

# Empirical Wavelet Transform

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- Introduction - EMD
- 1D Empirical Wavelets
  - Definition
  - Experiments
- 2D Extensions
  - Tensor product case
  - Ridgelet case
  - Experiments

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- Hilbert-Huang transform (EMD + Hilbert transform)

# Empirical Mode Decomposition (EMD): Principle

Goal: decompose a signal  $f(t)$  into a finite sum of Intrinsic Mode Functions (IMF)  $f_k(t)$ :

$$f(t) = \sum_{k=0}^N f_k(t)$$

where an IMF is an AM-FM signal:

$$f_k(t) = F_k(t) \cos(\varphi_k(t)) \quad \text{where } F_k(t), \varphi'_k(t) > 0 \quad \forall t.$$

Main assumption:  $F_k$  and  $\varphi'_k$  vary much slower than  $\varphi_k$ .

Huang et al.<sup>1</sup> propose a pure algorithmic method to extract the different IMF.

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<sup>1</sup>The empirical mode decomposition and the Hilbert spectrum for nonlinear and non-stationary time series



# Empirical Mode Decomposition (EMD): Algorithm

Initialization:  $f^0 = f$

**while** all IMF are no extracted **do**

$$r_0^k = f^k$$

**while**  $r_n^k$  is not an IMF (Sifting process) **do**

Upper envelope  $\bar{u}(t)$  (maxima + spline) of  $r_n^k(t)$

Lower envelope  $\underline{l}(t)$  (minima + spline) of  $r_n^k(t)$

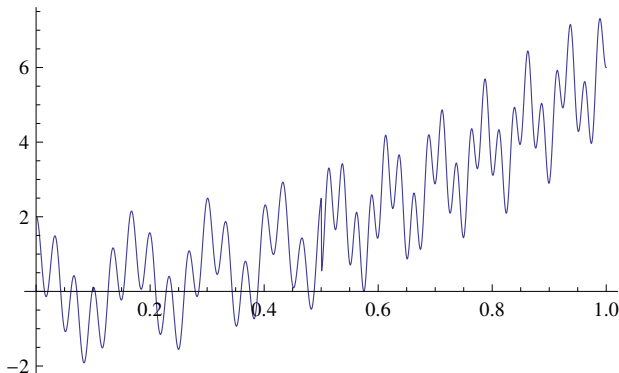
Mean envelope  $m(t) = (\bar{u}(t) + \underline{l}(t))/2$

IMF candidate  $r_{n+1}^k(t) = r_n^k(t) - m(t)$

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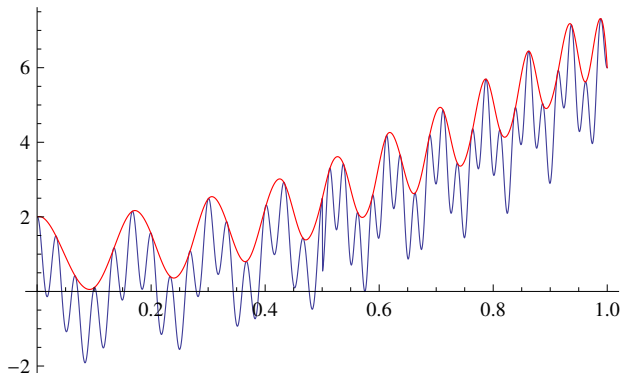
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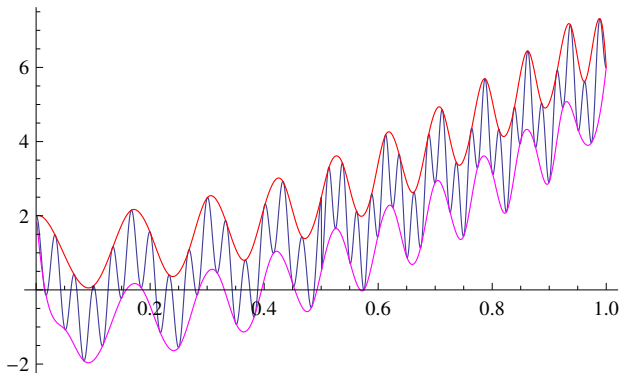
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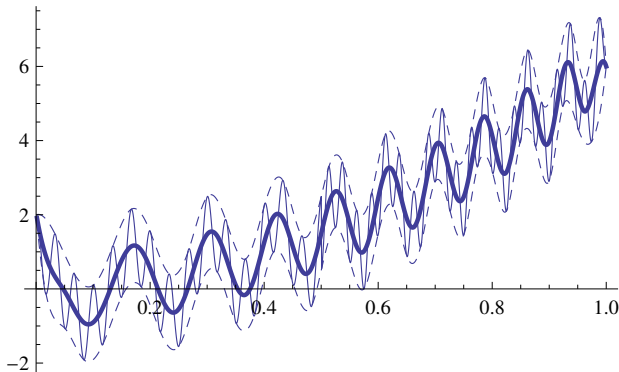
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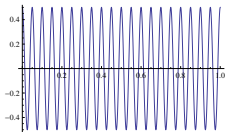
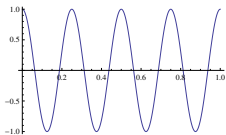
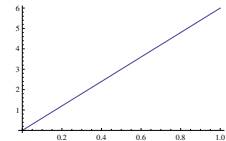
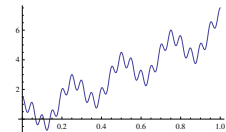
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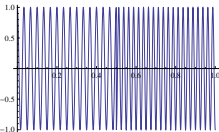
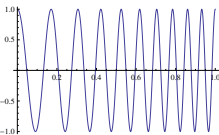
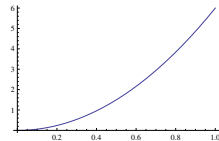
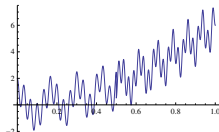


# Example of EMD: input signals

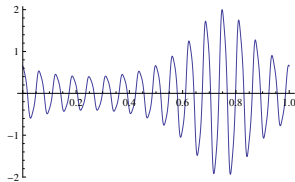
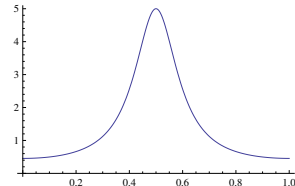
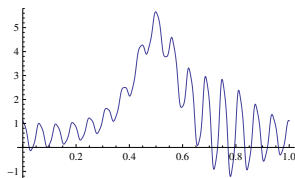
$$f_{Sig1}(t) = 6t + \cos(8\pi t) + 0.5\cos(40\pi t)$$



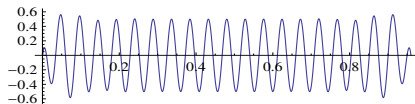
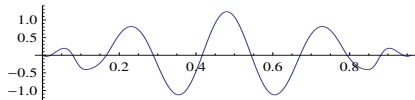
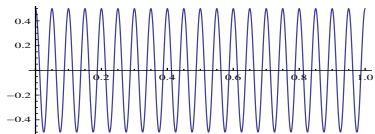
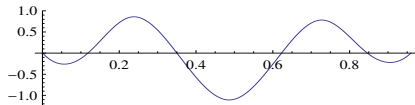
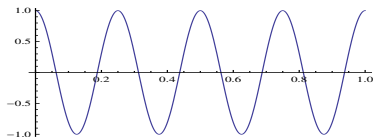
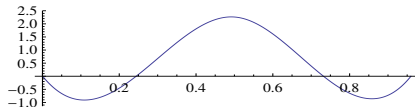
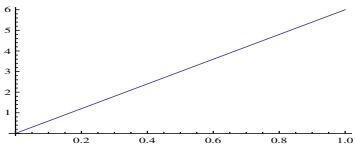
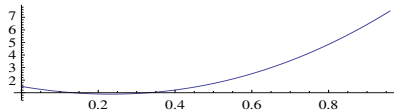
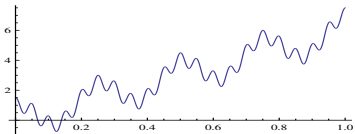
$$f_{Sig2}(t) = 6t^2 + \cos(10\pi t) + 10\pi t^2 + \chi(t > 0.5) \cos(80\pi t - 15\pi) + \chi(t \leq 0.5) \cos(60\pi t)$$



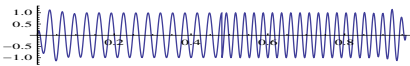
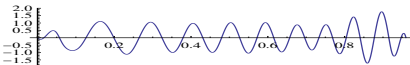
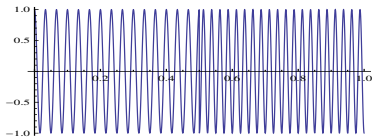
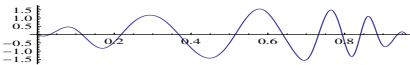
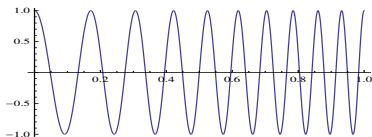
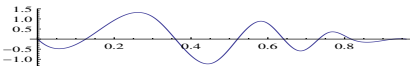
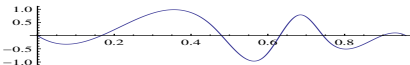
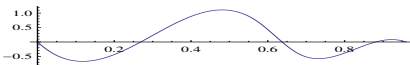
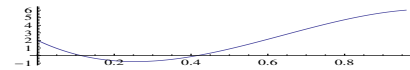
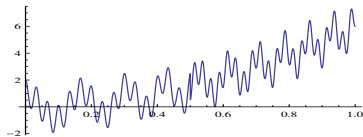
$$f_{Sig3}(t) = \frac{1}{1.2 + \cos(2\pi t)} + \frac{1}{1.5 + \sin(2\pi t)} \cos(32\pi t + \cos(64\pi t))$$



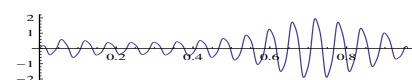
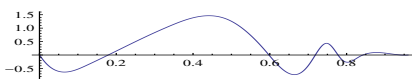
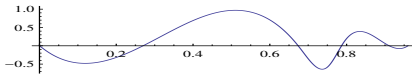
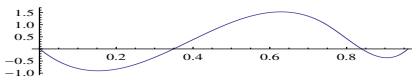
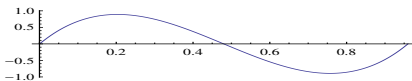
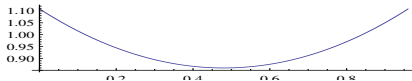
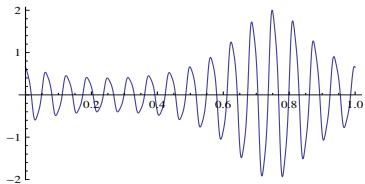
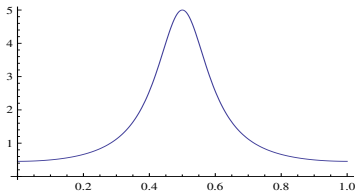
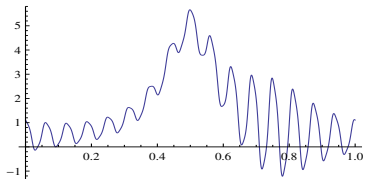
# Example of EMD: $f_{Sig1}$



# Example of EMD: $f_{\text{Sig2}}$

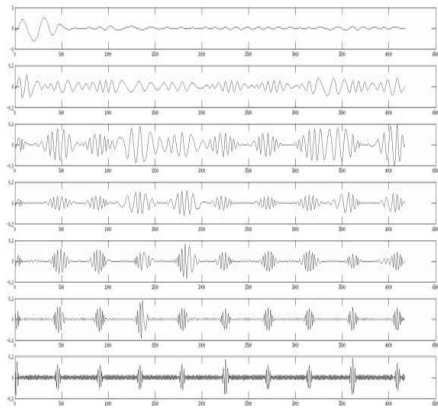
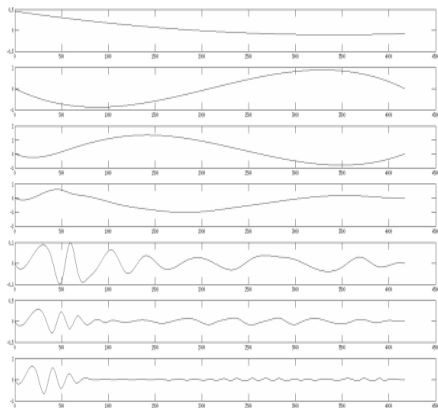
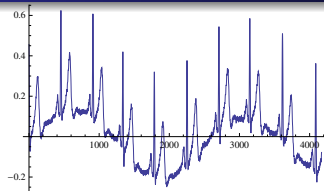


# Example of EMD: $f_{Sig3}$





# Example of EMD: $f_{Sig4}$ - ECG



# Hilbert-Huang Transform

## Hilbert transform

$$\mathcal{H}_f(t) = \frac{1}{\pi} p.v. \int_{-\infty}^{+\infty} \frac{f(\tau)}{t - \tau} d\tau$$

Property: if  $f_k(t) = F_k(t) \cos(\varphi_k(t))$  then

$$f_k^*(t) = f_k(t) + i\mathcal{H}_{f_k}(t) = F_k(t)e^{i\varphi_k(t)}$$

$\Rightarrow$  we can extract  $F_k(t)$  and the instantaneous frequency  $\frac{d\varphi_k}{dt}(t)$ .

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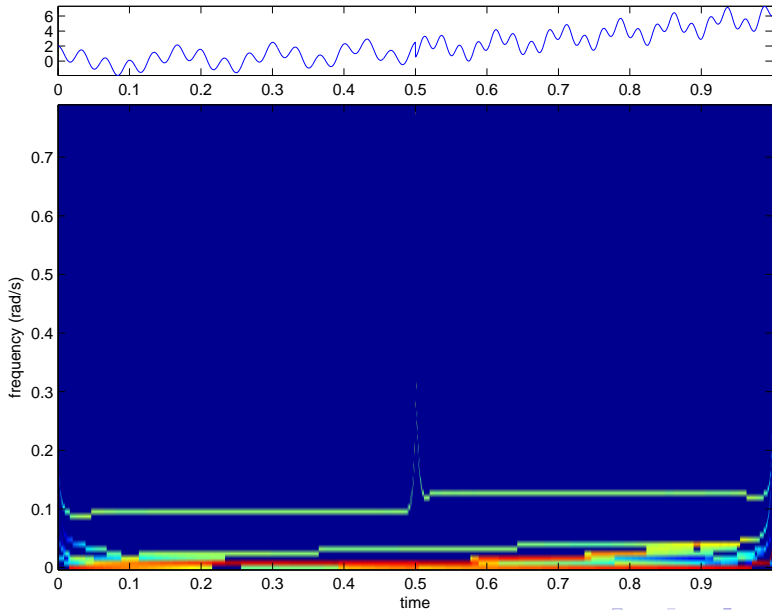
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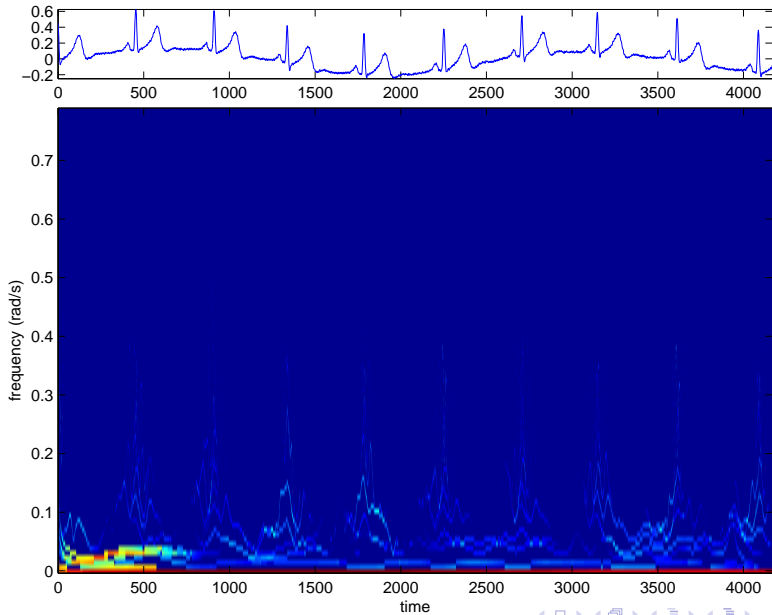
## HHT

For each IMF  $k$ , we extract  $F_k$  and  $\frac{d\varphi_k}{dt}(t)$  and accumulate the information in the time-frequency plane.

# HHT of $f_{sig2}$



# HHT of $f_{sig4}$ - ECG



# EMD: Issues and Properties

- Useful to analyze real signals.
- Implementation dependent.
- Problem: it's a nonlinear algorithm which has no mathematical theory  $\Rightarrow$  difficult to predict and understand its output and behavior in the general case.
- Experimental property: seems to behave as an adaptive filter bank (Flandrin et al.<sup>2</sup>)

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<sup>2</sup>Empirical mode decomposition as a filter bank, IEEE Signal Processing Letters, vol.11, No.2, pp.112–114,

# Key ideas about wavelets

## Wavelets $\Leftrightarrow$ filtering

$$\begin{aligned}\mathcal{WT}_f(m, n) &= a_0^{-m/2} \int f(t) \psi(a_0^{-m}t - nb_0) dt \\ &= a_0^{-m/2} \int f(t) \psi\left(\frac{t - na_0^m b_0}{a_0^m}\right) dt \\ &= (f \star \psi_m)(na_0^m b_0)\end{aligned}$$

where  $\psi_m(s) = \psi\left(\frac{-s}{a_0^{-m}}\right)$ .

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$\Rightarrow$  Wavelets can be built both in the temporal or Fourier domains.



# Empirical wavelet transform (EWT): Concept

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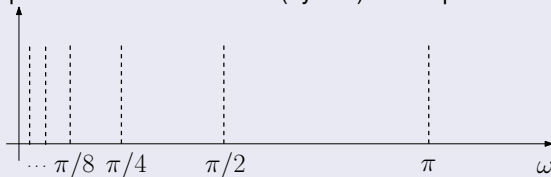
Combining the strength of wavelet's formalism with the adaptability of EMD.

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Wavelets are equivalent to filter banks  $\rightarrow$  (dyadic) decomposition of the Fourier line

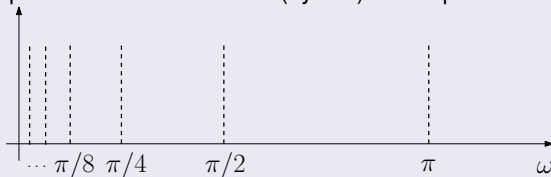


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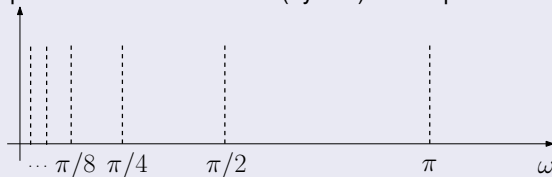
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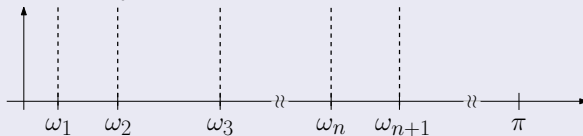
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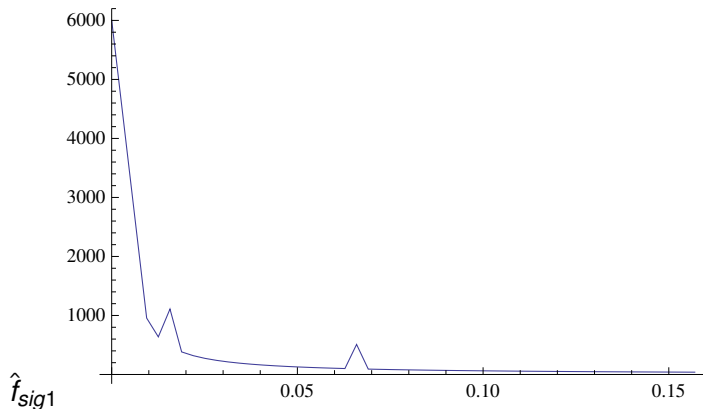
EWT  $\rightarrow$  adaptive decomposition of the Fourier line



# EWT: finding the modes

## Fourier spectrum segmentation:

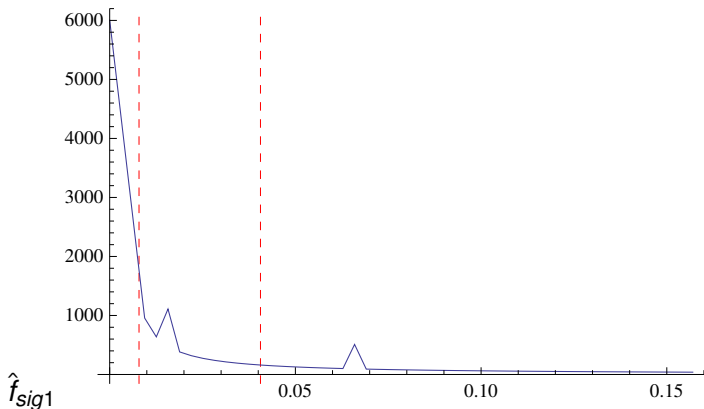
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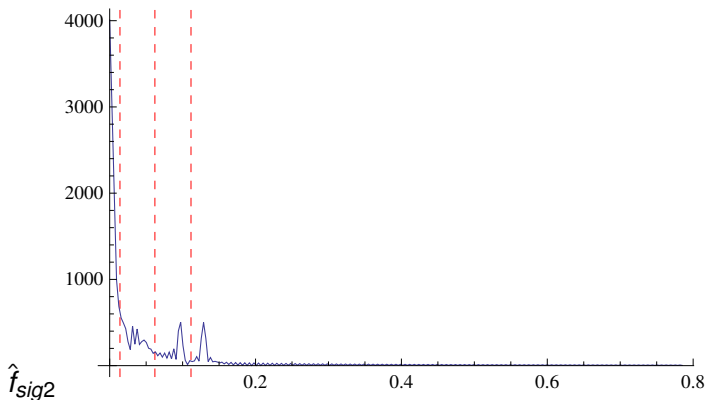
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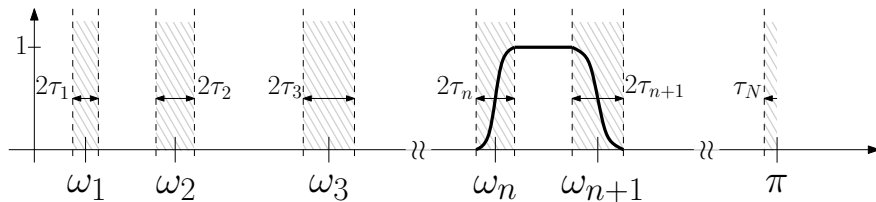
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# EWT: filter bank construction (1/3)

## Notations

- $\omega_n$ : support boundaries
- $\tau_n$ : half the length of the “transition phase”

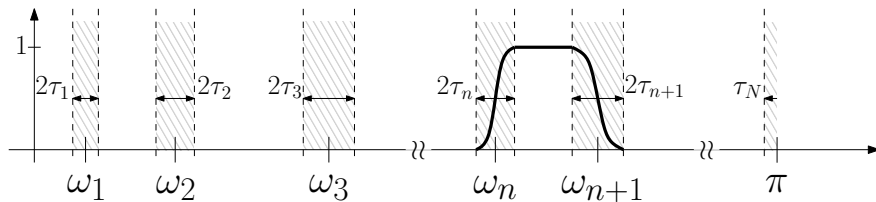




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In practice we choose  $\tau_n = \gamma \omega_n$

# EWT: filter bank construction (2/3)

## Scaling function spectrum

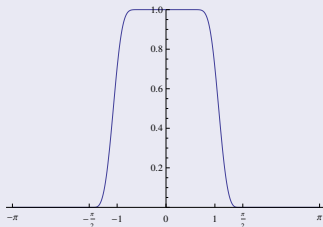
$$\hat{\phi}_n(\omega) = \begin{cases} 1 & \text{if } |\omega| \leq (1 - \gamma)\omega_n \\ \cos \left[ \frac{\pi}{2} \beta \left( \frac{1}{2\gamma\omega_n} (|\omega| - (1 - \gamma)\omega_n) \right) \right] & \text{if } (1 - \gamma)\omega_n \leq |\omega| \leq (1 + \gamma)\omega_n \\ 0 & \text{otherwise} \end{cases}$$

## Wavelet spectrum

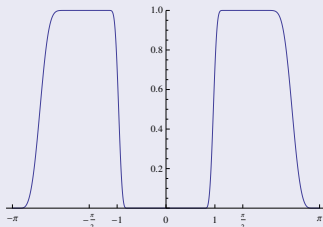
$$\hat{\psi}_n(\omega) = \begin{cases} 1 & \text{if } (1 + \gamma)\omega_n \leq |\omega| \leq (1 - \gamma)\omega_{n+1} \\ e^{-i\frac{\omega}{2}} \cos \left[ \frac{\pi}{2} \beta \left( \frac{1}{2\gamma\omega_{n+1}} (|\omega| - (1 - \gamma)\omega_{n+1}) \right) \right] & \text{if } (1 - \gamma)\omega_{n+1} \leq |\omega| \leq (1 + \gamma)\omega_{n+1} \\ e^{-i\frac{\omega}{2}} \sin \left[ \frac{\pi}{2} \beta \left( \frac{1}{2\gamma\omega_n} (|\omega| - (1 - \gamma)\omega_n) \right) \right] & \text{if } (1 - \gamma)\omega_n \leq |\omega| \leq (1 + \gamma)\omega_n \\ 0 & \text{otherwise} \end{cases}$$

# EWT: filter bank construction (3/3)

Scaling function spectrum for  $\omega_n = 1$  and  $\gamma = 0.5$



Wavelet spectrum for  $\omega_n = 1$ ,  $\omega_{n+1} = 2.5$  and  $\gamma = 0.2$



# EWT: property and example (1/2)

## Proposition

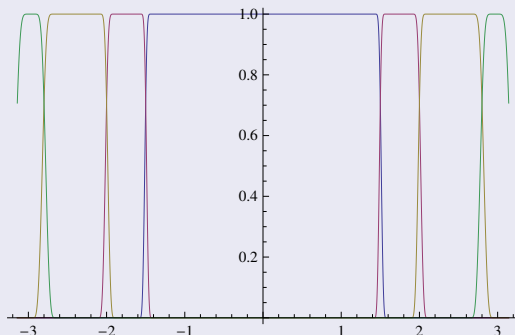
If  $\gamma < \min_n \left( \frac{\omega_{n+1} - \omega_n}{\omega_{n+1} + \omega_n} \right)$ , then the set  $\{\phi_1(t), \{\psi_n(t)\}_{n=1}^N\}$  is an orthonormal basis of  $L^2(\mathbb{R})$ .

# EWT: property and example (1/2)

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If  $\gamma < \min_n \left( \frac{\omega_{n+1} - \omega_n}{\omega_{n+1} + \omega_n} \right)$ , then the set  $\{\phi_1(t), \{\psi_n(t)\}_{n=1}^N\}$  is an orthonormal basis of  $L^2(\mathbb{R})$ .

Filter Bank for  $\omega_n \in \{0, 1.5, 2, 2.8, \pi\}$  with  $\gamma = 0.05 < 0.057$



# EWT: property and example (2/2)

Detail coefficients:

$$\begin{aligned}\mathcal{W}_f^{\mathcal{E}}(n, t) &= \langle f, \psi_n \rangle = \int f(\tau) \overline{\psi_n(\tau - t)} d\tau \\ &= \left( \hat{f}(\omega) \overline{\hat{\psi}_n(\omega)} \right)^{\vee},\end{aligned}$$

Approximation coefficients (convention  $\mathcal{W}_f^{\mathcal{E}}(0, t)$ ):

$$\begin{aligned}\mathcal{W}_f^{\mathcal{E}}(0, t) &= \langle f, \phi_1 \rangle = \int f(\tau) \overline{\phi_1(\tau - t)} d\tau \\ &= \left( \hat{f}(\omega) \overline{\hat{\phi}_1(\omega)} \right)^{\vee},\end{aligned}$$

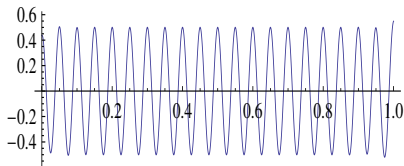
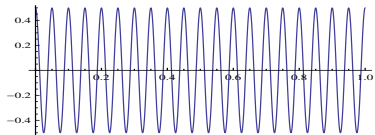
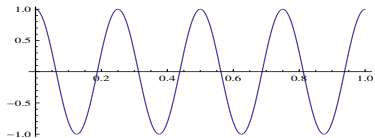
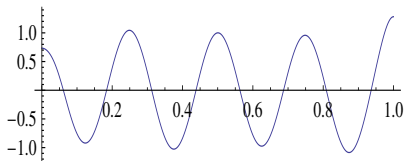
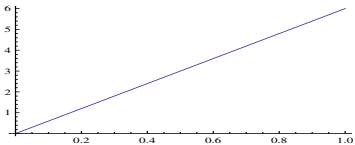
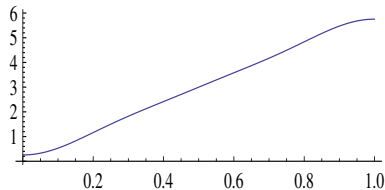
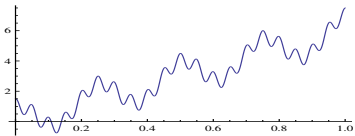
The reconstruction:

$$\begin{aligned}f(t) &= \mathcal{W}_f^{\mathcal{E}}(0, t) \star \phi_1(t) + \sum_{n=1}^N \mathcal{W}_f^{\mathcal{E}}(n, t) \star \psi_n(t) \\ &= \left( \widehat{\mathcal{W}_f^{\mathcal{E}}}(0, \omega) \hat{\phi}_1(\omega) + \sum_{n=1}^N \widehat{\mathcal{W}_f^{\mathcal{E}}}(n, \omega) \hat{\psi}_n(\omega) \right)^{\vee}.\end{aligned}$$

Input:  $f$ ,  $N$  (number of scales)

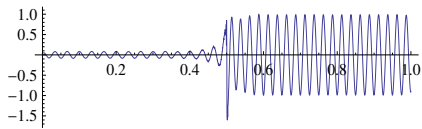
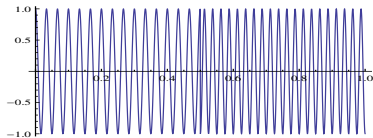
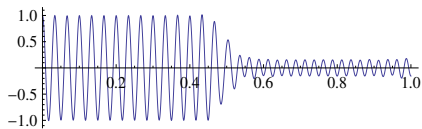
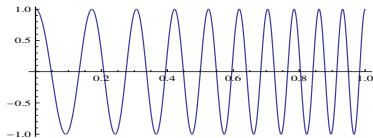
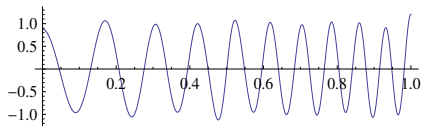
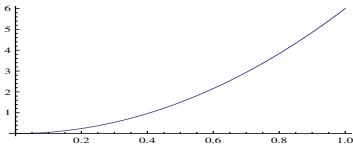
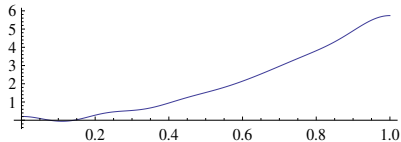
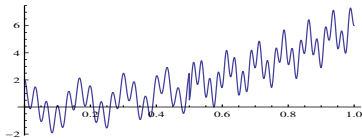
- 1 Fourier transform of  $f \rightarrow \hat{f}$ .
- 2 Compute the local maxima of  $\hat{f}$  on  $[0, \pi]$  and find the set  $\{\omega_n\}$ .
- 3 Choose  $\gamma < \min_n \left( \frac{\omega_{n+1} - \omega_n}{\omega_{n+1} + \omega_n} \right)$ .
- 4 Build the filter bank.
- 5 Filter the signal to get each component.

# Experiment: $f_{Sig1}$

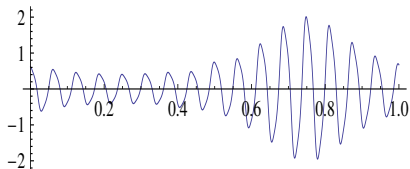
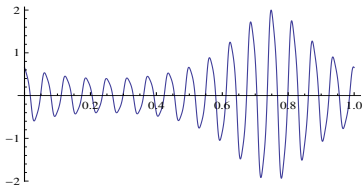
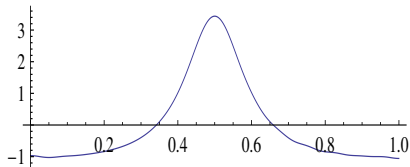
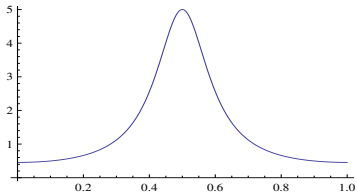
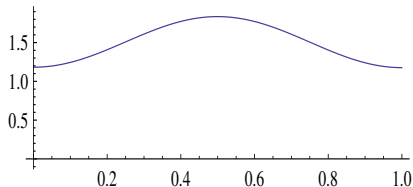
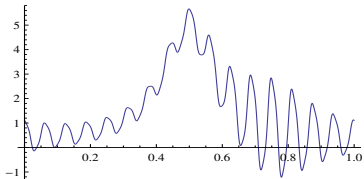




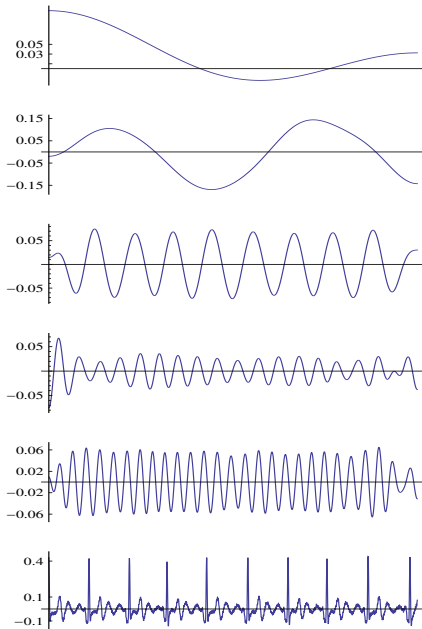
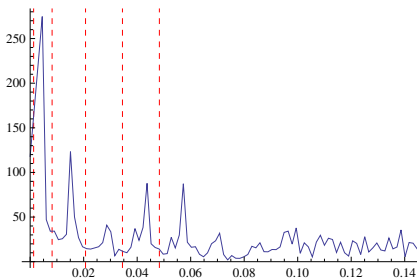
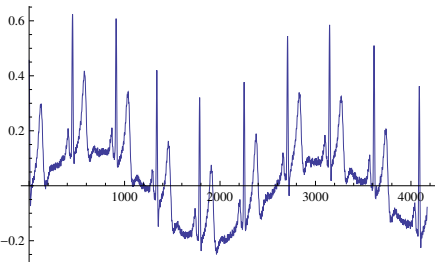
# Experiment of EMD: $f_{Sig2}$



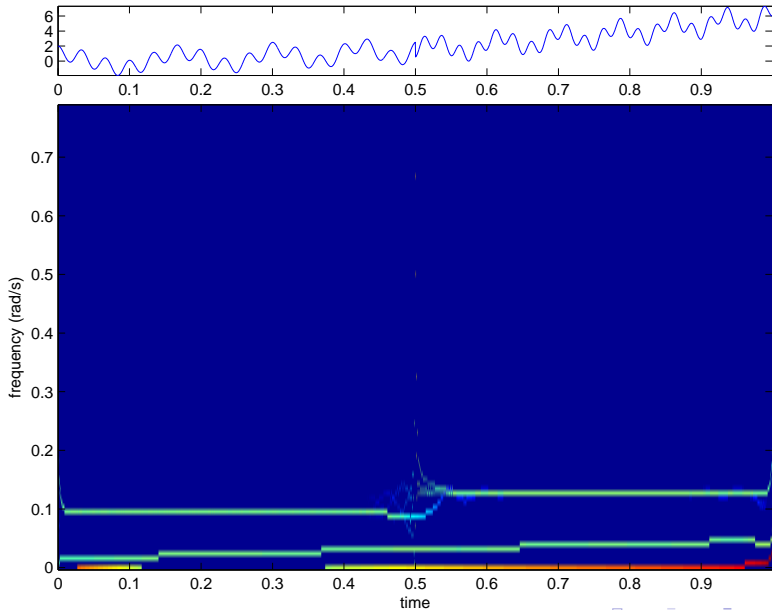
# Experiment of EMD: $f_{Sig3}$



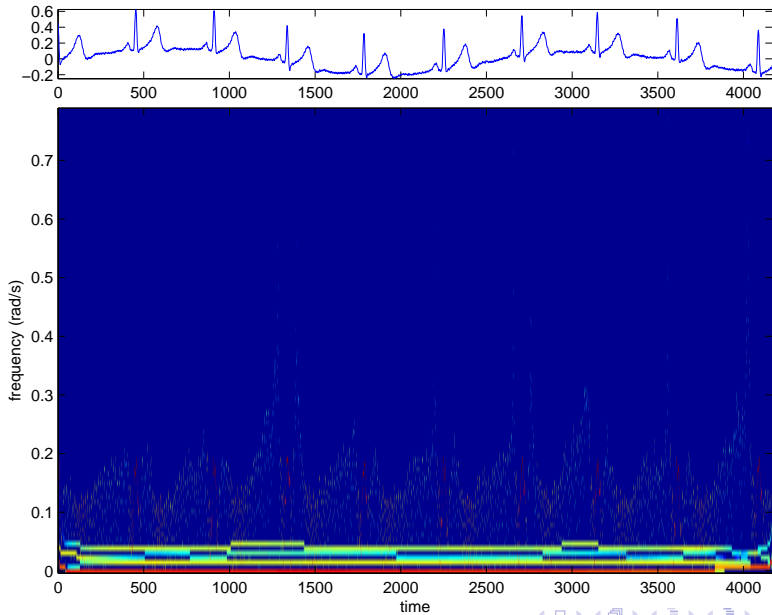
# Experiment of EMD: ECG



# Time-Frequency representation of $f_{sig2}$



# Time-Frequency representation of $f_{sig4}$



# 2D - Extension

joint work with Giang Tran and Stan Osher

## 2D Extension - “Tensor product” approach

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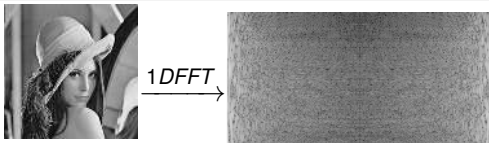
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$\Rightarrow$  Idea: “Mean Filter Banks”

# 2D Extension - Tensor product algorithm



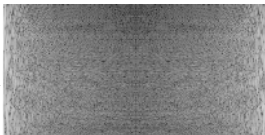
# 2D Extension - Tensor product algorithm



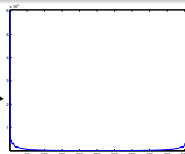
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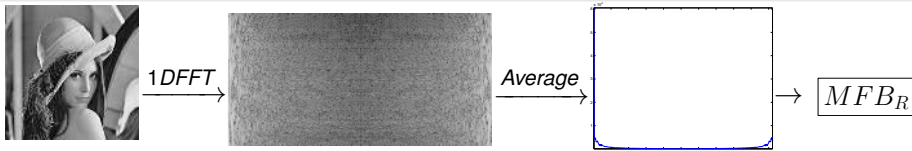
$1DFFT$



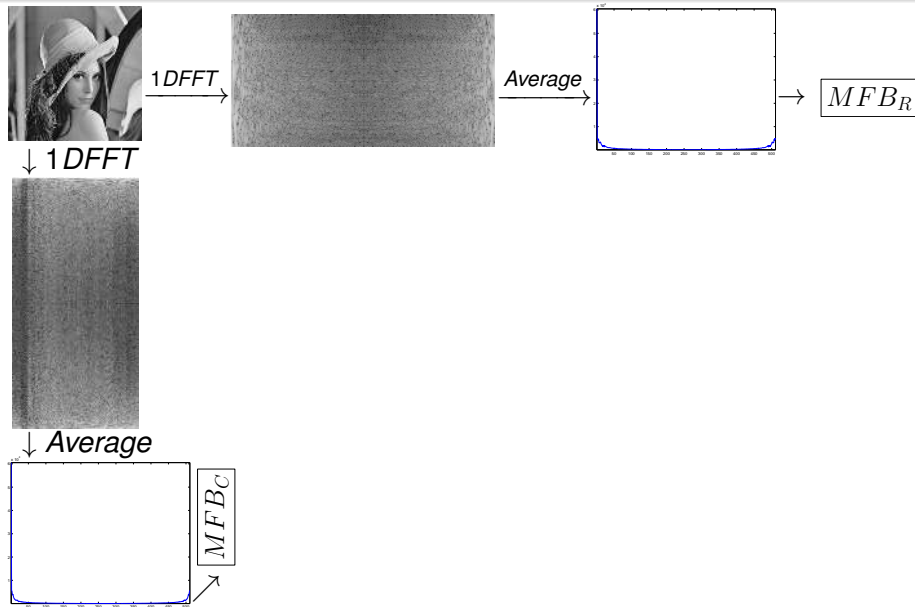
*Average*



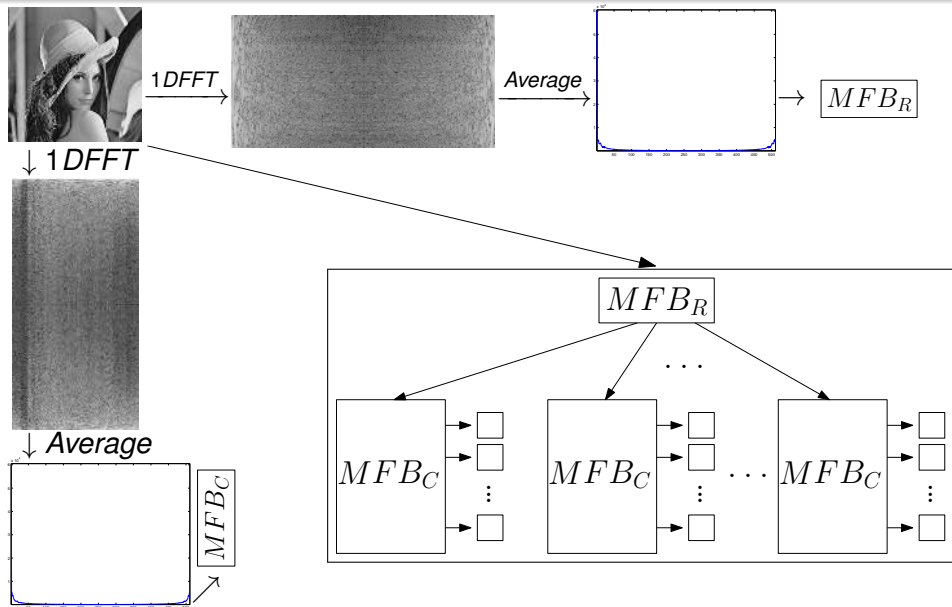
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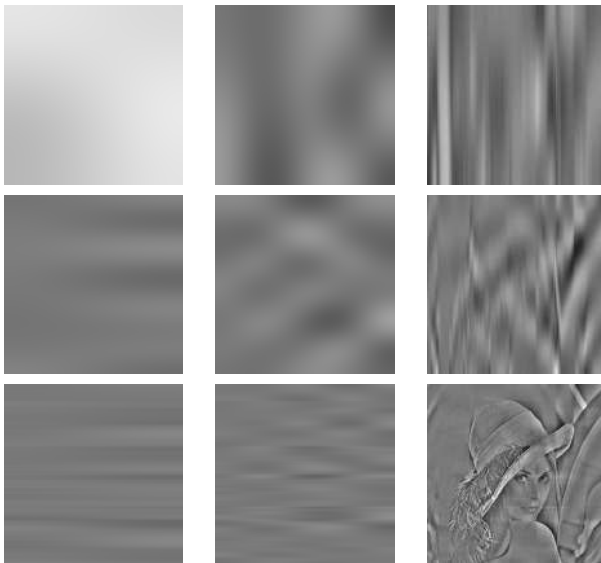


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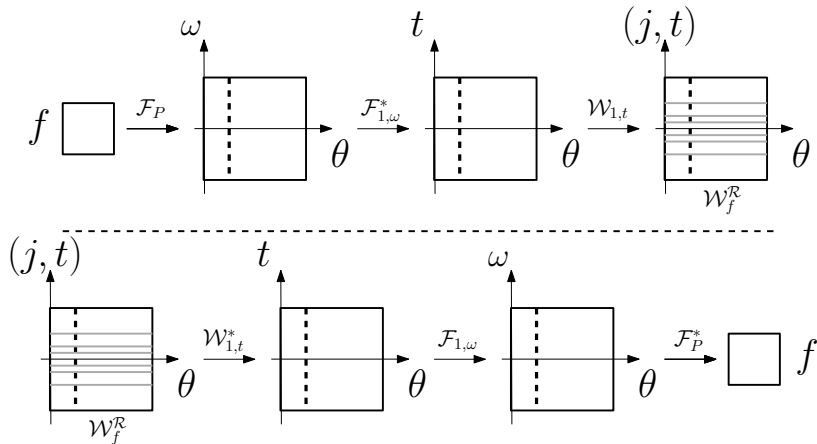


# 2D Extension - Example



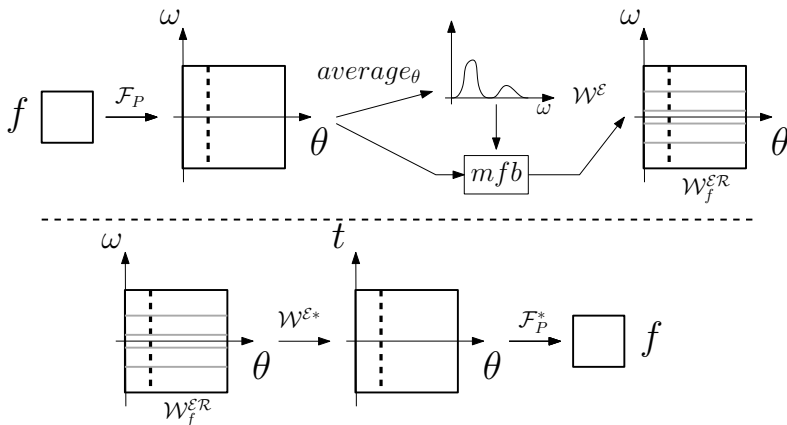
# 2D Extension - Ridgelet approach

## Classic Ridgelets

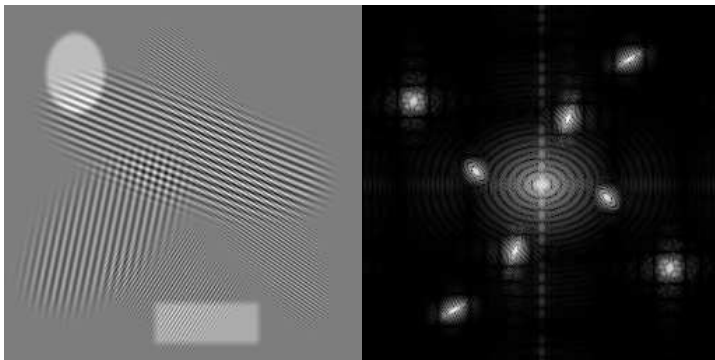


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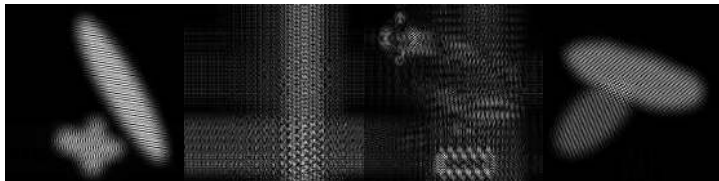
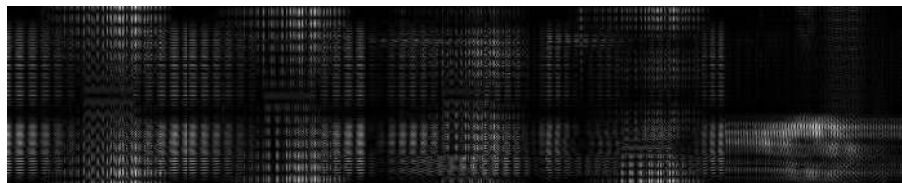
## Empirical Ridgelets



## 2D Extension - Ridgelet: a first example

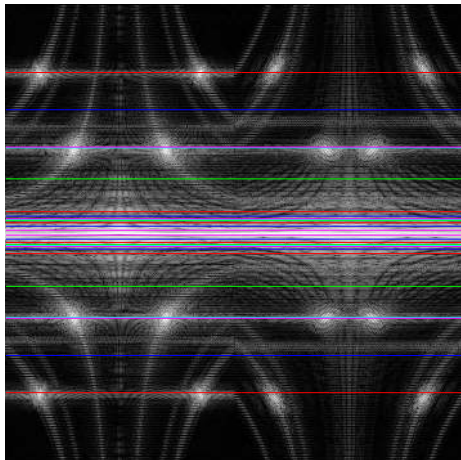


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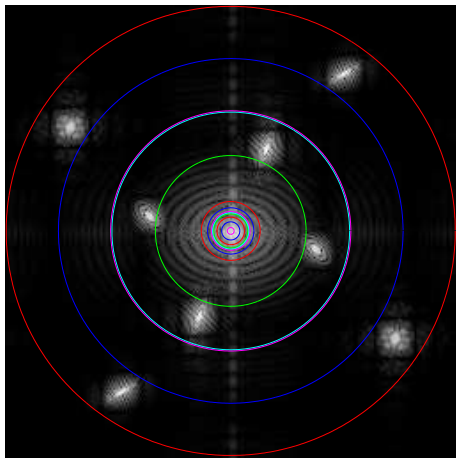
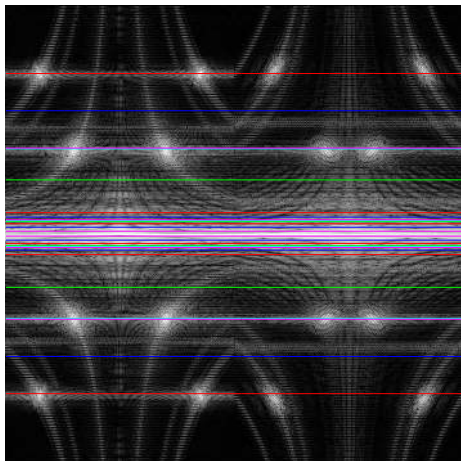


0-1-2-3-4  
5-6-7-8-9  
10-11-12-13

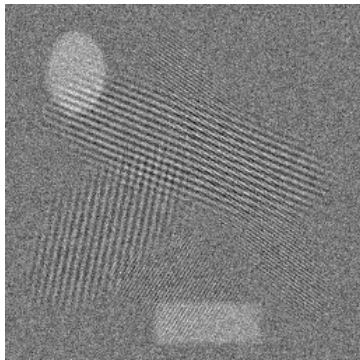
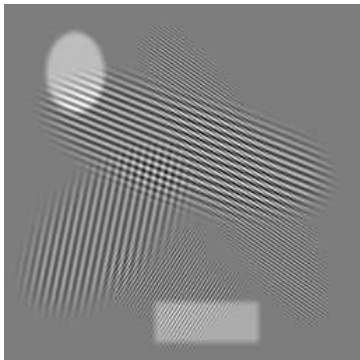
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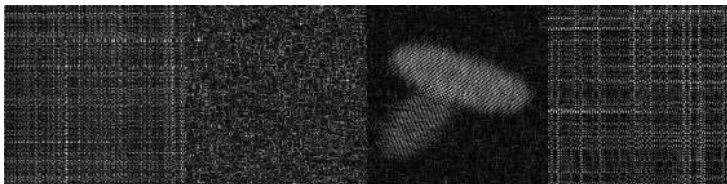
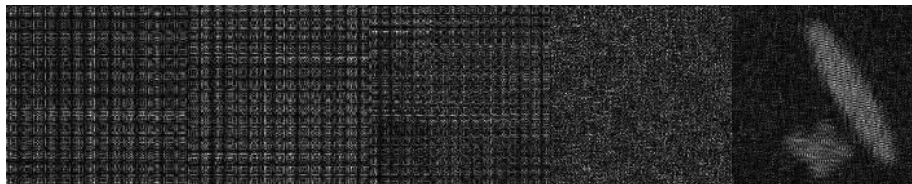
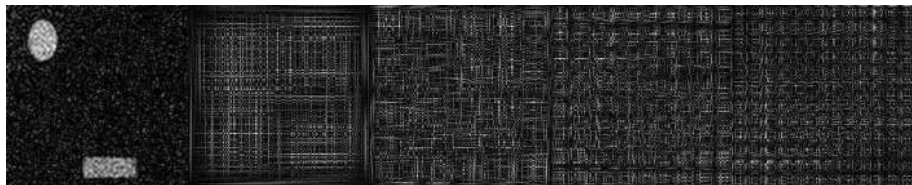


# 2D Extension - Ridgelet: a noisy example



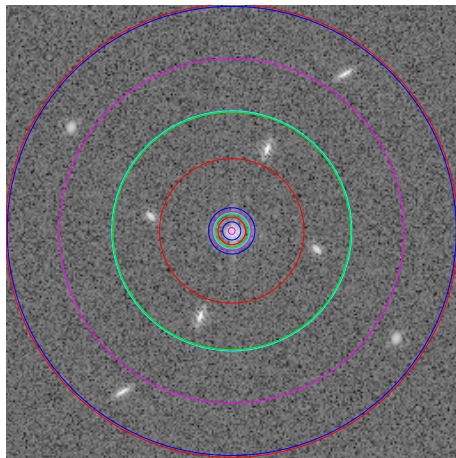
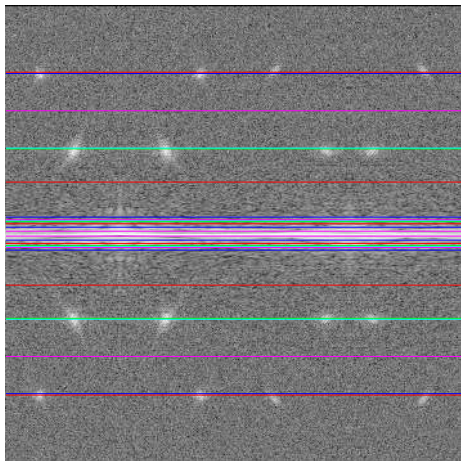


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# THANK YOU!

PS: Jack, I'm from UCLA and on the job market ;-)