# **Differential Equations of Ideal Memristors**

Zdenek BIOLEK<sup>1</sup>, Dalibor BIOLEK<sup>1,2</sup>, Viera BIOLKOVA<sup>3</sup>

zdenek.biolek@gmail.com, dalibor.biolek@unob.cz, biolkova@feec.vutbr.cz

Abstract. Ideal memristor is a resistor with a memory, which adds dynamics to its behavior. The most usual characteristics describing this dynamics are the constitutive relation (i.e. the relation between flux and charge), or Parameter-vs-state map (PSM), mostly represented by the memristance-to-charge dependence. One of the so far unheeded tools for memristor description is its differential equation (DEM), composed exclusively of instantaneous values of voltage, current, and their derivatives. The article derives a general form of DEM that holds for any ideal memristor and shows that it is always a nonlinear equation of the first order; the PSM forms are found for memristors which are governed by DEMs of the Bernoulli and the Riccati types; a classification of memristors according to the type of their dynamics with respect to voltage and current is carried out.

### **Keywords**

Memristor, differential equation, parameter-vs-state map (PSM)

#### 1. Introduction

In [1], the memristor was originally defined as a one-port establishing a unique relation between charge  $q = \int idt$  and flux  $\varphi = \int vdt$ , where i and v are memristor voltage and current. The unalterable relations  $\varphi = \varphi(q)$  and  $q = q(\varphi)$ , called constitutive relations of the memristor, are exclusive attributes of the element. The memristor always enforces their validity, irrespective of the instantaneous voltage and current relations, and independently of how it is connected to the rest of the circuit. The result is that the memristor appears as a resistor dependent only on the charge q passing through or on the flow  $\varphi$ . That is why two mutually dual forms of Ohm's law apply to the memristor

$$v = R_M(q)i, i = G_M(\varphi)v$$
 (1)

where

$$R_M(q) = \frac{d\varphi}{dq}, G_M(\varphi) = \frac{dq}{d\varphi}$$
 (2)

are the differential memristances and memductances, which can be calculated as the slopes of the corresponding

constitutive relations at the current operating points. The functions (2) are two mutually compatible forms of the PSM (Parameter-vs-state map) characteristics. The parameter is memristance or memductance, and the state of the memristor is the charge or flux as the integral quantity.

Over time, the designation "memristor" began to be used also for memristive systems originally introduced in [2]. They are resistors that are non-linear with respect to voltage/current and that are also dependent on a general vector state of the associated dynamical system. According to the terminology introduced in [3], memristive systems are considered *extended memristors*. This paper concerns the memristor in the sense of [1], which we will call, in accordance with [3], an *ideal memristor*.

A general dynamical one-port can be described by the differential equation

$$F\left(i, v, \frac{di}{dt}, \frac{dv}{dt}, \frac{d^2i}{dt^2}, \frac{d^2v}{dt^2}, \dots\right) = 0$$
 (3)

where v(t) is the one-port voltage and i(t) is the current flowing through it. Type (3) equation is the standard basis of modeling techniques in the circuit theory. It does not have to be limited to one-ports, i.e. v and i can be vectors of the input and output variables. In [4], an algorithm is described that converts the state and the output equations of a nonlinear dynamical system into an equivalent description, consisting of three sets of equations. The first one describes the dynamics of the unobservable part of the system; the second expresses the observable states as functions of input and output variables; the third defines the external behavior of the system using a general vector equation of the form (3).

Research into differential equations of memristor (DEM) is of practical importance. A specific type of equation (3) may lead to a structure of the state-based model which can exhibit better properties than the state space model based on a physical variable which is directly connected with the memory mechanism [5], [6]. Finding the type of DEM can also answer the question whether the memristor response to a known excitation can be computed analytically. However, there is no satisfactory explanation as to what equation (3) looks like in the case of an ideal memristor. Probably the first significant attempt in this respect was made in [7]. This work started a series of pub-

DOI: 10.13164/re.2015.0369 CIRCUITS

Dept. of Microelectronics, Brno University of Technology, Technická 10, 616 00 Brno, Czech Republic
 Dept. of Electrical Engineering, University of Defense, Kounicova 65, 662 10 Brno, Czech Republic

<sup>&</sup>lt;sup>3</sup> Dept. of Radio Electronics, Brno University of Technology, Technická 12, 616 00 Brno, Czech Republic

lications about the so-called Bernoulli dynamics of memristor, and it is the source of the still prevailing opinion [8], [9] that the Bernoulli differential equation (BDE) is a characteristic of the ideal memristor and, consequently, that "ideal memristors obey the Bernoulli dynamics". On the contrary, other publications [10], [11] point out that the BDE is only compatible with a certain special class of ideal memristors.

This paper deals with the problem of the differential equations of ideal memristors as follows. The general form of DEM of ideal memristor is derived in Sec. 2, starting from the memristor PSM in the form of (2). Section 3 shows examples of memristors with various PSMs, leading to various DEMs. It is demonstrated that the type of DEM depends on what variable, voltage or current is used as the input or output variable. Examples of the PSMs are given that lead to linear differential equations with a variable parameter, to the Bernoulli and the Riccati differential equations, to the Abel differential equation of the first kind, and also to equations of other types. It is proved in Sec. 4 that the so-called Bernoulli dynamics has its limits, and PSM forms are found that must comply with memristors exhibiting such dynamics. Sections 5 and 6 are devoted to similar issues but for cases of the Riccati and the Abel dynamics. Section 7 provides the classification of memristors in terms of their dynamics which is exbibited by their voltage and current.

# 2. General Form of Differential Equation of Ideal Memristor

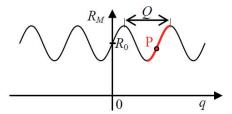
Consider a memristor with current i(t) flowing through it and with voltage v(t) across its terminals. Depending on how the charge q(t) and the flux  $\varphi(t)$  change, its memristance and memductance also vary. Their derivatives with respect to time will be

$$\frac{dR_{M}}{dt} = \frac{dR_{M}(q)}{dq}i, \frac{dG_{M}}{dt} = \frac{dG_{M}(\varphi)}{d\varphi}v, \qquad (4)$$

i.e. the rate of change of  $R_M$  or  $G_M$  is directly proportional to the slope of the respective PSM at the current operating point. Expressing the instantaneous memristance and memductance using the instantaneous values of voltage and current according to (1), performing differentiations on the left-hand sides of (4), and making slight arrangements, we get

$$\frac{dR_M}{dq}i^3 + v\frac{di}{dt} - i\frac{dv}{dt} = 0, \frac{dG_M}{d\varphi}v^3 + i\frac{dv}{dt} - v\frac{di}{dt} = 0.$$
 (5)

It is not yet possible to say that equations (5) are in the form of (3) because the slopes of PSMs are generally dependent on charge or flow. Suppose now that the current operating point is in such a position of the PSM function that its inverse function exists in some its neighborhood, i.e. the function  $q = q(R_M)$  or  $\varphi = \varphi(G_M)$ . The inverse PSM



**Fig. 1.** Example of PSM of  $R_M(q) = R_0 + R \sin(2\pi q/Q)$ . There exists an SPM characteristic  $q = q(R_M)$  in the red marked area around the operating point.

is in fact a state-to-parameter map. We will therefore refer to it as SPM. The layout is shown in Fig. 1.

For each operating point in whose neighborhood the SPM exists (5) can be rewritten to the form

$$f\left(\frac{v}{i}\right)i^3 + v\frac{di}{dt} - i\frac{dv}{dt} = 0, \ g\left(\frac{i}{v}\right)v^3 + i\frac{dv}{dt} - v\frac{di}{dt} = 0 \ (6)$$

where

$$f\left(\frac{v}{i}\right) = \left(\frac{dq}{dR_M}\right)^{-1}, \ g\left(\frac{i}{v}\right) = \left(\frac{d\varphi}{dG_M}\right)^{-1}. \tag{7}$$

Equations (6) are ordinary differential equations of the type (3).

Both forms of DEM (6) are equivalent and one equation follows from the other. We can check this easily by adding up the left sides and the right sides of two original equations (5). We get directly the well-known conversion [11] between the slopes of both dual forms of the PSM

$$\frac{dR_M}{dq}i^3 + \frac{dG_M}{d\varphi}v^3 = 0, \qquad (8)$$

which can be used for a quick transition from one form of equation (6) to the other. In practice, we use an option that better matches the given PSM.

Equations (6) represent the general form of the DEM of any ideal memristor. It is therefore obvious that the voltage and current of the ideal memristor with arbitrary PSM always satisfy the first-order nonlinear differential equation. It follows from (6) that if the unknown quantity is voltage, the DEM may be of a different type than in the case when the unknown quantity is current.

The following section presents several examples clarifying the process of obtaining the DEM from the PSM of an ideal memristor.

# 3. Examples of DEMs of Ideal Memristors

The so-called HP memristor is one of the frequently discussed types of memristors [12]–[14]. Its linear model assumes that the memristance linearly depends on the charge that passed through. Its PSM can therefore be expressed, for example, as follows:

$$R_M(q) = R + kq \tag{9}$$

where R and k are constants. In order to derive DEM (6), it is necessary to determine the form of the SPM in the first step. In this case it is the inversion of a linear PSM, which also gives a linear SPM

$$q(R_M) = \frac{1}{k} (R_M - R). \tag{10}$$

The reciprocal value of the slope of SPM (10) gives the function f() from (7)

$$f(R_M) = k \,, \tag{11}$$

thus DEM (6) has the form

$$kt^3 + v\frac{di}{dt} - i\frac{dv}{dt} = 0. ag{12}$$

It is obvious that if the current i(t) is the known variable and the voltage v(t) is unknown, the memristor is governed by a linear differential equation with time-varying coefficients

$$\frac{dv}{dt} + \alpha(t)v + \beta(t) = 0 \tag{13}$$

where  $\alpha(t) = -i^{-1}(t) di(t)/dt$ ,  $\beta(t) = -ki^2(t)$ . A memristor whose voltage is governed by a linear equation of the type of (13) will be denoted a class <sub>v</sub>L memristor. Equation (13) has an analytical solution [8], [11].

If the known quantity is the voltage v(t) and the unknown variable is the current i(t), the memristor is governed by the Bernoulli differential equation [15]

$$\frac{di}{dt} + \chi(t)i + \delta(t)i^{n} = 0$$
 (14)

where n = 3,  $\chi(t) = -v^{-1}(t)dv(t)/dt$ ,  $\delta(t) = kv^{-1}(t)$ . A memristor whose current is governed by the Bernoulli equation of the type of (14) will be denoted a memristor of class  ${}_{i}\mathbf{B}$ . This equation also has an analytical solution [8], [11]. Memristors belonging simultaneously to the classes  ${}_{v}\mathbf{L}$  and  ${}_{i}\mathbf{B}$  will be denoted memristors of the ( ${}_{v}\mathbf{L}$   ${}_{i}\mathbf{B}$ ) type. This and other examples are described in Tab. 1.

The original ideas of how the HP memristor works led to a simple model of the linear dependence of resistance on the charge that passed through [12], [13]. This corresponds with row 1 in Tab. 1, which summarizes the results of (13) and (14) regarding the types of DEM. The second row represents a memristor whose conductance is directly proportional to the flow. For this memristor, the voltage is governed by the BDE and the current by the LDE. The third row of Tab. 1 shows another variant, when, unlike with the "HP memristor", the conductance and not the resistance is linearly dependent on the charge. When excited by a known current, the memristor would, the same as the "HP memristor", comply with the Bernoulli differential equation with respect to voltage. Compared to the classical variant, however, the constant n from (14) would change from 3 to 2. Row 4 refers to a memristor with linear resistance-to-flow characteristic. It is noteworthy that if we interchange the resistance and the conductance and simultaneously the charge and the flux in the starting conditions, the result leads to the same type of DEM. The PSM of the  $(a + bq)^n$  type leads to memristors of the type of  $({}_{\nu}\mathbf{B}_{i}\mathbf{B})$  if  $n \notin \{-1; -0.5; 1\}$ ; for n = -1 a n = -1/2 we obtain  $({}_{v}\mathbf{B}_{i}\mathbf{L})$ -

PSM		DE	Class of DE for the output as		Type of memristor
			voltage	current	memi istoi
1	$R_{M}(q) = R + kq$	$ki^3 - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	LDE	BDE	$({}_{\scriptscriptstyle U}\!$
2	$G_{M}(\varphi) = G - k\varphi$	$kv^3 - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	BDE	LDE	$(_{v}\mathbf{B}_{i}\mathbf{L})$
3	$G_{M}(q) = G - kq$	$kv^2i - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	LDE	BDE	$(_{v}\mathbf{L}_{i}\mathbf{B})$
4	$R_{M}(\varphi) = R + k\varphi$	$ki^2v - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	BDE	LDE	$(\mathbf{J}_{i}\mathbf{B}_{v})$
5	$R_{M}(q) = (a+bq)^{n}$	$bnv^{1-\frac{1}{n}i^{2+\frac{1}{n}}} - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	$\mathop{\mathrm{BDE}}_{n\neq 1}$	BDE n∉{-1;-0.5}	
6	$R_{M}(q) = R + ke^{Aq}$	$-ARi^3 + Avi^2 - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	LDE	ADE	$(_{v}\mathbf{L}_{i}\mathbf{A})$
7	$R_{M}(q) = \frac{R}{\sqrt{k + e^{Aq}}}$	$\frac{A}{2R^2}kv^3 - \frac{A}{2}vi^2 - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	BDE	RDE	$(_{\nu}\mathbf{B}_{i}\mathbf{R})$
8	$R_{M}(q) = a + bq^{n}, a \neq 0$	$nb^{\frac{1}{n}} \left( \frac{v}{i} - a \right)^{1 - \frac{1}{n}} t^3 - i \frac{dv}{dt} + v \frac{di}{dt} = 0$	XDE	XDE	(v? i?)
9	$R_M(q) = R_0 + R\sin(Aq)$	$A\sqrt{R^{2}i^{2} - (v - R_{0}i)^{2}}i^{2} - i\frac{dv}{dt} + v\frac{di}{dt} = 0$	XDE	XDE	( <sub>v</sub> ? <sub>i</sub> ?)

LDE The Linear Differential Equation

BDE The Bernoulli Differential Equation

XDE Another Differential Equation

Tab. 1. Relation between PSM and class of DEM.

RDE The Riccati Differential Equation

ADE The Abel (first kind) Differential Equation

type memristor, and for n = 1 it would be a ( $_{v}$ **L**  $_{i}$ **B**)-type memristor.

Table 1 shows that there are memristors whose DEMs are not of the BDE type with respect to both voltage and current. As an example, one can mention PSMs that lead to the Abel differential equation of the first kind (ADE, row 6) or to an equation of quite a different type (rows 8 and 9). A frequently held view [7], [8] that ideal memristors are inseparably adherent to the Bernoulli dynamics, is therefore incorrect. The next section provides a derivation of the limits of the Bernoulli dynamics, i.e. the conditions for the PSM of the memristor such that the dynamics of its voltage or current would comply with the Bernoulli differential equation.

### 4. Bernoulli Dynamics

Denote  $_{\nu}\mathbf{B}$  (or  $_{i}\mathbf{B}$ ) a class of all the memristors that exhibit the Bernoulli dynamics with respect to voltage (or current), i.e. when equation (6) is a Bernoulli differential equation for voltage (or current). The question of what type of PSM all the memristors in classes  $_{\nu}\mathbf{B}$  and  $_{i}\mathbf{B}$  must have can be changed to another question: What conditions must the respective functions f() or g() from equation (6) meet?

Just like the PSM of ideal memristor (2) does not explicitly depend on the voltage or current but exclusively on the charge or flow, its SPM also depends exclusively on the memristance or memductance. The function  $f(\cdot)$  or  $g(\cdot)$  from (6) is therefore always a function of the ratio v/i, i.e. voltage and current cannot figure therein separately or in any other form than the above ratio. This fact makes the next steps easier, because those forms of  $f(\cdot)$  or  $g(\cdot)$  are sought which cover all the possibilities of generating the BDE from equations (6) with respect to voltage or current.

The function f() or g() will generally have different forms for memristors of classes  $_{\nu}\mathbf{B}$  and  $_{i}\mathbf{B}$ . We will denote the first one  $f_{\nu}()$  or  $g_{\nu}()$  and the other  $f_{i}()$  or  $g_{i}()$ . If equations (6) are to be BDEs with respect to current i, the function  $f_{i}()$  must fulfill the condition

$$f_i\left(\frac{v}{i}\right) = k_1 \left(\frac{v}{i}\right)^{3-n} + k_2 \left(\frac{v}{i}\right)^2 \tag{15}$$

where n is the constant from (14),  $k_1, k_2 \in R$ ,  $k_1 \ne 0$ . The first summand in (15), after multiplying by  $i^3$  from (6), generates the element  $k_1 v^{3-n} i^n$ . For a BDE with respect to i it is required that  $n \notin \{0;1\}$  [15]. The second summand in (15) may or may not contribute to the element (idv/dt) in (6) with a contribution  $k_2 v^2 i$  without violating the format of the BDE (14) with respect to i. Taking into account the condition (7) and the fact that  $v/i = R_M$ , then

$$\frac{dR_M}{da} = k_1 R_M^{3-n} + k_2 R_M^2 \,. \tag{16}$$

Relation (16) is a differential equation that generates a PSM of the  $R_M(q)$  type of all the memristors of class  ${}_{i}\mathbf{B}$ .

After the separation of variables, the appropriate SPM function can be expressed as the integral

$$q(R_M) = \int \frac{dR_M}{k_1 R_M^{3-n} + k_2 R_M^2}$$
 (17)

whose general solution can be found in terms of the hypergeometric function

$$q(R_M) = -\frac{1}{k_2 R_M} {}_2F_1 \left( 1, -\frac{1}{1-n}; -\frac{n}{1-n}; -\frac{k_1}{k_2} R_M^{1-n} \right). (18)$$

The resulting form of the PSM for the special case of  $k_2 = 0$  can be expressed in terms of standard functions:

$$R_{M}(q) = [k_{1}(n-2)(q-q_{0})]^{\frac{1}{n-2}}$$
(19)

where  $q_0 \in R$ .

The general form of the PSM of  $R_M(q)$  type that would cover the entire set of  ${}_i\mathbf{B}$  cannot be expressed via the elementary functions, but we can get its dual form  $G_M(\varphi)$ . Utilizing relations (8) and (16) and a simple rearrangement yield that all the PSMs of the  $G_M(\varphi)$  type that lead to memristors of class  ${}_i\mathbf{B}$  must be solutions of the differential equation

$$\frac{dG_M}{d\omega} = -k_1 G_M^n - k_2 G_M \tag{20}$$

which is a Bernoulli differential equation with constant parameters. Substituting the function  $g_i() = -k_1()^n - k_2()$  into the second equation in (6), we get the DEM form for any memristor of class  $_i$ **B** 

$$\frac{di}{dt} = -k_1 v^{2-n} i^n + \left( -k_2 v + \frac{1}{v} \frac{dv}{dt} \right) i.$$
 (21)

It follows from (21) that each memristor of class <sub>i</sub>**B** must with respect to voltage exhibit the dynamics governed by the differential equation

$$\frac{dv}{dt} = k_1 i^{n-1} v^{3-n} + k_2 v^2 + \frac{v}{i} \frac{di}{dt}.$$
 (22)

It is an equation that, for special combinations of  $k_2$  and n, becomes the BDE, RDE, or ADE of the first kind, or the LDE with time-varying parameters.

Equation (20) has an analytic solution. For  $k_2 = 0$  it is

$$G_{M}(\varphi) = [k_{1}(n-1)(\varphi - \varphi_{0})]^{\frac{1}{1-n}}$$
 (23)

where  $\varphi_0 \in R$ .

The solution of (20) for  $k_2 \neq 0$  is as follows:

$$\varphi - \varphi_0 = \frac{\ln \left| k_1 + k_2 G_M^{1-n} \right|}{k_2 (n-1)}.$$
 (24)

For some combinations of the constants  $k_1$  and  $k_2$ , the function (24) will show singularities at points where the argument of the abs () function acquires zero. Therefore, the resulting formula for  $G_M(\varphi)$  is ambiguous:

$$G_{M}(\varphi) = \left[\frac{\pm e^{-k_{2}(n-1)(\varphi-\varphi_{0})} - k_{1}}{k_{2}}\right]^{\frac{1}{1-n}}.$$
 (25)

The sign of the exponential term depends on the sign of the argument of the abs () function in (24).

Memristors of class iB with the BDE (21) must therefore have the  $G_M(\varphi)$  characteristic in the form of (23) or (25). Since the constants  $k_1$ ,  $k_2$ ,  $\varphi_0$  and n are, with discrete exceptions (e.g.  $k_1 \neq 0$ , etc.), arbitrary real numbers, the basic forms (23) and (25) can be transposed, with the help of other constants, into other forms. All adjustments that lead to the original structure of the source equation (20) but with different, mutually independent constants are acceptable, i.e.

$$\frac{dG_M}{d\varphi} = k_1' G_M^{n'} + k_2' G_M . \tag{26}$$

For example, considering the term 1/(1-n) in (25) as an independent constant N, then, due to the fact that  $n \notin \{0,1\}$ , there is only one limitation for N, namely  $N \neq 1$ . Similarly, since  $k_2$  in (25) is an arbitrary nonzero constant, the expression  $A = -k_2(n-1)$  can be considered a nonzero independent constant. Based on these facts, (23) and (25) can be converted to the basic forms given in Tab. 2.

In Tab. 2, we can also find analogous results for class <sub>v</sub>**B** memristors, which we would get via the same procedure we used to get conclusions (15) $\div$ (25). Note that current i and voltage v, charge q and flux  $\varphi$ , and memristance  $R_M$ and memductance  $G_M$  mutually interchange their roles. Table 2 contains all the possible forms of  $G_M(\varphi)$  of mem-

Memristors of class <sub>i</sub> B				
DE	$\frac{di}{dt} = -k_1 v^{2-n} i^n + \left( -k_2 v + \frac{1}{v} \frac{dv}{dt} \right) i = 0$			
Generating equation for $G_M$	$\frac{dG_M}{d\varphi} = -k_1 G_M^n - k_2 G_M$			
$G_{M}(\varphi)$ form	$G_{M}(\varphi) = G_{i}(\varphi - \varphi_{0})^{N}, N=1/(1-n), N\neq 1$ $G_{M}(\varphi) = G(a + e^{A(\varphi - \varphi_{0})})^{N}, N=1/(1-n), N\neq 1, a\neq 0$			
Generating equation for $R_M$	$\frac{dR_{M}}{dq} = k_{1}R_{M}^{3-n} + k_{2}R_{M}^{2}$			
$R_M(q)$ form	Non-standard function $R_{M}(q) = R_{i}(q - q_{0})^{N}, N=1/(n-2), N \notin \{-1/2;-1\}$			
Memristors of class <sub>v</sub> B				
DE	$\frac{dv}{dt} = k_1 i^{2-n} v^n + \left(k_2 i + \frac{1}{i} \frac{di}{dt}\right) v = 0$			
Generating equation for $G_M$	$\frac{dR_M}{dq} = k_1 R_M^n + k_2 R_M$			
$R_M(q)$ form	$R_{M}(q) = R_{v}(q - q_{0})^{N}, N=1/(1-n), N\neq 1$ $R_{M}(q) = R(b + e^{A(q-q_{0})})^{N}, N=1/(1-n), N\neq 1, b\neq 0$			
Generating equation for $G_M$	$\frac{dG_{M}}{d\varphi} = -k_{1}G_{M}^{3-n} - k_{2}G_{M}^{2}$			
$G_{M}(\varphi)$ form	Non-standard function $G_{M}(\varphi) = G_{v}(\varphi - \varphi_{0})^{N}, N=1/(n-2), N \notin \{-1/2;-1\}$			

**Tab. 2.** Boundary of the Bernoulli dynamics of ideal

ristors of class  ${}_{i}\mathbf{B}$  and all the possible forms of  $R_{M}(q)$  of memristors of the  $_{\nu}\mathbf{B}$  class. The general form of  $R_{M}(q)$  for class  ${}_{i}\mathbf{B}$  and  $G_{M}(\varphi)$  for class  ${}_{v}\mathbf{B}$  cannot be expressed via standard functions. We will succeed only with special combinations of parameters in the generating equations.

The linear model of the HP memristor [13] not employing any window functions is the well-known representative of the "Bernoulli dynamics". According to our classification, it is a memristor of the  $({}_{\nu}\mathbf{L}_{i}\mathbf{B})$  type with the BDE constant n = 3, see row 1 in Tab. 1.

### 5. Riccati Dynamics

Row 7 in Tab. 1 represents the case of a memristor whose current is governed by the Riccati differential equation. The class of memristors complying with this dynamics will be denoted by the symbol  ${}_{i}\mathbf{R}$ . Let us examine the requirements for the PSM that lead to the classes R and  $i\mathbf{R}$ .

In accordance with [15], regarded as a Riccati differential equation is an equation of the type of

$$\frac{dy}{dt} = f_2(t)y^2 + f_1(t)y + f_0(t). \tag{27}$$

For our purposes, y(t) will be either the memristor voltage or current.

On the basis of (6), (7) and (27), it can easily be found that belonging to the dynamics of class  $_{\nu}\mathbf{R}$  is equivalent to the condition

$$\frac{dR_M}{dq} = k_1 R_M^2 + k_2 R_M + k_3, (28)$$

 $k_1, k_2, k_3 \in \mathbb{R}, k_1 \neq 0, k_3 \neq 0$ . Then the voltage will comply with the differential equation

$$\frac{dv}{dt} = k_1 v^2 + \left(k_2 i + \frac{1}{i} \frac{di}{dt}\right) v + k_3 i^2,$$
 (29)

which is formally identical to the Riccati equation (27). The SPM  $q = q(R_M)$  can be calculated as the integral

$$q(R_M) = \int \frac{dR_M}{k_1 R_M^2 + k_2 R_M + k_3} \,. \tag{30}$$

The result depends on the sign of the discriminant

$$D = k_2^2 - 4k_1k_3 \tag{31}$$

as follows:  

$$q(R_M) = q_0 + \begin{cases} \frac{2}{\sqrt{-D}} \tan^{-1} \left( \frac{2k_1 R_M + k_2}{\sqrt{-D}} \right), & D < 0 \\ -\frac{2}{\sqrt{D}} \tanh^{-1} \left( \frac{2k_1 R_M + k_2}{\sqrt{D}} \right), & D > 0 \end{cases}$$

$$-\frac{1}{k_1 (k_1 R_M + k_2)}, \quad D = 0$$

The PSM is derived by inverting the formula (32)

$$R_{M}(q) = \begin{cases} \frac{1}{2k_{1}} \left( \sqrt{-D} \tan \left( \frac{\sqrt{-D}}{2} (q - q_{0}) \right) - k_{2} \right), & D < 0 \\ \frac{1}{2k_{1}} \left( \sqrt{D} \tanh \left( -\frac{\sqrt{D}}{2} (q - q_{0}) \right) - k_{2} \right), & D > 0 \\ -\frac{1}{k_{1}} \left( k_{2} + \frac{1}{k_{1} (q - q_{0})} \right), & D = 0 \end{cases}$$
(33)

Relations (33) represent the complete set of the forms of  $R_M(q)$  leading to the Riccati dynamics  $_{\nu}\mathbf{R}$ .

Since the hyperbolic tangent can be expressed using the exponential functions, the memristance for D > 0 can also be written as

$$R_M(q) = -\frac{k_2}{2k_1} + \frac{\sqrt{D}}{2k_1} \left( 1 - \frac{2}{1 + e^{-\sqrt{D}(q - q_0)}} \right). \tag{34}$$

Note that this relation employs the well-known sigmoid function [5].

Applying (8) to (28) yields an equation generating the PSMs of the  $G_M(\varphi)$  type

$$\frac{dG_{M}}{d\omega} = -k_{1}G_{M} - k_{2}G_{M}^{2} - k_{3}G_{M}^{3}$$
 (35)

whose general solution lies outside the area of standard functions. A particular solution, however, can be found for  $k_2 = 0$ ,  $k_1k_3 < 0$  in the form of

$$G_M(\varphi) = \sqrt{\frac{k_1}{k_1(1 - e^{2k_1(\varphi - \varphi_0)})}}$$
 (36)

where  $\varphi_0 \in R$ .

Another solution is available for D = 0 in the form of

$$G_{M}(\varphi) = \frac{1}{k_{1}k_{2}(\varphi - \varphi_{0})} \frac{1}{1 + W^{-1}(k_{1}k_{2}e^{k_{1}^{2}(\varphi - \varphi_{0})})}$$
(37)

where W() is the Lambert function.

General forms of the PSM that belong to memristors of classes  $_{\nu}\mathbf{R}$  and  $_{i}\mathbf{R}$  are summarized in Tab. 3. Table 3 contains all the possible forms of the  $G_{M}(\varphi)$  of the class  $_{i}\mathbf{R}$  memristors and all the possible forms of the  $R_{M}(q)$  of the class  $_{\nu}\mathbf{R}$  memristors. The general form of the  $R_{M}(q)$  for class  $_{i}\mathbf{R}$  and the  $G_{M}(\varphi)$  for class  $_{\nu}\mathbf{R}$  cannot be expressed using standard functions. We will only succeed for special combinations of the parameters in the generating equations.

The well-known model of the HP memristor with the Joglekar window function with the parameter p = 1 [13] exhibits the Riccati dynamics with respect to voltage. Its PSM is derived in [5]

$$R_M(q) = R_{off} - \frac{R_{off} - R_{on}}{ae^{-4kq} + 1}, \ a = \frac{R_{ini} - R_{on}}{R_{off} - R_{ini}}$$
 (38)

where  $R_{off}$ ,  $R_{on}$ , and  $R_{ini}$  are the maximum, minimum, and the initial memristor resistances, and k is the technological constant. According to Tab. 3, it is the  $R_M(q)$  of the memristor of class  $_{\nu}\mathbf{R}$ .

Memristors of class AR				
DE	$\frac{di}{dt} = -k_1 i^2 + \left(-k_2 v + \frac{1}{v} \frac{dv}{dt}\right) i - k_3 v^2$			
Generating equation for $G_M$	$\frac{dG_M}{d\varphi} = k_1 G_M^2 + k_2 G_M + k_3$			
$G_{M}(arphi)$ form	$egin{aligned} G_{_0} + G_{_1}  an (a(arphi - arphi_0)) \ G_{_M} ig( arphi ig) &= G_{_0} + rac{G_{_1}}{a + e^{A(arphi - arphi_0)}}, \ G_0  eq 0 \ G_{_0} + rac{G_{_1}}{arphi - arphi_0}, \ G_0  eq 0 \end{aligned}$			
Generating equation for $R_M$	$\frac{dR_{M}}{dq} = -k_{1}R_{M} - k_{2}R_{M}^{2} - k_{3}R_{M}^{3}$			
$R_M(q)$ form	$Non-standard function \ rac{R}{q-q_0}rac{1}{1+W^{-1}ig(ke^{A(q-q_0)}ig)}, \ W() \ is \ the \ Lambert function \ rac{R}{\sqrt{a+e^{A(q-q_0)}}}, \ a eq 0$			
	Memristors of class , R			
DE	$\frac{dv}{dt} = k_1 v^2 + \left(k_2 i + \frac{1}{i} \frac{di}{dt}\right) v + k_3 i^2$			
Generating equation for $G_M$	$\frac{dR_M}{dq} = k_1 R_M^2 + k_2 R_M + k_3$			
$R_M(q)$ form	$R_{0} + R_{1} \tan(a(q - q_{0}))$ $R_{M}(q) = R_{0} + \frac{R_{1}}{b + e^{A(q - q_{0})}}, R_{0} \neq 0$ $R_{0} + \frac{R_{1}}{q - q_{0}}, R_{0} \neq 0$			
Generating equation for $G_M$	$\frac{dG_{M}}{d\varphi} = -k_{1}G_{M} - k_{2}G_{M}^{2} - k_{3}G_{M}^{3}$			
$G_{\!\scriptscriptstyle M}\!(arphi)$ form	Non-standard function $\frac{G}{\varphi - \varphi_0} \frac{1}{1 + W^{-1} \left(k e^{A(\varphi - \varphi_0)}\right)},$ $W() is the Lambert function \frac{G}{\sqrt{b + e^{A(\varphi - \varphi_0)}}}, b \neq 0$			

Tab. 3. Boundary of the Riccati dynamics of ideal memristor.

### 6. Abel Dynamics

Consider the differential equation of the form

$$\frac{dy}{dt} = f_0(t) + f_1(t)y + f_2(t)y^2 + ... + f_n(t)y^n$$
 (39)

where n is a positive integer. Equation (39) is the generalized Abel differential equation (ADE) of the first kind [16]. Its basic form is given for n = 3 [15]. If n = 2 and simultaneously  $f_0(t) \neq 0$ , (39) changes to the Riccati differential equation. For n > 1 and simultaneously  $f_i(t) = 0$ ,  $i \in \{0,2,3,...,n-1\}$ , (39) becomes the Bernoulli differential equation for integer-type n.

If the current/voltage of the memristor is governed by the basic form of (39), i.e. for n = 3, then it belongs to the

 $_{i}\mathbf{A}/_{\nu}\mathbf{A}$  class. A specific case of the  $_{i}\mathbf{A}$  class memristor is shown in row 6 of Tab. 1.

Let us examine the requirements for the PSM of the memristors of class  $_{\nu}\mathbf{A}$ . The generating equation for the PSM of  $R_{M}(q)$  type can be obtained the same way as for the Bernoulli and the Riccati dynamics. It is the ADE of the first kind

$$\frac{dR_{M}}{da} = k_{0} + k_{1}R_{M} + k_{2}R_{M}^{2} + ... + k_{n}R_{M}^{n}, \qquad (40)$$

with the voltage conforming to the differential equation

$$\frac{dv}{dt} = k_0 i^2 + \left(k_1 i + i^{-1} \frac{di}{dt}\right) v +$$

$$k_2 v^2 + k_3 i^{-1} v^3 + k_4 i^{-2} v^4 ... + k_n i^{2-n} v.^n$$
(41)

Note that this equation is of type (39). The current will be the solution of the equation

$$\frac{di}{dt} = -k_0 \frac{1}{v} i^3 - k_1 i^2 - \left(k_2 v - \frac{1}{v} \frac{dv}{dt}\right) i - k_2 v^2 - k_4 v^3 i^{-1} - k_5 v^4 i^{-2} \dots - k_n v^{n-1} i^{3-n}$$
(42)

which is of type (39) only for n = 3.

The generating equations for the PSM of memristors of classes  ${}_{i}\mathbf{A}$  and  ${}_{\nu}\mathbf{A}$ , together with the relevant differential equations, are shown in Tab. 4. The general closed-form solution of the generating equations is complicated, being available only in particular cases. That is why it is not given in Tab. 4.

Memristors of class A				
DE	$\frac{di}{dt} = -k_0 v^2 + \left(-k_1 v + v^{-1} \frac{dv}{dt}\right) i - k_2 i^2 - k_3 v^{-1} i^3$			
Generating equation for $G_M$	$\frac{dG_{M}}{d\varphi} = k_{0} + k_{1}G_{M} + k_{2}G_{M}^{2} + k_{3}G_{M}^{3}$			
$G_M(\varphi)$ form	Non-standard function			
Generating equation for $R_M$	$\frac{dR_M}{dq} = -k_0 R_M^3 - k_1 R_M^2 - k_2 R_M - k_3$			
$R_M(q)$ form	Non-standard function			
Memristors of class vA				
DE	$\frac{dv}{dt} = k_0 i^2 + \left(k_1 i + i^{-1} \frac{di}{dt}\right) v + k_2 v^2 + k_3 i^{-1} v^3$			
Generating equation for $G_M$	$\frac{dR_M}{dq} = k_0 + k_1 R_M + k_2 R_M^2 + k_3 R_M^3$			
$R_M(q)$ form	Non-standard function			
Generating equation for $G_M$	$\frac{dG_{M}}{d\varphi} = -k_{0}G_{M}^{3} - k_{1}G_{M}^{2} - k_{2}G_{M} - k_{3}$			
$G_M(\varphi)$ form	Non-standard function			

**Tab. 4.** Boundary of the Abel dynamics of ideal memristor.

The modification of the HP memristor proposed in [11], which is based on the exponential change in the cross-section of the memristor, leads to the memristor of the ( $_{\nu}\mathbf{L}_{i}\mathbf{A}$ ) type, see row 6 in Tab. 1. Neither the voltage nor the current of the memristor is in compliance with the BDE.

# 7. Possible Types of Memristor in Terms of DEM

Consider the differential equation

$$\frac{dy}{dt} = f_R(t)y^R + f_0(t) + f_1(t)y + f_2(t)y^2 + ... + f_N(t)y^N$$
(43)

where n is a positive integer and R is a non-integer real number. If some of the functions  $f_R()$ ,  $f_0()$ , ...,  $f_N()$  were identically zero, then (43) could become the LDE, BDE, RDE, ADE or generalized ADE.

If y(t) is the voltage, then the generating equation for the PSM of the  $R_M(q)$  type will be

$$\frac{dR_M}{dq} = k_R R_M^R + k_0 + k_1 R_M + k_2 R_M^2 + ... + k_N R_M^N,$$
 (44)

with the voltage conforming to the differential equation

$$\frac{dv}{dt} = k_R i^{2-R} v^R + k_0 i^2 + \left( k_1 i + i^{-1} \frac{di}{dt} \right) v +$$

$$k_2 v^2 + k_3 i^{-1} v^3 + k_4 i^{-2} v^4 ... + k_N i^{2-N} v^N.$$
(45)

Note that this equation is of type (43). The current will be a solution of the equation

$$\frac{di}{dt} = -k_R v^{R-1} i^{3-R} - k_0 \frac{1}{v} i^3 - k_1 i^2 - \left( k_2 v - \frac{1}{v} \frac{dv}{dt} \right) i - (46)$$

$$k_3 v^2 - k_4 v^3 i^{-1} - k_5 v^4 i^{-2} \dots - k_N v^{N-1} i^{3-N}.$$

This equation is of type (43) only for n = 3.

Analogous results hold for y(t) representing the current. The analysis of these results yields all the possible combinations of classes that are represented as a matrix in Fig. 2.

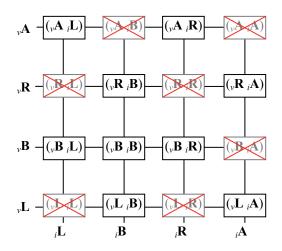


Fig. 2. Overview of the types of memristor according to the type of differential equation. The types that are fundamentally impossible are crossed out.

Some of the combinations of classes are not realizable in principle. If, for example, the memristor voltage is governed by the Bernoulli differential equation, then its current cannot be governed by a type of equation other than LDE, BDE or RDE. If we observe that the memristor voltage is governed by a linear differential equation, then we can be confident that the current must follow either the Bernoulli differential equation or the Abel differential equation of the first kind. If the voltage and current are to be governed by the same type of DEM, then it must be only the Bernoulli differential equation.

It is evident that there are memristors of different types outside the matrix. Examples can be found in Tab. 1 in rows 8 and 9. However, the scheme exhibits the symmetry  $(_{\nu}X,_{i}Y) \leftrightarrow (_{\nu}Y,_{i}X)$  with regard to the main diagonal. It means that the classes  $_{\nu}X$  and  $_{i}Y$ , presented in the matrix, cannot occur for any memristor that lies outside the matrix. To put it concisely: if, for example, the memristor current complies with a linear differential equation with timevarying coefficients, then the voltage cannot be governed by other than the Bernoulli or the Abel of the first kind dynamics. Likewise, the voltage or current of a memristor whose type is outside the matrix cannot exhibit linear dynamics (i.e. dynamics governed by a linear differential equation).

#### 8. Conclusions

The voltage and current of the ideal memristor are, under all circumstances, governed by an ordinary differential equation of the first order (6). This equation can be linear with variable coefficients, Bernoulli, Riccati, Abel of the first kind or a differential equation of quite a different type.

The methodology for determining the type of the differential equation from the memristor PSM is given by equations (6) to (8). This methodology is successful if it is possible to invert the given PSM analytically in terms of standard functions. It was used for determining the differential equations of all the types of memristors that are given in Tab. 1.

A method has been found for generating the PSM which must apply to memristors with a given type of dynamics with respect to voltage or current. For a prescribed type of the dynamics, the generating differential equation is found which generates all the PSMs sought. It is noteworthy that the generating differential equation is of the same type as the original differential equation that determines the type of the dynamics, but it is simpler in the sense that it has constant parameters. It should also be noted that the generating equations generate all the possible PSM functions that would lead to the prescribed type of the dynamics. In the case of the synthesis, for example, of only passive memristors, the free constants in the generating equations or directly in the resulting PSMs would be correspondingly limited. Another remarkable observation: it is possible to find all the closed-form PSMs of the  $R_M(q)/G_M(\varphi)$  type in terms of the standard functions for the Bernoulli and the Riccati dynamics with respect to

voltage/current. Finding the corresponding counterpart from the pair  $R_M(q)$  -  $G_M(\varphi)$  analytically is generally impossible, as shown in [11].

The overview of the memristor types in Fig. 2 indicates one of the possible reasons why the Bernoulli dynamics is still mistakenly considered a feature of every ideal memristor. This dynamics is mentioned if it governs at least one quantity of the pair of variables voltage - current [7], [8]. It is apparent from the matrix in Fig. 2 that if the memristor voltage (current) exhibits the linear, Bernoulli, or Riccati dynamics, it can lead to the Bernoulli dynamics with respect to the current (voltage). The current (voltage) cannot exhibit the Bernoulli dynamics only if the voltage (current) is governed by the Abel differential equation of the first kind. As a result, three of the four types of dynamics with respect to the current (voltage) can lead to the Bernoulli dynamics. The Bernoulli dynamics is therefore likely to be abundantly represented in textbook examples. For instance, there is a mention in [17] of a Bernoulli memristor with  $R_M(q) = \alpha q^2$ . According to Tab. 1, row 5, it is really of the  $({}_{v}\mathbf{B}_{i}\mathbf{B})$  type, thus the memristor exhibits the Bernoulli dynamics even with respect to voltage and current simultaneously. But a "slight" change of the PSM to  $R_M(q) = 1 + \alpha q^2$  would cause that the memristor would not only cease to be "Bernoulli" but its dynamics would put it completely outside the matrix of dynamics in Fig. 2.

As is obvious from Tab. 2, only memristors with  $R_M(q)$  of the  $R(q-q_0)^N$  or  $R(a+\exp(q-q_0))^N$  types or with  $G_M(\varphi)$  of the  $G(\varphi-\varphi_0)^N$  or  $G(a+\exp(\varphi-\varphi_0))^N$  types,  $N \neq 1$ ,  $a \neq 0$ , are governed by the  ${}_{\nu}\mathbf{B}/{}_{l}\mathbf{B}$  Bernoulli dynamics with respect to voltage/current. Coincidentally, according to the same table, the HP memristor from [13] with the PSM of the  $R_M(q) = R(q-q_0)$  type also belongs to the class  ${}_{l}\mathbf{B}$ . The belief that ideal memristors must obey the Bernoulli dynamics is nothing but a chimera, which is buried symbolically by the example of one of the most discussed types of memristor: The HP memristor with the Joglekar window function [13] for p=1 is a memristor of the  $({}_{\nu}\mathbf{R}_{l}\mathbf{A})$  type, thus neither its voltage nor its current exhibits the Bernoulli dynamics.

## Acknowledgments

This paper originated as part of the activities of the COST Action IC1401 "Memristors – Devices, Models, Circuits, Systems and Applications (MemoCIS)". This work has been supported by the Czech Science Foundation under grant No 14-19865S – Generalized higher-order elements. The research described was performed in laboratories supported by the SIX project; registration number CZ.1.05/2.1.00/03.0072, the operational program Research and Development for Innovation.

The research was also supported by the Project for the development of K217 Department, UD Brno.

#### References

- [1] CHUA, L. O. Memristor The missing circuit element. *IEEE Transactions on Circuit Theory*, 1971, vol. CT-18, no. 5, p. 507 to 519. DOI: 10.1109/TCT.1971.1083337
- [2] CHUA, L. O., KANG, S-M. Memristive devices and systems. Proc. of the IEEE, 1976, vol. 64, no. 2, p. 209–223. DOI: 10.1109/PROC.1976.10092
- [3] CHUA, L. O. If it's pinched it's a memristor. Semiconductor Science and Technology, 2014, vol. 29, p. 104001. DOI: 10.1088/0268-1242/29/10/104001
- [4] VAN DER SCHAFT, A. J. Representing a nonlinear state space system as a set of higher-order differential equations in the inputs and outputs. *Systems and Control Letters*, 1989, vol. 12, no. 2, p. 151–160. ISSN: 0167-6911
- [5] BIOLEK, D., DI VENTRA, M., PERSHIN, Y. V. Reliable SPICE simulations of memristors, memcapacitors and meminductors. *Radioengineering*, 2013, vol. 22, no. 4, p. 945–968. ISSN: 1805-9600
- [6] BIOLEK, D., BIOLEK, Z., BIOLKOVÁ, V., KOLKA, Z. Reliable modeling of ideal generic memristors using state variable transformation. *Radioengineering*, 2015, vol. 24, no. 2, p. 393–407. DOI: 10.13164/re.2015.00393
- [7] DRAKAKIS, E. M., YALIRAKI, S. N., BARAHONA, M. Memristors and Bernoulli dynamics. In *Proceedings of the 12th International Workshop on Cellular Nanoscale Networks and Their Applications (CNNA)*. Berkeley (CA, USA), 2010, p. 1–6. DOI: 10.1109/CNNA.2010.5430324
- [8] GEORGIOU, P. S., BARAHONA, M., YALIRAKI, S. N., DRAKAKIS, E. M. On memristor ideality and reciprocity. *Microelectronics Journal*, 2014, vol. 45, no. 11, p. 1363–1371. DOI: 10.1016/j.mejo.2014.08.003
- [9] MAZUMDER, P., KANG, S. M., WASER, R. Memristors: Devices, Models, and Applications [Scanning the Issue]. *Proc. of the IEEE*, 2012, vol. 100, no. 6, p. 1911–1919. DOI: 10.1109/JPROC.2012.2190812
- [10] CAI, W., ELLINGER, F., TETZLAFF, R., SCHMIDT, T. Abel dynamics of titanium dioxide memristor based on nonlinear ionic drift model. *Arxiv* 1105:2668, 2011, p. 1–5.
- [11] BIOLEK, Z., BIOLEK, D., BIOLKOVÁ, V. Analytical solution of circuits employing voltage- and current-excited memristors. *IEEE Transactions on Circuits and Systems-I: Regular Papers*, 2012, vol. 59, no. 11, p. 2619–2628. DOI: 10.1109/TCSI.2012.2189058
- [12] STRUKOV, D. B., SNIDER, G. S., STEWART, D. R., WILLIAMS, R. S. The missing memristor found. *Nature*, 2008, vol. 453, p. 80–83. DOI: 10.1038/nature06932
- [13] JOGLEKAR, Y. N., WOLF, S. J. The elusive memristor: properties of basic electrical circuits. *European Journal of Physics*, 2009, vol. 30, no. 4, p. 661–675. DOI: 10.1088/0143-0807/30/4/001
- [14] BIOLEK, Z., BIOLEK, D., BIOLKOVÁ, V. Spice model of memristor with nonlinear dopant drift. *Radioengineering*, 2009, vol. 18, no. 2, p. 210–214. ISSN: 1805-9600
- [15] POLYANIN, A. D., ZAITSEV, V. F. Handbook of Exact Solutions for Ordinary Differential Equations. 2nd ed. Chapman & Hall/CRC Press, 2002. ISBN: 9781584882978
- [16] GOLUBEV, V. V. Lectures on Differential Equations in the Complex Domain (in German: Vorlesungen über Differential-

- gleichungen in Komplexen). VEB Deutscher Verlag der Wissenschaften, 1958. ASIN: B0000BIOOA
- [17] GEORGIOU, P. S. A mathematical framework for the analysis and modelling of memristor nanodevices. *PhD Thesis*, 2013, Imperial College London, p. 212.

#### **About the Authors ...**

Zdeněk BIOLEK was born in Ostrava, Czech Republic, in 1959. He received the PhD degree in Electronics and Informatics from Brno University of Technology, Czech Republic, in 2001. He is currently with the Dept. of Microelectronics, Brno University of Technology (BUT), Czech Republic. Until the year 1993 he worked as independent researcher in semiconductor company TESLA Rožnov. He is the author of unique electronic instruments associated with IC production and testing. He is also the author of several papers from the area of the utilization of variational principles in theoretical electrical engineering and stability testing of resistive circuits, and also from the field of memristors and mem-systems. Dr Z. Biolek is also the co-author of two books about memristive systems and PSpice modeling and simulation of special electronic circuits including switched-capacitor filters, switched DC-DC converters, and memristors.

Dalibor BIOLEK (M'99) received the MSc degree in Electrical Engineering from Brno University of Technology, Czech Republic, in 1983, and the PhD degree in Electronics from the Military Academy Brno, Czech Republic, in 1989, focusing on algorithms of the symbolic and numerical computer analyses of electronic circuits with a view to the linear continuous-time and switched filters. He is currently with the Dept. of EE, University of Defense Brno (UDB), and with the Dept. of Microelectronics, Brno University of Technology (BUT), Czech Republic. His scientific activity is directed to the areas of general circuit theory, frequency filters, mem-systems, and computer simulation of electronic systems. He has published over 400 papers and is author of several books on circuit analysis and simulation. At present, he is professor at BUT and UDB in the field of Theoretical Electrical Engineering. Prof. Biolek is a member of the CAS/COM Czech National Group of IEEE. He also serves as an Area Editor of the International Journal of Electronics and Communications (AEU).

Viera BIOLKOVÁ was born in Partizánske, Slovakia. She received her MSc degree in Electrical Engineering from Brno University of Technology, Czech Republic, in 1983. She joined the Dept. of Radio Electronics in 1985, and is currently working as a Research Assistant at the Dept. of Radio Electronics, Brno University of Technology (BUT), Czech Republic. Her research and educational interests include signal theory, analog signal processing, and digital electronics.