



Electric Vehicles Charging Technology Review and Optimal Size Estimation

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Abstract

Many different types of electric vehicle (EV) charging technologies are described in literature and implemented in practical applications. This paper presents an overview of the existing and proposed EV charging technologies in terms of converter topologies, power levels, power flow directions and charging control strategies. An overview of the main charging methods is presented as well, particularly the goal is to highlight an effective and fast charging technique for lithium ions batteries concerning prolonging cell cycle life and retaining high charging efficiency. Once presented the main important aspects of charging technologies and strategies, in the last part of this paper, through the use of genetic algorithm, the optimal size of the charging systems is estimated and, on the base of a sensitive analysis, the possible future trends in this field are finally valued.

Keywords Electric vehicle · Plug-in electric vehicles · Battery charger · Charging infrastructure · Vehicle-to-grid · Grid-to-vehicle · Charging methods

1 Introduction

Growing concern of carbon dioxide emissions, greenhouse effects and rapid depletion of fossil fuels raise the necessity to produce and adopt new eco-friendly sustainable alternatives to the internal combustion engine (ICE) driven vehicles. For this reason, in the last decade, EVs have become in some way widespread, principally because of their negligible flue gas emissions and lesser reliance on oil. It is estimated that by 2022, EVs will be over 35 million in the World. However, a critical problem associated with EVs is that their high penetration causes significant issues on the power distribution grid such as: power quality deterioration, enhanced damaged of line, downturn of distribution transformers, increased distortion and higher fault current [1, 2].

One efficient approach to relieve the effect is to integrate local power generation such as renewable energy sources (RESs) into the EV charging infrastructure [3–5].

Batteries can be charged through conductive or inductive methods [6]. Inductive charger are wireless charging systems (WCS). WCS can be stationary, which means that they can only be utilized when the car is parked or in stationary modes, such as in car parks, garages, or at traffic signals, or they can be dynamic. This latter method allows battery charging while the vehicle is in motion. In general, WCS can bring some advantages in the form of aesthetic quality, reliability, durability and user friendliness. Anyway, due to some challenges such as electromagnetic compatibility (EMC) issues, limited power transfer, bulky and expensive structures, shorter range, and lower efficiency [7, 8], inductive chargers are not largely commercialized and employed as the conductive ones. In this paper only conductive charging strategies are analyzed.

Based on their power ratings EV battery chargers can be divided into level 1, level 2 and level 3. Table 1 summarize the characteristics of the three different power levels [9, 10].

A battery charger can allow a unidirectional or bidirectional power flow at all power levels. The bidirectional power flow adds to the grid-to-vehicle interaction (G2V) also the vehicle-to-grid (V2G) mode [11]. This latter technology can bring significant improvement in the overall reliability of the

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Table 1 Charging power levels

Power level	Charger location	Typical use	Typical power	Charging time	Connector
Level 1	On-board	Home	2 kW	4–11 h	SAE J1772
Level 2	On-board	Public	20 kW	1–4 h	SAE J1772
Level 3	Off-board	DC Fast	100 kW	< 30 min	CHAdeMO/ CCS COMBO 2

distribution grid, since in case of system failure, peak load demand or other unexpected scenarios, with a bidirectional power flow, the EVs can be used as back up generation, supplying the energy back to the grid when needed [12]. With V2G, as all the energy storage systems, EVs battery can be used not only as back up resource but also to improve the power quality, the stability and the operating cost of distribution network. Moreover, in the long run, V2G could reduce investment in new power generation infrastructure [13–16]. All the just listed reasons are increasing researchers' interest in this technology.

Not only the choice of the charging technology, but also the selection of the correct charging method is a feature that has to be considered during the charging procedure. The most popular charging strategies to recharge Li-ion batteries are constant-current/constant-voltage (CC/CV) and pulse current charging methods [17, 18]. However these methods do not take into account the several internal process of the battery which influence its charging capability and aging. As a result, some promising charging strategies which are based on more complete models of lithium-ion batteries are now under research [19].

EV battery and power electronic devices costs are steadily falling [20, 21]. This rapid decline is mostly due to a growing manufacturing industry, which is constantly increasing the knowledge, the number of applications and the improvements of both these technologies. As the costs are falling, the trends of EV battery energy density, gross weight and semiconductor devices performances are following exactly the opposite direction; in fact, with equal capacities, batteries are becoming ever smaller and lighter and the power electronic devices ever more performing [22]. All this impacts the choice and the size of the charging systems. The sizing procedure of a suitable charging system is made even more difficult by the presence of many different technologies of the onboard and off-board chargers and also different cost, dimensions, weights, power rating; and so on.

This the paper is organized as follow. In Sects. 2 and 3, respectively, on-board and off-board charger most common architectures are presented and their operating principles are explained. In Sect. 4, the concept of fast charging stations is introduced. The available and most suitable charging methods are listed in Sect. 5 with particular attention to the charging methods most suitable for direct-current (DC) fast

charging. Finally, a genetic algorithm is used in Sect. 6 to estimate the optimal charging system size and its possible future trends.

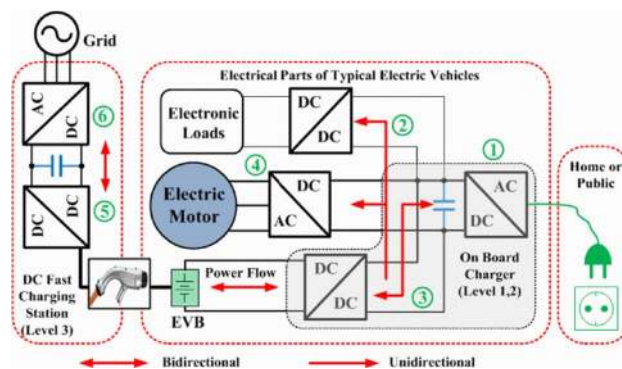
2 Onboard Charger

Battery chargers can be implemented inside (on-board) or outside (off-board) the vehicle. Onboard battery chargers (OBC) are limited by size, weight and volume [23] for this reason they are usually compatible with level 1 and level 2 chargers. They usually have unidirectional power transfer capability; nevertheless in some case the configuration, a bidirectional power transfer can be achieved. Figure 1 shows the typical architecture of an electric vehicle charging system, in such figure both the on-board charger and the off-board one are represented.

2.1 Two Stage

Onboard chargers are typically composed by two stages: a front-end AC–DC stage and a back-end DC–DC stage. Very different topologies are proposed in literature for both the converters.

The front-end rectifier usually contains a boost power factor correction (PFC) converter to achieve high power factor and low harmonic distortion. The rectifier stage can be performed by a half-bridge, full-bridge or multilevel diode bridge. Half-bridge rectifier is less expensive since it

**Fig. 1** Charging system configuration for electric vehicle

contains less number of diodes/switches, full-bridge rectifier is more complex but the components are subjected to lower stresses. If instead higher power ratings should be achieved a good choice for the ac–dc converter is a multilevel configuration. Figure 2 shows a full-bridge diode rectifier with a conventional PFC boost converter [24, 25]. By substituting all the diodes with active switches, a bidirectional power flow can be obtained.

In PFC converter the interleaved boost converter is becoming more and more popular. As shown in Fig. 3, an interleaved boost converter simply consists in two boost converters in parallel, operating 180° out of phase [26]. The main aim of this interleaving is to increase the output current by reducing the input current ripple and hence by reducing the overall volume of the input ElectroMagnetic Interference (EMI) filter and of the boost inductor [23, 27–29]. On the other hand, interleaving means increasing the cost and the complexity of the design.

The front-end converter is followed by a second dc–dc converter. Resonant power converters are very common to perform this second stage, because of the potential to achieve at the same time both higher switching frequency and lower switching losses [29]. Among all the resonant converter, LLC configuration, shown in Fig. 4, is achieving resounding interest, thanks to its several advantages over other resonant topologies, such as: (1) the ability to operate at zero-voltage switching (ZVS) or zero-current switching (ZCS), (2) containing a high frequency transformer it performs the galvanic isolation between the grid and the EV, (3) a wide output voltage regulation is possible, (4) the output filter consists only in a capacitor and not in an LC filter [29–31].

However in literature other types of dc–dc converter have been proposed. In [32] a bidirectional buck-boost non-isolated dc/dc converter is proposed whose principal aim is to allow a reduction of the dc-link capacitor. In [33] is proposed for the dc–dc stage a cascade structure of a high-frequency LLC converter followed by a buck converter. The advantage of this configuration consists in a constant switching frequency operation of the active switches of the LLC

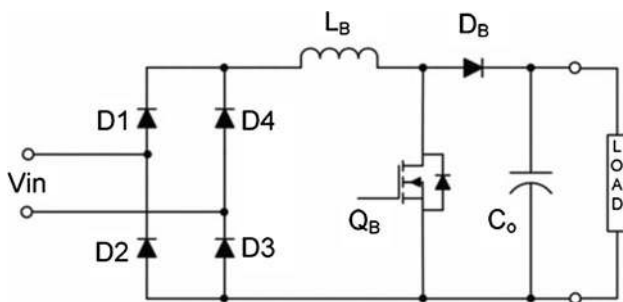


Fig. 2 Full-bridge rectifier with conventional PFC boost converter

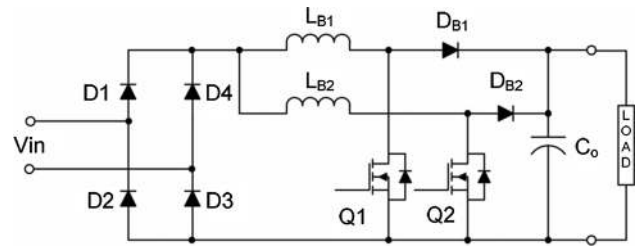


Fig. 3 Full-bridge rectifier with interleaved PFC boost converter

converter while the charge control is performed by the buck converter. On the other and by adding an additional stage to the dc–dc converter, the complexity and the cost of the converter increases. Finally, in [34] a dc–dc non-isolated buck converter is employed. This latter topology is easy to implement but it works only if the dc-link voltage is higher than the battery pack voltage.

2.2 Single Stage

If the ac–dc rectifier is combined with the dc–dc converter, a single stage battery charger is obtained. This topology of battery charger is used if lower cost and size are required [23, 34], in fact single stage battery charger allows the elimination of some bulky and expensive components such as inductors and dc-link capacitors [35, 36] which instead are required in two-stage charger. However, the drawback is that single stage battery chargers with non-isolated converter suffer from a limited conversion ratio, which limits their application for the wide range of output voltage. If instead a high frequency isolation is present, as in the OCB configuration proposed in [35], the low frequency component generated by the rectification stage pass through the high frequency transformer leading to large magnetizing current. Moreover to achieve power factor correction, a large number of diodes and active switches could be necessary [36], increasing in this way the complexity of the configuration and hence decreasing the reliability of the overall charger.

2.3 Integrated

To maximize the reduction of components number and hence to further reduce the size, weight and cost of the

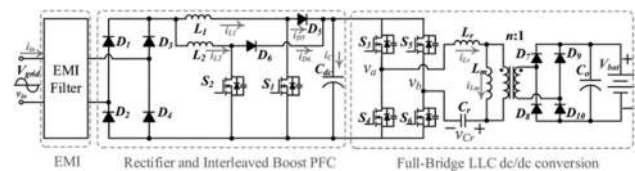


Fig. 4 Example of two-stage onboard charger with LLC converter

battery charger several integrated topologies have been proposed and studied in literature [23, 37–40]. The concept of integration consists of reusing some of the drivetrain components (inverter and motor windings) to implement the onboard charging system. However, some problems may be born from this combination: the configurations proposed in [37, 38] required access to inaccessible points of the motor windings, in [39] a rearrangement of the motor windings is necessary during the transition between different operation modes. Finally, by using the charger configuration proposed in [40], even if neither access to the neutral point of the motor windings nor their rearrangement are required, the control of the active switches becomes more difficult.

2.4 Multifunctional

The last type of proposed OBC are the so called multifunctional OBCs. In this type of battery charger some components are shared to accomplish different aims. In this way higher fuel efficiency can be reached by smaller and lighter design. In [41], the proposed multifunctional battery charger can charge the auxiliary battery via the propulsion battery when the vehicle is in a driving state, acting in this way as an OBC and as low-voltage dc-to-dc converter (LDC) jointly. In [42], a similar configuration, shown in Fig. 5, with the same duties is presented.

3 Off-Board Charger

Level 3 charger, because of their rating powers, are usually installed outside the vehicle (off-board). Also for level 3 off-board charger a large amount of different solutions is studied in literature [43–62]. Since it is mandatory to guarantee galvanic isolation between the AC supply circuit and the DC output circuit according to the IEC EN 61,851–23 standard, in this paragraph only isolated off-board charger have been presented.

The off-board charging system is most commonly composed of two stages: a grid-facing AC/DC converter followed by a DC/DC converter providing an interface to EV

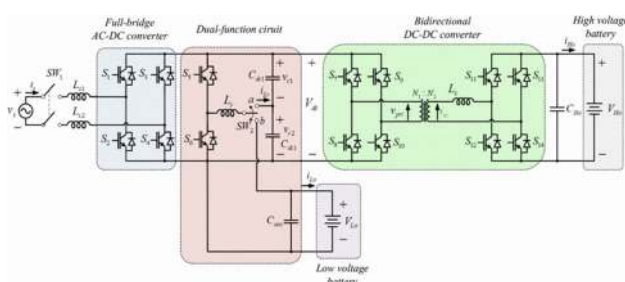


Fig. 5 Multifunctional OBC proposed in [42]

battery. Based on the converter topology, both these stages can allow unidirectional or bidirectional power flow.

3.1 Bidirectional AC/DC Converter

One of the most widely used bidirectional AC/DC converter is the three-phase LCL active rectifier [43–45], whose scheme is reported in Fig. 6.

The advantages of this type of converter are low harmonic input currents, bidirectional power flow and power factor (PF) regulation.

In [46] and [47] the front end ac–dc conversion is performed by a neutral-point-clamped (NPC) three-phase three-level converter. This converter has been used to increase the power density and to achieve low current harmonics distortion. Another advantage is that it allows the creation of a bipolar dc bus which can be used for the implementation of partial-power converters. However, the NPC causes imbalance of power and, as a result, voltage balancing problem across the DC bus capacitors.

3.2 Unidirectional AC/DC Converter

The most common unidirectional AC/DC converter used in off-board charging system is the Vienna rectifier [48–50]. It has advantages such as low voltage stress on each switch and high efficiency. However, the main limitations are the restricted reactive power control and the need of a dc-link capacitor voltage balancing. In [48] the authors propose a 25 kW off-board charger prototype composed by a single-switch Vienna rectifier, as depicted in Fig. 7, and four three level dc/dc modules parallel connected.

3.3 Bidirectional DC/DC Converter

The main isolated DC/DC converter used in case of bidirectional power flow is the dual active bridge (DAB) [51–54], reported in Fig. 8, and its variants (resonant DAB [53], multilevel DAB [54]). In particular, this topology is gaining

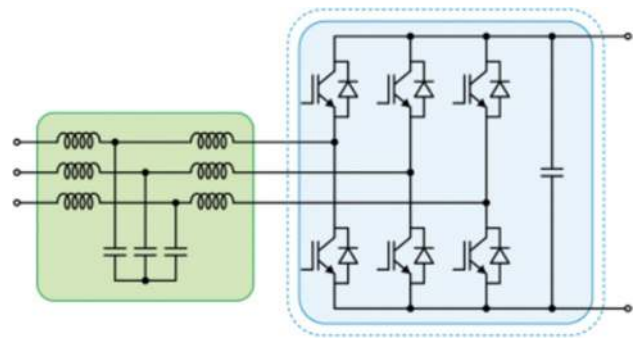


Fig. 6 Three-phase LCL active rectifier [44]

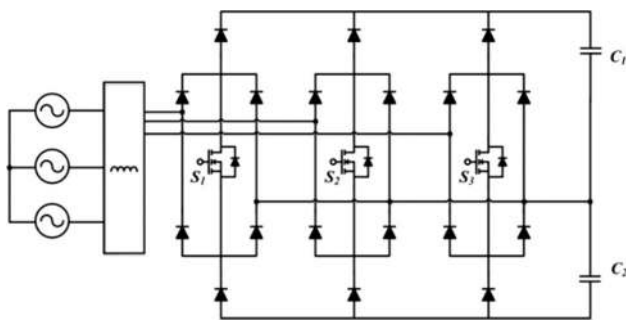


Fig. 7 Vienna rectifier scheme proposed in [48]

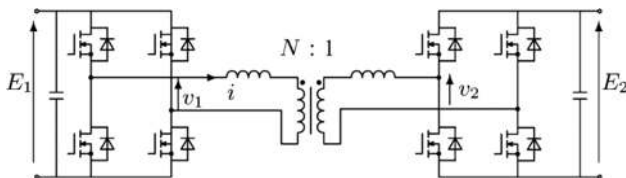


Fig. 8 Dual active bridge scheme

interest thanks to the capabilities of the new wide-bandgap semiconductor (Gan/SiC) devices which enabled the converter efficiency and power density improvements [55].

3.4 Unidirectional DC/DC Converter

If unidirectional power flow is required, a LLC resonant converter is chosen, in most cases, as power interface between the dc bus of the AC/DC converter and the EV battery because of its advantages over other resonant topologies [56, 57], such as: the ability to operate at zero-voltage switching (ZVS) or zero-current switching (ZCS), it allows a wide output voltage regulation, the output filter consists only of a capacitor and not of an inductor and capacitor (LC) filter [58].

Another unidirectional DC/DC converter used in case of unidirectional off-board charger is the phase shifted full bridge converter [59, 60], whose scheme is reported in Fig. 9. This type of converter has different advantages such as high power density, low magnetic interference and high efficiency which make it well suitable for the implementation in battery chargers [61].

4 Fast Charging Stations

To reduce the driving range anxiety and hence to support a stronger increase of the penetration of EVs worldwide there is the need of a charging system which is able to replace the current existing oil station. A fast charging station (FCS) can allow the charging of an EV at 80% within a half of

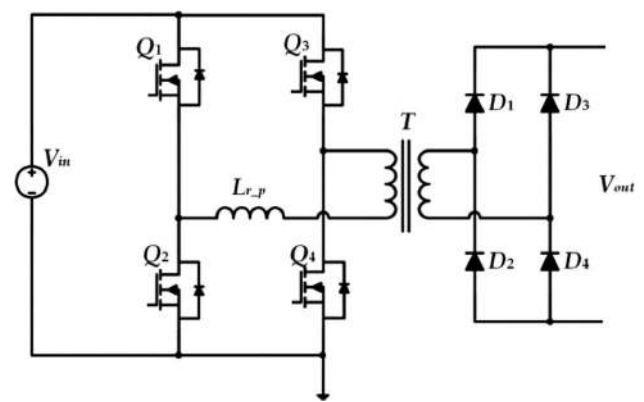


Fig. 9 Phase shift full bridge converter

hour from its depletion, but to reduce the charging time from 7–8 h to 30 min, FCS requires high power from the grid and for this reason they are usually connected to the MV network [63–65], even if some FCS connected to the LV grid are proposed too [66]. The connection of such charging stations requires a huge capital investment and it could easily overload the distribution network. Another critical aspect to be considered consists in the voltage drop that the connection of FCS can cause along the lines of the distribution networks, which according to the standard EN50160 has to remain lower than 10%.

According to [67] the impact of fast charging stations on distribution MV grid can be mitigated with the use of energy storage systems (ESSs) which can shave peak power demand and provide additional network services. Moreover, ESS can also increase the voltage level in case of too high voltage drop along the lines, this service requires the implementation of a voltage control.

In order to further minimize the impact of the FCS on the grid, renewable energy resources can be integrated within the FCS too [68]. In fact, in normal operation, during day-time, the EV batteries can be charged from the solar PV by reducing in this way the possibility of overloading the MV network. In night-time, instead, when solar energy is not available the EV batteries can be charged from the grid. EVs also can support to the grid at the peak load demand if needed. By this way, the grid will never become unstable with a high pulse power of charging from EVs.

4.1 Architecture

The configuration chosen in [63] for the ultra-fast charging station is shown in Fig. 10. The isolation between the AC side and the DC side is performed by the line-frequency transformer. The ac–dc stage is common to all the EV, in fact the output of the Cascade H Bridge (CHB) multilevel converter creates a unipolar common DC voltage bus at

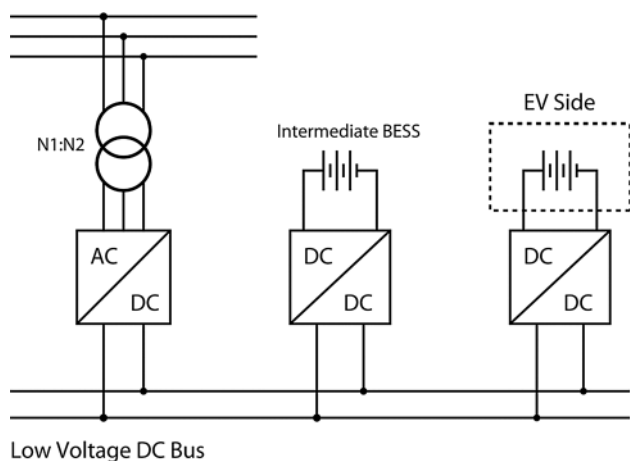


Fig. 10 Fast charging station with unipolar dc voltage bus proposed in [63]

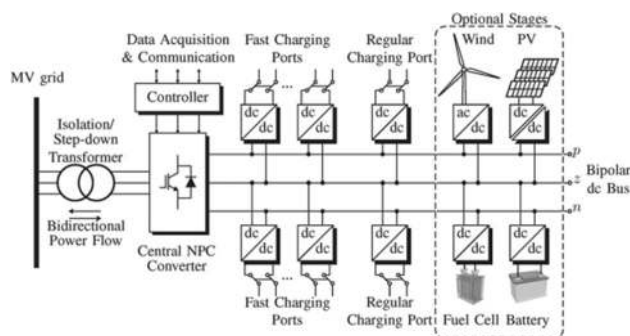


Fig. 11 Fast charging station with bipolar dc voltage bus proposed in [64]

which all dc–dc converters are connected. The power interface between the dc bus and the integrated storage or the EV battery is performed by a LLC resonant converter.

In [64] a bipolar dc bus architecture is proposed. The charging station here adopted is shown in Fig. 11. The bipolar dc bus is necessary to feed the three-level dc–dc converter. On one side, this architecture increase the capability of the FC station and reduce the THD; on the other hand, this system requires a DC power balance management mechanism [65].

4.2 Power Control Strategies

A number of studies have been carried out also on the control strategy of EV charging station [69–72]. The first difference among them concerns the choice of uncoordinated charging or coordinated charging. In uncoordinated charging scheme the EV battery starts charging immediately when it is plugged in or after a fixed start delay chosen by the user. If the EVs are charged according to this scheme, their

impact on the grid will result in very high peak demand and hence in huge grid issues such as: feeders and transformers overloading, high losses, high voltage drops, and more cost [69]. For all the aforementioned reasons, the research has focused on EV coordinated charging strategies, in particular. Authors in [70] propose a coordinated control strategy for an FCS with an ESS, their principal aim is that of reducing the electricity purchase cost and flatten the peak load. In [71] the FCS contains both ESSs and RESs, the coordinated energy management proposed here is obtained by using a Fuzzy Logic based control defined by three inputs and two outputs. Other control strategies based on very different and numerous algorithms have been proposed in literature. Authors in [72] investigate and compare three different control strategies for a charging station. The charging strategies coordinate in different ways the energy flow of the ESS and PV panels. The results prove that the control strategy which uses the energy of the ESS with in response to the number of EVs plugged in and takes into account also the amount of generated renewable energy has the best performance. Furthermore according to this control method the SoC of the ESS is restored during night and during excess of solar energy.

5 Charging Methods

The advantages of a lithium-ion battery over other types of energy storage devices such as high energy and power density, low memory effect and resulting capacity loss, make this type of battery the best candidate for the field of electric vehicles [73]. However, li-ion battery charging must be carried out very carefully, since the charging method greatly affects how actively electrochemical side reactions occur inside the battery, and hence the cycle life of the battery itself [74]. For this reason, finding the optimal technique to charge a battery in the shortest period of time with high efficiency and without damaging the cells, has become a new challenge for many researchers [75].

In this paragraph the main charging methods for Li-ion batteries are discussed.

5.1 Constant Current-Constant Voltage (CC–CV)

In this method, represented in Fig. 12, both an initial constant current and a final constant voltage are used. The charging process start with a constant current until a certain voltage value, known as cut-off voltage, is reached. For Li-ion with the traditional cathode materials of cobalt, nickel, manganese and aluminium typically the cut-off voltage value is around 4.20 V/cell. The tolerance is ± 50 mV/cell. Battery charging continues with a constant voltage just equal to the cut-off value. Full charge is reached when the

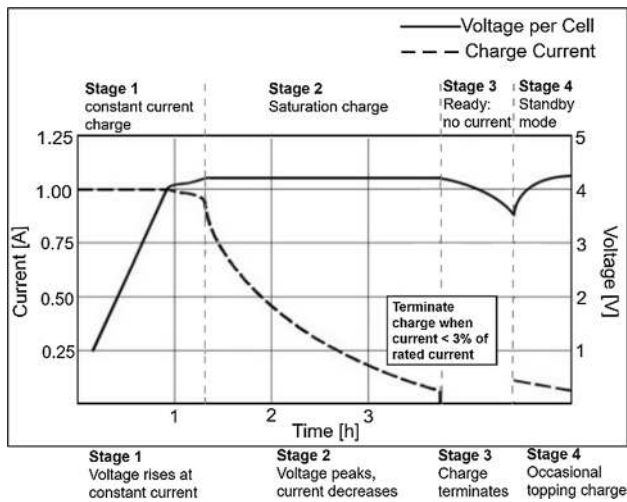


Fig. 12 CC–CV charge stages for a Li-ion battery

current decreases to between 3 and 5% of the rated current. Trickle or float charge at full charge is not suitable for a Li-ion battery; since it would cause plating of metallic lithium and compromise safety. Instead of trickle charge, a topping charge can be applied when the voltage drops below a set value.

In [77], a little bit different variant is proposed, in fact the first stage consists now in a trickle charge. This stage is activated only if the battery is deeply discharged, i.e. the cell voltage is below 3.0 V and after it the CC–CV method keeps place with the aforementioned way.

5.2 Five-Step Charging Pattern

An alternative method is here described in order to obtain faster and safer charging and longer battery cycle life [77, 78]. The five-step charging pattern consists in a multistage (five stages) constant-current charging method, in which the charging time is divided into five steps. In each stage, the charging current is set to a constant threshold value. During charging, the voltage of the battery will increase and when it reaches the pre-set limit voltage, the stage number will increase and a new charging current set value will be applied accordingly. This process will continue until the stage number reaches 5. Figure 13 illustrates the concept of the five-step constant current charging pattern.

To find the charging current in each steps different algorithms can be used, however it could be difficult and time ineffective find the optimal charging pattern.

5.3 Pulse Charging Method

With this charging strategy the charging current is injected into the battery in form of pulses, so that a rest period is

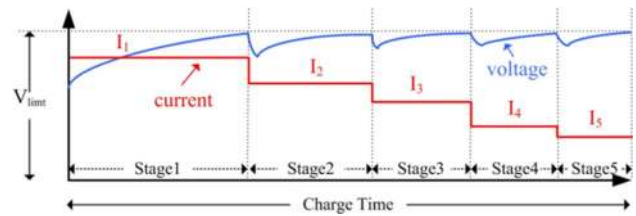


Fig. 13 Five-step charging pattern for a Li-ion battery

provided for the ions to diffuse and neutralize. The charging rate, which depends on the average current, can be controlled by varying the width of the pulses. It is claimed [77, 79] that this method can really speed up the charging process, slow down the polarization effect and increase life cycles. As shown in Fig. 14, every pulse charge current that is applied to the battery is characterized by the following factors: peak amplitude I_{pk} , a duty cycle $D = t_{on}/T_p$, and frequency f .

Two different pulse charging methods exist: duty-fixed and duty-varied pulse-charge strategy. According to [79], the duty-varied strategy can increase the charging speed and the charging efficiency with respect to the conventional duty-fixed method.

5.4 Charging Strategies Based on Battery Physical-Based Models

In all the charging methods reported so far, only current and voltage limits are considered. Anyway, these limits do not take into consideration the aging process and the side reactions occurring inside the battery; and hence they might result too conservative for new batteries and possibly dangerous for aged batteries due to the altered behavior and characteristics. This, hence, motivates the development and the research of innovative charging algorithms which tackle the issue of the charging impact on battery state-of-health (SoH) and aging [76, 80]. In fact, recently, many researches are focusing on develop new charging methods which minimize the charging time and extend the battery life at the same time [75, 80–83].

This new category of charging strategies employs the electrochemical lithium-ion battery models to calculate quantitatively and almost precisely the amount of battery aging and to directly minimize the aging in a given charge

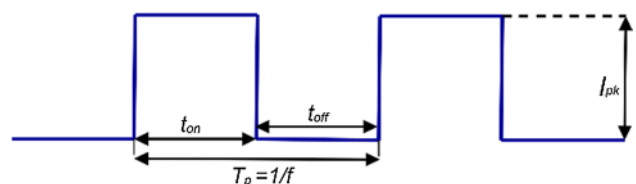


Fig. 14 Pulse charge current parameters

time. The electrochemical battery model estimates the internal states of a battery in a more accurate way since it is built starting from the internal microstructure of the lithium-ion battery. However, it is very arduous to precisely estimate the parameters because the electrochemical model is composed of complicated coupled partial differential equations (PDEs), and involving a large number of parameters and boundary conditions it requires a large computation burden [82]. Therefore the complexity of these models often leads to the necessity for more memory and computational effort and thus they may not be practically implemented in the fast and real-time computations of EV BMS [86].

In recent years, many works focus on identifying simplified internal electrochemical models of the battery for efficiently determining the battery operation and hence finding the optimal charging profile.

For example, in [75] Klein et al. aim at achieving a fast charging rate while not excessively aging the cell, using non-linear model predictive control (NMPC) techniques founded on a single particle model (SPM) of the lithium ion battery. In the single particle model (SPM) the electrodes are represented by two single spherical porous particles in which intercalation and de-intercalation phenomena take place. However, the variations in the electrolyte concentration and in the potential are ignored. This electrochemical model is exhaustively explained in [83, 84].

In [81], a method to optimize the charging current for Li-ion batteries using a soft actor-critic (SAC) reinforcement learning algorithm is proposed. As battery aging model Kim et al. use the single particle model augmented by electrolyte and thermal dynamics (SPMeT). Anyway, even if this model is more exhaustive than the SPM one, it is subject to a complicated mathematical structure including PDEs, ordinary-differential equations and algebraic equations. Therefore the corresponding model based algorithms can be computationally too arduous to be implemented in typical battery management systems (BMSs) [80]. In light of this, in [80], Zou et al. propose and validate a simplified ODE-based SPMeT model to achieve fast charging control. Once validated the reduced model, the authors present a charging strategy based on model predictive control (MPC) techniques which aims to charge the battery in the fastest possible manner, without excessively degrading the battery's SOH.

Another widely employed electrochemical battery model is the pseudo two-dimensional model (P2D). This model is constructed based on the assumption that electrodes are seen as an aggregation of spherical particles (2D representation) in which the Li⁺ ions are inserted. The first spatial dimension of this model, represented by variable x , is the horizontal axis. The second spatial dimension is the particle radius r . The cell is comprised of three regions that imply four distinct boundaries. The full description of this model can be found in [85]. Finally, in [82], the authors present

an effective method to estimate the parameters of the P2D model using a neural network-based estimation scheme.

6 Genetic Algorithm for Optimal Charging Technology Size and Forecasting of Future Trends

Concerning the onboard charger, as seen in paragraph II, different technologies exist in literature and in industrial applications. The cost, dimension and weight of this product is strongly influenced by its power rating; nowadays many sizes are proposed such as: 0.75, 3.3, 6.6, 11 and 22 kW. Many different power ratings are described also for off-board charger and consequently the cost varies as a function of the power and quality.

Generally, the wide range of power ratings, the fast improvements and the lack of standardization in e-mobility technologies make quite difficult estimating the actual optimal power rate of charging technology and forecasting future tendencies for large interval of time.

In this paragraph, initially, using a genetic algorithm (GA) single-objective model, the best size of both the onboard and off-board charging technology is estimated; then, through a sensitive analysis, we try to evaluate the possible future trends in the e-mobility as the costs of the battery and of the charging technology vary.

GA is an heuristic search algorithm, which mimics the natural selection process and it is used to solve optimization problems. GAs are being used to solve a wide variety of optimization problems in the field of EVs such as: charging scheduling [87–89], charging station planning and location [90, 91] and drive train control strategy implementation [92–93].

6.1 Objective Function

For managing the EV charging technology, a single-objective optimization is used to determine the optimal size of the charging technology both on-board and off-board and to determine a suitable battery capacity. The proposed optimization allows to find the optimal trade-off between the onboard and off-board charger power rate.

The aim of the sizing procedure is to minimize the overall costs coming from the charging of electric vehicles. Therefore, the optimization problem is modeled in the expressions (1), (2) and (3).

First of all, the objective function representing the cost of the charging system is computed in (1)

$$\begin{aligned}
 f_c(C_{bat}, P_{on}, P_{off}) = & k_{bat} \cdot C_{bat}^{1.1} + k_{on} \cdot [\max((P_{on} - 6.5), 0)]^{1.2} \\
 & + k_{off} \cdot P_{off}^{1.2} - \{\beta_{on} \cdot \log[\max((P_{on} - 6.5), 0)] \\
 & + \beta_{bat-off} \cdot \sqrt{P_{off}} \cdot \log C_{bat}\}
 \end{aligned}
 \tag{1}$$

The coefficients k_i represent the cost of the i -th component allocated to the vehicle, while coefficients β_i embody the benefits. The coefficient k_{bat} represents the cost of the battery per kWh, more precisely it also includes an amount which consider the costs arising from the space occupied by the battery inside the vehicle, its positioning and all the other aspects which are influenced by the battery capacity. Therefore, the first factor of the function describes the cost trend as a function of the battery size. This behaviour is not perfectly linear since we assume the fact that as the battery capacity increases the payload available inside the vehicle decreases. Coefficient k_{on} includes the costs in € per kW of all the components influenced by the rated power of the onboard charger: the cost of power electronics, cost of space, cost of control and so on. Also for the cost trend of the onboard battery charger a non-linear behaviour has been assumed for the same reasons explained for the battery. In particular for this component, it can be seen that a minimum level of 6.5 kW is taken, since we assume that a smaller battery charger will not be used anymore. The computation of the cost coming from the off-board charger power rating follows the same concept used for the onboard charger, with the difference that a minimum value has not been imposed.

Once the cost of each single technology has been estimated, the next passage consists in weighing its benefits, which will be added in the cost function with a negative sign and two coefficients: β_{on} and $\beta_{bat-off}$, which respectively represent the benefit in terms of €/kW for the on-board battery charger and the benefits in terms of €/(kW kWh). The coefficient related to the capacity of the battery and to the power rating of the off-board charger is expected to assume high values, since it concerns the reduction of the driving range anxiety, which we know to be one of the major cause limiting the penetration of the EV worldwide. Starting from the onboard charger, for this parameter a logarithm behaviour has been assumed, since we suppose that an increasing of the P_{on} specially takes advantage passing from 6.5 kW to values such as 22 kW, an additional increase doesn't weigh so much on the overall system. Finally, the benefits of the off-board charger power rating P_{off} and of the battery capacity C_{bat} are introduced. The advantages brought by these two variable are linked together and considered in the last term of the function. An increase in the capacity of the battery should be compensated by an increase in the rating of the off-board charger, otherwise it is not fully appreciated, and vice versa if the off-board charger becomes more powerful, the capacity of the battery has to limit the stresses to which the battery is exposed.

6.2 Optimal Size of the Charging Technology

As aforementioned, finding the minimum of the function f_c (2) means find those values of C_{bat} , P_{on} and P_{off} which are such as to minimise as much as possible the cost of the overall charging system. In particular, the minimum of the cost function has been performed through the use of genetic algorithms coded in Matlab. The bounds listed in (3) have been used.

$$\min_{f_c} (C_{bat}, P_{on}, P_{off}) \tag{2}$$

$$\begin{cases} 0 < C_{bat} \leq 200 \text{ kWh} \\ 0 < P_{on} < 40 \text{ kW} \\ 0 < P_{off} < 1500 \text{ kW} \end{cases} \tag{3}$$

Figures 15, 16 and 17 given below report the minimum of the function once the value of the coefficients changes. The starting condition considers the actual prices and benefits reported in Table 2. By using these values for the coefficients, a battery capacity of 60 kWh, an off-board charger of 170 kW and an onboard one of 14 kW are obtained from the optimization of the cost function.

6.3 Forecast of Future Trends

In the following second step of the study, the values of coefficients are varied, one at a time, to perform a sensitivity analysis.

It is possible to notice that the capacity of the battery and the power rating of the off-board charger are not influenced by the variation of the cost of the onboard charger per kW.

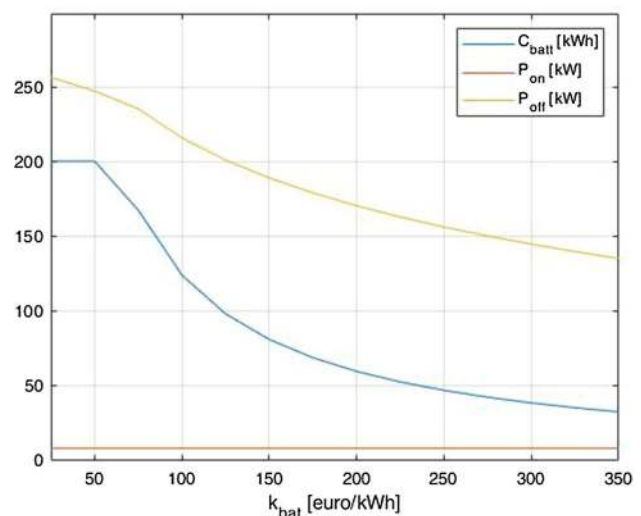


Fig. 15 C_{bat} , P_{on} , C_{bat} , P_{on} , C_{bat} , P_{on} , C_{bat} , P_{on} and P_{off} trends as the coefficient k_{bat} varies

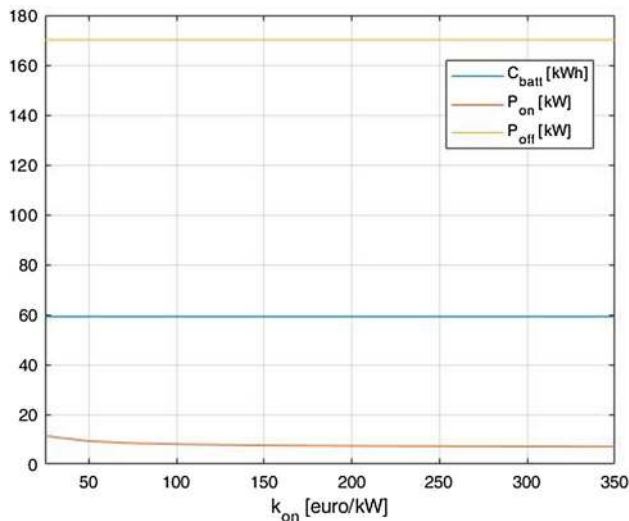


Fig. 16 C_{bat} , P_{on} and P_{off} trends as the coefficient k_{on} varies

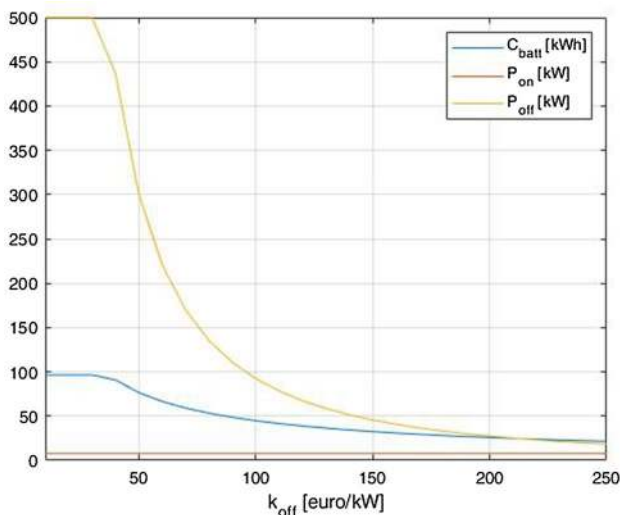


Fig. 17 C_{bat} , P_{on} and P_{off} trends as the coefficient k_{off} varies

Table 2 Starting values of coefficients

Coefficient	Value
k_{bat}	200_{kWh}
k_{on}	150_{kW}
k_{off}	70_{kW}
β_{on}	200_{kW}
$\beta_{bat-off}$	$1500_{\text{kWh kW}}$

The power rating of the onboard charger seems to stay quite constant around values of 14 kW, however if the cost fell below 100 € each kW, P_{on} would start to increase until it reaches 18 kW at 25 €/kW.

Table 3 Wide bandgap materials properties

Properties	Si	SiC-4H	GaN
Band gap (eV)	1.1	3.2	3.4
Critical field (10^6 V/cm)	0.3	3	3.5
Electron Mobility ($\text{cm}^2/\text{V s}$)	1450	900	2000
Electron saturation velocity (10^6 cm/s)	10	22	25
Thermal conductivity ($\text{W}/\text{cm}^2 \text{K}$)	1.5	5	1.3

As expected, the value of the power of the onboard charger does not change as k_{bat} changes since the benefits and the costs of this parameter has not be tied to the cost of the battery technology.

The coefficient k_{off} , represented in Fig. 17, includes all the costs related to the off-board charger weighed from the point of view of the owner. The real cost per kW of this technology is currently around 2500 €, by considering all the aspects such as: power electronics, control, connection to MV grid, connection and communication with the vehicle, installation and maintenance.

Finally, a forecast is done about the semiconductor devices. In this field, two wide bandgap materials are showing great promise for the future: Gallium Nitride (GaN) and Silicon Carbide (SiC). Thanks to all their characteristics, compared in Table 3, wide bandgap semiconductors allow for greater power efficiency, smaller size, lighter weight and lower overall cost. In particular, by comparing SiC and GaN properties, it can be concluded that GaN is the most suitable for low power, high frequency application while SiC material is more appropriate for high power high frequency applications [92].

7 Conclusion

The first part of this paper reviews the current conditions of EV battery charging technologies. The most common topologies which are suitable candidates for each level of an EV charger have been presented. Level 1 and level 2 charger topologies are usually mounted inside the vehicle forming in this way the so called onboard charger. On the other side level 3 chargers are installed off board the vehicles, in this way their collection leads to the creation of the so called FCSs which are promising candidates for future EV high penetration. The most common technologies used in a FCS are multilevel converters which have a high power density and a lower current harmonic distortion. To reduce the impact on the grid, almost all the FCSs are integrated with RESs and ESSs.

Li-ion batteries can be recharged according many different charging techniques which can more or less complicate the charger architecture and control. In particular,

the standard charging strategy are simplest since they don't require model information to charge the battery. Furthermore, they can be realized with very basic circuits, keeping the costs of the charger to a minimum. On the other hand, the charging strategies based on electrochemical models, taking into account the internal dynamics of the battery, consider also the aging of the battery and other constraints, hence resulting in greater accuracy and. All this is at the expense of cost and computational difficulty.

A long-term forecasting of future trends in the field of EV charging systems is a very tough task for different many reasons including the lack of standards and the continuous improvements. In the last section of this paper a possible forecasting and estimation based on GA is described. According to the results, with the actual costs, the capacity of the battery of an electric vehicle should be around 60 kWh, the power rating of the onboard charger around 14 kW and that of the off-board fast charger around 170 kW.

Finally, the silicon switching devices are expected to be replaced by wide bandgap silicon carbide (SiC) devices in order to allow a remarkable reduction in charger's weight and volume.

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