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Observer-based Event-triggered Boundary Control of a Class of Reaction-Diffusion PDEs

Bhathiya Rathnayake, Mamadou Diagne, *Member, IEEE*, Nicolás Espitia, and Iason Karafyllis

Abstract—This paper presents an observer-based event-triggered boundary control strategy for a class of reaction-diffusion PDEs with Robin actuation. The observer only requires boundary measurements. The control approach consists of a backstepping output feedback boundary controller, derived using estimated states, and a dynamic triggering condition, which determines the time instants at which the control input needs to be updated. It is shown that under the proposed observer-based event-triggered boundary control approach, there is a minimal dwell-time between two triggering instants independent of initial conditions. Furthermore, the well-posedness and the global exponential convergence of the closed-loop system to the equilibrium point are established. A simulation example is provided to validate the theoretical developments.

Index Terms—Backstepping control design, event-triggered control, linear reaction-diffusion systems, output feedback.

I. INTRODUCTION

EVENT-triggered control (ETC) is a control implementation technique that closes the feedback loop only if an event indicates that the control input error exceeds an appropriate threshold. When an event occurs, the controller computes the control value and transmits it to the actuator, completing the feedback path. Therefore, unlike periodic sampled-data control [1], [2], ETC only requires the control input to be updated aperiodically (only when needed). For networked and embedded control systems, the periodic computation and transmission of the control inputs are sometimes not desirable due to task scheduling limitations and bandwidth constraints in the communication [3]–[6]. Despite the developments in aperiodic sampled-data control [7]–[10], they lack explicit criteria for selecting appropriate sampling schedules. ETC, on the other hand, provides a rigorous *resource-aware* method of implementing the control laws into digital platforms [11]–[13].

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In general, ETC consists of two main components: a feedback control law that stabilizes the system and an event-triggered mechanism, which determines when the control value has to be computed and sent toward the actuator. In the literature, one can find two main approaches to the control law design: *emulation* (e.g. [11]) and *co-design* (e.g. [5]). The former requires a pre-designed continuous feedback controller applied to the plant in a Zero-Order-Hold fashion between two event times. In co-design, the feedback control law and the event-triggered mechanism are simultaneously designed to obtain the desired stability properties. An important property that every ETC design should possess is the non-existence of *Zeno behavior* [13]; otherwise, it will lead to the triggering of an infinite number of control updates over a finite period, making the design infeasible for digital implementation. Usually, ETC designs are ensured to be Zeno-free by showing the existence of a guaranteed lower bound for the time between two consecutive events, known as the minimal dwell-time.

In the case of finite-dimensional systems, ETC has grown to be a mature field of research. During the past decade, many significant results related to ETC have been reported on systems described by linear and nonlinear ODEs in both full-state and output feedback settings (see [11], [14]–[25]). Recently, there has been a growing interest in periodic ETC [16], [22], [23], which employs a sampled-data event-triggered mechanism instead of the usual continuous-time event trigger. With this, the system states and the triggering condition need to be monitored and evaluated only at the sampling instants, making ETC more realistic for digital implementation. Despite some recent developments such as in [26]–[37], ETC strategies for PDE systems have not reached the level of maturity seen by finite-dimensional systems yet. Only recently, even the notions of solutions of linear hyperbolic and parabolic PDEs under sampled-data control have been clarified [8], [9].

One can find several recent works that employ event-triggered boundary control strategies based on the emulation approach for hyperbolic PDEs [27], [29], [33] and parabolic PDEs [32], [34], [36]. The authors of [27] propose an output feedback event-triggered boundary controller for 1-D linear hyperbolic systems of conservation laws, using Lyapunov techniques. By utilizing a dynamic triggering condition, an event-based backstepping boundary controller is designed for a coupled 2×2 hyperbolic system in [29] with Zeno-free guarantees. This work is extended in [33] to obtain an event-triggered output feedback boundary controller for a similar system. In the case of 1D reaction-diffusion PDEs with constant parameters and Dirichlet boundary actuation, the work [34]

proposes an event-triggered boundary control strategy with a suitable state-dependent triggering condition that determines when the continuous boundary backstepping controller has to be sampled/updated. By following ISS properties (w.r.t the input holding error) and small-gain arguments, it is proved the global exponential stability of the closed-loop system. Furthermore, the avoidance of the Zeno phenomena through the derivation of a uniform minimal dwell-time (independent of the initial condition) is guaranteed. To the best of our knowledge, [34] is the first work to report event-based infinite dimensional *boundary* control design for parabolic PDEs.

The design of event-triggered control strategies for PDEs appear to be more realistic, and in turn challenging, as long as both the control and the triggering condition depend on estimated states for which the need to design observers is a crucial task. Indeed, the importance of event-triggered output feedback control cannot be stressed enough as the use of full state measurements is either impossible or prohibitively expensive for many practical applications. Some studies such as [26], [38], [39] report several event-triggered output feedback designs for parabolic PDEs. All these works, nevertheless, rely on in-domain control and distributed observation. Event-based boundary control of parabolic PDEs with boundary sensing only is quite challenging, and only few works such as [32], [36] relying on modal decomposition have dealt with it. To the best of our knowledge, [32], [36] are the first contributions to come up with sampled-data and observer-based event triggered boundary control for 1D reaction-diffusion systems in the presence of time-varying input delays. While relying on modal decomposition, those contributions propose an observer-based event-triggered strategy that includes a novel switching-based dynamic triggering condition depending on the finite modes of the estimated state. The triggering condition includes a suitable waiting-time parameter (for time regularization) allowing the avoidance of the Zeno phenomena. Nevertheless, an infinite-dimensional control approach such as backstepping [40] for observer-based event-triggered boundary control of parabolic PDEs is missing. The possibility of avoiding Zeno behavior in ETC of general parabolic PDE systems with boundary actuation and boundary observation is challenging and, in general, not known. The choice of the actuation type when using backstepping is critical as both Dirichlet and Neumann actuation pose a severe impediment in establishing a minimal dwell-time and hence well-posedness and convergence results, due to unbounded local terms. However, we have identified that a class of reaction-diffusion equations with Robin actuation is conducive for event-triggered backstepping boundary control with boundary sensing. This paper proposes an observer-based event-triggered backstepping boundary controller for the class of PDEs mentioned above using emulation. The observer only requires boundary observation. The main contributions are as follows:

- We consider a class of reaction-diffusion systems with Robin actuation. We perform emulation on an observer-based backstepping boundary control design and propose a dynamic triggering condition under which Zeno behavior is excluded. It is proved the existence of a minimal-dwell-time independent of the initial conditions.

- We prove the well-posedness of the closed-loop system and its global exponential convergence to the equilibrium point in L^2 -sense.

The paper is organized as follows. Section 2 introduces the class of linear reaction-diffusion system and the continuous-time output feedback boundary control. Section 3 presents the observer-based event-triggered boundary control and some properties. In Section 4, we discuss the main results of this paper. We provide a numerical example in Section 5 to illustrate the results and conclude the paper in Section 6.

Notation: \mathbb{R}_+ is the nonnegative real line whereas \mathbb{N} is the set of natural numbers including zero. By $C^0(A; \Omega)$, we denote the class of continuous functions on $A \subseteq \mathbb{R}^n$, which takes values in $\Omega \subseteq \mathbb{R}$. By $C^k(A; \Omega)$, where $k \geq 1$, we denote the class of continuous functions on A , which takes values in Ω and has continuous derivatives of order k . $L^2(0, 1)$ denotes the equivalence class of Lebesgue measurable functions $f : [0, 1] \rightarrow \mathbb{R}$ such that $\|f\| = (\int_0^1 |f(x)|^2)^{1/2} < \infty$. Let $u : [0, 1] \times \mathbb{R}_+ \rightarrow \mathbb{R}$ be given. $u[t]$ denotes the profile of u at certain $t \geq 0$, i.e., $(u[t])(x) = u(x, t)$, for all $x \in [0, 1]$. For an interval $J \subseteq \mathbb{R}_+$, the space $C^0(J; L^2(0, 1))$ is the space of continuous mappings $J \ni t \rightarrow u[t] \in L^2(0, 1)$. $I_m(\cdot)$, and $J_m(\cdot)$ with m being an integer respectively denote the modified Bessel and (nonmodified) Bessel functions of the first kind.

II. OBSERVER-BASED BACKSTEPPING BOUNDARY CONTROL AND EMULATION

Let us consider the following 1-D reaction-diffusion system with constant coefficients:

$$u_t(x, t) = \varepsilon u_{xx}(x, t) + \lambda u(x, t), \quad (1a)$$

$$u_x(0, t) = 0, \quad (1b)$$

$$u_x(1, t) + qu(1, t) = U(t), \quad (1c)$$

and the initial condition $u[0] \in L^2(0, 1)$, where $\varepsilon, \lambda > 0$, $u : [0, 1] \times [0, \infty) \rightarrow \mathbb{R}$ is the system state and $U(t)$ is the control input.

Assumption 1: The plant's parameters q, λ , and ε satisfy the following inequality:

$$q > (\lambda + \varepsilon)/2\varepsilon.$$

Remark 1: Assumption 1 is required to avoid a trace term for which it is impossible to obtain a useful bound on its rate of change. In order to overcome this, it is required that $q - \lambda/2\varepsilon$ is sufficiently large. Furthermore, It should be mentioned that an eigenfunction expansion of the solution of (1) with $U(t) = 0$ (zero input) shows that the system is unstable when $\lambda > \varepsilon\pi^2/4$, no matter what $q > 0$ is.

We propose an observer for the system (1) using $u(0, t)$ as the available measurement/output. Note that the output is anticollocated with the input. The observer consists of a copy of the system (1) with output injection terms, which is stated

as follows:

$$\hat{u}_t(x, t) = \varepsilon \hat{u}_{xx}(x, t) + \lambda \hat{u}(x, t) + p_1(x)(u(0, t) - \hat{u}(0, t)), \quad (2a)$$

$$\hat{u}_x(0, t) = p_{10}(u(0, t) - \hat{u}(0, t)), \quad (2b)$$

$$\hat{u}_x(1, t) + q\hat{u}(1, t) = U(t), \quad (2c)$$

and the initial condition $\hat{u}[0] \in L^2(0, 1)$. Here, the function $p_1(x)$ and the constant p_{10} are observer gains to be determined. Let us denote the observer error by $\tilde{u}(x, t)$, which is defined as

$$\tilde{u}(x, t) := u(x, t) - \hat{u}(x, t). \quad (3)$$

By subtracting (2) from (1), one can see that $\tilde{u}(x, t)$ satisfies the following PDE:

$$\tilde{u}_t(x, t) = \varepsilon \tilde{u}_{xx}(x, t) + \lambda \tilde{u}(x, t) - p_1(x)\tilde{u}(0, t), \quad (4a)$$

$$\tilde{u}_x(0, t) = -p_{10}\tilde{u}(0, t), \quad (4b)$$

$$\tilde{u}_x(1, t) + q\tilde{u}(1, t) = 0. \quad (4c)$$

Proposition 1: Subject to Assumption 1 and the invertible backstepping transformation

$$\tilde{u}(x, t) = \tilde{w}(x, t) - \int_0^x P(x, y)\tilde{w}(y, t)dy, \quad (5)$$

where

$$\begin{aligned} P(x, y) = & \frac{q\lambda/\varepsilon}{\sqrt{\lambda/\varepsilon + q^2}} \\ & \times \int_0^{x-y} e^{-q\tau/2} I_0\left(\sqrt{\lambda(2-x-y)(x-y-\tau)/\varepsilon}\right) \\ & \times \sinh\left(\frac{\sqrt{\lambda/\varepsilon + q^2}}{2}\tau\right) d\tau \\ & - \frac{\lambda}{\varepsilon}(1-y) \frac{J_1\left(\sqrt{\lambda((1-y)^2 - (1-x)^2)/\varepsilon}\right)}{\sqrt{\lambda((1-y)^2 - (1-x)^2)/\varepsilon}}, \end{aligned} \quad (6)$$

for $0 \leq y \leq x \leq 1$, then the observer error system (4) with the gains $p_1(x)$ and p_{10} chosen as

$$p_1(x) = \varepsilon P_y(x, 0), \quad p_{10} = P(0, 0) = -\frac{\lambda}{2\varepsilon}, \quad (7)$$

gets transformed to the following globally L^2 -exponentially stable observer error target system

$$\tilde{w}_t(x, t) = \varepsilon \tilde{w}_{xx}(x, t), \quad (8a)$$

$$\tilde{w}_x(0, t) = 0, \quad (8b)$$

$$\tilde{w}_x(1, t) = -q\tilde{w}(1, t). \quad (8c)$$

Proof: See Appendix A.

The inverse transformation of (5) can be shown to be as follows:

$$\tilde{w}(x, t) = \tilde{u}(x, t) + \int_0^x Q(x, y)\tilde{u}(y, t)dy, \quad (9)$$

where $Q(x, y)$ is

$$\begin{aligned} Q(x, y) = & \frac{q\lambda/\varepsilon}{\sqrt{-\lambda/\varepsilon + q^2}} \\ & \times \int_0^{x-y} e^{-q\tau/2} J_0\left(\sqrt{\lambda(2-x-y)(x-y-\tau)/\varepsilon}\right) \\ & \times \sinh\left(\frac{\sqrt{-\lambda/\varepsilon + q^2}}{2}\tau\right) d\tau \\ & - \frac{\lambda}{\varepsilon}(1-y) \frac{J_1\left(\sqrt{\lambda((1-y)^2 - (1-x)^2)/\varepsilon}\right)}{\sqrt{\lambda((1-y)^2 - (1-x)^2)/\varepsilon}}, \end{aligned} \quad (10)$$

for $0 \leq y \leq x \leq 1$.

Proposition 2: The invertible backstepping transformation

$$\hat{w}(x, t) = \hat{u}(x, t) - \int_0^x K(x, y)\hat{u}(y, t)dy, \quad (11)$$

where

$$K(x, y) = -\frac{\lambda}{\varepsilon}x \frac{I_1\left(\sqrt{\lambda(x^2 - y^2)/\varepsilon}\right)}{\sqrt{\lambda(x^2 - y^2)/\varepsilon}}, \quad (12)$$

for $0 \leq y \leq x \leq 1$, and a control law $U(t)$ chosen as

$$U(t) = \int_0^1 \left(rK(1, y) + K_x(1, y)\right)\hat{u}(y, t)dy, \quad (13)$$

map the system (2) with the gains $p_1(x)$ and p_{10} chosen as in (7), into the following target system:

$$\hat{w}_t(x, t) = \varepsilon \hat{w}_{xx}(x, t) + g(x)\hat{w}(0, t), \quad (14a)$$

$$\hat{w}_x(0, t) = -\frac{\lambda}{2\varepsilon}\hat{w}(0, t), \quad (14b)$$

$$\hat{w}_x(1, t) = -r\hat{w}(1, t), \quad (14c)$$

with

$$g(x) = p_1(x) - \frac{\lambda}{2}K(x, 0) - \int_0^x K(x, y)p_1(y)dy, \quad (15)$$

and

$$r = q - \frac{\lambda}{2\varepsilon}. \quad (16)$$

Proof: See Appendix B.

The inverse transformation of (11) can be shown to be as follows:

$$\hat{u}(x, t) = \hat{w}(x, t) + \int_0^x L(x, y)\hat{w}(y, t)dy, \quad (17)$$

where

$$L(x, y) = -\frac{\lambda}{\varepsilon}x \frac{J_1\left(\sqrt{\lambda(x^2 - y^2)/\varepsilon}\right)}{\sqrt{\lambda(x^2 - y^2)/\varepsilon}}, \quad (18)$$

for $0 \leq y \leq x \leq 1$.

Proposition 3: Subject to Assumption 1, the closed-loop system which consists of the plant (1) and the observer (2) with the continuous-time control law (13), is globally exponentially stable in L^2 -sense.

Proof: See Appendix C.

A. Emulation of the Observer-based Backstepping Boundary Control

We strive to stabilize the closed-loop system containing the plant (1) and the observer (2) while sampling the continuous-time controller $U(t)$ given by (13) at a certain sequence of time instants $(t_j)_{j \in \mathbb{N}}$. These time instants will be given a precise characterization later on based on an event trigger. The control input is held constant between two successive time instants and is updated when a certain condition is met. Therefore, we define the control input for $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$ as

$$U_j := U(t_j) = \int_0^1 \left(rK(1, y) + K_x(1, y) \right) \hat{u}(y, t_j) dy. \quad (19)$$

Accordingly, the boundary conditions (1c) and (2c) are modified, respectively, as follows:

$$u_x(1, t) + qu(1, t) = U_j, \quad (20)$$

$$\hat{u}_x(1, t) + q\hat{u}(1, t) = U_j. \quad (21)$$

The deviation between the continuous-time control law and its sampled counterpart, referred to as the input holding error, is defined as follows:

$$d(t) := \int_0^1 \left(rK(1, y) + K_x(1, y) \right) (\hat{u}(y, t_j) - \hat{u}(y, t)) dy. \quad (22)$$

for $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$. It can be shown that the backstepping transformation (11), applied on the system (2a),(2b),(21) between t_j and t_{j+1} , yields the following target system, valid for $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$:

$$\hat{w}_t(x, t) = \varepsilon \hat{w}_{xx}(x, t) + g(x) \tilde{w}(0, t), \quad (23a)$$

$$\hat{w}_x(0, t) = -\frac{\lambda}{2\varepsilon} \tilde{w}(0, t), \quad (23b)$$

$$\hat{w}_x(1, t) = -r\hat{w}(1, t) + d(t), \quad (23c)$$

where $g(x)$ and r are given by (15) and (16), respectively.

It is straightforward to see that the observer error system \tilde{u} for $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$ under the modified boundary conditions (20) and (21) will still be the same as (4). Therefore, the application of the backstepping transformation (5) on \tilde{u} between t_j and t_{j+1} yields the following observer error target system

$$\tilde{w}_t(x, t) = \varepsilon \tilde{w}_{xx}(x, t), \quad (24a)$$

$$\tilde{w}_x(0, t) = 0, \quad (24b)$$

$$\tilde{w}_x(1, t) = -q\tilde{w}(1, t), \quad (24c)$$

valid for $t \in [t_j, t_{j+1})$, $j \in \mathbb{N}$.

B. Well-posedness Issues

Proposition 4: For every given initial data $u[t_j], \hat{u}[t_j] \in L^2(0, 1)$, there exist unique mappings $u, \hat{u} \in C^0([t_j, t_{j+1}]; L^2(0, 1)) \cap C^1((t_j, t_{j+1}) \times [0, 1])$ with $u[t], \hat{u}[t] \in C^2([0, 1])$ which satisfy (1b),(2b),(19)-(21) for $t \in [t_j, t_{j+1})$ and (1a), (2a) for $t \in (t_j, t_{j+1})$, $x \in (0, 1)$.

Proof: The initial condition $\tilde{w}[t_j]$ for the system (24) can be uniquely determined by using (3) and (9) once $u[t_j]$ and $\hat{u}[t_j]$ are given. Therefore, from the straightforward

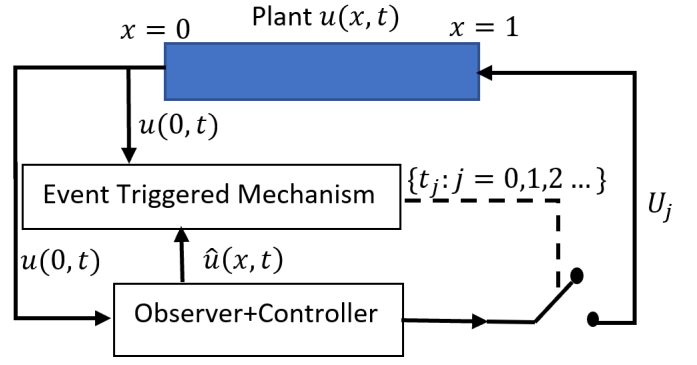


Fig. 1: Event-triggered observer-based closed-loop system.

application of Theorem 4.11 in [41], we can show that there exist unique mappings $u, \tilde{w} \in C^0([t_j, t_{j+1}]; L^2(0, 1)) \cap C^1((t_j, t_{j+1}) \times [0, 1])$ with $u[t], \tilde{w}[t] \in C^2([0, 1])$ which satisfy (1b),(19),(20),(24b),(24c) for $t \in (t_j, t_{j+1})$ and (1a),(24a) for $t \in [t_j, t_{j+1})$, $x \in (0, 1)$. Further, due to (3) and the transformation (5), there also exists a unique mapping $\hat{u} \in C^0([t_j, t_{j+1}]; L^2(0, 1)) \cap C^1((t_j, t_{j+1}) \times [0, 1])$ with $\hat{u}[t] \in C^2([0, 1])$ which satisfy (2b),(19),(21) for $t \in (t_j, t_{j+1})$ and (2a) for $t \in [t_j, t_{j+1})$, $x \in (0, 1)$. \square

III. OBSERVER-BASED EVENT-TRIGGERED BOUNDARY CONTROL

Let us now present the observer-based event-triggered boundary control approach considered in this work. It consists of two components: 1) an event-triggered mechanism which decides the time instants at which the control value needs to be sampled/updated and 2) the observer-based backstepping output feedback controller. The structure of the closed-loop system consisting of the plant, the observer-based controller, and the event trigger is illustrated in Fig. 1. The event-triggering condition is based on the square of the input holding error $d(t)$ and a dynamic variable $m(t)$ which depends on the information of the systems (23) and (24).

Definition 1: Let $\gamma, \eta, \rho, \beta_1, \beta_2, \beta_3 > 0$. The observer-based event-triggered boundary control strategy consists of two components.

- 1) (The event-trigger) The set of event times $I = \{t_0, t_1, t_2, \dots\}$ with $t_0 = 0$ forms an increasing sequence via the following rules:

- if $\{t \in \mathbb{R}_+ | t > t_j \wedge d^2(t) > -\gamma m(t)\} = \emptyset$ then the set of the times of the events is $\{t_0, \dots, t_j\}$.
- if $\{t \in \mathbb{R}_+ | t > t_j \wedge d^2(t) > -\gamma m(t)\} \neq \emptyset$ then the next event time is given by:

$$t_{j+1} = \inf\{t \in \mathbb{R}_+ | t > t_j \wedge d^2(t) > -\gamma m(t)\}, \quad (25)$$

where $d(t)$ is given by

$$d(t) = \int_0^1 \left(rK(1, y) + K_x(1, y) \right) (\hat{u}(y, t_j) - \hat{u}(y, t)) dy, \quad (26)$$

for all $t \in [t_j, t_{j+1})$ and $m(t)$ satisfies the ODE

$$\begin{aligned} \dot{m}(t) = & -\eta m(t) + \rho d^2(t) - \beta_1 \|\hat{w}[t]\|^2 \\ & - \beta_2 |\hat{w}(1, t)|^2 - \beta_3 |\tilde{w}(0, t)|^2, \end{aligned} \quad (27)$$

for all $t \in (t_j, t_{j+1})$ with $m(t_0) = m(0) < 0$ and $m(t_j^-) = m(t_j) = m(t_j^+)$.

2) (The control action) The output boundary feedback control law

$$U_j = \int_0^1 \left(rK(1, y) + K_x(1, y) \right) \hat{u}(y, t_j) dy, \quad (28)$$

for all $t \in [t_j, t_{j+1}), j \in \mathbb{N}$.

In Definition 1, it is worth noting that the initial condition for $m(t)$ in each time interval has been chosen such that $m(t)$ is time-continuous.

Proposition 4 allows us to define the solution of the closed-loop system under the observer-based event-triggered boundary control (25)-(28) in the interval $[0, F)$, where $F = \sup(I)$.

Lemma 1: Under the definition of the observer-based event-triggered boundary control (25)-(28), it holds that $d^2(t) \leq -\gamma m(t)$ and $m(t) < 0$, for $t \in [0, F)$ where $F = \sup(I)$.

Proof: From Definition 1, the events are triggered to guarantee $d^2(t) \leq -\gamma m(t), t \in [0, F)$. This inequality in combination with (27) yields:

$$\begin{aligned} \dot{m}(t) \leq & -(\eta + \gamma\rho)m(t) - \beta_1 \|\hat{w}[t]\|^2 \\ & - \beta_2 |\hat{w}(1, t)|^2 - \beta_3 |\tilde{w}(0, t)|^2, \end{aligned} \quad (29)$$

for $t \in (t_j, t_{j+1}), j \in \mathbb{N}$. Thus, considering the time-continuity of $m(t)$, we can obtain the following estimate:

$$\begin{aligned} m(t) \leq & m(t_j) e^{-(\eta + \gamma\rho)(t - t_j)} \\ & - \int_{t_j}^t e^{-(\eta + \gamma\rho)(t - \tau)} (\beta_1 \|\hat{w}[\tau]\|^2 \\ & + \beta_2 |\hat{w}(1, \tau)|^2 + \beta_3 |\tilde{w}(0, \tau)|^2) d\tau, \end{aligned} \quad (30)$$

for $t \in [t_j, t_{j+1}), j \in \mathbb{N}$. From Definition 1, we have that $m(t_0) = m(0) < 0$. Therefore, it follows from (30) that $m(t) < 0$ for all $t \in [0, t_1]$. Again using (30) on $[t_1, t_2]$, we can show that $m(t) < 0$ for all $t \in [t_1, t_2]$. Applying the same reasoning successively to the future intervals, it can be shown that $m(t) < 0$ for $t \in [0, F)$. \square

Lemma 2: For $d(t)$ given by (26), it holds that

$$d^2(t) \leq \rho_1 d^2(t) + \alpha_1 \|\hat{w}[t]\|^2 + \alpha_2 |\hat{w}(1, t)|^2 + \alpha_3 |\tilde{w}(0, t)|^2, \quad (31)$$

for some $\rho_1, \alpha_1, \alpha_2, \alpha_3 > 0$, for all $t \in (t_j, t_{j+1}), j \in \mathbb{N}$.

Proof: From (26), we can show for $t \in (t_j, t_{j+1}), j \in \mathbb{N}$

$$\dot{d}(t) = - \int_0^1 k(y) \hat{u}_t(y, t) dy, \quad (32)$$

where

$$k(y) = rK(1, y) + K_x(1, y). \quad (33)$$

Using (2a) on (32) and integrating by parts twice in the interval $(t_j, t_{j+1}), j \in \mathbb{N}$, we can show that

$$\begin{aligned} \dot{d}(t) = & -\varepsilon \int_0^1 k(y) \hat{u}_{yy}(y, t) dy - \lambda \int_0^1 k(y) \hat{u}(y, t) dy \\ & - \int_0^1 k(y) p_1(y) dy \tilde{u}(0, t) \\ = & -\varepsilon k(1) \hat{u}_x(1, t) + \varepsilon k(0) \hat{u}_x(0, t) + \varepsilon \frac{dk(x)}{dx} \Big|_{x=1} \hat{u}(1, t) \\ & - \varepsilon \frac{dk(x)}{dx} \Big|_{x=0} \hat{u}(0, t) - \varepsilon \int_0^1 \frac{d^2 k(y)}{dy^2} \hat{u}(y, t) dy \\ & - \lambda \int_0^1 k(y) \hat{u}(y, t) dy - \int_0^1 k(y) p_1(y) dy \tilde{u}(0, t). \end{aligned} \quad (34)$$

Furthermore, using (2b), (21), (26), and (28), we can show that

$$\begin{aligned} \dot{d}(t) = & -\varepsilon k(1) d(t) + \left(\varepsilon qk(1) + \varepsilon \frac{dk(x)}{dx} \Big|_{x=1} \right) \hat{u}(1, t) \\ & - \int_0^1 \left(\varepsilon \frac{d^2 k(y)}{dy^2} + \varepsilon k(1) k(y) + \lambda k(y) \right) \hat{u}(y, t) dy \\ & - \left(\frac{\lambda k(0)}{2} + \int_0^1 k(y) p_1(y) dy \right) \tilde{u}(0, t). \end{aligned} \quad (35)$$

It is worth mentioning that above we have used the fact that $dk(x)/dx \Big|_{x=0} = 0$, which can be shown using (33) and (12). Using Young's and Cauchy-Schwarz inequalities on (17), we also can show that

$$\|\hat{u}[t]\|^2 \leq \left(1 + \left(\int_0^1 \int_0^x L^2(x, y) dy dx \right)^{1/2} \right)^2 \|\hat{w}[t]\|^2, \quad (36)$$

$$\hat{u}^2(1, t) \leq 2\hat{w}^2(1, t) + 2 \int_0^1 L^2(1, y) dy \|\hat{w}[t]\|^2. \quad (37)$$

Using Young's and Cauchy-Schwarz inequalities repeatedly on (35) along with (36) and (37), we can show that

$$\begin{aligned} \dot{d}^2(t) \leq & \rho_1 d^2(t) + \alpha_1 \|\hat{w}[t]\|^2 + \alpha_2 |\hat{w}(1, t)|^2 \\ & + \alpha_3 |\tilde{w}(0, t)|^2, \end{aligned} \quad (38)$$

where

$$\rho_1 = 6\varepsilon^2 k^2(1), \quad (39)$$

$$\begin{aligned} \alpha_1 = & 3 \left(1 + \left(\int_0^1 \int_0^x L^2(x, y) dy dx \right)^{1/2} \right)^2 \\ & \times \int_0^1 \left(\varepsilon \frac{d^2 k(y)}{dy^2} + \varepsilon k(1) k(y) + \lambda k(y) \right)^2 dy \\ & + 6 \left(\varepsilon qk(1) + \varepsilon \frac{dk(x)}{dx} \Big|_{x=1} \right)^2 \int_0^1 L^2(1, y) dy, \end{aligned} \quad (40)$$

$$\alpha_2 = 6 \left(\varepsilon qk(1) + \varepsilon \frac{dk(x)}{dx} \Big|_{x=1} \right)^2, \quad (41)$$

$$\alpha_3 = 6 \left(\frac{\lambda k(0)}{2} + \int_0^1 k(y) p_1(y) dy \right)^2. \quad (42)$$

\square

IV. MAIN RESULTS

Theorem 1: Under the observer-based event-triggered boundary control in Definition 1, with $\beta_1, \beta_2, \beta_3$ chosen as

$$\beta_1 = \frac{\alpha_1}{\gamma(1-\sigma)}, \quad \beta_2 = \frac{\alpha_2}{\gamma(1-\sigma)}, \quad \beta_3 = \frac{\alpha_3}{\gamma(1-\sigma)}, \quad (43)$$

where $\alpha_1, \alpha_2, \alpha_3$ given by (40)-(42) and $\sigma \in (0, 1)$, there exists a minimal dwell-time $\tau > 0$ between two triggering times, i.e., there exists a constant $\tau > 0$ such that $t_{j+1} - t_j \geq \tau$, for all $j \in \mathbb{N}$, which is independent of the initial conditions and only depends on the system and control parameters.

Proof: From Lemma 1, we have that $d^2(t) \leq -\gamma(1-\sigma)m(t) - \gamma\sigma m(t)$, where $\sigma \in (0, 1)$ and $m(t) < 0$ for $t \in [0, F)$, where $F = \sup(I)$. Let us define the function

$$\psi(t) := \frac{d^2(t) + \gamma(1-\sigma)m(t)}{-\gamma\sigma m(t)}. \quad (44)$$

Note that $\psi(t)$ is continuous in $[t_j, t_{j+1})$. A lower bound for the inter-execution times is given by the time it takes for the function ψ to go from $\psi(t_j)$ to $\psi(t_{j+1}^-) = 1$, where $\psi(t_j) < 0$, which holds since $d(t_j) = 0$. Here t_{j+1}^- is the left limit at $t = t_{j+1}$. Therefore, by the intermediate value theorem, there exists a $t'_j > t_j$ such that $\psi(t'_j) = 0$ and $\psi(t) \in [0, 1]$ for $t \in [t'_j, t_{j+1}^-]$. The time derivative of ψ on $[t'_j, t_{j+1})$ is given by

$$\dot{\psi}(t) = \frac{2d(t)\dot{d}(t) + \gamma(1-\sigma)\dot{m}(t)}{-\gamma\sigma m(t)} - \frac{\dot{m}(t)}{m(t)}\psi(t). \quad (45)$$

From Young's inequality, we have that

$$\dot{\psi}(t) \leq \frac{d^2(t) + \dot{d}^2(t) + \gamma(1-\sigma)\dot{m}(t)}{-\gamma\sigma m(t)} - \frac{\dot{m}(t)}{m(t)}\psi(t). \quad (46)$$

Using Lemma 2 and (27), we can show that

$$\begin{aligned} \dot{\psi}(t) &\leq \frac{(1 + \rho_1 + \gamma(1-\sigma)\rho)d^2(t)}{-\gamma\sigma m(t)} + \frac{(1-\sigma)\eta}{\sigma} \\ &+ \eta\psi(t) + \frac{\gamma\sigma\rho d^2(t)}{-\gamma\sigma m(t)}\psi(t) + \frac{(\alpha_1 - \gamma(1-\sigma)\beta_1)\|\hat{w}[t]\|^2}{-\gamma\sigma m(t)} \\ &+ \frac{(\alpha_2 - \gamma(1-\sigma)\beta_2)|\hat{w}(1, t)|^2}{-\gamma\sigma m(t)} \\ &+ \frac{(\alpha_3 - \gamma(1-\sigma)\beta_3)|\tilde{w}(0, t)|^2}{-\gamma\sigma m(t)} \\ &+ \frac{\beta_1\|\hat{w}[t]\|^2 + \beta_2|\hat{w}(1, t)|^2 + \beta_3|\tilde{w}(0, t)|^2}{m(t)}\psi(t). \end{aligned} \quad (47)$$

Let us choose $\beta_1, \beta_2, \beta_3$ as in (43), where $\alpha_1, \alpha_2, \alpha_3$ are given by (40)-(42), respectively. Also, note that the last term in the right hand side of (47) is negative. Therefore, we have

$$\begin{aligned} \dot{\psi}(t) &\leq \frac{(1 + \rho_1 + \gamma(1-\sigma)\rho)d^2(t)}{-\gamma\sigma m(t)} + \frac{(1-\sigma)\eta}{\sigma} \\ &+ \eta\psi(t) + \frac{\gamma\sigma\rho d^2(t)}{-\gamma\sigma m(t)}\psi(t). \end{aligned} \quad (48)$$

We also can write that

$$\begin{aligned} \dot{\psi}(t) &\leq \left(1 + \rho_1 + \gamma(1-\sigma)\rho\right) \frac{(d^2(t) + \gamma(1-\sigma)m(t))}{-\gamma\sigma m(t)} \\ &+ \frac{(1-\sigma)\eta}{\sigma} + \eta\psi(t) + \gamma\sigma\rho \frac{d^2(t) + \gamma(1-\sigma)m(t)}{-\gamma\sigma m(t)}\psi(t) \\ &+ \left(1 + \rho_1 + \gamma(1-\sigma)\rho\right) \frac{(1-\sigma)}{\sigma} + \gamma\rho(1-\sigma)\psi(t). \end{aligned} \quad (49)$$

We can rewrite (49) as

$$\dot{\psi}(t) \leq a_1\psi^2(t) + a_2\psi(t) + a_3, \quad (50)$$

where

$$a_1 = \gamma\sigma\rho > 0, \quad (51)$$

$$a_2 = 1 + \rho_1 + 2\gamma(1-\sigma)\rho + \eta > 0, \quad (52)$$

$$a_3 = (1 + \rho_1 + \gamma(1-\sigma)\rho + \eta) \frac{1-\sigma}{\sigma} > 0. \quad (53)$$

By the Comparison principle, it follows that the time needed for ψ to go from $\psi(t'_j) = 0$ to $\psi(t_{j+1}^-) = 1$ is at least

$$\tau = \int_0^1 \frac{1}{a_1 s^2 + a_2 s + a_3} ds. \quad (54)$$

Therefore, $t_{j+1} - t'_j \geq \tau$. As $t_{j+1} - t_j \geq t_{j+1} - t'_j$, we can conclude that $t_{j+1} - t_j \geq \tau$. Thus, τ can be considered a lower bound for the minimal dwell-time. Note that τ is independent of initial conditions and only depends on system and control parameters. \square

Corollary 1: For every given initial data $u[0], \hat{u}[0] \in L^2(0, 1)$, there exist unique mappings $u, \hat{u} \in C^0(\mathbb{R}_+; L^2(0, 1)) \cap C^1(J \times [0, 1])$ with $u[t], \hat{u}[t] \in C^2([0, 1])$ which satisfy (1b),(2b),(19)-(21) for all $t > 0$ and (1a), (2a) for all $t > 0, x \in (0, 1)$, where $J = \mathbb{R}_+ \setminus \{t_j \geq 0, j \in \mathbb{N}\}$. The increasing sequence $\{t_j \geq 0, j \in \mathbb{N}\}$ is determined by the set of rules given in Definition 1.

Proof: This is a straightforward consequence of Proposition 4 and Theorem 4.10 in [41]. The solutions are constructed iteratively between consecutive triggering times. \square

Theorem 2: Let $\gamma, \eta > 0$ be design parameters, $\sigma \in (0, 1)$, and $g(x)$ and r be given by (15) and (16), respectively, while $\beta_1, \beta_2, \beta_3$ are chosen according to (43). Further, subject to Assumption 1, let us choose parameters $B, \kappa_1, \kappa_2, \kappa_3 > 0$ such that

$$\begin{aligned} B \left(\varepsilon \min \left\{ r - \frac{1}{2}, \frac{1}{2} \right\} - \frac{\varepsilon}{2\kappa_1} - \frac{5\lambda}{8\kappa_2} - \frac{\|g\|^2}{\kappa_3} \right) \\ - 2\beta_1 - \beta_2 > 0, \end{aligned} \quad (55)$$

$A > 0$ such that

$$A \varepsilon \min \left\{ q - \frac{1}{2}, \frac{1}{2} \right\} - \frac{5\lambda\kappa_2 B}{8} - \frac{5\kappa_3 B}{4} - \frac{5\beta_3}{2} > 0, \quad (56)$$

and ρ as

$$\rho = \frac{\varepsilon\kappa_1 B}{2}. \quad (57)$$

Then, the closed-loop system which consists of the plant (1a),(1b),(20) and the observer (2a),(2b),(21) with the event-triggered boundary controller (25)-(28) has a unique solution

and globally exponentially converges to zero, i.e., $\|u[t]\| + \|\hat{u}[t]\| \rightarrow 0$ as $t \rightarrow \infty$.

Proof: From Corollary 1, the existence and the uniqueness of solutions to the plant (1a),(1b),(20) and the observer (2a),(2b),(21) are guaranteed. Now let us show that the closed-loop system is globally L^2 -exponentially convergent to zero.

Let us choose the following candidate Lyapunov function noting that $m(t) < 0$ for $t \geq 0$:

$$V = \frac{A}{2} \int_0^1 \tilde{w}^2(x, t) dx + \frac{B}{2} \int_0^1 \hat{w}^2(x, t) dx - m(t). \quad (58)$$

Here \hat{w} and \tilde{w} are the systems described by (23) and (24), respectively. We can show that for $t \in (t_j, t_{j+1})$, $j \in \mathbb{N}$

$$\begin{aligned} \dot{V} &= -A\varepsilon q \tilde{w}^2(1, t) - A\varepsilon \int_0^1 \tilde{w}_x^2(x, t) dx \\ &\quad - r\varepsilon B \hat{w}^2(1, t) + \varepsilon B d(t) \hat{w}(1, t) + \frac{\lambda B}{2} \hat{w}(0, t) \tilde{w}(0, t) \\ &\quad - \varepsilon B \int_0^1 \hat{w}_x^2(x, t) dx + B \int_0^1 g(x) \hat{w}(x, t) dx \tilde{w}(0, t) \\ &\quad - \dot{m}(t). \end{aligned} \quad (59)$$

From Young's and Cauchy-Schwarz inequalities, we can write that

$$\varepsilon B d(t) \hat{w}(1, t) \leq \frac{\varepsilon B}{2\kappa_1} \hat{w}^2(1, t) + \frac{\varepsilon \kappa_1 B}{2} d^2(t), \quad (60)$$

$$\frac{\lambda B}{2} \hat{w}(0, t) \tilde{w}(0, t) \leq \frac{\lambda B}{4\kappa_2} \hat{w}^2(0, t) + \frac{\lambda \kappa_2 B}{4} \tilde{w}^2(0, t), \quad (61)$$

$$B \int_0^1 g(x) \hat{w}(x, t) dx \tilde{w}(0, t) \leq \frac{\|g\|^2 B}{2\kappa_3} \|\hat{w}[t]\|^2 + \frac{\kappa_3 B}{2} \tilde{w}^2(0, t). \quad (62)$$

Therefore, using (60)-(62),(27),(57), we can write (59) as

$$\begin{aligned} \dot{V} &\leq -A\varepsilon q \tilde{w}^2(1, t) - A\varepsilon \|\tilde{w}_x[t]\|^2 - r\varepsilon B \hat{w}^2(1, t) \\ &\quad + \frac{\varepsilon B}{2\kappa_1} \hat{w}^2(1, t) + \frac{\lambda B}{4\kappa_2} \hat{w}^2(0, t) + \frac{\lambda \kappa_2 B}{4} \tilde{w}^2(0, t) \\ &\quad - \varepsilon B \|\hat{w}_x[t]\|^2 + \frac{\|g\|^2 B}{2\kappa_3} \|\hat{w}[t]\|^2 + \frac{\kappa_3 B}{2} \tilde{w}^2(0, t) \\ &\quad + \beta_1 \|\hat{w}[t]\|^2 + \beta_2 \hat{w}^2(1, t) \\ &\quad + \beta_3 \tilde{w}^2(0, t) + \eta m(t). \end{aligned} \quad (63)$$

From Agmon's and Young's inequalities, we have that

$$\tilde{w}^2(0, t) \leq \hat{w}^2(1, t) + \|\tilde{w}[t]\|^2 + \|\tilde{w}_x[t]\|^2, \quad (64)$$

$$\hat{w}^2(0, t) \leq \hat{w}^2(1, t) + \|\hat{w}[t]\|^2 + \|\hat{w}_x[t]\|^2. \quad (65)$$

Therefore, we can show from (63) that

$$\begin{aligned} \dot{V} &\leq -\left(A\varepsilon q - \frac{\lambda \kappa_2 B}{4} - \frac{\kappa_3 B}{2} - \beta_3\right) \tilde{w}^2(1, t) \\ &\quad - \left(A\varepsilon - \frac{\lambda \kappa_2 B}{4} - \frac{\kappa_3 B}{2} - \beta_3\right) \|\tilde{w}_x[t]\|^2 \\ &\quad + \left(\frac{\lambda \kappa_2 B}{4} + \frac{\kappa_3 B}{2} + \beta_3\right) \|\tilde{w}[t]\|^2 \\ &\quad - \left(r\varepsilon B - \frac{\varepsilon B}{2\kappa_1} - \frac{\lambda B}{4\kappa_2} - \beta_2\right) \hat{w}^2(1, t) \\ &\quad - \left(\varepsilon B - \frac{\lambda B}{4\kappa_2}\right) \|\hat{w}_x[t]\|^2 \\ &\quad + \left(\frac{\lambda B}{4\kappa_2} + \frac{\|g\|^2 B}{2\kappa_3} + \beta_1\right) \|\hat{w}[t]\|^2 + \eta m(t). \end{aligned} \quad (66)$$

From Poincaré Inequality, we have that

$$-\|\tilde{w}_x[t]\|^2 \leq \frac{1}{2} \tilde{w}^2(1, t) - \frac{1}{4} \|\tilde{w}[t]\|^2, \quad (67)$$

$$-\|\hat{w}_x[t]\|^2 \leq \frac{1}{2} \hat{w}^2(1, t) - \frac{1}{4} \|\hat{w}[t]\|^2. \quad (68)$$

Furthermore, we have from (55) and (56) that

$$A\varepsilon - \frac{\lambda \kappa_2 B}{4} - \frac{\kappa_3 B}{2} - \beta_3 > 0, \text{ and } \varepsilon B - \frac{\lambda B}{4\kappa_2} > 0. \quad (69)$$

Therefore, using (66)-(69), we can show that

$$\begin{aligned} \dot{V} &\leq -\left(A\varepsilon\left(q - \frac{1}{2}\right) - \frac{\lambda \kappa_2 B}{8} - \frac{\kappa_3 B}{4} - \frac{\beta_3}{2}\right) \tilde{w}^2(1, t) \\ &\quad - \left(\frac{A\varepsilon}{4} - \frac{5\lambda \kappa_2 B}{16} - \frac{5\kappa_3 B}{8} - \frac{5\beta_3}{4}\right) \|\tilde{w}[t]\|^2 \\ &\quad - \left(\varepsilon B\left(r - \frac{1}{2}\right) - \frac{\varepsilon B}{2\kappa_1} - \frac{\lambda B}{8\kappa_2} - \beta_2\right) \hat{w}^2(1, t) \\ &\quad - \left(\frac{\varepsilon B}{4} - \frac{5\lambda B}{16\kappa_2} - \frac{\|g\|^2 B}{2\kappa_3} - \beta_1\right) \|\hat{w}[t]\|^2 + \eta m(t). \end{aligned} \quad (70)$$

From (55) and (56), we can show that

$$A\varepsilon\left(q - \frac{1}{2}\right) - \frac{\lambda \kappa_2 B}{8} - \frac{\kappa_3 B}{4} - \frac{\beta_3}{2} > 0, \quad (71)$$

$$\varepsilon B\left(r - \frac{1}{2}\right) - \frac{\varepsilon B}{2\kappa_1} - \frac{\lambda B}{8\kappa_2} - \beta_2 > 0. \quad (72)$$

Thus, using (70)-(72), we can obtain that

$$\begin{aligned} \dot{V} &\leq -\left(\frac{A\varepsilon}{4} - \frac{5\lambda \kappa_2 B}{16} - \frac{5\kappa_3 B}{8} - \frac{5\beta_3}{4}\right) \|\tilde{w}[t]\|^2 \\ &\quad - \left(\frac{\varepsilon B}{4} - \frac{5\lambda B}{16\kappa_2} - \frac{\|g\|^2 B}{2\kappa_3} - \beta_1\right) \|\hat{w}[t]\|^2 + \eta m(t). \end{aligned} \quad (73)$$

Again, we have from (55) and (56) that

$$b_1 = \frac{A\varepsilon}{4} - \frac{5\lambda \kappa_2 B}{16} - \frac{5\kappa_3 B}{8} - \frac{5\beta_3}{4} > 0, \quad (74)$$

$$b_2 = \frac{\varepsilon B}{4} - \frac{5\lambda B}{16\kappa_2} - \frac{\|g\|^2 B}{2\kappa_3} - \beta_1 > 0. \quad (75)$$

Therefore, we have for $t \in (t_j, t_{j+1})$, $j \in \mathbb{N}$ that

$$\dot{V} \leq -\varrho V, \quad (76)$$

where

$$\varrho = \min \left\{ \frac{2b_1}{A}, \frac{2b_2}{B}, \eta \right\}. \quad (77)$$

Concentrating on this time interval, we can show that $V(t_{j+1}^-) \leq e^{-\rho(t_{j+1}^- - t_j^+)} V(t_j^+)$. Here t_j^+ and t_j^- are the right and left limits of $t = t_j$. Since $V(t)$ is continuous (as $m(t), \|\hat{w}[t], \|\tilde{w}[t]\|$ are continuous), we have that $V(t_{j+1}^-) = V(t_{j+1})$ and $V(t_j^+) = V(t_j)$, and therefore,

$$V(t_{j+1}) \leq e^{-\rho(t_{j+1} - t_j)} V(t_j). \quad (78)$$

Hence, for any $t \geq 0$ in $t \in [t_j, t_{j+1}), j \in \mathbb{N}$, we can also obtain the following:

$$\begin{aligned} V(t) &\leq e^{-\rho(t-t_j)} V(t_j) \\ &\leq e^{-\rho(t-t_j)} \times e^{-\rho(t_j-t_{j-1})} V(t_{j-1}) \\ &\leq \dots \leq \\ &\leq e^{-\rho(t-t_j)} \times \prod_{i=1}^{j-1} e^{-\rho(t_i-t_{i-1})} V(0) \\ &= e^{-\rho t} V(0). \end{aligned} \quad (79)$$

Thus, recalling that $m(0) < 0$ from Definition 1, we have that

$$\begin{aligned} \frac{A}{2} \|\tilde{w}[t]\|^2 + \frac{B}{2} \|\hat{w}[t]\|^2 - m(t) \\ \leq e^{-\rho t} \left(\frac{A}{2} \|\tilde{w}[0]\|^2 + \frac{B}{2} \|\hat{w}[0]\|^2 - m(0) \right). \end{aligned} \quad (80)$$

As $m(t) < 0$, we also have that

$$\begin{aligned} \frac{A}{2} \|\tilde{w}[t]\|^2 + \frac{B}{2} \|\hat{w}[t]\|^2 \\ \leq e^{-\rho t} \left(\frac{A}{2} \|\tilde{w}[0]\|^2 + \frac{B}{2} \|\hat{w}[0]\|^2 - m(0) \right). \end{aligned} \quad (81)$$

This implies that the systems \hat{w} and \tilde{w} given by (23) and (24), respectively, are globally exponentially convergent in L^2 -sense to zero. Using (5) and (17), we can show that $\|\tilde{u}[t]\|^2 \leq \tilde{P}^2 \|\tilde{w}[t]\|^2$ and $\|\hat{u}[t]\|^2 \leq \tilde{L}^2 \|\hat{w}[t]\|^2$, respectively. Here $\tilde{P} = 1 + \left(\int_0^1 \int_0^x P^2(x, y) dy dx \right)^{1/2}$ and $\tilde{L} = 1 + \left(\int_0^1 \int_0^x L^2(x, y) dy dx \right)^{1/2}$. Further from (3), we can show that $\|u[t]\|^2 \leq 2\|\tilde{u}[t]\|^2 + 2\|\hat{u}[t]\|^2$. Therefore, we can obtain from (81) that

$$\begin{aligned} \min \left\{ \frac{A}{\tilde{P}^2}, \frac{B}{\tilde{L}^2} \right\} \|u[t]\|^2 \\ \leq 4e^{-\rho t} \left(\frac{A}{2} \|\tilde{w}[0]\|^2 + \frac{B}{2} \|\hat{w}[0]\|^2 - m(0) \right), \end{aligned} \quad (82)$$

$$\frac{B}{\tilde{L}^2} \|\hat{u}[t]\|^2 \leq 2e^{-\rho t} \left(\frac{A}{2} \|\tilde{w}[0]\|^2 + \frac{B}{2} \|\hat{w}[0]\|^2 - m(0) \right). \quad (83)$$

Therefore, we can derive the following estimate:

$$\begin{aligned} \|u[t]\| + \|\hat{u}[t]\| \\ \leq \sqrt{\frac{12 \left(\frac{A}{2} \|\tilde{w}[0]\|^2 + \frac{B}{2} \|\hat{w}[0]\|^2 - m(0) \right)}{\min \left\{ \frac{A}{\tilde{P}^2}, \frac{B}{\tilde{L}^2} \right\}}} e^{-\frac{\rho t}{2}}, \end{aligned} \quad (84)$$

Using (9) and (11), we can show that $\|\tilde{w}[0]\|^2 \leq \tilde{Q}^2 \|\tilde{u}[0]\|^2$ and $\|\hat{w}[0]\|^2 \leq \tilde{K}^2 \|\hat{u}[0]\|^2$, respectively. Here

$$\begin{aligned} \tilde{Q} &= 1 + \left(\int_0^1 \int_0^x Q^2(x, y) dy dx \right)^{1/2} \text{ and } \tilde{K} = 1 + \\ &\left(\int_0^1 \int_0^x K^2(x, y) dy dx \right)^{1/2}. \text{ Then we have} \\ \|u[t]\| + \|\hat{u}[t]\| \\ &\leq \sqrt{\frac{12 \left(A\tilde{Q}^2 \|u[0]\|^2 + (A\tilde{Q}^2 + \frac{B\tilde{K}^2}{2}) \|\hat{u}[0]\|^2 - m(0) \right)}{\min \left\{ \frac{A}{\tilde{P}^2}, \frac{B}{\tilde{L}^2} \right\}}} e^{-\frac{\rho t}{2}}, \end{aligned} \quad (85)$$

which implies that $\|u[t]\| + \|\hat{u}[t]\| \rightarrow 0$ as $t \rightarrow \infty$. \square

Remark 2: In Theorem 2, we have established the global exponential convergence of the closed-loop system to the equilibrium point. It follows from (81) that we could have obtained global exponential stability if we chose $m(0) = 0$. However, if $m(0) = 0$, then $m(t) \leq 0$ (this can be shown by following the same arguments in the proof of Lemma 1). Then, the function $\psi(t)$ in (44) is not defined when $m(t) = 0$. Therefore, the existence of a minimal-dwell time cannot be proved easily by following the same arguments as in the proof of Theorem 1. Hence, $m(0)$ has to be chosen strictly negative.

Remark 3: The parameter $\eta > 0$ characterizes the decay rate of $m(t)$ governed by (27). Thus, η may be used to adjust the sampling speed of the event-triggered mechanism. The larger η , the faster is the sampling speed. The parameter γ can be chosen to scale up/down the values of $\beta_1, \beta_2, \beta_3$ given by (43), as required. We consider $\sigma \in (0, 1)$ as a free parameter that can be tuned appropriately such that the conditions for guaranteeing a minimal dwell-time are met.

Remark 4: We remark that if a periodic sampling scheme where the control value is periodically updated in a sampled-and-hold manner is to be used to stabilize the plant (1) and the observer (2), one can choose a sampling period T upper bounded by the minimal dwell-time τ (54). It will ensure the closed-loop system's global exponential convergence because the relation (79) is guaranteed to hold for all $T \leq \tau$. However, one should expect τ to be very small as the coefficients a_1, a_2 , and a_3 given by (51)-(53) are usually large. This issue, on the other hand, reinforces the motivation for ETC, that is sample and update only when required.

V. NUMERICAL SIMULATIONS

We consider a reaction-diffusion system with $\varepsilon = 1; \lambda = 3; q = 3$ and the initial conditions $u[0] = 10x^2(x-1)^2$ and $\hat{u}[0] = 5x^2(x-1)^2 + 5x^3(x-1)^3$. For numerical simulations, both plant and observer states are discretized with uniform step size of $h = 0.005$ for the space variable. Time discretization was done using the implicit Euler scheme with a step size $\Delta t = 0.001$ s.

The parameters for the event-trigger mechanism are chosen as follows: $m(0) = -0.5, \gamma = 1 \times 10^4, \eta = 1$ or 100 and $\sigma = 0.1$. It can be shown using (40)-(42) that $\alpha_1 = 2.8318 \times 10^3; \alpha_2 = 759.3750; \alpha_3 = 3.0115 \times 10^3$. Therefore, from (43), we can obtain $\beta_1 = 0.3146; \beta_2 = 0.0844; \beta_3 = 0.3346$. Finding that $\|g\|^2 = 16.3241$, let us choose $\kappa_1 = 2.1; \kappa_2 = 200; \kappa_3 = 1 \times 10^3$ and $B = 2.0571$ to satisfy (55). Then, from (57), we can obtain $\rho = 5.2$.

Fig. 2 shows the zero-input response of the plant and it is clear that the system is unstable. Fig. 3(a) shows the response of the plant under ETC with $\eta = 1$ and Fig. 3(b) shows the evolution of $\|u[t]\|$, $\|\hat{u}[t]\|$ and $\|\tilde{u}[t]\|$. The evolution of the control input when $\eta = 1$ is presented in Fig. 4(a) along with the corresponding continuous-time control input. The behavior of the functions associated with the triggering condition (25) for the case of $\eta = 1$ is depicted in Fig. 4(b) (for better visualization, the functions $(-m(t)^{0.2})$ and $(d^2(t)/\gamma)^{0.2}$ have been plotted). Fig. 5 compares the ETC control input for $\eta = 1$ and $\eta = 100$, and it can be seen that $\eta = 100$ results in faster sampling than $\eta = 1$.

Finally, we conduct simulations for 100 different initial conditions $u[0] = x^2(x-1)^2 \sin(n\pi x)$, $n = 1, \dots, 100$ and $\hat{u}[0] = 2u[t]$ on a time frame of 10 s. Next, we compute the inter-execution times between two events and compare the cases for slow and fast sampling, *i.e.*, $\eta = 1$ and $\eta = 100$, respectively. Fig. 6 shows the density of the inter-execution times plotted in logarithmic scale, and it can be stated that when η is smaller, the inter-executions times are larger and the sampling is less often. For $\eta = 1$, the inter-execution times are around 0.1 s, whereas for $\eta = 100$, the inter-execution times are typically in the range 0.01 s – 0.1 s. The minimal dwell time τ calculated using (54) when $\eta = 1$ and $\eta = 100$ are respectively 2.1305×10^{-6} s and 2.1241×10^{-6} s. Therefore, the fact that an analogous sampled-data controller guaranteeing exponential convergence has to be implemented using these conservative sampled periods indicates the importance and the need of ETC.

VI. CONCLUSION

This paper has proposed an event-triggered output feedback boundary control strategy for a class of reaction-diffusion systems with Robin boundary actuation. We have used a dynamic event triggering condition to determine when the control value needs to be updated. Under the proposed strategy, we have proved the existence of a minimal-dwell time between two updates independent of initial conditions, which excludes Zeno behavior. Further, we have shown the well-posedness of the closed-loop system and its global L^2 -exponential convergence to the equilibrium point.

In our future work, we may look into event-triggered boundary control of reaction-diffusion PDEs with in-domain point measurements in the presence of actuator delays. We also may consider periodic event-triggered boundary control (PETC) of reaction-diffusion systems. The idea is to evaluate the triggering condition periodically and to decide, at every sampling instant, whether the feedback loop needs to be closed. PETC is highly desirable as it not only guarantees a minimal dwell-time equal to the sampling period but also provides a more realistic approach toward digital implementations while reducing the utilization of computational resources.

APPENDIX A

PROOF OF PROPOSITION 1

Using the conventional Lyapunov analysis, we can prove the global L^2 -exponential stability for the system (8) for any $q > \frac{1}{2}$ (this is ensured by Assumption 1).

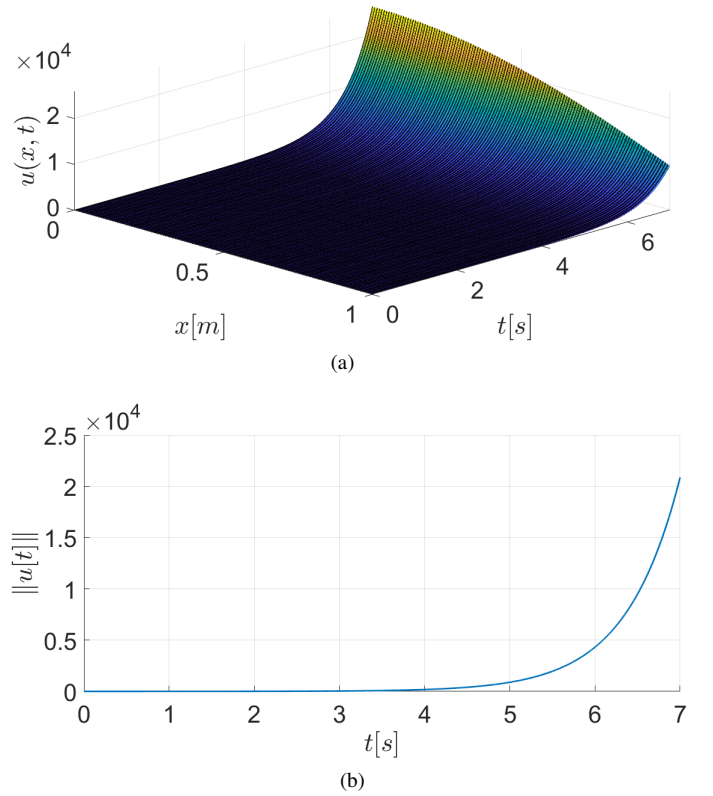


Fig. 2: Results for open-loop plant with $\varepsilon = 1$, $\lambda = 3$, $q = 3$ and $u[0] = 10x^2(x-1)^2$. (a) $u(x, t)$. (b) $\|u[t]\|$.

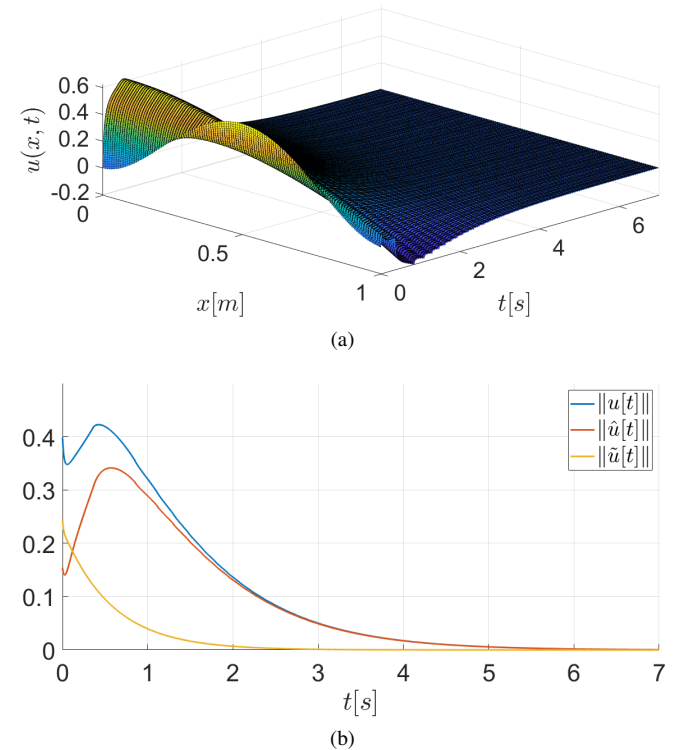


Fig. 3: Results for the event-triggered closed-loop system with $\varepsilon = 1$, $\lambda = 3$, $q = 3$, $m(0) = -0.5$, $\gamma = 1 \times 10^4$, $\eta = 1$, $u[0] = 10x^2(x-1)^2$ and $\hat{u}[0] = 5x^2(x-1)^2 + 5x^3(x-1)^3$. (a) $u(x, t)$. (b) $\|u[t]\|$, $\|\hat{u}[t]\|$ and $\|\tilde{u}[t]\|$.

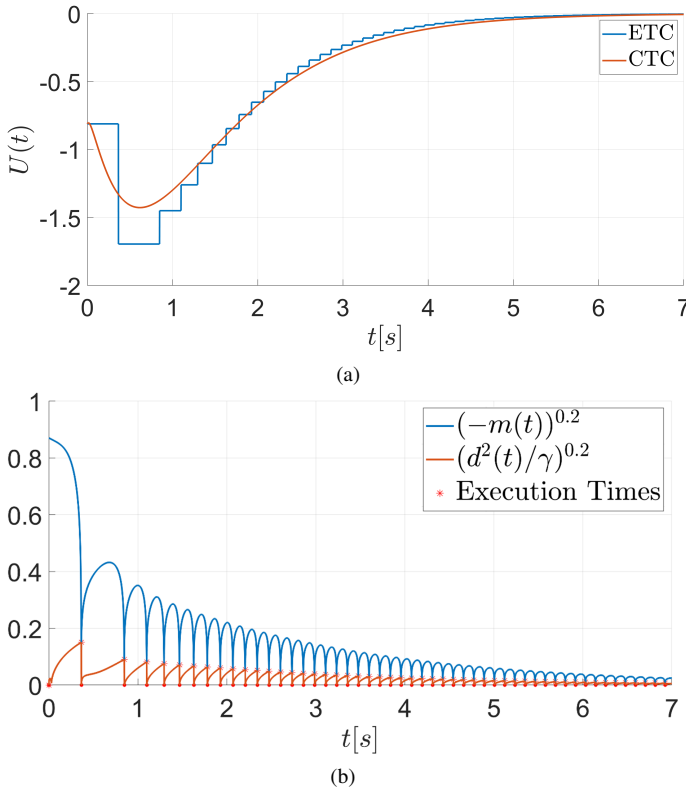


Fig. 4: (a) The event-triggered control (ETC) input for the system considered in Fig. 3 along with the corresponding continuous-time control (CTC) input. (b) Trajectories involved in the triggering condition (25) for the system in Fig. 3.

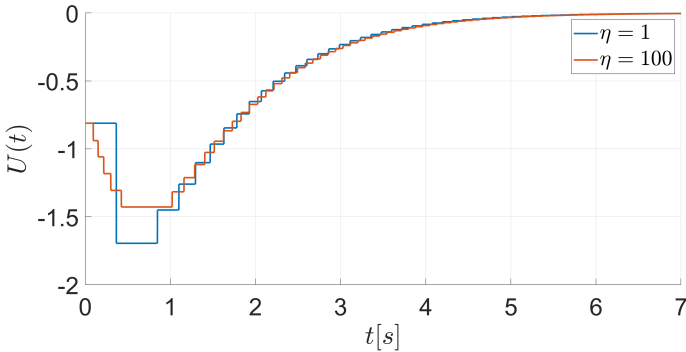


Fig. 5: Comparison of ETC input $U(t)$ for different η : $\eta = 1$ and $\eta = 100$, for the same system considered in Fig. 3.

Let us show the gain kernel $P(x, y)$ and the observer gains $p_1(x)$ and p_{10} , which transform (4) into (8) via (5), are given by (6) and (7), respectively. This proves Proposition 1.

Taking the time derivative of (5) along the solutions of (4) and applying integration by parts twice, we can show that

$$\begin{aligned} \tilde{u}_t(x, t) &= \tilde{w}_t(x, t) - \varepsilon P(x, x) \tilde{w}_x(x, t) \\ &\quad + \varepsilon P(x, 0) \tilde{w}_x(0, t) + \varepsilon P_y(x, x) \tilde{w}(x, t) \\ &\quad - \varepsilon P_y(x, 0) \tilde{w}(0, t) - \varepsilon \int_0^x P_{yy}(x, y) \tilde{w}(y, t) dy. \end{aligned} \quad (86)$$

Differentiating (5) w.r.t x and using Leibnitz differentiation

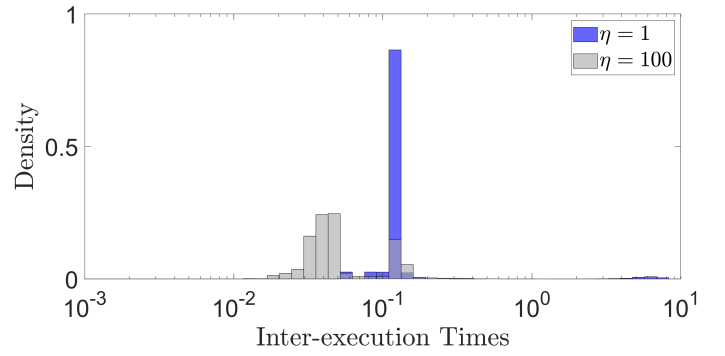


Fig. 6: Density of the inter-execution times (logarithmic scale) computed for 100 different initial conditions: $u[0] = x^2(x - 1)^2 \sin(n\pi x)$, $n = 1, \dots, 100$ and $\dot{u}[0] = 2u[t]$ on a time frame of 10 s, for the same system considered in Fig. 3 with $\eta = 1$ and $\eta = 100$.

rule, we can obtain that

$$\tilde{u}_x(x, t) = \tilde{w}_x(x, t) - P(x, x) \tilde{w}(x, t) - \int_0^x P_x(x, y) \tilde{w}(y, t) dy, \quad (87)$$

$$\begin{aligned} \tilde{u}_{xx}(x, t) &= \tilde{w}_{xx}(x, t) - \frac{dP(x, x)}{dx} \tilde{w}(x, t) - P(x, x) \tilde{w}_x(x, t) \\ &\quad - P_x(x, x) \tilde{w}(x, t) - \int_0^x P_{xx}(x, y) \tilde{w}(y, t) dy. \end{aligned} \quad (88)$$

Therefore, from (4a),(5),(86) and (88), we can show that

$$\begin{aligned} 0 &= (p_1(x) - \varepsilon P_y(x, 0)) \tilde{w}(0, t) - \left(\lambda - 2\varepsilon \frac{dP(x, x)}{dx} \right) \tilde{w}(x, t) \\ &\quad + \int_0^x \left(\varepsilon P_{xx}(x, y) - \varepsilon P_{yy}(x, y) + \lambda P(x, y) \right) \tilde{w}(y, t) dy. \end{aligned} \quad (89)$$

Let us choose $P_{xx}(x, y) - P_{yy}(x, y) = -\frac{\lambda}{\varepsilon} P(x, y)$, $\frac{dP(x, x)}{dx} = \frac{\lambda}{2\varepsilon}$, and $p_1(x) = \varepsilon P_y(x, 0)$ so that (89) is valid for any \tilde{w} . Further, let us choose $p_{10} = P(0, 0)$ so that the boundary conditions (4b) and (8b) are satisfied, and choose $P(1, 1) = 0$ and $P_x(1, y) = -qP(1, y)$ so that the conditions (4c) and (8c) are met. Therefore, the gain kernel $P(x, y)$ in (5) as a whole should satisfy the following PDE:

$$P_{xx}(x, y) - P_{yy}(x, y) = -\frac{\lambda}{\varepsilon} P(x, y), \quad (90a)$$

$$P_x(1, y) = -qP(1, y), \quad (90b)$$

$$P(x, x) = \frac{\lambda}{2\varepsilon} (x - 1). \quad (90c)$$

It can be shown that the change of variables $x = 1 - \bar{y}$ and $y = 1 - \bar{x}$ on (90) leads to the same PDE (108-110) in [42] to which the explicit solution has been obtained. Therefore, the solution to (90) can be shown to be given by (6). Above we have obtained $p_1(x) = \varepsilon P_y(x, 0)$ and $p_{10} = P(0, 0)$, which are the same as stated in Proposition 1. \square

APPENDIX B
PROOF OF PROPOSITION 2

Let us show that the gain kernel $K(x, y)$ and the control law $U(t)$, which transform (2) into (14)-(16) via (11), are given by (12) and (13), respectively. This proves Proposition 2.

Taking the time derivative of (11) along the solutions of (2) and applying integration by parts twice, we can show that

$$\begin{aligned} \hat{w}_t(x, t) &= \hat{u}_t(x, t) - \lambda \int_0^x K(x, y) \hat{u}(y, t) dy \\ &- \int_0^x K(x, y) p_1(y) dy \tilde{w}(0, t) - \varepsilon K(x, x) \hat{u}_x(x, t) \\ &+ \varepsilon K(x, 0) \hat{u}_x(0, t) + \varepsilon K_y(x, x) \hat{u}(x, t) \\ &- \varepsilon K_y(x, 0) \hat{u}(0, t) - \varepsilon \int_0^x K_{yy}(x, y) \hat{u}(y, t) dy, \end{aligned} \quad (91)$$

Differentiating (11) w.r.t x and using Leibnitz differentiation rule, we can obtain that

$$\hat{w}_x(x, t) = \hat{u}_x(x, t) - K(x, x) \hat{u}(x, t) - \int_0^x K_x(x, y) \hat{u}(y, t) dy, \quad (92)$$

$$\begin{aligned} \hat{w}_{xx}(x, t) &= \hat{u}_{xx}(x, t) - \frac{dK(x, x)}{dx} \hat{u}(x, t) - K(x, x) \hat{u}_x(x, t) \\ &- K_x(x, x) \hat{u}(x, t) - \int_0^x K_{xx}(x, y) \hat{u}(y, t) dy. \end{aligned} \quad (93)$$

Therefore, from (7),(11),(14a),(15),(91) and (93), we can show

$$\begin{aligned} 0 &= \left(\lambda + 2\varepsilon \frac{dK(x, x)}{dx} \right) \hat{u}(x, t) dy - \varepsilon K_y(x, 0) \hat{u}(0, t) \\ &+ \int_0^x \left(\varepsilon K_{xx}(x, y) - \varepsilon K_{yy}(x, y) - \lambda K(x, y) \right) \hat{u}(y, t) dy. \end{aligned} \quad (94)$$

Let us choose $K_{xx}(x, y) - K_{yy}(x, y) = \frac{\lambda}{\varepsilon} K(x, y)$, $K_y(x, 0) = 0$, and, $\frac{dK(x, x)}{dx} = -\frac{\lambda}{2\varepsilon}$ so that (94) holds for any \hat{u} . Further, let us choose $K(0, 0) = 0$, so that the boundary conditions (2b) and (14b) are met, and choose $U(t) = \int_0^1 \left(rK(1, y) + K_x(1, y) \right) \hat{u}(y, t) dy$ so that the boundary conditions (2c) and (14c) are satisfied. Therefore, the gain kernel $K(x, y)$ in (11) as a whole should satisfy the following PDE:

$$K_{xx}(x, y) - K_{yy}(x, y) = \frac{\lambda}{\varepsilon} K(x, y), \quad (95a)$$

$$K_y(x, 0) = 0, \quad (95b)$$

$$K(x, x) = -\frac{\lambda}{2\varepsilon} x. \quad (95c)$$

The solution to (95) is given by (12) [40]. Further, the control law obtained above is the same as (13). \square

APPENDIX C
PROOF OF PROPOSITION 3

Subject to Assumption 1, let us choose parameters $\delta_1, \delta_2 > 0$ such that

$$\varepsilon \min \left\{ r - \frac{1}{2}, \frac{1}{2} \right\} - \frac{5\lambda}{8\delta_1} - \frac{\|g\|^2}{\delta_2} \geq 0, \quad (96)$$

and $H > 0$ such that

$$H\varepsilon \min \left\{ q - \frac{1}{2}, \frac{1}{2} \right\} - \frac{5\lambda\delta_1}{8} - \frac{5\delta_2}{4} \geq 0. \quad (97)$$

Here $g(x)$ and r are given by (15) and (16), respectively. Note that $r > 1/2$ due to Assumption 1. Then, let us consider the following Lyapunov candidate

$$\mathcal{V} = \frac{H}{2} \int_0^1 \tilde{w}^2(x, t) dx + \frac{1}{2} \int_0^1 \hat{w}^2(x, t) dx, \quad (98)$$

where \tilde{w} and \hat{w} are the systems described by (8) and (14), respectively. We can show that

$$\begin{aligned} \dot{\mathcal{V}} &= -H\varepsilon q \tilde{w}^2(1, t) - H\varepsilon \int_0^1 \tilde{w}_x^2(x, t) dx \\ &- r\varepsilon \hat{w}^2(1, t) + \frac{\lambda}{2} \hat{w}(0, t) \tilde{w}(0, t) \\ &- \varepsilon \int_0^1 \hat{w}_x^2(x, t) dx + \int_0^1 g(x) \hat{w}(x, t) dx \tilde{w}(0, t). \end{aligned} \quad (99)$$

From Young's and Cauchy-Schwarz inequalities, we can obtain that

$$\begin{aligned} \frac{\lambda}{2} \hat{w}(0, t) \tilde{w}(0, t) &\leq \frac{\lambda}{4\delta_1} \hat{w}^2(0, t) + \frac{\lambda\delta_1}{4} \tilde{w}^2(0, t), \quad (100) \\ \int_0^1 g(x) \hat{w}(x, t) dx \tilde{w}(0, t) &\leq \frac{\|g\|^2}{2\delta_2} \|\hat{w}[t]\|^2 + \frac{\delta_2}{2} \tilde{w}^2(0, t). \end{aligned} \quad (101)$$

Therefore, using (100) and (101), we can write (99) as

$$\begin{aligned} \dot{\mathcal{V}} &\leq -H\varepsilon q \tilde{w}^2(1, t) - H\varepsilon \|\tilde{w}_x[t]\|^2 - r\varepsilon \hat{w}^2(1, t) \\ &+ \frac{\lambda}{4\delta_1} \hat{w}^2(0, t) + \frac{\lambda\delta_1}{4} \tilde{w}^2(0, t) \\ &- \varepsilon \|\hat{w}_x[t]\|^2 + \frac{\|g\|^2}{2\delta_2} \|\hat{w}[t]\|^2 + \frac{\delta_2}{2} \tilde{w}^2(0, t). \end{aligned} \quad (102)$$

From Agmon's and Young's inequalities, we have that

$$\tilde{w}^2(0, t) \leq \tilde{w}^2(1, t) + \|\tilde{w}[t]\|^2 + \|\tilde{w}_x[t]\|^2, \quad (103)$$

$$\hat{w}^2(0, t) \leq \hat{w}^2(1, t) + \|\hat{w}[t]\|^2 + \|\hat{w}_x[t]\|^2. \quad (104)$$

Therefore, we can show using (102) that

$$\begin{aligned} \dot{\mathcal{V}} &\leq - \left(H\varepsilon q - \frac{\lambda\delta_1}{4} - \frac{\delta_2}{2} \right) \tilde{w}^2(1, t) \\ &- \left(H\varepsilon - \frac{\lambda\delta_1}{4} - \frac{\delta_2}{2} \right) \|\tilde{w}_x[t]\|^2 \\ &+ \left(\frac{\lambda\delta_1}{4} + \frac{\delta_2}{2} \right) \|\tilde{w}[t]\|^2 - \left(r\varepsilon - \frac{\lambda}{4\delta_1} \right) \hat{w}^2(1, t) \\ &- \left(\varepsilon - \frac{\lambda}{4\delta_1} \right) \|\hat{w}_x[t]\|^2 + \left(\frac{\lambda}{4\delta_1} + \frac{\|g\|^2}{2\delta_2} \right) \|\hat{w}[t]\|^2. \end{aligned} \quad (105)$$

From Poincaré Inequality, we have that

$$-\|\tilde{w}_x[t]\|^2 \leq \frac{1}{2} \tilde{w}^2(1, t) - \frac{1}{4} \|\tilde{w}[t]\|^2, \quad (106)$$

$$-\|\hat{w}_x[t]\|^2 \leq \frac{1}{2} \hat{w}^2(1, t) - \frac{1}{4} \|\hat{w}[t]\|^2. \quad (107)$$

Furthermore, we have from (96) and (97) that

$$H\varepsilon - \frac{\lambda\delta_1}{4} - \frac{\delta_2}{2} > 0 \text{ and } \varepsilon - \frac{\lambda}{4\delta_1} > 0. \quad (108)$$

Therefore, using (105)-(108), we can show that

$$\begin{aligned} \dot{\mathcal{V}} \leq & - \left(H\varepsilon \left(q - \frac{1}{2} \right) - \frac{\lambda\delta_1}{8} - \frac{\delta_2}{4} \right) \hat{w}^2(1, t) \\ & - \left(\frac{H\varepsilon}{4} - \frac{5\lambda\delta_1}{16} - \frac{5\delta_2}{8} \right) \|\tilde{w}[t]\|^2 \\ & - \left(\varepsilon \left(r - \frac{1}{2} \right) - \frac{\lambda}{8\delta_1} \right) \hat{w}^2(1, t) \\ & - \left(\frac{\varepsilon}{4} - \frac{5\lambda}{16\delta_1} - \frac{\|g\|^2}{2\delta_2} \right) \|\hat{w}[t]\|^2. \end{aligned} \quad (109)$$

From (96) and (97), we have that

$$H\varepsilon \left(q - \frac{1}{2} \right) - \frac{\lambda\delta_1}{8} - \frac{\delta_2}{4} > 0 \text{ and } \varepsilon \left(r - \frac{1}{2} \right) - \frac{\lambda}{8\delta_1} > 0, \quad (110)$$

Thus, it follows from (109) that

$$\begin{aligned} \dot{\mathcal{V}} \leq & - \left(\frac{H\varepsilon}{4} - \frac{5\lambda\delta_1}{16} - \frac{5\delta_2}{8} \right) \|\tilde{w}[t]\|^2 \\ & - \left(\frac{\varepsilon}{4} - \frac{5\lambda}{16\delta_1} - \frac{\|g\|^2}{2\delta_2} \right) \|\hat{w}[t]\|^2. \end{aligned} \quad (111)$$

Again, we have from (96) and (97) that

$$\vartheta_1 = \frac{H\varepsilon}{4} - \frac{5\lambda\delta_1}{16} - \frac{5\delta_2}{8} > 0, \quad (112)$$

$$\vartheta_2 = \frac{\varepsilon}{4} - \frac{5\lambda}{16\delta_1} - \frac{\|g\|^2}{2\delta_2} > 0. \quad (113)$$

Therefore, (111) can be written as

$$\dot{\mathcal{V}} \leq - \min \left\{ \frac{2\vartheta_1}{H}, 2\vartheta_2 \right\} \mathcal{V}. \quad (114)$$

Hence, from standard arguments, we can state that the closed-loop system which consists of the plant (1) and the observer (2) with the continuous-time control law (13), is globally exponentially stable in L^2 -sense. \square

REFERENCES

- [1] T. Chen and B. A. Francis, "Input-output stability of sampled-data systems," *IEEE Transactions on Automatic Control*, vol. 36, no. 1, pp. 50–58, 1991.
- [2] K. J. Åström and B. Wittenmark, *Computer-controlled systems: theory and design*. Courier Corporation, 2013.
- [3] G. C. Walsh and H. Ye, "Scheduling of networked control systems," *IEEE control systems magazine*, vol. 21, no. 1, pp. 57–65, 2001.
- [4] X. Wang and M. D. Lemmon, "Event-triggering in distributed networked control systems," *IEEE Transactions on Automatic Control*, vol. 56, no. 3, pp. 586–601, 2010.
- [5] C. Peng and T. C. Yang, "Event-triggered communication and H_∞ control co-design for networked control systems," *Automatica*, vol. 49, no. 5, pp. 1326–1332, 2013.
- [6] C. Nowzari, E. Garcia, and J. Cortés, "Event-triggered communication and control of networked systems for multi-agent consensus," *Automatica*, vol. 105, pp. 1–27, 2019.
- [7] L. Hetel, C. Fiter, H. Omran, A. Seuret, E. Fridman, J.-P. Richard, and S. I. Niculescu, "Recent developments on the stability of systems with aperiodic sampling: An overview," *Automatica*, vol. 76, pp. 309–335, 2017.
- [8] I. Karafyllis and M. Krstic, "Sampled-data boundary feedback control of 1-d linear transport PDEs with non-local terms," *Systems & Control Letters*, vol. 107, pp. 68–75, 2017.
- [9] —, "Sampled-data boundary feedback control of 1-d parabolic PDEs," *Automatica*, vol. 87, pp. 226–237, 2018.
- [10] W. Kang and E. Fridman, "Distributed sampled-data control of kuramoto-sivashinsky equation," *Automatica*, vol. 95, pp. 514–524, 2018.
- [11] W. Heemels, K. H. Johansson, and P. Tabuada, "An introduction to event-triggered and self-triggered control," in *2012 51st IEEE Conference on Decision and Control (CDC)*. IEEE, 2012, pp. 3270–3285.
- [12] Z. Yao and N. H. El-Farra, "Resource-aware model predictive control of spatially distributed processes using event-triggered communication," in *52nd IEEE Conference on Decision and Control*. IEEE, 2013, pp. 3726–3731.
- [13] M. Lemmon, "Event-triggered feedback in control, estimation, and optimization," *Networked control systems*, pp. 293–358, 2010.
- [14] P. Tabuada, "Event-triggered real-time scheduling of stabilizing control tasks," *IEEE Transactions on Automatic Control*, vol. 52, no. 9, pp. 1680–1685, 2007.
- [15] M. C. F. Donkers and W. P. M. H. Heemels, "Output-based event-triggered control with guaranteed \mathcal{L}_∞ -gain and improved and decentralized event-triggering," *IEEE Transactions on Automatic Control*, vol. 57, no. 6, pp. 1362–1376, 2012.
- [16] W. H. Heemels, M. Donkers, and A. R. Teel, "Periodic event-triggered control for linear systems," *IEEE Transactions on Automatic Control*, vol. 58, no. 4, pp. 847–861, 2012.
- [17] N. Marchand, S. Durand, and J. F. G. Castellanos, "A general formula for event-based stabilization of nonlinear systems," *IEEE Transactions on Automatic Control*, vol. 58, no. 5, pp. 1332–1337, 2012.
- [18] P. Tallapragada and N. Chopra, "On event triggered tracking for nonlinear systems," *IEEE Transactions on Automatic Control*, vol. 58, no. 9, pp. 2343–2348, 2013.
- [19] R. Postoyan, P. Tabuada, D. Nešić, and A. Anta, "A framework for the event-triggered stabilization of nonlinear systems," *IEEE Transactions on Automatic Control*, vol. 60, no. 4, pp. 982–996, 2014.
- [20] A. Girard, "Dynamic triggering mechanisms for event-triggered control," *IEEE Transactions on Automatic Control*, vol. 60, no. 7, pp. 1992–1997, 2014.
- [21] M. Abdelrahim, R. Postoyan, J. Daafouz, and D. Nešić, "Stabilization of nonlinear systems using event-triggered output feedback controllers," *IEEE Transactions on Automatic Control*, vol. 61, no. 9, pp. 2682–2687, 2015.
- [22] D. P. Borgers, R. Postoyan, A. Anta, P. Tabuada, D. Nešić, and W. Heemels, "Periodic event-triggered control of nonlinear systems using overapproximation techniques," *Automatica*, vol. 94, pp. 81–87, 2018.
- [23] W. Wang, R. Postoyan, D. Nešić, and W. Heemels, "Periodic event-triggered control for nonlinear networked control systems," *IEEE Transactions on Automatic Control*, vol. 65, no. 2, pp. 620–635, 2019.
- [24] P. Zhang, T. Liu, and Z.-P. Jiang, "Event-triggered stabilization of a class of nonlinear time-delay systems," *IEEE Transactions on Automatic Control*, 2020.
- [25] —, "Systematic design of robust event-triggered state and output feedback controllers for uncertain nonholonomic systems," *IEEE Transactions on Automatic Control*, 2020.
- [26] A. Selivanov and E. Fridman, "Distributed event-triggered control of diffusion semilinear PDEs," *Automatica*, vol. 68, pp. 344–351, 2016.
- [27] N. Espitia, A. Girard, N. Marchand, and C. Prieur, "Event-based control of linear hyperbolic systems of conservation laws," *Automatica*, vol. 70, pp. 275–287, 2016.
- [28] N. Espitia, A. Tanwani, and S. Tarbouriech, "Stabilization of boundary controlled hyperbolic PDEs via lyapunov-based event triggered sampling and quantization," in *2017 IEEE 56th Annual Conference on Decision and Control (CDC)*. IEEE, 2017, pp. 1266–1271.
- [29] N. Espitia, A. Girard, N. Marchand, and C. Prieur, "Event-based boundary control of a linear 2×2 hyperbolic system via backstepping approach," *IEEE Transactions on Automatic Control*, vol. 63, no. 8, pp. 2686–2693, 2017.
- [30] J.-W. Wang, "Observer-based boundary control of semi-linear parabolic PDEs with non-collocated distributed event-triggered observation," *Journal of the Franklin Institute*, vol. 356, no. 17, pp. 10405–10420, 2019.
- [31] L. Baudouin, S. Marx, and S. Tarbouriech, "Event-triggered damping of a linear wave equation," *IFAC-PapersOnLine*, vol. 52, no. 2, pp. 58–63, 2019.
- [32] R. Katz, E. Fridman, and A. Selivanov, "Network-based boundary observer-controller design for 1D heat equation," in *2019 IEEE 58th Conference on Decision and Control (CDC)*. IEEE, 2019, pp. 2151–2156.
- [33] N. Espitia, "Observer-based event-triggered boundary control of a linear 2×2 hyperbolic systems," *Systems & Control Letters*, vol. 138, pp. 104668, 2020.
- [34] N. Espitia, I. Karafyllis, and M. Krstic, "Event-triggered boundary control of constant-parameter reaction-diffusion PDEs: a small-gain approach," *Automatica*, vol. 128, p. 109562, 2021.

- [35] M. Diagne and I. Karafyllis, “Event-triggered control of a continuum model of highly re-entrant manufacturing system,” *arXiv preprint arXiv:2004.01916*, 2020.
- [36] R. Katz, E. Fridman, and A. Selivanov, “Boundary delayed observer-controller design for reaction-diffusion systems,” *IEEE Transactions on Automatic Control*, 2020.
- [37] W. Kang, L. Baudouin, and E. Fridman, “Event-triggered control of korteweg–de vries equation under averaged measurements,” *Automatica*, vol. 123, p. 109315, 2021.
- [38] D. Xue and N. H. El-Farra, “Output feedback-based event-triggered control of distributed processes with communication constraints,” in *2016 IEEE 55th Conference on Decision and Control (CDC)*. IEEE, 2016, pp. 4296–4301.
- [39] F. Ge and Y. Chen, “Observer-based event-triggered control for semilinear time-fractional diffusion systems with distributed feedback,” *Nonlinear Dynamics*, vol. 99, no. 2, pp. 1089–1101, 2020.
- [40] M. Krstic and A. Smyshlyaev, *Boundary control of PDEs: A course on backstepping designs*. SIAM, 2008.
- [41] I. Karafyllis and M. Krstic, *Input-to-state stability for PDEs*. Springer, 2019.
- [42] A. Smyshlyaev and M. Krstic, “Closed-form boundary state feedbacks for a class of 1-d partial integro-differential equations,” *IEEE Transactions on Automatic Control*, vol. 49, no. 12, pp. 2185–2202, 2004.