LARGE APERTURE LOW ABERRATION ASPHERIC DIELECTRIC LENS ANTENNA FOR W-BAND QUASI-OPTICS

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Abstract—An advanced aspheric and asymmetric large aperture dielectric lens antenna is proposed firstly here for high resolution at W-band frequency. Large aperture and aspheric lens provides minimum focusing error and high resolution in millimeter wave quasi-optics application. To the best of the authors' knowledge we design first time 500 mm large aperture lens for W-band quasi optics application. Near field radiation pattern, beam size and focal length of the lens are obtained theoretically and experimentally as well. Dielectric rod waveguide antenna is also designed and employed as a source antenna for the lens. The measured and simulated results of the DRW antenna also show very good performance at W-band frequency, and it has 15.3 dB gain with $-22.5 \, \mathrm{dB}$ sidelobe levels at 94 GHz.

1. INTRODUCTION

Recently, millimeter wave applications are rapidly increasing in modern wireless systems such as uncompressed HD-TV transmission, radio astronomy, collision avoidance devices and image sensing radiometer. At these frequencies lenses become adequate in size and weight for many applications. Lens antennas consist of two main parts: the feeding antenna that can be any type of antenna or antenna array and lens that collimate incident divergent energy to prevent it from spreading in undesired directions. This kind of arrangement at millimeter wave frequency is known as quasi-optics [1–3]. At millimeter wave, the beam radius is smaller in wavelengths, and therefore the beam diffracts strongly. The critical parameters of the quasi-optics

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include antenna gain, beam size, and beam quality over the range of scanned angles. Various types of feeding antennas of the lens, such as horns, dipoles, planar, dielectric rod antenna, have been studied widely and used for achieving a compact size and high radiation efficiency. So far, the most practical structure is a dielectric rod antenna due to its low cross polarization, relatively high gain and potential for a very dense packaging [4–6]. In the case of lens antenna design, primarily two different lenses are used for realizing different goals. Canonical (hyperbolical, bi-hyperbolical, elliptical, hemispherical ...) or shaped lens antennas are used for collimating the radiated energy, or, in the case of shaped designs, for shaping the beam to required radiation patterns, while cylindrical and spherical lenses are mostly used for beam scanning with single, or multiple feed possibility [7–9].

The main objective of this paper is to present very large aperture (500 mm) aspheric dielectric lens antenna to obtain high resolution beam which fulfills compact quasi-optics requirements for image sensing. It has aspheric bi-convex surfaces to keep the beam quality from the aberration. The proposed lens is a design using commercial software CODE-V provided by optical research associates [10]. Further, the prototype of the lens and DRW antenna is fabricated and measured. The near field pattern, focal length and beam size of the lens are obtained experimentally.

High gain and very low side lobes dielectric rod waveguide (DRW) antenna is used as a feed of the lens. Measured and simulated radiation patterns and return loss of the DRW antenna are presented in the further sections.

2. ASPHERIC LENS ANTENNA DESIGN

Aspheric dielectric lens antenna geometry is shown in Fig. 1. To obtain low aberration loss of the lens, it is designed with biconvex aspheric surfaces using optics simulator, CODE V [8]. Eq. (1) shows the design formula for side-I and side-II aspheric surface of the proposed lens. This empirical equation is obtained by curve fitting of the lens surfaces.

$$z = \frac{y^2}{R(1 + \sqrt{1 - (1 + k)y^2/R^2})} + Ay^4 + By^6 + Cy^8 + Dy^{10}$$
(1)

where, y and z are the design coordinates of the surfaces. A, B, C, D, R and k are the constants for both surfaces. The values of the constants for both surfaces are given in Table 1. The centre thickness of the lens is about 145 mm, and edge thickness of the lens is 20 mm as shown in Fig. 1. Fabricated lens antenna is shown in Fig. 2. Material of the lens is HDPE (high density polyethylene) which has dielectric constant

2.3 and loss tangent $0.9 \times 10e - 3$ at 94 GHz. The lens is designed of the diameter approximately 500 mm, corresponding to 150λ at centre frequency 94 GHz. The next section describes the DRW antenna design which is applied as a source for the lens.



Figure 1. Aspheric lens antenna geometry.



Figure 2. Fabricated aspheric lens (a) Side-I and (b) Side-II.

Constants	Side-I	Side-II
R	365.2447	-623.1636
K	0.33277	-81.42037
A	1.73E - 08	1.44E - 08
В	-4.53E - 13	-5.33E - 13
C	4.55E - 18	8.02E - 18
D	-1.53E - 23	-4.11E - 23

Table 1. Constants value of the Eq. (1).





Figure 3. Dielectric rod wave guide antenna.

Figure 4. Return Loss Dielectric rod wave guide antenna at 94 GHz.

3. DRW ANTENNA DESIGN AND MEASUREMENTS

A modified version of the tapered DRW antenna is shown in Fig. 3. The antenna is built of a 7λ long (at Centre frequency 94 GHz) tapered dielectric rod and a conventional metallic waveguide (WR-10) with a junction or matching part between the rod and waveguide. Inner dimension of the junction part of the metallic waveguide is slightly wider $(2.54 \times 1.5 \text{ mm}^2)$ than the remaining waveguide $(2.54 \times 1.27 \text{ mm}^2)$ to improve the return loss. The dielectric rod is constructed by HDPE, same material characteristic as lens dielectric. The above mentioned antenna is also simulated using CST electromagnetic simulator. Details of the DRW antenna design have been described in [11].

Simulated and measured return loss of the above mentioned feed antenna is shown in Fig. 4 at W-band frequency. It shows -15 dB return-loss over the W-band (75–110 GHz) and -21 dB return



Figure 5. (a) *E*-plane, (b) *H*-plane. Radiation gain of DRW at 94 GHz.

loss at desired 90 to 100 GHz frequency band. It shows a well matching between the measured and simulated results except frequency shifted towards higher side of the measured return loss due to imprecise electrical parameter of the HDPE material at millimeter wave frequency. The E and H-field radiation patterns are depicted in Fig. 5(a) and Fig. 5(b) at 94 GHz frequency. Apart from the slightly higher side lobes in the measurement results, good agreement between measured and simulated results can be observed. It can be observed from the figures that the antenna has gain about $15.3 \,\mathrm{dB}$ and $10 \,\mathrm{dB}$ beam-width of 53.2° and 48.6° at each E and H-plane. The measured side lobe labels of the DRW antenna differ from the simulation results in Fig. 5 due to fabrication error of the dielectric rod. The molding of the dielectric rod highly influence the radiation pattern of the DRW antenna at W-band. The following section evaluates the quasi-optics performance of the above mentioned lens and DRW antenna.

4. DESIGN AND ANALYSIS OF QUASI-OPTICS

Quasi-optics configuration of the lens and DRW is shown in Fig. 6. The focal plane array of quasi-optics for millimeter-wave imaging system is composed of an array of the dielectric rod antennas used as a feed antenna of a receiver. We report here the optimum focusing of the incident radiation beam from the lens to the dielectric rod feed antenna. The intensity distribution of the beam is assumed to be Gaussian which is relatively easy to obtain and widely used. The best condition to couple the incident beam with the feed antenna is obtained when the beam shape of the lens coincides with the antenna pattern.

The incident radiation beam on a feed antenna usually has a Gaussian distribution of intensity in the cross-section. The Gaussian beam which is focused by a lens with an angle ψ_w has intensity distribution expressed as

$$q(\psi) \propto e^{-2(\psi/\psi_W)^2} \tag{2}$$

The transformation coefficient between the intensity distribution $q(\psi)$ of the Gaussian beam and feed antenna *E*-field pattern $f(\psi)$ is shown in Fig. 5(a) as follows

$$\alpha = \frac{\int_{-\pi/2}^{\pi/2} q(\psi) \cdot f(\psi) d\psi}{\sqrt{\int_{-\pi/2}^{\pi/2} q(\psi)^2 d\psi \int_{-\pi/2}^{\pi/2} f(\psi)^2 d\psi}}$$
(3)

Calculated results of α are indicated in Fig. 7 as a function of ψ_w , where Eq. (3) was solved numerically by MATLAB with the simulated radiation pattern of the DRW feed antenna. The maximum quasioptical transformation efficiency is obtained at the focusing angle of ψ_w equal to 52° which means the angle between the axial line of the lens and the extended line from the center of the feed antenna aperture to the edge of the lens aperture. At the focusing angle the relative gain level to the maximum gain of the feed antenna is about 10 dB which is the edge taper of the coverage from the feed to the lens which is usually applied to get the maximum transformation efficiency. In the case of our quasi-optic system the 10 dB edge taper angle of the feed antenna should be 52° for low loss from the lens to the feed antenna.



Figure 6. Quasi-optics configuration of the lens and DRW.



Figure 7. Transformation coefficient between the intensity distribution $q(\psi)$ of the Gaussian beam and the feed antenna *E*-field pattern $f(\psi)$.

5. QUASI-OPTICS MEASUREMENTS

Figure 8 shows the quasi-optics experiment setup. Experiment setup was established for near-field pattern measurement of the quasi-optics for different scanning angles. W-band source generator and a standard pyramidal gain horn (Millitech Co, aperture size of $24 \times 18 \text{ mm}^2$) applied as a transmitter and combination of the lens with a dielectric rod antenna was used as a receiver. W-band power meter was attached to the dielectric rod antenna to measure received power.

As an initial experiment set-up for the focal point measurement, distance $S_1 = 2380 \text{ mm}$ between lens and transmitter was fixed. Dielectric rod antenna (feed) was separated about 2400 mm from the lens and scanned on the optical axis of the lens until the maximum output power of the feed antenna was observed by W-band power meter. Maximum received power was obtained at feed position Δ (or $S_2 = 520 \text{ mm}$), and it is considered as a beam focal point of the lens as shown in Fig. 9. Actual value of the focal length (F) about 427 mm is depicted using thin lens equation as follows,

$$\frac{1}{S_1} + \frac{1}{S_2} = \frac{1}{F} \tag{4}$$

where, S_1 and S_2 show a transmitter and feed antenna positions from the lens, respectively, and F is the focal length of the lens. There is a difference between focal length of the measured and expected values that can be caused by the difficulty to measure the correct dielectric constant of HDPE at W-band and misalignment on testing the quasi optics.

Further, the feed antenna (DRW) is positioned at the focus point of the lens, and then the measurement of the beam pattern was performed. The measurement results, shown in Fig. 10, demonstrate the H-plane beam pattern of the received power of the aspheric lens



Figure 8. Quasi-optics Measurement setup.



Figure 9. Relative received power at various feed distance distance position (S_2) .



Figure 10. *H*-plane radiation pattern of the lens, feed at focal length (∇) .

antenna at 90, 94 and 98 GHz frequencies. It is obtained by horizontal movement (x-axis) of the feed (DRW) at focal point of the lens. Field patterns are almost the same for all frequencies, and 3 dB beam size of the received power is about 3 mm. The results represent that the quasi optics is independent of the frequency because the beam patterns are very similar at all frequencies.

6. CONCLUSION

Large aperture aspheric dielectric lens is proposed for quasi-optics design at millimeter wave W-band focal plane imaging system. Prototype of the lens and DRW antenna was fabricated and measured. DRW antenna has gain of 15.3 dB; 10 dB beam-width is 48.6° and SLL of -22.5 dB in *H*-plane at 94 GHz. Its beam pattern is nearly circular. The *E*- and *H*-field patterns are almost the same.

Measured result of the quasi-optics shows that the dielectric lens antenna has focal length about 427 mm and H-plane 3 dB beam size about 3 mm.

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