of weights (gain and delay)

## ❖ Complex sample calculation (channel function)

$$\text{tapline}_{ij} = \text{delayline}_{\text{index}}$$

$$\text{taps}_i = \sum_{j=0}^{N,K} \text{tapline}_{ij} * \text{weight}_{ji}$$

$$so_{r_i} = \sum_{i=0}^N \text{gains}_{r_i} * \text{taps}_{r_i} - \text{gains}_{i_i} * \text{taps}_{i_i}$$

$$so_{i_i} = \sum \text{gains}_{r_i} * \text{taps}_{i_i} - \text{gains}_{i_i} * \text{taps}_{r_i}$$

Path delays

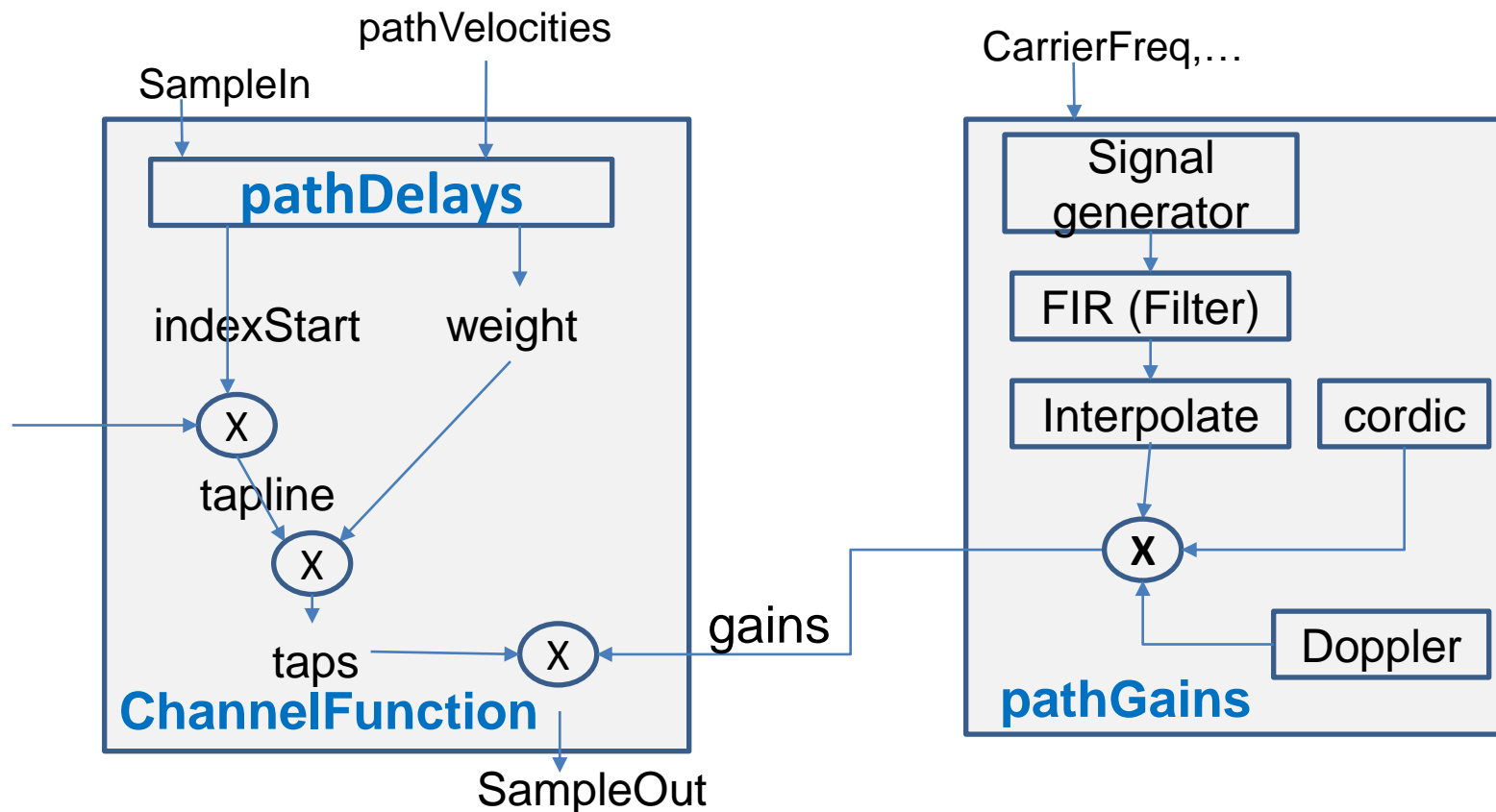
Path gains

channelFunction

# Block Diagram of WCE

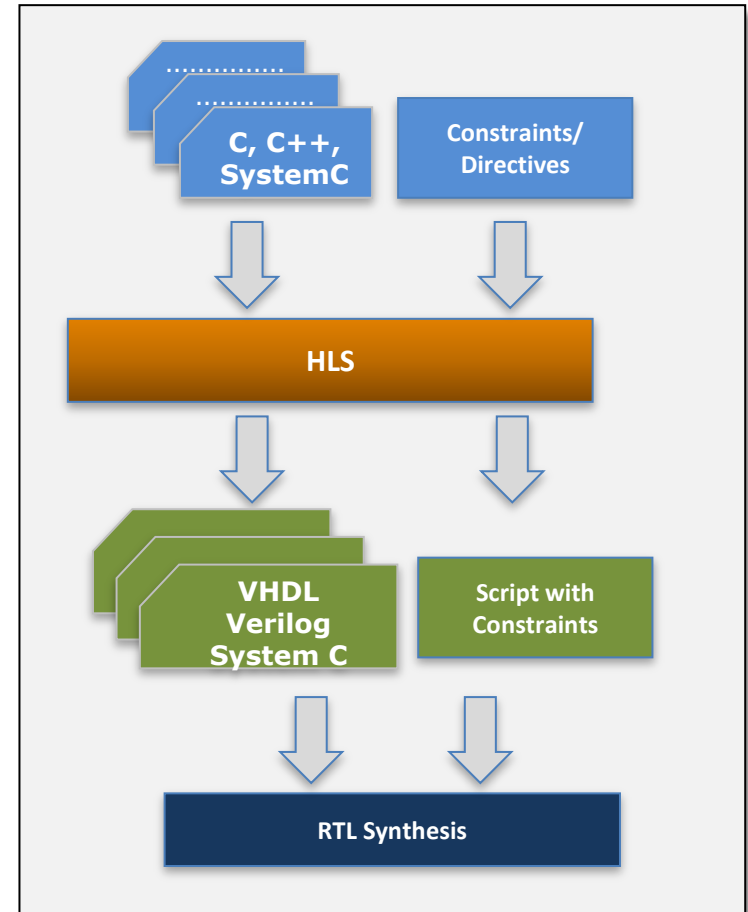
## ❖ Target performance

- ❖ pathDelays and ChannelFunction: 30 Mhz
- ❖ pathGains: 40kHz



# High Level Synthesis

- ❖ High Level Synthesis
  - ❖ Creates an RTL implementation from C level source code
- ❖ Why use HLS ?
  - ❖ A good digital WCE has to handle wide range of dynamically changing parameters such as Doppler effect, fast fading, and multipath
  - ❖ HLS provides easy design space exploration with different parameters
    - ❖ E.g., varying number of paths in a channel



*Courtesy to Xilinx*



# Process of WCE design with HLS

## 1. Baseline design

- ❖ Synthesizable C code

## 2. Restructured design

- ❖ Manually optimizing C code for HW

## 3. Bit accurate design

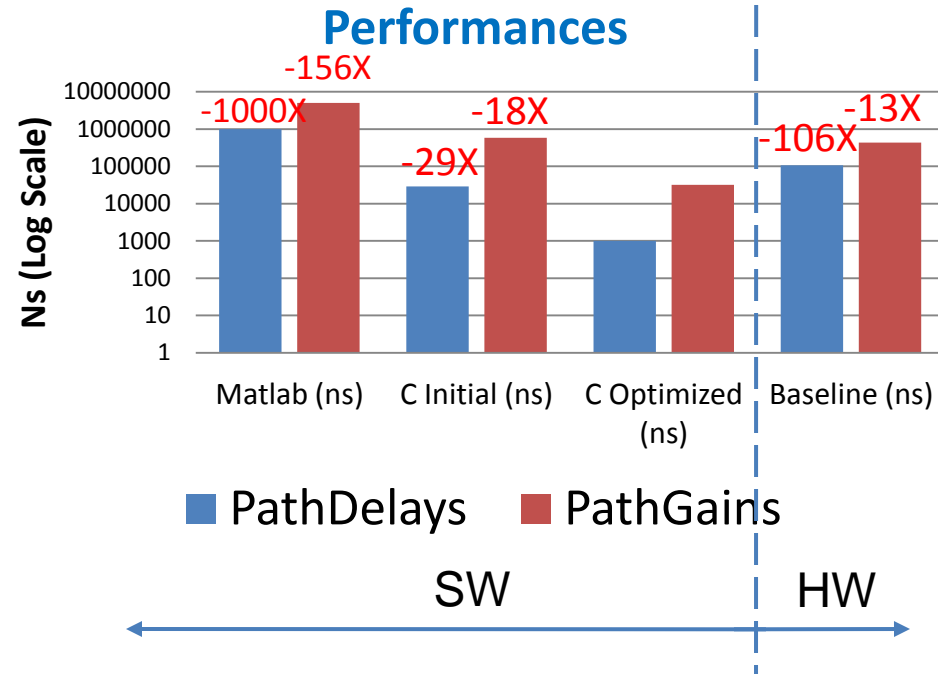
- ❖ Bit-width optimization

## 4. Pipelining, Unrolling and Partitioning (PUP)

- ❖ Parallelizing computation

# 1. Baseline design

- ❖ Main goal:
  - ❖ Synthesizable
- ❖ Things to be done
  - ❖ Matlab → Initial C
  - ❖ Initial C → Optimized C
  - ❖ Remove dependencies
  - ❖ Remove dynamic memory...



# 1. Baseline - Results

## ❖ Baseline

- ❖ PathDelays: **3180X** slower than target (**30 Mhz**)
- ❖ PathGains: **17X** slower than target (**40 kHz**)

## 2. Restructured Design

- ❖ Two goals:
  - ❖ To optimize the code itself without using any HLS pragmas
  - ❖ To write a “C code” targeting the architecture
- ❖ E.g.,
  - ❖ Loop merging
  - ❖ Expression balancing
  - ❖ Loop unrolling
  - ❖ ...

## 2. Restructured Design - Example

### ❖ Clock cycle reduction of pathGains module

#### ❖ Loop merging

```

for (int p = 0; p < N; p++){
  for(int i=0; i < SIZE; i++){
    t1=t1+js[i]*FIRin[p][i][0];
  }
  for(int i=0; i < SIZE; i++){
    t2=t1+js[i]*FIRin[p][i][1];
  }
}

```

→

```

for (int p = 0; p < N; p++){
  for(int i=0; i < SIZE; i++){
    t1=t1+js[i]*FIRin[p][i][0];
    t2=t1+js[i]*FIRin[p][i][1];
  }
  ...
}

```

50215 → 41250

#### ❖ Expression balancing

```

t1=t1+js[i]*FIRin[p][i][0];
t2=t1+js[i]*FIRin[p][i][1];

```

→

```

t1_1=js[i]*FIRin[p][i][0];
t1=t1+t1_1;
t2_1=js[i]*FIRin[p][i][1];
t2=t2+t2_1;

```

41250 → 29730

#### ❖ Loop unrolling

```

for (int p = 0; p < N; p++){
  for (int i = 0; i < SIZE; i++){
    for (int j = 0; j < SIZE2; j++){
      FIRinputs[p][i][j] = ...
    }
  }
}

```

→

```

for (int p = 0; p < N; p++){
  for (int i = 0; i < SIZE; i++){
    FIRinputs[p][i][0] = ...
    FIRinputs[p][i][1] = ...
  }
}

```

29730 → 20785

## 2. Restructured Design – Results

- ❖ Restructured design vs. Target
  - ❖ PathDelays: **523X** slower than target (**30 Mhz**)
  - ❖ PathGains: **7X** slower than target (**40 kHz**)
  - ❖ ChannelFunction: **229X** slower than target (**30 Mhz**)

## 3. Bit accurate design

- ❖ By default, HLS C/C++ have standard types
  - ❖ E.g., char (8-bit), int (32-bit),...
- ❖ Minimizing bit widths will result in smaller & faster hardware
  - ❖ E.g., ap\_fixed and ap\_int
- ❖ Bit accurate design of PathGains module
  - ❖ 55 types are set to use fixed point type

```

#ifdef BIT_ACCURATE
typedef ap_uint<6> AP_UINT6;
typedef ap_uint<12> AP_UINT12;
typedef ap_uint<33> AP_UINT33;
typedef ap_uint<32> AP_UINT32;
typedef ap_int<32> AP_INT32;
typedef ap_uint<20> AP_UINT20;
typedef ap_fixed<15,2,AP_RND > AP_FIXED20_3;
typedef ap_fixed<12,10,AP_RND > AP_FIXED12_10;
typedef ap_fixed<10,9,AP_RND > AP_FIXED10_9;

```

```

#else
typedef unsigned int AP_UINT6;
typedef unsigned int AP_UINT12;
typedef unsigned long long int AP_UINT33;
typedef unsigned long int AP_UINT32;
typedef int AP_INT32;
typedef unsigned int AP_UINT20;
typedef double AP_FIXED20_3;
typedef double AP_FIXED12_10;
typedef double AP_FIXED10_9;
#endif

```

```

typedef ap_fixed<8,4,AP_RND > AP_FIXED8_4;
typedef ap_fixed<15,5,AP_RND > AP_FIXED15_5;
typedef ap_fixed<22,16,AP_RND > AP_FIXED22_16;
typedef ap_fixed<18,16,AP_RND > AP_FIXED18_16;

```

```

typedef ap_fixed<33,1,AP_RND > AP_FIXED33_1;
typedef ap_fixed<32,1,AP_RND > AP_FIXED32_1;
typedef ap_fixed<27,1,AP_RND > AP_FIXED27_1;
typedef ap_fixed<20,1,AP_RND > AP_FIXED20_1;
typedef ap_fixed<10,1,AP_RND > AP_FIXED10_1;
typedef ap_fixed<6,1,AP_RND > AP_FIXED6_1;
typedef ap_fixed<6,5,AP_RND > AP_FIXED6_5;
typedef ap_fixed<20,15,AP_RND > AP_FIXED20_15;
typedef ap_fixed<18,14,AP_RND > AP_FIXED18_14;

```

```

typedef double AP_FIXED45_1;
typedef double AP_FIXED33_1;
typedef double AP_FIXED32_1;
typedef double AP_FIXED27_1;
typedef double AP_FIXED20_1;
typedef double AP_FIXED20_1;
typedef double AP_FIXED10_1;
typedef double AP_FIXED6_1;
typedef double AP_FIXED6_5;
typedef double AP_FIXED20_15;
typedef double AP_FIXED18_14;
#endif

```

## 3. Bit accurate design - Results

- ❖ Bit accurate design vs. Target
  - ❖ PathDelays: **47X** slower than target (**30 Mhz**)
  - ❖ PathGains: **3X** slower than target (**40 kHz**)
  - ❖ ChannelFunction: **133X** slower than target (**30 Mhz**)



## 4. Pipelining and Partitioning

- ❖ On top of bit accurate design, PUP is applied
- ❖ Pipeline
  - ❖ Improves throughput
  - ❖ Default: Target initiation interval( $II$ ) of 1  
 $II=2, II=3, \dots$
- ❖ Partition
  - ❖ BRAMs limit pipelining  $\rightarrow$  Partition large BRAMs into smaller BRAMS or into registers

## 4. Pipelining and Partitioning - Example

### ❖ pathDelays

- ❖ Optimizations: Partition: 5 BRAM, Pipeline: II=1
- ❖ DSP48: 19 → 30 (57%),
- ❖ FF: 424 → 786 (85%),
- ❖ LUT: 563 → 4230 (651%)
- ❖ Throughput: 47X than bit accurate design

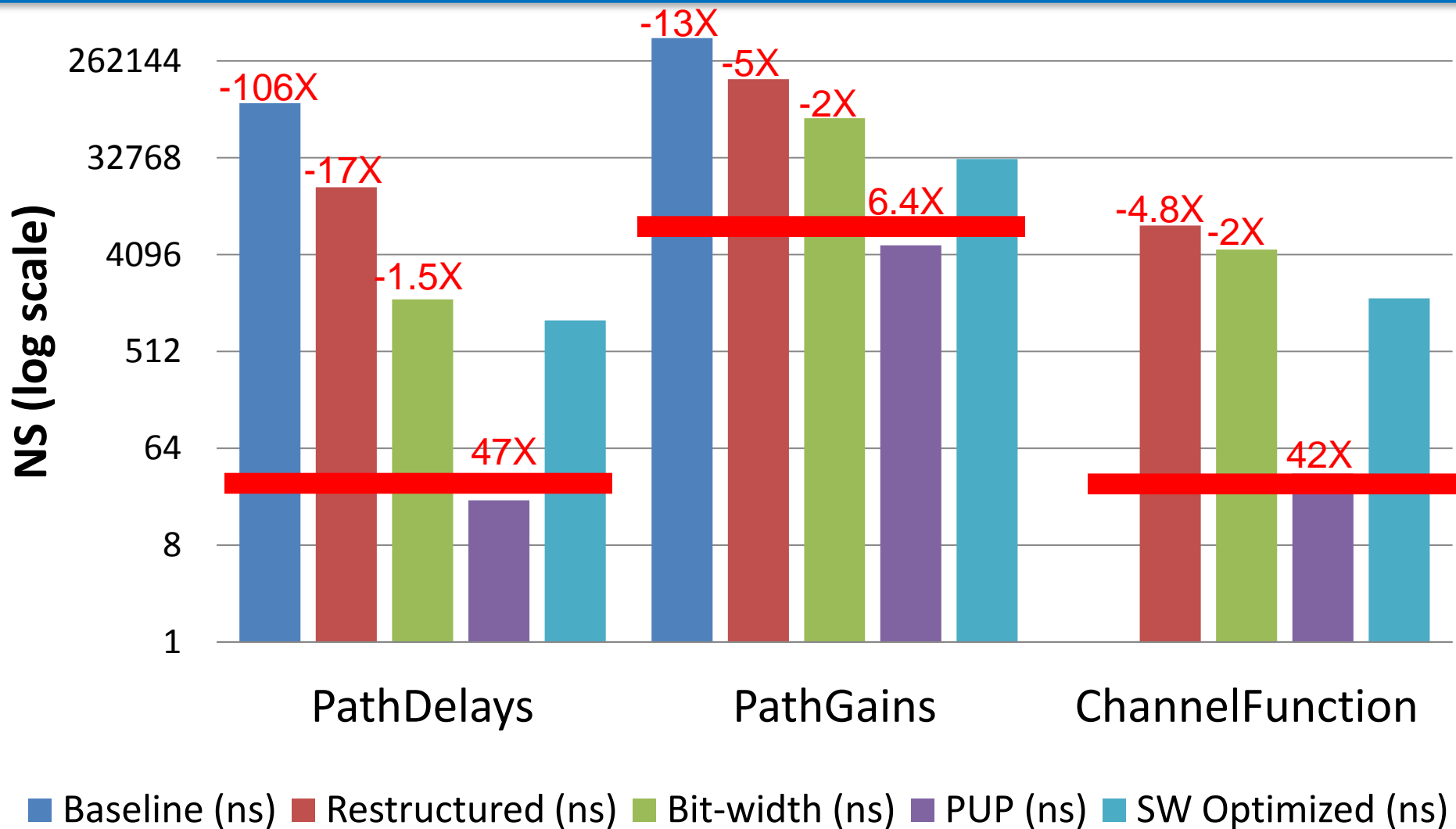
### ❖ pathGains

- ❖ Optimizations: Partition: 12 BRAM of 42, Pipeline/Unroll
- ❖ DSP48E: 47 → 86 (82%),
- ❖ FF: 20399 → 34421 (68%),
- ❖ LUT: 22121 → 38893 (75%)
- ❖ Throughput: 15X than bit accurate design

## 4. Pipelining and Partitioning -Results

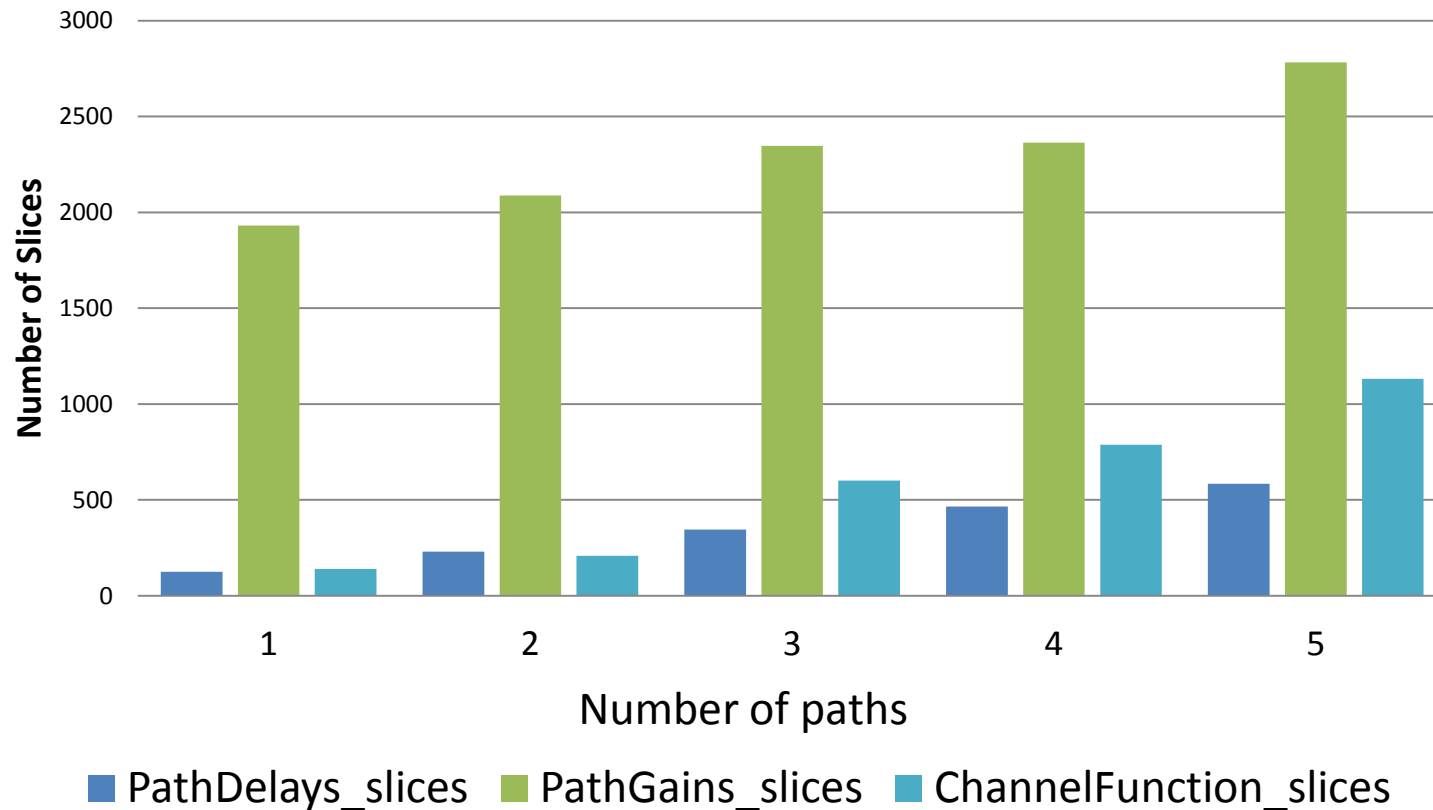
- ❖ Bit accurate design vs. Target
  - ❖ PathDelays **0.6X** slower than target (**30 Mhz**)
  - ❖ PathGains **0.2X** slower than target (**40 kHz**)
  - ❖ ChannelFunction **1.1X** slower than target (**30 Mhz**)

# Final Results



# DSE of WCE for Different Number of Paths

- ❖ HLS allows easy DSE of WCE for different parameters
- ❖ E.g., number of paths 1,2,3,4,5



# Results for Five Paths

## ❖ Resource (xc6vlx240t)

	Slices	LUT	FF	DSP48E	BRAM
PathDelays	584	1843	411	30	0
PathGains	2783	8756	7044	53	30
ChannelFunction	1131	3469	1798	40	0

## ❖ Performance

	Clock Cycles	Clock Period (ns)/ Frequency	Latency(ns)/Throughput
PathDelays	4	5.394 /184 Mhz	21 / 47 Mhz
PathGains	501	9.97 / 100 Mhz	4994 /0.2 Mhz
ChannelFunction	6	6.62 /151 Mhz	37 /26 Mhz

# Design Effort

Design	Days spend	Tasks
Baseline	<b>25.9%</b> (28 hours)	Understanding the code Converting matlab to C++, Removing library dependency, Writing HLS synthesizable code
Restructured code	<b>22.2%</b> (24 hours)	Manually loop merging, Expression balancing, Loop unrolling
Bit Accurate Design	<b>29.6%</b> (32 hours)	Calculation of 57 fixed point type widths (pathGains: 36, PathDelays:19, ChannelFunction: 2)
Optimized Design	<b>7.4%</b> (8 hours)	Optimizing using directives
Collecting Results/DSE/Presenting	<b>14.8%</b> (16 hours)	DSE, Collecting results, Presenting
<b>Total</b>	<b>~108 hours</b>	

# Conclusion

- ❖ Designed single channel wireless emulator using HLS tool.
  - ❖ HLS provides easy parameterization of WCE design.
- ❖ We plan to extend this work to multiple channel emulator and make end-to-end system
- ❖ Lessons Learned
  - ❖ Achieving target performance and area depends
    - ❖ Writing a “C Code” targeting architecture is essential
    - ❖ Application and code size
    - ❖ 2 optimization pragmas (**pipeline, partition out of 33**) + Restructured code+ Bit Width → Target goal