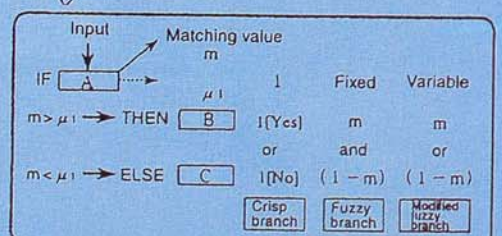
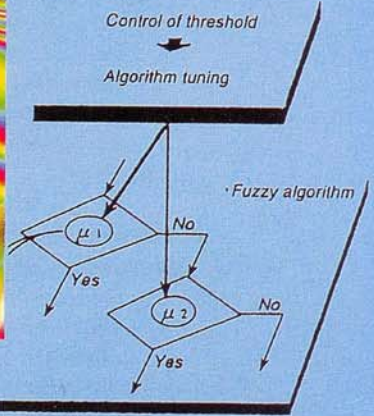
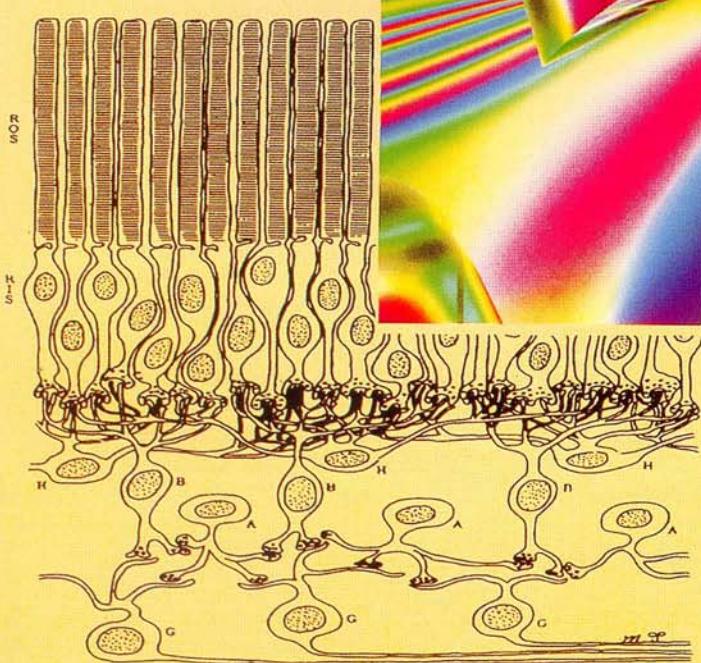
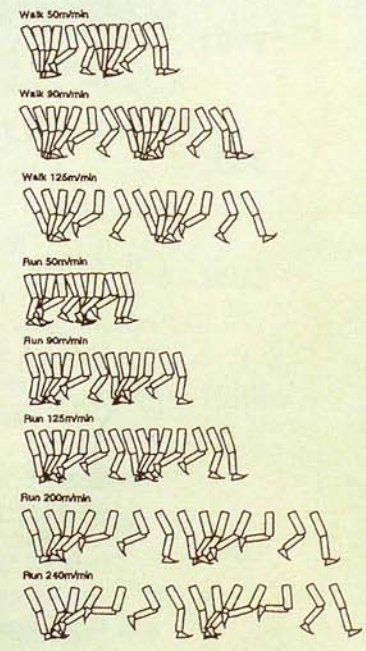
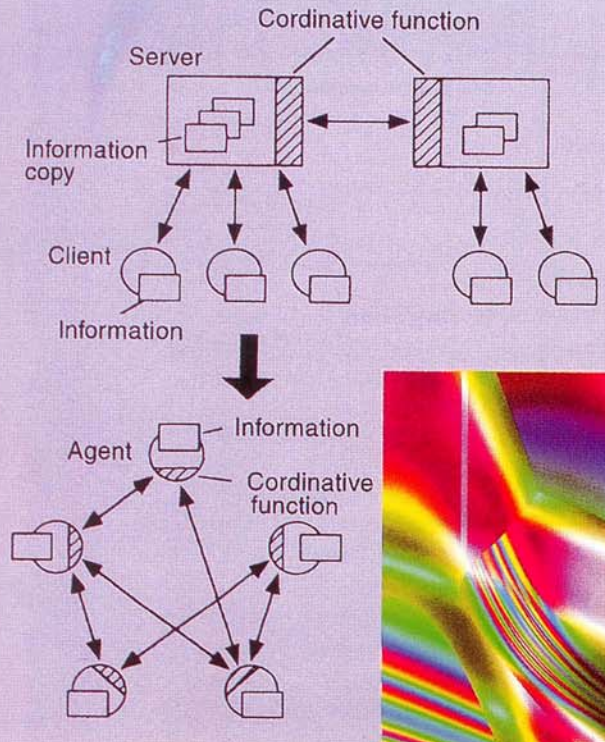


Journal of Advanced Computational Intelligence and Intelligent Informatics

Vol.11 No.9
Nov. 2007

In Cooperation with
International Fuzzy Systems Association (IFSA),
Japan Society for Fuzzy Theory and intelligent informatics (SOFT),
Brazilian Society of Automatics (SBA),
Society of Instrument and Control Engineers (SICE, Japan),
John von Neumann Computer Society (NJSZT),
Vietnamese Fuzzy Systems Society (VFSS),
Fuzzy Systems and Intelligent Technologies Research Society of Thailand (FIRST),
The Korean Fuzzy Logic and Intelligent System Society (KFIS), and
The Taiwanese Association for Artificial Intelligence (TAAI)



Paper:

Experimental Signal Analysis of Robot Impacts in a Fractional Calculus Perspective

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[Received February 5, 2007; accepted June 14, 2007]

This paper presents a study of the signals captured during impacts and vibrations of a mechanical manipulator. In order to acquire and study the signals an experimental setup is implemented. The system acquires data from the sensors, in real time, and processes it through an off-line analysis package. The study is developed based on a set of signal processing tools, such as the Fast Fourier Transform and the windowed Fourier transform. The signals whose Fourier spectrum presents a non integer behavior are pointed out and analyzed using several time windows. The experimental results provide useful information that can assist in the design of a control system to deal with the unwanted effects of vibrations.

Keywords: windowed Fourier transform, vibrations, impacts, robotics, fractional order systems

1. Introduction

The advent of lightweight arm manipulators, mainly in the aerospace industry, where weight is an important issue, leads to the problem of intense vibrations. On the other hand, robots interacting with the environment often generate impacts that propagate through the mechanical structure and produce also vibrations.

This paper presents a fractional calculus (FC) perspective in the study of the robotic signals captured during an impact phase of the manipulator. In order to analyze these phenomena an acquisition system was developed. The manipulator motion produces vibrations, either from the structural modes or from end-effector impacts. The instrumentation system acquires signals from multiple sensors that capture the axis positions, mass accelerations, forces and moments and electrical currents in the motors. Afterwards, an analysis package, running off-line, reads the data recorded by the acquisition system and examines them. Bearing these ideas in mind, this paper is organized as follows. Section 2 addresses the motivation for this work. Section 3 describes briefly the robotic system en-

hanced with the instrumentation setup. Section 4 presents the experimental results. Finally, section 5 draws the main conclusions and points out future work.

2. Motivation

Reference [13] mentions several techniques for reducing vibrations and its implementation either at the robot manufacturing stage or at the operational stage. Briefly, the techniques can be enumerate as: (i) conventional compensation, (ii) structural damping or passive vibration absorption, (iii) control based on the direct measurement of the absolute position of the gripper, (iv) control schemes using the direct measurement of the modal response, (v) active control, driving out energy of the vibration modes, (vi) use a micromanipulator at the endpoint of the larger manipulator and (vii) adjustment of the manipulator command inputs so that vibrations are eliminated.

The work presented here is a step towards the implementation of the sixth technique. In recent years the use of micro/macro robotic manipulators has been proposed for space applications and nuclear waste cleanup. Several authors have studied this technique [14], namely [4, 9] that adopted a command filtering approach in order to position the micromanipulator. Also, Ref. [4] used inertial damping techniques taking advantage of a micro manipulator located at the end of a flexible link.

The experiments described in this paper use a macro manipulator, with a low bandwidth, that is compensated through a much faster micromanipulator inserted at the robot endpoint. In this perspective, in order to eliminate or reduce the effect of the vibration is fundamental to study variables, for implementing an adequate control of the macro/micro system.

Bearing these ideas in mind, is presented a study of the robotic signals, in a FC perspective. In fact, the study of feedback fractional order systems has been receiving considerable attention [8] due to the facts that many physical systems are well characterized by fractional-order models [12]. With the success in the synthesis of real noninteger differentiator and the emergence of new electrical cir-

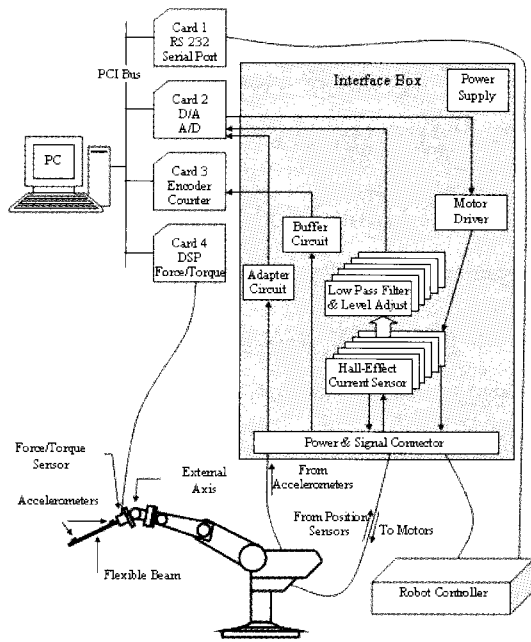


Fig. 1. Block diagram of the hardware architecture.

cuit element called “fractance” [3], and fractional-order controllers [2], have been designed and applied to control a variety of dynamical processes [11]. Therefore the study presented here can assist in the design of the control system to be used.

3. Experimental Platform

The developed experimental platform has two main parts: the hardware and the software components [6]. The hardware architecture is shown in Fig. 1. Essentially it is made up of a robot manipulator, a Personal Computer (PC) and an interface electronic system. The interface box is inserted between the robot arm and the robot controller, in order to acquire the internal robot signals; nevertheless, the interface captures also external signals, such as those arising from accelerometers and force/torque sensors, and controls the external micro-arm. The modules are made up of electronic cards specifically designed for this work. The function of the modules is to adapt the signals and to isolate galvanically the robot’s electronic equipment from the rest of the hardware required by the experiments.

The software package runs in a Pentium 4, 3.0 GHz PC and, from the user’s point of view, consists on two applications:

- The acquisition application is a real time program responsible for acquiring and recording the robot signals.
- The analysis package runs off-line and handles the recorded data. This program allows several signal processing algorithms such as, Fourier transform (FT), correlation, time synchronization, etc.

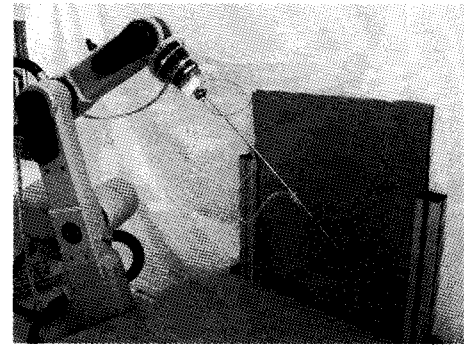


Fig. 2. Steel rod impact against a rigid surface.

Table 1. Physical properties of the flexible beam.

Characteristic	Steel Rod
Density [kg m^{-3}]	7.86×10^3
Elasticity Modulus [N m^{-2}]	200×10^9
Mass [kg]	0.107
Length [m]	0.475
Diameter [m]	5.75×10^{-3}

4. Experimental Results

In the experiment is used a steel rod flexible link. To test impacts, the link consists on a long, thin, round, flexible steel rod clamped to the end-effector of the manipulator. The robot motion is programmed in a way such that the rod moves against a rigid surface. Fig. 2 depicts the robot with the flexible link and the impact surface. The physical properties of the flexible beam are shown in Table 1.

During the motion of the manipulator the clamped rod is moved by the robot against a rigid surface. An impact occurs and several signals are recorded with a sampling frequency of $f_s = 500$ Hz. The signals come from different sensors, such as accelerometers, force and torque sensor, position encoders and current sensors.

4.1. Time Domain

A typical time evolution of some variables is shown in Figs. 3-6 corresponding to: (i) the impact of the rod on a rigid surface and (ii) without impact [7].

In this example, the signals present clearly a strong variation at the instant of the impact that occurs, approximately, at $t = 4$ sec. Consequently, the effect of the impact force, shown in Fig. 5, is reflected in the current required by the robot motors (Fig. 3). Fig. 6 shows the accelerations at the rod free-end (accelerometer 1), where the impact occurs, and at the rod clamped-end (accelerometer 2). The amplitudes of the accelerometers signals are higher near the rod impact side. The two signals are superimposed in Fig. 6. The first acceleration peak (accelerometer 1), due to the impact, corresponds to the rigid surface (case i) while the second peak corresponds to the care of no impact (case ii).

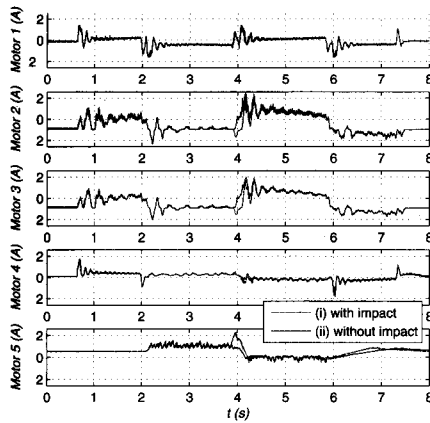


Fig. 3. Electrical currents of robot axis motors.

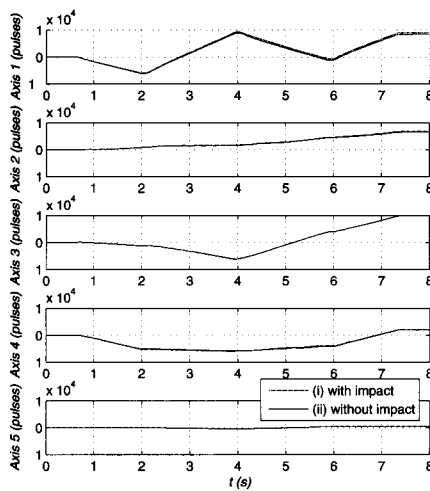


Fig. 4. Robot axis positions.

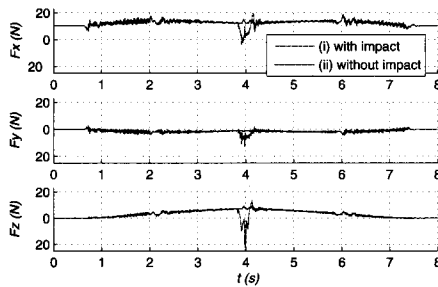


Fig. 5. Forces at the gripper sensor.

4.2. Fourier Transform

In order to study the behavior of the signal FT, a trendline can be superimposed over the spectrum. The trendline is based on a power law approximation:

$$|\mathcal{F}\{f(t)\}| \approx c\omega^m \dots \dots \dots (1)$$

where \mathcal{F} is the signal FT, $c \in \mathcal{R}$ is a constant that depends on the amplitude, ω is the frequency and $m \in \mathcal{R}$ is the slope.

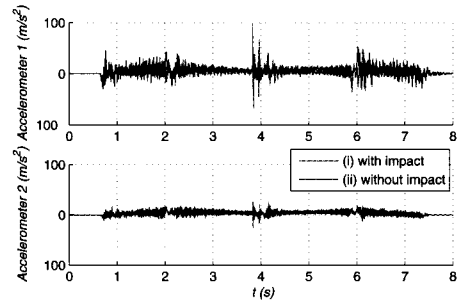


Fig. 6. Rod accelerations.

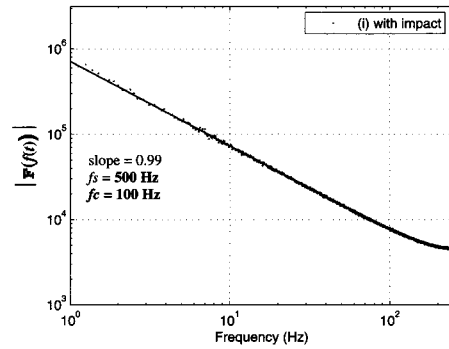


Fig. 7. Spectrum of the axis 1 position.

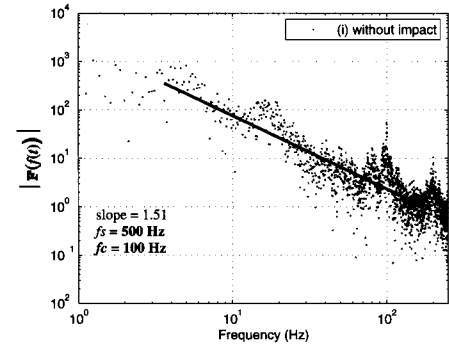
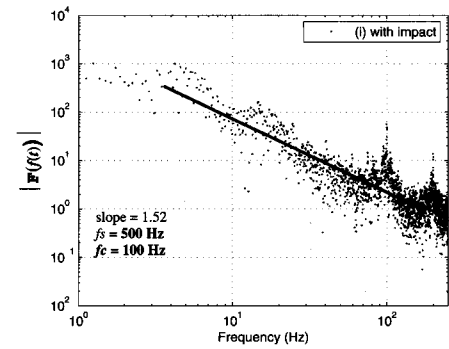


Fig. 8. Spectrum of the axis 3 motor current.

Figure 7 shows the amplitude of the Fast Fourier Transform (FFT) of the axis 1 position signal. The trendline (1) leads to a slope $m = -0.99$ revealing, clearly, the integer order behavior. The others position signals were studied, revealing also an integer behavior, both under impact and no impact conditions. Fig. 8 shows the amplitude of the

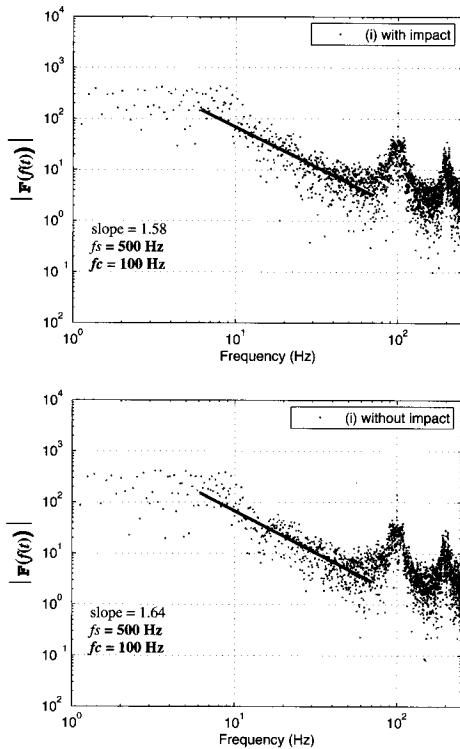


Fig. 9. Spectrum of the axis 4 motor current.

FFT of the electrical current for the axis 3 motor. The spectrum was also approximated by trendlines in a frequency range larger than one decade. These trendlines (Fig. 8) have slopes of $m = -1.52$ and $m = -1.51$ under impact (i) and without impact (ii) conditions, respectively. The lines present a fractional order behavior in both cases. Fig. 9 depicts the amplitude of the FFT of the electrical current for the axis 4 motor. Here the trendlines present slopes that vary slightly (slopes $m = -1.58$ with impact and $m = -1.64$ without impact) but, in both cases, continue to reveal a fractional order behavior. The others axis motor currents were studied, as well. Some of them, for a limited frequency range, present also fractional order behavior while others have a complicated spectrum difficult to approximate by one trendline. According to the robot manufacturer specifications the loop control of the robot has a cycle time of $t_c = 10$ ms. This fact is observed approximately at the fundamental ($f_c = 100$ Hz) and multiple harmonics in all spectra of motor currents (Figs. 8 and 9). Fig. 10 shows, as example, the spectrum of the F_z force. This spectrum is not so well defined in a large frequency range. All force/moments spectra present identical behavior and, therefore, it is difficult to define accurately the behavior of the signals. Finally, Fig. 11 depicts the spectrum of the signal captured from the accelerometer 1 located at the rod free-end of the beam.

Like the spectrum from the other accelerometer, this spectrum is spread and complicated. Therefore, it is difficult to define accurately the slope of the signal and consequently its behavior in terms of integer or fractional system.

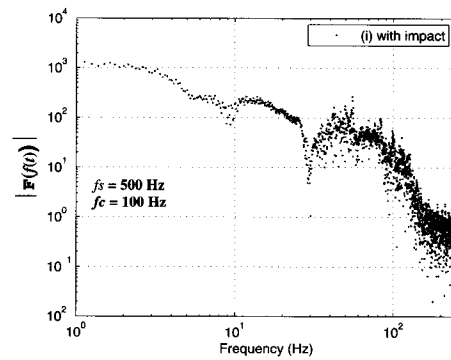


Fig. 10. F_z force spectrum with impact.

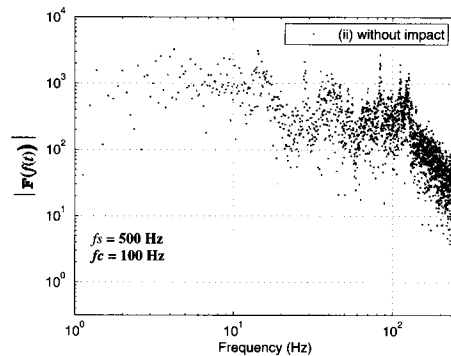


Fig. 11. Acceleration spectrum of the rod free-end without impact.

4.3. Windowed Fourier Transform

Several spectra of the signals captured during approximately 8 seconds are represented in Figs. 7-11. The signal spectra are scattered and, therefore, in order to obtain smoother curves is used a multiwindow algorithm. If we time slice the signals and calculate the Fourier transform, then for each obtained section of the signal, the resulting spectrum is a smoother curve. One way of obtaining the time-dependent frequency content of a signal is to take the Fourier transform of a function over an interval around an instant τ , where τ is a variable parameter [5]. The Gabor Transform accomplishes this by using the Gaussian window. The windowed Fourier transform (WFT), also known as short time Fourier transform (STFT), generalizes the Gabor transform by allowing a general window function [1]. The concept of this mathematical tool is very simple. We multiply $x(t)$, which is the signal to be analyzed, by an analysis moving window $x(t)g(t - \tau)$, and then we compute the Fourier transform of the windowed signal $x(t)g(t - \tau)$. Each FT gives a frequency domain 'slice' associated with the time value at the window centre. Actually, windowing the signal improves local spectral estimates [10]. Considering the window function centered at time τ , the WFT is represented analytically by:

$$F_w(\tau, \omega) = \int_{-\infty}^{+\infty} x(t)w(t - \tau)e^{-j\omega t} dt \dots \dots (2)$$

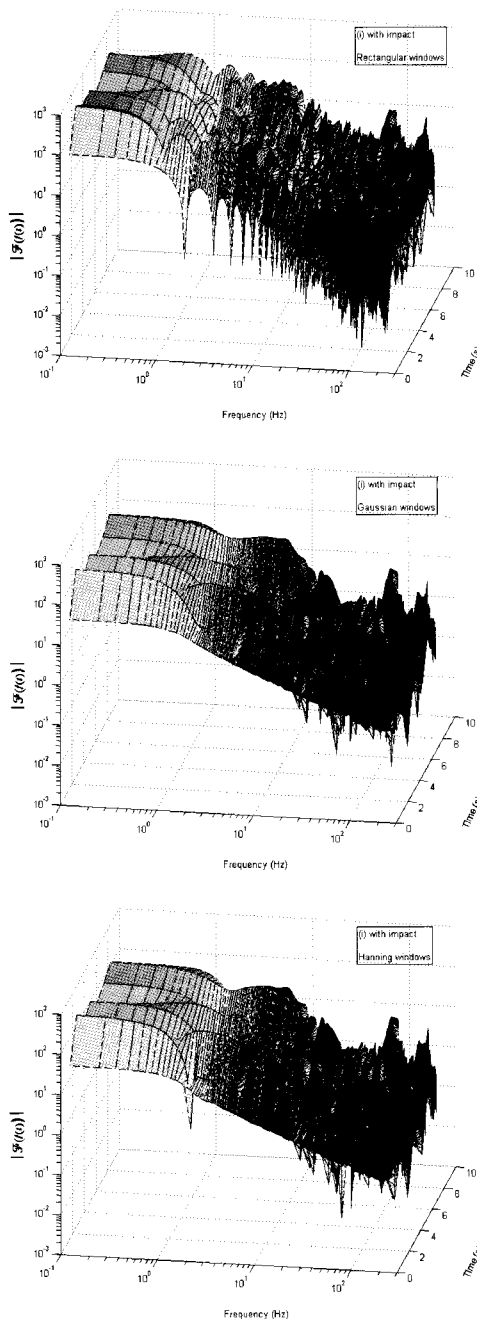


Fig. 12. WFT of axis 4 motor current using the window {rectangular, Gaussian, Hanning}, for $\delta = 1$ sec and $t_w = 1$ sec.

where $\omega = 2\pi f$ is the frequency.

Each window has a width of t_w , where $t_{min} < t_w < t_{max}$ and the time between the centers of two successive windows is δ . Therefore, the frequencies of the analyzing signal $f < 1/t_w$ are rejected by the WFT. Diminishing t_w produces a reduction of the frequency resolution and an increase in the time resolution. Augmenting t_w has the opposite effect. Therefore, the choice of the WFT window entails a well-known duration-bandwidth tradeoff.

On the other hand, the window can introduce an unwanted side effect in the frequency domain. As a result of

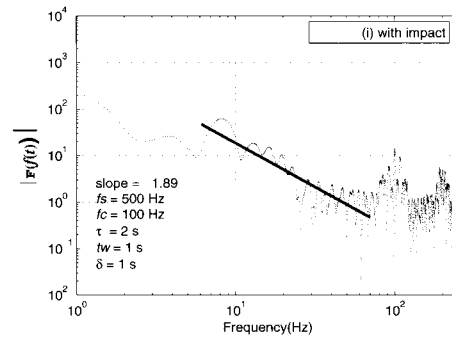


Fig. 13. Spectrum of the axis 4 motor current using a Gaussian window.

having abrupt truncations at the ends of the time domain caused by the window, specially the rectangular one, the spectrum of the FT will include unwanted “side lobes”. This gives rise to an oscillatory behavior in the FT results called the Gibbs Phenomenon [10]. In order to reduce this unwanted effect, generally is used a weighting window function that attenuate signals at their discontinuities. For that reason there are several others popular windows normally adopted in the WFT, namely, Hanning, Hamming, Gaussian, Blackman, etc.

In this line of thought, several experiments were developed with different windows shapes. Due to space limitations we are only depicting the more relevant features. In **Fig. 12** are shown the spectra versus time of the axis 4 motor current, using a WFT with rectangular, Gaussian and Hanning window, respectively. The WFT with the rectangular window presents clearly the Gibbs Phenomenon. The tests show that the WFT with the Hamming window present also this phenomenon, though to a smaller extend. The WFTs with the Hanning, Blackman and Gaussian windows present the best behavior, with slightly differences. Therefore, in the experiments described we adopt the rectangular, Gaussian and Hanning windows. The adequate time parameters of the window, namely the width (t_w) and the inter-spacing time (δ) for the WFT were determined by trial and error. **Fig. 13** depicts the amplitude of the FFT of the electrical current for the axis 4 motor under impact condition using a Gaussian window for $\tau = 2$ sec. In this example the trendline presents a slope of approximately $m = -1.89$. According with the position in time of the moving window, the slope of the trendline will varies revealing different spectral component along the acquisition time.

This fact can be seen in **Fig. 14**. These figures shows the calculated slope for the spectrum obtained applying Rectangular, Gaussian and Hanning windows with different widths t_w . The inter-space time of two successive windows consists on $\delta = 1$ sec.

The wider the window of time, the closer the value of the slope of each window is to the slope calculated by the classical FT, as would be expected. From **Fig. 14** it can be seen that a WFT using a window length of $t_w = 0.25$ sec presents a kind of an erratic behavior, showing therefore

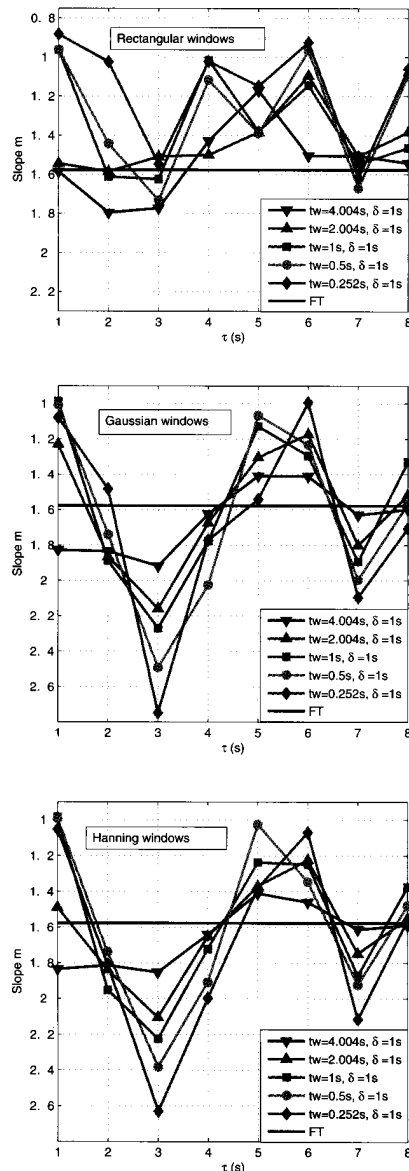


Fig. 14. Slopes of the spectrum of the axis 4 motor current under impact condition using {rectangular, Gaussian, Hanning} windows with $t_w = \{0.25, 0.5, 1, 2, 4\}$ [sec] and $\delta = 1$ sec.

that is an unsuitable window.

Moreover, in **Fig. 14** it can be seen the different behavior of the slopes obtained by the rectangular window comparing with the Gaussian and Hanning windows. Actually, as referred before, using the rectangular window the several slopes present an erratic behavior caused by the Gibbs phenomena.

On the other hand, the slopes obtained by the Gaussian and Hanning windows with suitable parameters, namely, $t_w = \{0.5, 1, 2, 4\}$, are similar, which shows the best behavior of these windows.

We tried the same approach to others acquisitions from the same signal (the axis 4 motor current) at identical conditions. The conclusions were almost identical, which reveals the consistent results presented here.

5. Conclusions

In this paper an experimental study was conducted to investigate several robot signals, in a FC perspective. The use of the WFT confirms the fractional nature of the signals whose behavior was analyzed by the classical FT also. This study provides useful information that can assist in the design of a control system to be used in eliminating or reducing the effect of vibrations.

The next stage of development of the software and hardware apparatus is to reduce the vibrations and its effect upon the robot structure. In this line of thought, is under development a micromanipulator, with a higher frequency response than the main manipulator, mounted at the end-effector and actively counter-acting the undesirable dynamics.

Acknowledgements

This work was supported by the GECAD - Instituto Superior de Engenharia do Porto, 4200-072 Porto, Portugal.

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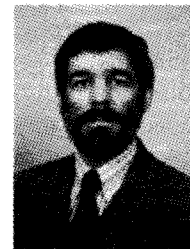
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