



# Article Features and Always-On Wake-Up Detectors for Sparse Acoustic Event Detection

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**Abstract:** The need to understand and manage our surroundings has led to increased interest in sensor networks for the continuous monitoring of events and processes of interest. To reduce the power consumption required for continuous monitoring, dedicated always-on wake-up detectors have been designed, with an emphasis on their low power consumption, simple and robust design, and reliable and accurate detection. An especially interesting application of these wake-up detectors is in detecting acoustic signals. In this paper, we present a study on the features and detectors applicable for the detectors based on the features they utilize and their implementations. This analysis shows that acoustic wake-up detectors predominantly utilize spectro-temporal (56%) and temporal features (36%). Following the state-of-the-art analysis, we select two detector architecture candidates for a case study on passing motor vehicle detection. We utilize our previously developed spectro-temporal decomposition detector and develop a novel level-crossing rate detector. The results of the case study shows that the proposed level-crossing rate detector has lower component count (44 compared to 70) and power consumption (9.1  $\mu$ W compared to 34.6  $\mu$ W) and is an optimal solution for SNRs over 0 dB.

**Keywords:** low power; state-of-the-art analysis; wake-up detector architecture; embedded electronics; case study; motor vehicle detection

## 1. Introduction

The growing need to better understand and manage our surroundings has led to increased interest in the continuous monitoring of events and processes, utilizing sensor networks consisting of hundreds or thousands of small, robust sensor nodes [1–4]. However, having a complex system continuously monitoring for events of interest consumes a lot of power [5,6]. To reduce this power consumption, dedicated always-on low-power wake-up detectors have been designed that wake up the more complex circuits with higher power consumption only when an event of interest is detected [3,7]. Such detectors determine the presence of event candidates by performing low-power extraction and analysis of the sensor signal's features [8–11].

The key emphasis in the design of wake-up detectors is on low power consumption, cheap, simple design, and accurate detection [7,11–14] to ensure low false detection rates, even in the most adverse conditions, as false event detections increase the overall system's power consumption by causing unnecessary activations of the main stage.

Wake-up detectors are often employed in acoustic event recognition because acoustic signals contain a lot of easily extracted information [15–17]. Because of this, they have been utilized in many fields, including safety and security [5,18–21], biomedical and health monitoring [22–24], environmental monitoring [12,25–27], Internet of Things (IoT) applications [2,8], structural health monitoring, non-destructive testing and machinery diagnosis [24], speech or voice activity detection [24,28–33], and others.



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**Copyright:** © 2022 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). In this paper, we present a study on the signal features and wake-up detector architectures applicable for the detection of sporadic, rarely occurring transient acoustic events that appear in the lower end of the acoustic spectrum (up to a few kHz), such as passing motor vehicles.

Our contributions include a review of the state-of-the-art (SOTA) acoustic detectors, the selection of detector architectures of interest, and a comparison of their performance in a case study of motor vehicle (speedboat) detection. A dditionally, we develop a novel implementation of a level-crossing rate acoustic wake-up detector and analyze its performance.

The rest of the paper is organized as follows. Section 2 details the SOTA acoustic wake-up detector analysis. Stemming from the SOTA analysis, in Section 3, the detector selection is performed, and the principles of operation and generalized block schematics of the selected detectors are presented. In Section 4, a case study experiment is presented to evaluate the performance of the selected detectors in the detection of passing motor vehicles. Section 5 concludes the paper and presents future work.

### 2. State-of-the-Art Acoustic Wake-Up Detector Analysis

# 2.1. Methodology

To select the applicable detectors, we perform an analysis of SOTA acoustic wake-up detectors. We explore detector implementations, feature extraction domains (analog, digital, or mixed), power consumptions, and detection accuracies (true and false positive rates). The detector implementation is divided into embedded and integrated, and both are further divided into analog, digital, and mixed-signal detectors. The embedded implementations utilize commercial off-the-shelf (COTS) components, while the integrated implementations are custom-made.

In this analysis, we group the detectors by the acoustic signal features they utilize. To enable this grouping, we devise a feature categorization (Table 1) by analyzing the literature on acoustic signal features [34–36]. While a detailed description of each feature used in acoustic event detection would go beyond the scope of this paper, readers interested in a more detailed explanation of any mentioned feature can find detailed explanations in the literature focused on acoustic feature analysis [34–36].

Temporal	Spectral	Spectro-temporal	Cepstral	Other	
Level-crossing rate-based	Spectral shape-based	Spectro-temporal decomposition-based	Mel-frequency cepstral coefficient-based	Eigenspace-based	
Temporal amplitude-based	Brightness-based	Hurst parameter-based	Other cepstral coefficient-based	Acoustic environment-based	
Temporal power-based	Tonality-based	MP-based Gabor features	-	-	
Rhythm-based	Chroma-based	Sparse coding tensor-based		-	
Correlation-based	-	-	-	-	

Table 1. Acoustic signal feature categorization.

# 2.2. Results

As we can see from the results of the SOTA acoustic wake-up detector analysis (presented in Table 2), six of the categorized acoustic feature subgroups are used in powerconstrained wake-up detector event detection.

Feature Group	Feature	T fund	Pof	Detector Im-	Feature	Power (µW)	<b>Detection Accuracy</b>	
eature Group	Subgroup	Feature	Ref.	plementation	Extraction Domain	rower (μνν)	TP (%)	FP (%)
		Spectro-temporal envelope	[37,38]	Embedded mixed	Analog	7.33; 34.92	90.91	Not stated
	Spectro- temporal decomposi- tion -		[8,15]	Embedded mixed	Analog	26.89	98.67; 100	14; 0
			[16]	Integrated mixed	Analog	43	100	0
		Spectro-temporal energy	[39]	Integrated mixed	Analog	1.01	Not stated	Not stated
Spectro- temporal			[32]	Integrated digital	Digital	~100	96.63	2.33
		Spectro-temporal power	[40]	Integrated mixed	Mixed	0.142	90–91.5	Not stated
		Spectro-temporal RMS	[41]	Integrated mixed	Analog	6	89	Not stated
		Spectro-temporal (absolute) voltage	[17]	Integrated mixed	Digital	0.012	96–98	0
			[30]	Integrated analog	Analog	2.5	Not stated	Not stated
			[28,42]	Integrated mixed	Analog	1; 27.77	~85; ~80	Not stated 0
		Spectro-temporal instant rate of change	[43]	Integrated digital	Digital	0.148	85–99	1–18
		Zero-crossing rate	[13]	Embedded analog	Analog	34	Not stated	Not stated
			[8]	Embedded digital	Digital	~600	Not stated	Not stated
Temporal	Level-crossing rate	Zero-crossing rate with peak amplitude (ZCPA)	[44]	Integrated digital	Digital	Not stated	98	Not stated
		Zero-crossing with short-time magnitude difference	[45]	Embedded digital	Digital	30.71	91	Not stated
	Correlation	Autocorrelation -	[31]	Integrated digital	Digital	24.4	55–95	5–20
			[46]	Integrated mixed	Digital	0.835	97	0
		Cross-correlation	[47]	Integrated mixed	Mixed	1.5	92	7
	Short-time energy	Short-time energy difference	[31]	Integrated digital	Digital	8.5	55–95	5–20
	Multiple	Rise time, min/max, energy	[48]	Embedded mixed	Digital	8.7	100	Not stated
Spectral	Spectral shape	Power spectrum density	[49]	Integrated mixed	Digital	4.7	Not stated	Not stated
Cepstral	Cepstral coefficients	Mel-frequency CC	[33]	Integrated digital	Digital	0.51	97.3	2–2.3

Table 2. Acoustic wake-up detectors.

The spectro-temporal decomposition feature subgroup implies the filtering of the input signals into sub-bands, and the continuous extraction of each sub-band's feature of interest (envelope, energy, power, root mean square (RMS)). After extraction, the feature values are quantified and converted into a binarized spectro-temporal template. A classifier determines this template's resemblance to a preset template, defined by the event of interest. These detectors are usually implemented as mixed-signal detectors, with feature extraction and processing performed in the analog and classification in the digital domain.

The level-crossing rate feature subgroup entails converting the input signal's crossings of a predefined level into pulses of fixed length and amplitude, estimating the number of those pulses in a defined time interval, quantifying it and, therefore, quantifying the level crossing rate and determining if it is within the bounds of level-crossing rates specific for the event of interest. These detectors are usually implemented fully digitally, but they can also be implemented completely in the analog domain. The correlation subgroup requires the input signal to be compared to a delayed version of itself (autocorrelation) or to a preset template representing the event of interest (crosscorrelation). Autocorrelation can also be employed to estimate the input signal's spectral content by examining and locating the local maxima of the autocorrelation function, which appear at delay times equal to periods of the input signal's dominant spectral components. These detectors are usually implemented as digital, because of the impracticality of the analog implementation of some required elements, such as delay lines or memories for storing templates.

The short-time energy feature subgroup implies measuring the input signal's energy in short time windows and comparing it to a preset template. While the SOTA detector utilizing this feature subgroup ([31]) is implemented as digital, a mixed-signal implementation similar to the spectro-temporal decomposition could also be considered.

The spectral shape feature subgroup requires the signal spectrum to be determined, and then for certain parameters of its shape to be examined and quantified. To obtain a detailed enough spectrum representation, these detectors must be implemented as digital.

The cepstral coefficients entail estimating the signal's spectrum, calculating the logarithm of the spectral amplitude, and then performing the discrete cosine transformation on it, generating a cepstrum. The amplitudes of the cepstrum peaks represent cepstral coefficients. Detectors utilizing these features can be implemented as mixed-signal detectors, employing analog domain filtering specific for the cepstral coefficients of interest, followed by digital domain cepstral coefficient estimation.

#### 2.3. Discussion

As can be seen from Table 2, most acoustic wake-up detectors utilize spectro-temporal (56% of all analyzed detectors) and, to a lesser extent, temporal features (36% of all analyzed detectors). Furthermore, of those detectors, spectro-temporal-decomposition-based (61%), level-crossing rate-based (17%) and correlation-based (13%) detectors constitute the majority.

Next, we can see that the integrated custom designs account for 68% of all analyzed wake-up detector designs, while embedded implementations utilizing COTS components constitute around 32%.

There is approximately the same number of detectors that extract the features in the analog and digital domains, with only a few detectors extracting features in both domains simultaneously (8%).

Acoustic wake-up detectors have high detection accuracies (over 90% true positives and under 15% false positives, where stated) and their power consumptions vary from around 10 nW to around 600  $\mu$ W, greatly depending on the detector implementation and utilized feature. Integrated mixed-signal and digital spectro-temporal decomposition detectors can reach sub- $\mu$ W power consumptions, while embedded level-crossing rate detectors reach tens or even hundreds of  $\mu$ W.

## 3. Wake-Up Detector Selection

# 3.1. Criteria

Motivated by the passing motor vehicle use case scenario, we aim to develop a wakeup detector of sporadic, transient acoustic events, lasting for several seconds, with the bandwidth spanning up to 2 kHz.

We focus on wake-up detector architectures implementable with COTS components, operating on analog-domain signals, designed for direct interfacing with acoustic sensors with little or no amplification, and avoiding power-hungry analog-to-digital (AD) conversion [16]. Hence, the detector should be able to reliably operate with weak electric input signals (on the order of 10 mV).

Finally, a wake-up detector must have high detection accuracy, both in terms of high true positive rates, as a detector should not miss events, and low false positive rates, as

false detections lead to wasting power due to unnecessary activations of the power-hungry main stage.

#### 3.2. Results

As we can see from Table 3, wake-up detectors utilizing level-crossing rate and spectrotemporal decomposition meet all our selection criteria.

Table 3. Detector Selection.

	Criteria				
Detectors Utilizing	Applicable for Signals of Interest	Embedded Implementation	No AD Conversion	Detection Accuracy	
Autocorrelation/cross- correlation	✓	×	×	1	
Level-crossing rate	✓	✓	✓	✓	
Spectro-temporal decomposition	✓	1	✓	1	
Other features	1	×	×	$\checkmark$	

# 3.3. Selected Detectors

3.3.1. Spectro-Temporal Decomposition

The generalized architecture of the spectro-temporal decomposition wake-up detector is shown in Figure 1.

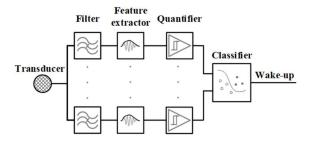


Figure 1. Spectro-temporal decomposition wake-up detector generalized architecture.

The spectro-temporal decomposition detector consists of a filter for spectral decomposition, a feature extractor, a quantifier for quantifying the extracted features, and a classifier to determine if the input signal is from an event of interest. It is usually implemented with multiple channels.

#### 3.3.2. Level-Crossing Rate

The level-crossing rate wake-up detector general architecture is shown in Figure 2.

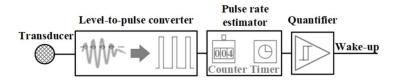


Figure 2. Level-crossing rate wake-up detector generalized architecture.

The level-crossing rate detector consists of a level-to-pulse converter that detects level crossings and converts them into pulses, a pulse rate estimator for estimating the level-crossing rate, and a quantifier for quantifying the level-crossing rate. It is usually implemented as a single-channel detector.

## 4. Motor Vehicle Passing Detection

In this section, we present a case study in which we evaluate the performance of the two selected detectors in the detection of passing motor vehicles through the analysis of their power consumption, minimal input voltage, detection accuracy, and component count (estimate of hardware complexity).

### 4.1. Motor Vehicle Passing Event and Signal

For our dataset, we used 11 prerecorded signals of a twin-engine speedboat passing over a hydrophone submerged approximately 1 m under the surface in shallow water [50]. A representative signal and its spectrogram are shown in Figure 3a,b, respectively.

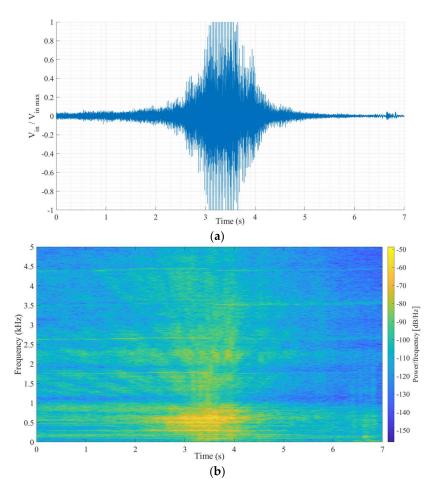


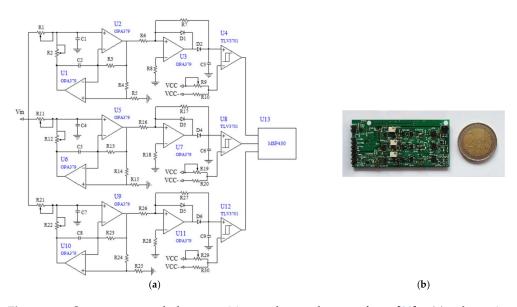
Figure 3. (a) Speedboat passing signal and (b) its spectrogram.

In case of the wake-up detector utilizing spectro-temporal decomposition, the passage of the speedboat can be detected by detecting and tracking the duration of the presence of the signal in the characteristic frequency band (e.g., 100 Hz to 1 kHz for typically 0.5–5 s). On the other hand, a similar type of information is obtained by the level-crossing wake-up detector by tracking the rate at which the signal passes a predefined level in a set time interval.

#### 4.2. Detector Implementations

#### 4.2.1. Spectro-Temporal Decomposition Detector Implementation

We utilize an embedded spectro-temporal decomposition detector that we first presented in [38] (schematic and photograph shown in Figure 4).

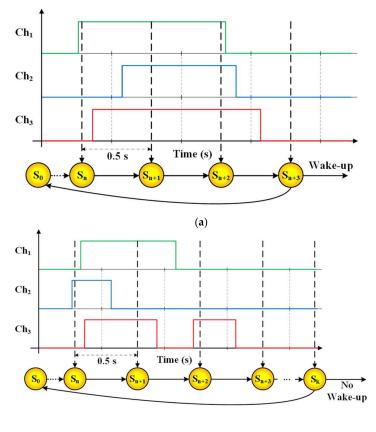


**Figure 4.** Spectro-temporal decomposition wake-up detector from [38]: (**a**) schematic and (**b**) photograph.

As can be seen from Figure 4a, the implemented detector consists of three channels. Each channel extracts information on signal presence within its frequency band. Each channel filters the input signal by a digitally programmable active bandpass filter in the general impedance converter (GIC) topology, implemented with two MCP6142 operational amplifiers. The first channel spans the frequency range from 200 Hz to 500 Hz, the second from 500 Hz to 1 kHz, and the third from 1 kHz to 2.5 kHz. The central frequency and bandpass are programmable (within set limits) in 256 steps by digitally adjustable AD5144 potentiometers.

After filtering, the envelope is extracted utilizing an active voltage doubler, consisting of an MCP6141 operational amplifier and two diodes. The envelope is then quantified using a TLV3701 comparator, with a digitally adjustable threshold, adjusted by another AD5144 potentiometer.

Classification is implemented by binary template matching. A template representing the signal of interest is programed into an MSP430F2013 low-power microcontroller, which also implements a three-channel digital sequence recognition state machine. If the spectro-temporal envelopes' relations match the predefined template, a wake-up signal triggers a more power-hungry digital audio signal processing stage. To achieve this, the microcontroller implements a state machine, which in each state  $S_0, \ldots, S_k$  compares the binary outputs of the three comparators to the prestored three-channel template. The change from  $S_0$  to  $S_1$  is asynchronous, and occurs upon the first change of comparator state (started by an interrupt), while  $S_1$  to  $S_k$  each last 0.5 s up to the maximal sequence length. A more detailed explanation on the basics of the state machine implementation can be found in [9]. For this experiment, the sequence either ends without a wake-up after 7 s (in  $S_{14}$ ) if there is no template match, or with a wake-up signal if all three channels' comparators are simultaneously in a high state for a duration between 0.5 s and 4 s (preset template). This sequence description leads to the state machine implementation with states  $S_0$  to  $S_{14}$ . The state machine functionality is also illustrated in Figure 5.





**Figure 5.** Microcontroller state machine implementation and event detection scheme.  $Ch_1$  to  $Ch_3$  and the three colored lines (green, blue and red) represent each channel's comparator output, and  $S_0$  to  $S_k$  are the state machine states, each lasting 0.5 s. (a) Event detected: at least 2 and no more than 9 consecutive states have all 3 comparator outputs in a high state, and a wake-up signal is generated. (b) No event detected: the total  $S_0$  to  $S_{14}$  sequence passes without meeting the detection condition, no wake-up signal is generated.

### 4.2.2. Level-Crossing Rate Detector Implementation

We developed and utilized a novel embedded level-crossing rate detector, which is an adapted version of a similar detector presented in [13] (schematic and photograph shown in Figure 6).

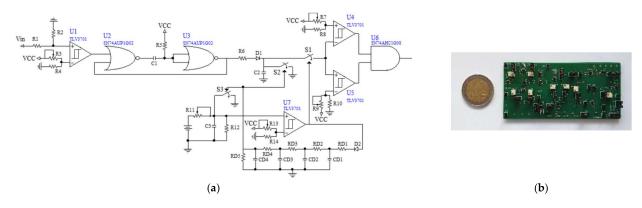


Figure 6. Novel level-crossing rate wake-up detector: (a) schematic and (b) photograph.

The level-crossing detector consists of three main parts. In the first part, each levelcrossing is detected with a TLV3701 comparator. The level is set by adjusting the comparator threshold and, for this experiment, it is set to 2.2 mV. Every time the comparator output changes to a high state, a monostable (consisting of two SN74AUP1G02 NOR gates) is triggered to generate a fixed-length pulse. These pulses are summed by a passive RC circuit, whose output, thus, represents the number of level crossings.

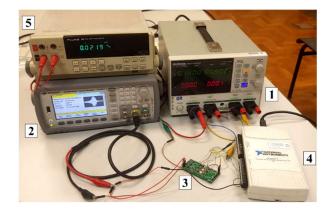
The second part is the timer that consists of a capacitor connected to a TLV3701 comparator with an adjustable threshold. The capacitor is charged by a fixed voltage source over a trimmer resistor. The trimmer resistor and comparator threshold values determine the capacitor charge time, which is set to around 600 ms for this experiment. When the capacitor voltage reaches the comparator threshold, the comparator output changes and closes the S1 switch (TMUX1101) to propagate the RC circuit voltage to the final detector part. After an interval determined by the delay line, the reset switches S2 and S3 (TMUX1101) of the RC circuit and the timer close, allowing their capacitors to discharge to the ground. During the reset, the switch S1 opens, disconnecting the RC circuit from the final detector part. After the RC circuit and timer resets are complete, the reset switches open, and a new level crossing counting interval starts. For this experiment, the delay of the reset signal is set to around 5 ms.

The final part consists of two TLV3701 comparators with adjustable thresholds and an AND logic gate. If the RC circuit voltage is both higher than the lower threshold and lower than the higher one, the level-crossing rate is within the set bounds, an event of interest is detected, and a wake-up pulse is generated at the AND gate output. For this experiment, the lower and upper bounds are set to 100 mV and 625 mV, respectively.

4.3. Experimental Setup and Procedure

# 4.3.1. Experimental Setup

The experimental setup utilized in this case study is shown in Figure 7.



**Figure 7.** Experimental setup, with marked components: (1) power source, (2) waveform generator, (3) tested detector PCB, (4) data acquisition card, and (5) multimeter.

The experimental setup (Figure 7) consists of a GW INSTEK GPD-4303S power source (1), a Keysight 33500B waveform generator (2), the tested detector PCBs (spectro-temporal decomposition or level-crossing rate detector) (3), a National Instruments USB-6211 data acquisition card (4), and a Fluke 45 multimeter (5).

## 4.3.2. Experimental Procedure

The prerecorded speedboat signals were processed in MATLAB, cropped to a duration of 7 s, and then attenuated to determine the lowest input signal with which each detector is operational. Then, for the detection accuracy test, the input signals are scaled to 10 mV and 20 mV peak-to-peak for the level-crossing and spectro-temporal detector, respectively. This voltage scaling adjusts the input signal peak-to-peak voltages to adequately represent the signals generated by passing speedboats on passive hydrophones. Additionally, both detectors are operational with higher voltage levels, with threshold adjustments. However,

if the approximate expected voltage levels are not known for a given application, or the input signal dynamic range would expectedly exceed around 40 dB, an additional automatic gain control (AGC) amplifier would have to be added to each detector's input to ensure correct operation.

To compare the performance of the two wake-up detectors, different levels of white noise were added to each signal to achieve signal-to-noise ratio (SNR) levels from -15 dB to 15 dB with a 5 dB step. Such signals were then stored in the waveform generator used as a signal source for the detectors. The detector outputs are recorded by a data acquisition card, and the recordings are processed using MATLAB.

For the spectro-temporal detector, a successful wake-up was recorded when its comparator outputs matched the predefined binary template and the detector generated a wake-up signal. On the other hand, for the level-crossing rate detector, successful detection is recorded if the RC circuit capacitor voltage was within predefined bounds, generating a wake-up signal.

The detector's power consumption was assessed by measuring its supply current (using a multimeter) and multiplying it with the detector's supply voltage, consumed in the steady state, while listening for the acoustic event.

#### 4.4. Results

In Table 4 and Figure 8, we present the case study results, showing each detector's power consumption, minimal input voltage, component count, and detection accuracy.

Table 4. Wake-up detector's power consumption, minimal input voltage and component count.

	Power (µW)	Minimal Input Voltage (mVpp)	Number of Components			
Detector			Active	Passive	Diode	Total
Spectro-temporal envelope detector	34.6	20	12 + μC	51	6	70
Level-crossing rate detector	9.1	10	7	35	2	44

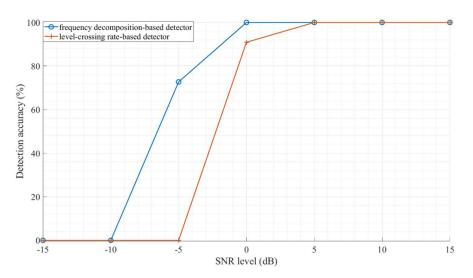


Figure 8. Comparison of speedboat passing detections at given SNR with selected detectors.

As we can see from Table 4 and Figure 8, the level-crossing rate detector has a significantly lower component count and would require around a 40% smaller area to implement (hardware complexity), has lower power consumption, and is operational with lower input voltages, while the spectro-temporal detector has slightly better performance with low-SNR signals, being operational even with -5 dB SNR, as opposed to the level-crossing rate one, which requires at least an SNR of 0 dB for the examined implementation.

# 5. Conclusions

In this paper, we presented a study on low-power always-on sporadic acoustic event wake-up detector designs. To determine the employable detectors and features for this application, we performed a SOTA acoustic wake-up detector analysis, which showed that most acoustic wake-up detectors utilize spectro-temporal (56%) and temporal features (36%), and that the dominant detector implementation is integrated custom-made detectors (68%). Following the SOTA analysis, we presented criteria and selected spectro-temporal decomposition and level-crossing rate as features that allow for the design of a low-power, embedded, always-on wake-up detector operating in the analog domain. These two wakeup detector designs were compared on a case study on passing marine motor vehicle detection. This case study showed that the level-crossing rate detector can be made with a significantly lower component count (44 compared to 70) and power consumption (9.1  $\mu$ W compared to 34.6  $\mu$ W), but a slightly narrower SNR range of operation (minimum of 0 dB SNR compared to -5 dB) than the spectro-temporal detector. In future work, the possibilities of utilizing features not utilized previously in wake-up detectors will be examined, and a more detailed study of the novel level-crossing rate detector will be performed.

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