

Magneto-Thermal Analysis of an Axial-Flux Permanent-Magnet-Assisted Eddy-Current Brake at High-Temperature Working Conditions

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Magneto-Thermal Analysis of an Axial-Flux Permanent-Magnet-Assisted Eddy-Current Brake at High-Temperature Working Conditions

Abstract-This paper proposes an analytic coupled magneto-thermal analysis of an axial-flux (AF) permanentmagnet-assisted (PMA) eddy-current brake (ECB) at hightemperature working conditions. In the topology investigated, permanent magnets (PMs) are placed into stator slot openings to increase the braking torque production capability. This modification enables to control the magnet flux by altering the DC excitation current. However, the utilization of PMs will make the construction vulnerable at high operating temperatures simply because the magnet properties and the brake capability are strongly dependent on temperature. Such problems require complex coupled multiphysics finite element analyses (FEA) to obtain the actual brake performance. The proposed approach offers a simple and effective solution that consists of magnetic and thermal models, which are coupled to each other in the time domain. The nonlinear electrical, magnetic and thermal properties are influenced by the temperature variation in time. An AF-PMA-ECB prototype is manufactured to validate the proposed coupled models and the experimental studies confirm that the proposed approach provides very practical results to determine the working conditions of the AF-PMA-ECB at high-temperature operations.

Index Terms—Axial flux machines, eddy current brakes, permanent magnet assisted eddy current brake, magneto-thermal modeling.

I. INTRODUCTION

 $E_{\rm future\ in\ such\ a\ way\ that\ efficient\ and\ innovative\ products}$ will be dominant in the market. AF-ECBs have been used as a frictionless braking component over a century, and it seems that they will be used more intensively to fulfill different braking demands. ECBs will have a significant role especially in the traction market [1]-[4]. AF-PM and PMA-ECBs provide higher braking torque characteristic compared to traditional magnetless AF-ECB topologies. However, PM-ECBs need an external system to control the braking torque and PM flux [5]-[6]. An external system is also needed for almost all PMA-ECBs. In PMA-ECBs, the magnets can be placed in several places in the magnetic system, for example in the stator yoke or stator teeth or in the stator slot openings. Among these alternatives, the braking torque can be fully controlled only if the magnets are placed in the stator slot openings, where the PM flux can complete its path over the yoke when there is no

excitation. PMs provide an advantage over other topologies but the magnets are close to the brake disc, where a great amount of heat is released during braking resulting in a fast temperature rise. NdFeB PMs are very sensitive to high temperatures. When the PMs are exposed to high temperatures in ECBs, their magnetic properties will change causing some reduction in the braking torque. In addition, the temperature rise in the magnets should be checked to eliminate the danger of demagnetization and to determine the allowed braking time. Therefore, special attention should be given in modeling and analyses of the AF-PMA-ECB coupling magnetic and thermal simulations. Such coupled problems can be solved by commercial multiphysics FEAs software, but existing software require far too long computation time even for practical design cases. A very limited number of papers focus on this issue in the literature. Ye et al. [6] introduced a novel electro-magnetic-thermal-fluid multi-physics coupling model for a radial-flux permanent magnet ECB by using commercial software. Jin et al. [7] presented an analytical thermal analysis of a hybrid excitation linear ECB. The authors built separate magnetic and thermal models and calculated the thermal contacts and airflow by a commercial computational fluid dynamics software. According to the understanding of the authors, no prior work on magnetothermal analysis for ECBs has been reported in literature.

In this work, a new analytical coupled magneto-thermal analysis approach is proposed for AF-PMA-ECB to determine the working conditions at high temperature operating conditions as a fast and practical solution contrary to multiphysics FEAs software. Magnetic analysis consists of 2dimensional (2D) reluctance network modeling, Newton-Raphson nonlinear solver and coupled Faraday's Law and Ampere's Law. In thermal analysis, 3D lumped parameter model is used to determine the temperature rise of each component. The proposed approach runs in the time domain and all nonlinear parameters are updated by temperature rise in each time step.

II. INVESTIGATED AF-PMA-ECB

The investigated AF-PMA-ECB has a single-rotor (brake disc)single-stator configuration and its structure is illustrated in Fig. 1-(a). PM magnetizations, excitation coils and flux paths are provided in Fig. 1-(b). The magnet driven flux completes its path over yoke without exaction current. PM flux can be altered by DC excitation during braking operation in this topology and it is possible to fully control the braking torque without external mechanisms as to [8]-[13]. The key parameters of the investigated AF-PMA-ECB are provided in Table I. The brake has 180 mm/100 mm of outer-to-inner diameter ratio with an axial air gap, g, of 1 mm. The brake disc has a thickness of 15 mm. The N30UH type NdFeB magnets are used in the slot openings of the brake. Low carbon steel (ST37) is selected as brake disc and stator material [14] due to its low material cost. The maximum braking speed of the brake is selected as 2000 min⁻¹ because of the limitations of the test system.

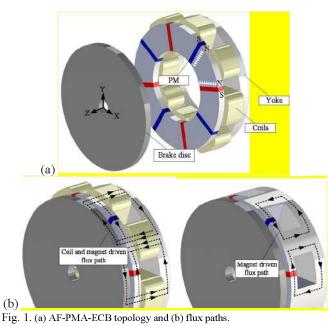


TABLE I

Stator teeth-to-pole ratio

Number of slots & magnets

Axial air gap, g, [mm]

Turn number in the slot

KEY PARAMETERS OF THE AF-PMA-ECB

0.472

2×172

Stator and braking disc material	Low carbon steel (ST37)
Brake disc thickness [mm]	15
PM Type	N30UH
Magnet Size	3.1×10×40 mm ³
PMB_r at room temp.	1.1 T
$PM H_c$ at room temp.	860 A/m

III. PROPOSED COUPLED MAGNETO-THERMAL ANALYSIS APPROACH

An analytic coupled magneto-thermal modeling is developed for the proposed AF-PMA-ECB to determine the actual brake behavior. The proposed coupled magneto-thermal analysis approach aims to determine the braking torque variation, braking time and the temperature rise in time. The flowchart of the proposed approach is given in Fig. 2. The approach has two main stages: The first stage covers the magnetic analysis and the second stage deals with the thermal analysis. In the magnetic analysis, 2D reluctance network modeling and Newton-Raphson nonlinear solver are firstly applied to obtain the static air gap magnetic flux density and skin depth. Then, Faraday's Law and Ampere's Law are solved jointly to get the dynamic air gap magnetic flux density variation and the braking power. In the thermal analysis, a nonlinear 3D lumped parameter model is developed and temperature rise for all components are calculated.

The proposed approach starts with setting the initial parameters such as ambient temperature, operating speed, DC excitation current, *BH*-curve characteristic, PM hysteresis curve, resistivity, heat capacity, the braking time and critical temperature of the brake disc and magnets. In the first time step, magnetic analysis is performed by the initial values and the braking power is calculated. Next, computed braking power is used as an input in the thermal analysis. Temperature rise for all nodes is calculated and the critical temperature of the brake disc and magnets is monitored at the end of the thermal analysis.

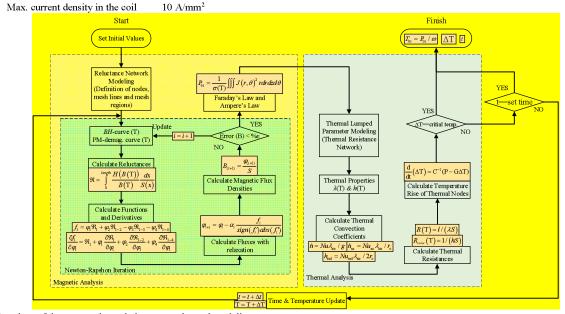


Fig. 2. Flowchart of the proposed coupled magneto-thermal modeling.

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Before updating the time and the temperature for the next time step, the braking time is compared with the set value and the process continues until the critical temperature of the brake disc or magnets is reached, or the braking time is completed. The material properties depended on temperature are updated in each time step. The *BH*-curve, PM hysteresis curve, resistivity and thermal properties are defined as a function of temperature in the proposed approach.

A. Magnetic Analysis

The investigated AF-PMA-ECB has 8 slots and as a result of symmetry, it can be modeled by concentrating on only one slot. The symmetric 2D model of the AF-PMA-ECB is illustrated in Fig. 3-(a). 2D model of the brake is obtained by a z-axis cut plane, which is applied in the middle of the inner-outer diameter. In this case, the model depth is equal to the difference between the inner and the outer radius [15]-[16]. The black circles represent the mesh points and the red dashed lines represent the mesh lines. The air gap in the model is meshed by two horizontal layers to improve the accuracy of the results. The yoke and the slot regions are simply meshed to avoid heavy computation. The reluctance network model is given in Fig. 3-(b). The 22×22 reluctance matrix and the 1×22 current linkage vector are generated to solve the matrix equation so as to find the mesh flux vector. The reluctances are defined by their x and y axes mesh information as described in Fig. 3-(b). To give an example, the PM reluctance as function of temperature is defined by

$$\Re_{\rm PM}(T) = \frac{H_{\rm cint}(100+b)}{B_{\rm rint}(100+a)} \frac{x_{\rm PM} - x_0}{(y_4 - y_3)d}$$
(1)

where B_{rint} is the remanent flux density at room temperature, $H_{\text{c}}_{\text{int}}$ is the intrinsic coercivity, a is the temperature coefficient of the remanence ($a = -0.09\Delta T$ for UH magnet type), b is the temperature coefficient of intrinsic coercivity ($b = -0.52\Delta T$ for UH magnet type) and d is the model depth, which equals to outer-to-inner radius difference.

The saturable reluctances of stator and brake disc are defined as functions of magnetic field strength in the network model. For example, the stator yoke reluctance is defined as

$$\Re_{x_{0}-1,y_{0}-1} = \frac{H_{x_{0}-1,y_{0}-1}\left(B\left(T\right)\right)}{B_{x_{0}-1,y_{0}-1}\left(T\right)} \frac{x_{1}-x_{0}}{\left(y_{2}-y_{1}\right)d}.$$
(2)

Magnetic properties of low carbon steel depend on temperature, B(T) and H(T) which are given in the Appendix [17]. The saturable reluctances are expressed as functions of H since the nonlinear analysis converges the solution faster with this approach. The *HB*-curve of the low carbon steel used here is presented in Fig. 4.

The Newton-Raphson method is used as a nonlinear solver for the reluctance network modeling. The Newton-Raphson method may have converging problems when the model has numerous initial parameters. Because of the selection of arbitrary initial parameters, the model could have nontrivial dynamics. The reluctance network model given in Fig. 3-(b) has the number of 37 saturable reluctances and classical Newton-Raphson method cannot converge a solution.

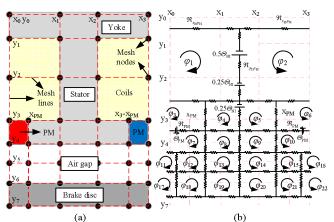


Fig. 3. (a) 2D symmetry model of the proposed AF-PMA-ECB with mesh nodes and mesh lines and (b) reluctance network model of the proposed AF-PMA-ECB. The current linkage Θ is divided in three proportions. The coordinates of y_5 and y_6 are not shown for better view.

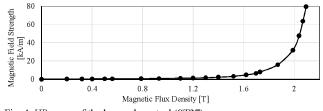


Fig. 4. HB-curve of the low carbon steel (ST37).

Therefore, the Newton-Raphson method is relaxed by

$$x_{i+1} = x_i - \alpha_i \frac{f_i(x)}{sign(f'_i(x))abs(f'_i(x))}$$
(3)

where *i* is the iteration number, *a* is the relaxation factor, which is set to 0.5 in this study. The aim of this modification is to follow the derivative of the function to assign the slope sign into the convergence problem. The Newton-Raphson method is an iteration-based root finding algorithm which starts with the initialization of magnetic flux densities and fluxes. Next step, magnetic fields are obtained based on *HB*-curve and then the reluctances are calculated. Following, functions and theirs derivatives are computed and mesh fluxes are calculated. As a last computational step, magnetic flux densities are calculated and comparison is performed. In each iteration, saturable reluctances are updated by the *HB*-curve until the desired error is obtained. One has to point out that the *HB*-curve is preferred since it is more suitable for the Newton-Raphson functions derivatives than the *BH*-curve.

The important stage of the Newton-Raphson method is the calculation of functions and their derivatives. As seen in Fig. 3-(b), the mesh fluxes have symmetry so that the 1^{st} flux mesh is equal to the 2^{nd} flux mesh and the 3^{rd} flux mesh is equal to the 6^{th} flux mesh and etc. Therefore, the definition of the left-side or the right-side Newton-Raphson functions and derivatives is adequate for the entire model. The 1^{st} function and its derivative are described as

$$f_1 = \varphi_1 \Re_1 + \varphi_2 \Re_{1-2} - \varphi_3 \Re_{1-3} - \varphi_4 \Re_{1-4} + 0.75 \Theta_{\rm NI}$$
(4)

$$\frac{\partial f_1}{\partial \varphi_1} = \Re_1 + \varphi_1 \frac{\partial \Re_1}{\partial \varphi_1} + \varphi_2 \frac{\partial \Re_{1-2}}{\partial \varphi_1} + \varphi_4 \frac{\partial \Re_{1-4}}{\partial \varphi_1}$$
(5)

where φ is the mesh flux, \Re_1 notation represents the sum of the

reluctances in the first mesh flux while \Re_{1-2} notation represents the mutual reluctances of 1st and 2nd mesh fluxes. The other functions and their derivatives are described in similar manner. At the end of Newton-Raphson method, magnetic flux densities for all reluctances and skip depth are calculated.

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Static air gap magnetic flux density variation is defined as [18]

$$B_{s}(\theta) = \begin{cases} B_{a}e^{(\theta - 0.5\beta)/g}, & -\pi/2p < \theta < -0.5\beta \\ B_{a}, & -0.5\beta \le \theta \le 0.5\beta \\ B_{a}e^{(-\theta - 0.5\beta)/g}, & 0.5\beta < \theta < \pi/2p \end{cases}$$
(6)

where θ is electrical brake disc position, *p* is the number of pole pairs, $\beta = \pi (x_3-2x_{PM})/x_3p$ and B_a *is* the amplitude of the air gap magnetic flux density which is defined as

$$B_{a} = \left(B_{y5-4,x1-PM} + B_{y5-6,x1-PM}\right)/2 \tag{7}$$

The dynamic magnetic field in the air gap consists of static magnetic field and the induced magnetic field defined by [19], [20]

$$B_{\rm d}\left(\theta\right) = B_{\rm s}\left(\theta\right) + B_{\rm i}\left(\theta\right) \tag{8}$$

where B_i is induced magnetic flux density by eddy currents. According to Faraday's Law, the induced current density is written as

$$J(r,\theta) = \sigma(T)r\omega B_{d}(\theta)$$
(9)

where $\sigma(T)$ is conductivity of the brake disc as a function of temperature (given in the Appendix), *r* is the mean radius and ω is the angular speed. Based on Ampere's Law, the current linkage generated by eddy current path can be explained by

$$\oint H dl = \int_{\theta_1}^{\theta_2} \int_{0}^{\theta_2} J(r,\theta) r dz d\theta$$
(10)

where δ is the skin depth defined by

$$\delta = \sqrt{\frac{2H_0}{\sigma(T)\omega B_{\rm st}(H_0,T)}} \tag{11}$$

where H_0 is the peak value of applied surface magnetic field and B_{st} is the saturation value of magnetic flux density. It should be pointed out that contrary to the classical skin depth theory for nonsaturating materials, if time varying magnetic field is applied to conductive region, the magnetic flux density will be either $-B_{st}$ or $+B_{st}$. [21], [22].

The permeability of the stator and the brake disc is much greater than the air gap. Therefore, the current linkage drops of these parts can be neglected. Thus, (10) can be expressed by

$$B_{i}(\theta) = m \int_{\theta_{i}}^{s} (B_{s}(\theta) + B_{i}(\theta)) d\theta$$
(12)

where *m* is $\mu_0 \sigma(T) \omega \delta r^2/2g$. Differentiating (12) with respect to the brake disc position, θ , the below differential formula is obtained.

$$\frac{\mathrm{d}B_{\mathrm{i}}(\theta)}{\mathrm{d}\theta} - mB_{\mathrm{i}}(\theta) - mB_{\mathrm{s}}(\theta) = 0.$$
(13)

 θ_{γ}

General solution of B_i can be derived as

$$B_{i}(\theta) = \begin{cases} e^{m\theta}c_{1} + e^{m\theta}\frac{amB_{a}e_{1}}{-1 + am}, & -\pi/2p < \theta < -0.5\beta \\ e^{m\theta}c_{2} + e^{m\theta}B_{a}e_{2}, & -0.5\beta \le \theta \le 0.5\beta. \\ e^{m\theta}c_{3} + e^{m\theta}\frac{amB_{a}e_{3}}{1 + am}, & 0.5\beta < \theta < \pi/2p \end{cases}$$
(14)

where the parameters of e_1 , e_2 and e_3 are given in the Appendix. According to Ampere's Law, the total current is equal in the intervals $[-\pi/2p, \theta_0]$ and $[\theta_0, \pi/2p]$ defined as

$$\int_{-\pi/2p}^{\theta_0} \int_{0}^{\delta} J(r,\theta) r dz d\theta = \int_{\theta_0}^{\pi/2p} \int_{0}^{\delta} J(r,\theta) r dz d\theta$$
(15)

where θ_0 is an special point where $e^{m\theta}c_2 + e^{m\theta}B_ae_2 = 0$. Other interface conditions are defined by

$$e^{m\theta}c_2 + e^{m\theta}B_ae_2 = e^{m\theta}c_1 + e^{m\theta}\frac{amB_ae_1}{-1+am}, \quad \theta = -0.5\beta$$
 (16)

$$e^{m\theta}c_2 + e^{m\theta}B_a e_2 = e^{m\theta}c_3 + e^{m\theta}\frac{amB_a e_3}{1+am}, \quad \theta = 0.5\beta$$
 (17)

The coefficients of c_1 , c_2 , and c_3 are obtained by solving the interface conditions and defined by

$$c_{1} = B_{a} \left(e^{-m\theta_{0}} + \frac{e^{m0.5\beta} - ame^{\frac{-(\pi/2) + am(\pi/2) + 0.5\beta p}{ap}}}{am - 1} \right)$$
(18)

$$c_{2} = B_{a} \left(e^{-m\theta_{0}} - e^{m0.5\beta} \right)$$
(19)
$$c_{3} = B_{a} \left(e^{-m\theta_{0}} - e^{-m0.5\beta} \right)$$
(20)

By solving (15) with interface conditions, θ_0 is calculated as

$$\theta_0 = -m^{-1} \ln\left(\cosh(\frac{m\pi}{2p} - m0.5\beta) / \cosh(\frac{m\pi}{2p})\right)$$
(21)

Finally, dynamic air gap magnetic flux density is defined by $\int_{-\infty}^{\infty} dr dr$

$$B_{\rm d}\left(\theta\right) = \begin{cases} \frac{B_{\rm a}e^{-m_0}d_1}{am-1}, & -\pi/2p < \theta < -0.5\beta \\ B_{\rm a}e^{m(\theta-\theta_0)}, & -0.5\beta \le \theta \le 0.5\beta \\ \frac{B_{\rm a}d_2}{am+1}, & 0.5\beta < \theta < \pi/2p \end{cases}$$
(22)

where the parameters of d_1 and d_2 are defined by

$$d_{1} = e^{-m(\theta + \theta_{0} + 0.5\beta)} - e^{(0.5\beta + am\theta_{0} + \theta)/a} + e^{m\theta} (am - 1)$$
(23)

$$d_{2} = e^{(0.5\beta - \theta)/a} - e^{k(-0.5\beta + \theta)} + e^{m(\theta - \theta_{0})} \left(am + 1\right)$$
(24)

After the dynamic air gap magnetic flux density is obtained, the braking power can be calculated by

$$P_{\rm br} = \frac{1}{\sigma(T)} \int_{-\pi/2p}^{\pi/2p} \int_{0}^{\delta} \int_{r_{\rm i}}^{r_{\rm o}} J(r,\theta)^2 r dr dz d\theta$$

$$= \frac{1}{\sigma(T)} \int_{-\pi/2p}^{\pi/2p} \int_{0}^{\delta} \int_{r_{\rm i}}^{r_{\rm o}} \left[\sigma(T) r \omega \left(B_{\rm s}(T) + B_{\rm i}(T) \right) \right]^2 r dr dz d\theta$$
(25)

where r_i and r_o are inner and outer radius respectively. Finally, the braking torque is obtained by

$$T_{\rm br} = P_{\rm br} \,/\, \omega \tag{26}$$

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B. Thermal Analysis

Thermal analysis of the proposed AF-PMA-ECB is performed by nonlinear 3D thermal lumped parameter model. The braking power calculated in the magnetic analysis is used as an input in the thermal analysis. The schematic drawing and thermal resistance network of the proposed ECB are given in Fig. 5. Due to the AF geometry and 3D temperature flow, Tequivalent circuit modeling is preferred [23], [24]. The thermal network consists of eleven nodes: 1) disc, 2) air gap, 3) PM, 4) teeth tips, 5) stator teeth, 6) slot windings, 7) yoke, 8) frame and 9-10-11) shaft.

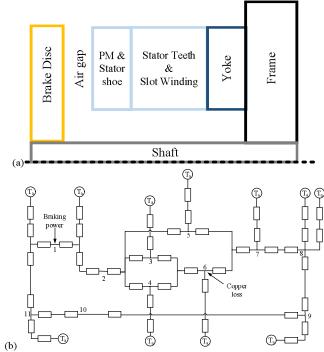


Fig. 5. (a) The schematic drawing and (b) thermal resistance network of the proposed AF-PMA-ECB. $T_{\rm a}$ represents the ambient temperature.

For the transient analysis, the temperature rise of each node is calculated with the matrix equation defined as

$$\frac{\mathrm{d}}{\mathrm{d}t}(\Delta \mathbf{T}) = \mathbf{C}^{-1} \left(\mathbf{P} - \mathbf{G} \cdot \Delta \mathbf{T} \right)$$
(27)

where t is the time, $\Delta \mathbf{T}$ is the temperature rise vector, **C** is the thermal capacitance matrix, **P** is the power loss vector and **G** is the thermal conductance matrix. The thermal capacitance matrix is defined by

$$\mathbf{C} = \begin{bmatrix} Cp_1 & 0 & 0 & \cdots & 0 \\ 0 & Cp_2 & 0 & \cdots & 0 \\ 0 & 0 & Cp_3 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \cdots & Cp_n \end{bmatrix}$$
(28)

where n is the node number and C is the thermal capacitance of an element defined by

$$\boldsymbol{C}\boldsymbol{p} = \boldsymbol{c}\left(\boldsymbol{T}\right)\boldsymbol{\rho}_{\mathrm{d}}\left(\boldsymbol{T}\right)\boldsymbol{V} \tag{29}$$

where c is the specific heat, ρ_d is the mass density (given in the Appendix) and V is the volume. The power loss vector is defined by

$$P = \begin{bmatrix} P_1 & \cdots & P_6 & \cdots & P_n \end{bmatrix}$$
(30)

Here, P_1 is braking power calculated in (25) and P_6 is the copper loss defined by

$$P_{\rm Cu} = J^2 A_{\rm Cu} l_{\rm Cu} \rho_{\rm Cu} \left(T\right) \tag{31}$$

where \underline{A}_{Cu} is the cross sectional area of the copper, l_{Cu} is total length of the copper wire and ρ_{Cu} is the resistivity of copper (given in the Appendix). The thermal conductance matrix is defined by

$$G = \begin{bmatrix} \sum_{i=1}^{n} \frac{1}{R_{i,i}} & -\frac{1}{R_{1,2}} & -\frac{1}{R_{i,3}} & \cdots & -\frac{1}{R_{i,n}} \\ -\frac{1}{R_{2,1}} & \sum_{i=1}^{n} \frac{1}{R_{2,i}} & -\frac{1}{R_{2,3}} & \cdots & -\frac{1}{R_{2,n}} \\ -\frac{1}{R_{3,1}} & -\frac{1}{R_{3,2}} & \sum_{i=1}^{n} \frac{1}{R_{3,i}} & \cdots & -\frac{1}{R_{3,n}} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ -\frac{1}{R_{n,1}} & -\frac{1}{R_{n,2}} & -\frac{1}{R_{n,3}} & \cdots & \sum_{i=1}^{m} \frac{1}{R_{n,i}} \end{bmatrix}$$
(32)

Here, R is the thermal resistance. The subscripts in R define the total thermal resistance between the two nodes. In the nonlinear thermal modelling, conduction and convection are considered and radiation is neglected. The conductive and convective thermal resistances are defined by

$$R = \frac{l}{\lambda(T)S}$$
(33)

$$R_{\rm conv} = \frac{1}{hA} \tag{34}$$

where l is the length of the heat flow, λ is the thermal conductivity (parameters are given in the Appendix), h is the convection coefficient, S is the cross-sectional area. The convection coefficient in the air gap between the brake disc and stator is calculated by

$$h_{\rm air} = \frac{Nu\lambda_{\rm air}}{g} \tag{35}$$

where λ_{air} is the thermal conductivity of air and *Nu* is the Nusselt numbers. The convection coefficient between the brake disc and the ambient has radial and axial directions. The convection coefficient in axial direction is calculated by the following set of equations [23], [24]

$$h_{\rm ax} = \frac{N u_{\rm ax} \lambda_{\rm air}}{r_{\rm o}}$$
(36)

$$Nu_{\rm ax} = \frac{Pr \ Re_{\rm ax} Cm_{\rm ax}}{2\pi}$$
(37)

Laminar Flow
$$\left(Re_{ax} < 3 \times 10^5\right)$$
, $Cm_{ax} = \frac{3.87}{Re_{ax}^{0.5}}$ (38)

Turbulent Flow
$$\left(Re_{ax} > 3 \times 10^5\right)$$
, $Cm_{ax} = \frac{0.146}{Re_{ax}^{0.2}}$ (39)

where, Re is the Reynolds number, Cm_{ax} is the axial flow coefficient and Pr is the average Prandt number. The convection coefficient in radial direction is calculated by the following set of equations

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$$Gr = \frac{\rho_{d_a u}^2 F_g \beta_c \Delta T \left(L_c\right)^2}{\mu_d^2}$$
(40)

$$Nu_{\rm rad} = 0.18 \left(Pr \left(0.5 Re_{\rm rad}^2 + Gr \right) \right)^{0.315}$$
(41)

where Gr is the Grashof number, ρ_{d_a} is the mass density of the air, F_g is the gravitational attraction force, β_c coefficient of cubical expansion, L_c is the characteristic length. Finally, the convection coefficient in radial direction is defined as

$$h_{\rm rad} = \frac{N u_{\rm rad} \lambda_{\rm air}}{2r_{\rm c}} \tag{42}$$

The thermal analysis has two inputs as the braking loss and the copper loss and starts with setting the initial parameters. The coefficients are derived by current temperature value and (26) is solved to calculate the temperature rise for each nodes. The critical temperature of the brake disc and the magnets are checked and if either is larger than the desired value, the analysis is stopped. If not, the calculated temperature data is shared with the magnetic analysis.

IV. 3D-FEA OF THE AF-PMA-ECB

Electromagnetic 3D-FEA of the AF-PMA-ECB is investigated to prove the proposed magnetic analysis. The symmetric 3D-FE model, mesh structure and magnetic flux density profile for maximum excitation current of 7.7A DC (equals to 10 A/mm² in the coil) are given in Fig. 6. Air gap is meshed with 2 layers to increase the accuracy of analyses. The magnetic flux density value of the yoke, the tooth and the brake disc are also illustrated in the figure. The comparison of the magnetic flux density values obtained from 3D-FEA and the proposed static magnetic analysis is summarized in Table II. Close results are obtained between the 3D-FEA and the proposed static magnetic analysis even in the saturation region.

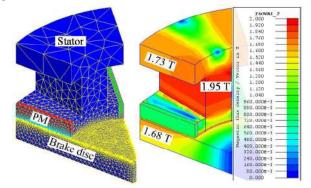


Fig. 6. 3D-FE model, mesh structure and magnetic flux density profile of the AF-PMA-ECB for 7.7 A DC excitation current at the time zero.

 TABLE II

 COMPARISON OF RESULTS OBTAINED FROM 3D-FEA AND STATIC MAGNETIC

 ANAL VSIS

ANAL Y SIS					
	Yoke	Tooth	Brake Disc		
Magnetic Analysis	1.77 T	1.88 T	1.65 T		
3D-FEA	1.73 T	1.95 T	1.68 T		

The key component to calculate the braking torque is definitely the air gap magnetic flux density variation. In ECBs, the air gap magnetic flux density alters by the speed because of armature reaction phenomenon. The influence of armature reaction on the air gap magnetic flux density has a great impact on braking torque. To verify the proposed magnetic analysis, 3D-FEAs are carried out for various speeds (500, 1000 and 2000 min⁻¹) at maximum excitation current. Air gap magnetic flux density variations obtained from 3D-FEA and proposed magnetic analysis are given in Fig. 7. The static air gap magnetic flux density variation is given in Fig. 7-(a) and the magnetic flux density variations obtained from proposed approach and 3D-FEA are in perfect agreement. Air gap magnetic flux density variations at the speeds of 500, 1000 and 2000 min⁻¹ are provided in Fig. 7-(b-c-d). Air gap flux density variations obtained from magnetic analysis are in good agreement with 3D-FEA results. The effect of armature reaction is getting dominant as the speed is increased and it is clearly seen by the figure. The average magnetic flux density values are given in Table III. The average magnetic flux density values obtained from analytical analysis and 3D-FEA are in bigh harmony. As the speed is increased, the average magnetic flux density closes to zero. It can be stated that the proposed magnetic analysis provides practical dynamic air gap magnetic flux density variation.

TABLE III Average Magnetic Flux Density Value Obtained from 3D-FEA and Magnetic Analysis

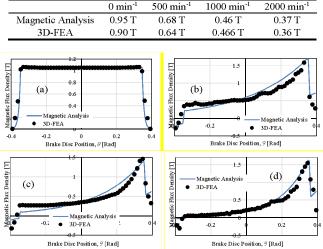


Fig. 7. Dynamic air gap magnetic flux density variations obtained from magnetic analysis and 3D-FEA at (a) zero speed, (b) the speed of 500 min⁻¹, (c) the speed of 1000 min^{-1} and (d) the speed of 2000 min^{-1} .

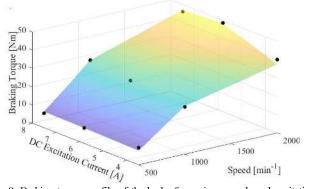


Fig. 8. Braking torque profile of the brake for various speeds and excitation currents. The black points represent the 3D-FEA results.

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Various simulations are also performed at different speeds (500, 1000 and 2000 min⁻¹) and excitation currents (3.5, 5.8 A DC) so as to obtain the braking torque profile for validation. The braking profile obtained from proposed magnetic analysis approach is given in Fig. 8. 3D-FEAs results are also provided in the figure with black spots and it is seen that the proposed approach works extremely well for the proposed ECB system.

V. PROTOTYPE, TEST SETUP AND EXPERIMENTAL STUDIES

The prototype of the investigated AF-PMA-ECB was manufactured and is illustrated in Fig. 9 where stator with magnets, windings and the brake disc are clearly seen. The same parameters provided in Table I are used in the prototype ECB. The prototype has 180-to-100 mm outer-to-inner diameter. ST37, low carbon steel, material is used as the stator core and the brake disc material in the prototype. The manufactured brake disc has 15 mm thickness and outer diameter of 180mm. Concentrated windings are employed and connected in series. N30UH type magnets were glued in the slot openings.

The test setup with temperature measuring and capturing tools is displayed in Fig. 10. The test setup consists of a drive motor, a prototype and a torque meter. The drive motor has 47.5 Nm rated torque with 3000 min⁻¹ of maximum speed. The torque meter has 0-50 Nm torque range with high accuracy. Thermal camera is used in the test system to capture the thermal images of the prototype during braking. During the tests, a thermal camera is located exactly in the same place as shown in the figure. The approximate distance is about 120 cm. A thermocouple is also attached to the magnet surface by special thermal paste and two laser-spot infrared thermometers are used to measure the temperature rise in the brake disc and the PM. The dSPACE control board is also used in the system to record the temperature rise, excitation current variation and the braking torque variations.

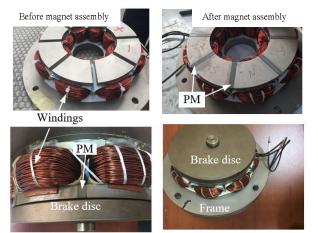


Fig. 9. Prototype of the investigated AF-PMA-ECB.

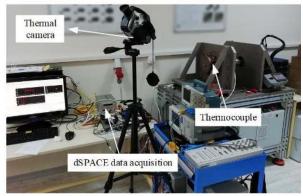
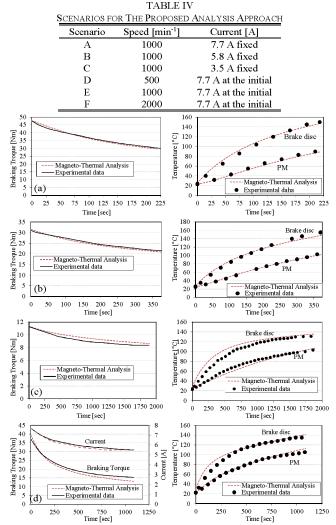


Fig. 10. Test setup with temperature measuring and capturing tools.

Six different test scenarios are explored during the tests. The studied experimental cases are shown in Table IV. The excitation current is fixed in the first 3 scenarios (A, B and C), while the excitation current is varied with temperature rise in scenarios D, E and F. The maximum temperature for the brake disc is limited to 160 °C, for the magnets it is limited to 120 °C during the tests. The magnets used in the prototype have maximum working temperature of 180 °C.



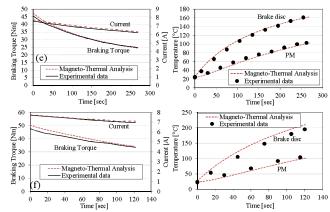


Fig. 11. Braking torque and temperature variation for (a) Scenario-A, (b) Scenario-B, (c) Scenario-C, (d) Scenario-D, (e) Scenario-E and (f) Scenario-F. (Current variation is valid for Scenario-D-E-F.)

Braking torque and temperature variations for all scenarios are provided in Fig. 11. Generally speaking, the results obtained from the experiments and the proposed coupled magnetothermal analysis are in good agreement. The following comments are drawn from the experiments:

- Scenario-A (Fig. 11-a): The braking torque is initially 47.5 Nm and is reduced to 30 Nm after 225 s. The brake disc temperature rise is 128.7 °C and the permanent magnet temperature rise is 77.7 °C after 225 s. The ECB can work nearly 225 s with this scenario.
- Scenario-B (Fig. 11-b): The braking torque is reduced from 31 Nm to 21.5 Nm after 376 s. The temperature rise for the brake disc is 129 °C and for the magnets it is 77.7 °C. The ECB can run approximately 376 s with this scenario.
- Scenario-C (Fig. 11-c): Initially 11.3 Nm of braking torque is reduced to 8.4 Nm after 1890 s. The temperature rise for the brake disc is 106 °C and for the magnets it is 80 °C. The proposed AF-PMA-ECB can work approximately 2000 s in this scenario.
- Scenario-D (Fig. 11-d): The braking torque is initially 37.7 Nm and reduced to 15.3 Nm after 1088 s. The brake disc temperature rise is 112.5 °C and the magnets temperature rise is 82.5 °C after 1088 s. The excitation current is reduced by 28.6% during this period. The proposed ECB can run nearly 1100 s with this scenario.
- Scenario-E (Fig. 11-e): 45.5 Nm of braking torque at the beginning is reduced to 24.8 Nm after 270 s. The temperature rise for the disc is 136 °C and for the magnets it is 80 °C. The reduction of the excitation current is 17.7%. The brake can work approximately 270 s in this scenario.
- Scenario-F (Fig. 11-f): The braking torque is initially 47.5 Nm and is reduced to 33.4 Nm after 121 s. The temperature rise for the brake is 171 °C and for the magnets it is 81 °C. The excitation current reduction is 8.4%. The brake can operate approximately 100 s with this scenario. Thermal images of the prototype during Scenario-F operation are captured to check the proposed analysis approach (Fig. 12). The thermal images are taken in the beginning of the test, after 40, 74 and 120 seconds of braking, respectively. It is observed that the temperature values obtained from the

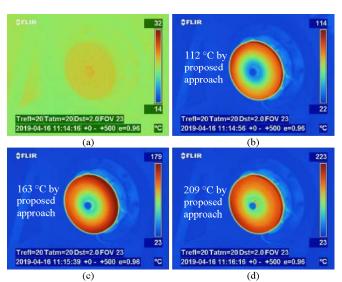


Fig. 12. Thermal image of the prototype for Scenario-F: (a) in the beginning, (b) after 40 seconds, (c) after 74 seconds and (d) after 120 seconds of braking.

The proposed approach results are also given in Table V. To give some comparison; the braking torque reduction percentage for Scenario-A is calculated to be 37.64% by the proposed approach while the experimental reduction was 36.8%. The temperature rise in the brake disc for Scenario-D is calculated to be 104 °C by the proposed approach and was found to be 112.5 °C during experiments. The temperature rise in the magnets for Scenario-F was obtained to be 78 °C by the proposed approach and 81 °C found in the experiments. It has to be marked that the braking torque obtained from the proposed approach is nearly 7.5% lower with respect to the experiment results in the beginning of Scenario-F. The main reason for the error is the high rotational speed and the 3D armature effect owing to the physical structure of the brake disc. Overall, it can be stated that the proposed approach works well in determining the total braking time and the temperature rise in the brake disc and the magnets.

TABLE V

PERFORMANCE OF SCENARIOS BASED ON PROPOSED APPROACH							
Scenario	Braking Time	Braking Torque Reduction %	Current Reduction %	Temperature Rise in Brake Disc	Temperature Rise in Magnets		
А	225 s	37.64 %	-	121.37	64.95		
В	376 s	33.7 %	-	121	76		
С	1890 s	24.1%	-	107	78.3		
D	1088 s	66.8%	28.6%	104	88.7		
Е	270 s	48.2%	17.7%	137	78.5		
F	121 s	32.4 %	8.4 %	183	78		

VI. CONCLUSIONS

In this paper, a new analytic magneto-thermal analysis approach is proposed for AF-PMA-ECB to define the working limitations at high temperature conditions. The proposed coupled analysis is a simple and effective approach and is an alternative option to time-consuming and complex multiphysics FEA software. A prototype AF-PMA-ECB was built and tested for several test scenarios. The results indicate that the proposed approach works well and provides very fast solutions compared

thermal images agree well with the results of the proposed coupled magneto-thermal analysis approach.

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2

to complex FEA simulations. The total operating time, maximum temperature rise and braking torque variation can accurately be predicted by the proposed approach. It can also be stated that the proposed approach can be implemented to any kind of PM-based electrical machine at high temperature conditions provided that magnetic modeling is modified based on magnetic and thermal modeling should be revised by the machine geometry.

APPENDIX

$$e_{1} = e^{\frac{\pi}{2p} - 0.5\beta} - e^{\frac{\pi}{a} - \frac{\pi}{2p}} - e^{\frac{\theta + 0.5\beta}{a} - m\theta}, \quad e_{2} = e^{m0.5\beta} - e^{-m\theta},$$
$$e_{3} = e^{-0.5\beta m} - e^{-m\theta + \frac{-\theta + 0.5\beta}{a}}$$

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The magnetic flux density variation and magnetic field strength of low carbon steel:

$$B_{\rm s}({\rm T}) = -0.003067e^{0.0091{\rm T}} + 1.63e^{0.0001313{\rm T}}$$

 $H(B(T)) = 182.1e^{1.534B(T)} + 0.07098e^{6.469B(T)}$

Conductivity of low carbon steel:

 $\sigma(T) = (1+0.003(T-20))^{-1}10^{7}$

Resistivity of copper: $\rho_{Cu}(T) = 16.8(1+0.00393(T-20))10^{-9}$

Specific heat capacity of steel, shaft and copper:

 $c_{\text{shaft}}(T) = 450 + 0.28T - 2.91T^2 10^{-4} + 1.34T^3 10^{-7}$

 $c_{\text{steel}}(T) = 425 + 0.773T - 1.69T^2 10^{-3} + 2.22T^3 10^{-6}$

 $c_{\rm Cu}(T) = 383 + 0.158T - 0.2T^2 10^{-3} + 0.2T^3 10^{-6}$

Thermal conductivity of steel, shaft and copper:

 $\lambda_{\text{steel}}(T) = 54 - 0.0333T$, $\lambda_{\text{shaft}}(T) = 14.6 - 0.0127T$,

 $\lambda_{cu}(T) = 400.56 - 0.0683T$

Mass density of low carbon steel:

 $\rho_{\rm d}(T) = 8050 / (1 + 10.8 \Delta T 10^{-6})$ in Kelvin.

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