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Title	Optin ization of linear and nonlinear interaction schemes for stable synchronization of weakly coupled limit tycle oscillators
Authors	NobuhiroWatanabe□YuzuruKato□ShoShirasaka□HiroyaNakao
C itation	PhysicalReview E□Vo□100□Issue 4□pp□042205
Pub⊡date	2019□10
DOI	http://dxidoiorg/https://doiorg/i01/103/PhysRevE/I000042205
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Optimization of linear and nonlinear interaction schemes for stable synchronization of weakly coupled limit-cycle oscillators

Nobuhiro Watanabe, ¹ Yuzuru Kato ¹, ^{1,*} Sho Shirasaka, ² and Hiroya Nakao ¹

¹Department of Systems and Control Engineering, Tokyo Institute of Technology, Tokyo 152-8552, Japan

²Department of Information and Physical Sciences, Graduate School of Information Science and Technology, Osaka University 1-5

Yamadaoka, Suita, Osaka 565-0871, Japan

(Received 9 January 2019; revised manuscript received 10 July 2019; published 4 October 2019)

Optimization of mutual synchronization between a pair of limit-cycle oscillators with weak symmetric coupling is considered in the framework of the phase-reduction theory. By generalizing our previous study [S. Shirasaka, N. Watanabe, Y. Kawamura, and H. Nakao, Optimizing stability of mutual synchronization between a pair of limit-cycle oscillators with weak cross coupling, Phys. Rev. E 96, 012223 (2017)] on the optimization of cross-diffusion coupling matrices between the oscillators, we consider optimization of mutual coupling signals to maximize the linear stability of the synchronized state, which are functionals of the past time sequences of the oscillator states. For the case of linear coupling, optimization of the delay time and linear filtering of coupling signals are considered. For the case of nonlinear coupling, general drive-response coupling is considered and the optimal response and driving functions are derived. The theoretical results are illustrated by numerical simulations.

DOI: 10.1103/PhysRevE.100.042205

I. INTRODUCTION

Synchronization of rhythmic dynamical elements exhibiting periodic oscillations is widely observed in the real world [1–3]. Collective oscillations arising from the mutual synchronization of dynamical elements often play important functional roles, such as the synchronized secretion of insulin from pancreatic beta cells [1,4] and synchronized oscillation of power generators [3,5,6]. Clarifying the mechanisms of collective synchronization and devising efficient methods of mutual synchronization are thus both fundamentally and practically important.

The stable periodic dynamics of rhythmic elements are often modeled as limit-cycle oscillators [1-3,7]. When mutual interactions between limit-cycle oscillators are weak, synchronization dynamics of the oscillators can be analyzed using the phase-reduction theory [1,8-12]. In this approach, nonlinear multidimensional dynamics of an oscillator is reduced to a simple approximate phase equation, characterized by the natural frequency and phase sensitivity of the oscillator. The phase-reduction theory facilitates systematic and detailed analysis of synchronization dynamics. It has been used to explain nontrivial synchronization dynamics of coupled oscillators, such as the collective synchronization transition of an ensemble of coupled oscillators [1,8-12]. Generalization of the method for nonconventional limit-cycling systems, such as time-delayed oscillators [13,14], hybrid oscillators [15,16], collectively oscillating networks [17], and rhythmic spatiotemporal patterns [18,19], has also been discussed.

Recently, the phase-reduction theory has been applied for the control and optimization of synchronization dynamics in oscillatory systems. For example, Moehlis *et al.* [20], Harada *et al.* [21], Dasanayake and Li [22], Zlotnik *et al.* [23–25], Pikovsky [26], Tanaka *et al.* [27–29], Wilson *et al.* [30], Pyragas *et al.* [31], and Monga *et al.* [32,33] have used the phase-reduction theory (as well as the phase-amplitude reduction theory) to derive optimal driving signals for the stable entrainment of nonlinear oscillators in various physical situations.

In a similar spirit, in our previous study [34], we considered a problem of improving the linear stability of synchronized state between a pair of limit-cycle oscillators by optimizing a cross-diffusion coupling matrix between the oscillators, where different components of the oscillators are allowed to interact. We also considered a pair of mutually interacting reaction-diffusion systems exhibiting rhythmic spatiotemporal patterns, and we derived optimal spatial filters for stable mutual synchronization [35].

In this study, we consider this problem in a more general setting, whereby the oscillators can interact not only by their present states but also through the time sequences of their past states. We first consider linear coupling with time delay or temporal filtering, and we derive the optimal delay time or linear filter. We then consider general nonlinear coupling with a mutual drive-response configuration, and we derive the optimal response function and driving function for stable synchronization. We argue that, although we consider general coupling that can depend on the past time sequences of the oscillators, the optimal mutual coupling can be obtained as a function of the present phase values of the oscillators in the framework of the phase-reduction approximation. The results are illustrated by numerical simulations using Stuart-Landau and FitzHugh-Nagumo oscillators as examples.

^{*}Author to whom all correspondence should be addressed: kato.y.bg@m.titech.ac.jp

This paper is organized as follows. In Sec. II, we introduce a general model of coupled limit-cycle oscillators and reduce it to coupled phase equations. In Sec. III, we consider the case with linear coupling and derive the optimal time delay and optimal linear filter for coupling signals. In Sec. IV, we consider nonlinear coupling of the drive-response type and derive the optimal response function and driving function. In Sec. V, a summary is provided.

II. MODEL

A. Pair of weakly coupled oscillators

In this study, we consider a pair of weakly and symmetrically coupled limit-cycle oscillators with identical properties, where the oscillators can mutually interact not only through their present states but also through their past time sequences. We assume that the oscillators are generally described by the following set of functional differential equations:

$$\dot{X}_{1}(t) = F(X_{1}(t)) + \epsilon \hat{H} \{X_{1}^{(t)}(\cdot), X_{2}^{(t)}(\cdot)\},
\dot{X}_{2}(t) = F(X_{2}(t)) + \epsilon \hat{H} \{X_{2}^{(t)}(\cdot), X_{1}^{(t)}(\cdot)\},$$
(1)

where $X_{1,2} \in \mathbb{R}^N$ are N-dimensional state vectors of the oscillators 1 and 2 at time $t, F : \mathbb{R}^N \to \mathbb{R}^N$ is a sufficiently smooth vector field representing the dynamics of individual oscillators, and $\epsilon \hat{H}$ represents weak mutual coupling between the oscillators. Here, $\hat{H} : C \times C \to \mathbb{R}^N$ (C is a function space of the time sequences of length L) is a sufficiently smooth *functional* of two vector functions, i.e., the past time sequences of $X_{1,2}(t)$, and $0 < \epsilon \ll 1$ is a small parameter representing the smallness of the mutual coupling. We assume that each isolated oscillator, $\hat{X} = F(X)$, has an exponentially stable limit cycle $\tilde{X}_0(t) = \tilde{X}_0(t+T)$ of period T and frequency $\omega = 2\pi/T$, and the deviation of the oscillator state from this limit cycle remains small even if weak perturbations due to the mutual coupling are applied.

We use the standard notation of functional differential equations [36] to represent the time sequences of $X_{1,2}$ in the coupling functional \hat{H} . Namely, the symbol $X_i^{(t)}(\cdot) \in C$ (i = 1, 2) represents the time sequence of X_i on the interval [t - L, t], defined by

$$\boldsymbol{X}_{i}^{(t)}(\sigma) = \boldsymbol{X}_{i}(t+\sigma) \quad (-L \leqslant \sigma \leqslant 0), \tag{2}$$

where the parameter σ runs from -L to 0. The length $L \geqslant 0$ of the time sequences used for the coupling is arbitrary as long as the assumption described later is satisfied. We abbreviate these time sequences as $X_i^{(t)}$ hereafter. In Eq. (1), the symbol (·) indicates that \hat{H} is a functional, which depends not only on the values of $X_{1,2}$ at a particular time but generally on the time sequences of $X_{1,2}$. We omit this symbol hereafter unless necessary.

B. Phase reduction and averaging

For weakly coupled limit-cycle oscillators, we can employ the standard method of phase reduction [1,8–12]. Let us consider a single isolated oscillator for the moment. We can introduce a phase function $\Theta(X): \mathbb{R}^N \to [0, 2\pi)$, which maps the oscillator state to a phase value, such that it satisfies $F(X) \cdot \nabla \Theta(X) = \omega$ in the whole basin of the limit cycle.

Using this $\Theta(X)$, the phase variable of the oscillator can be defined as $\theta = \Theta(X)$, which constantly increases with time as $\dot{\theta} = \dot{\Theta}(X) = \omega$ in the basin of the limit cycle (2π) is identified with 0). The oscillator state on the limit cycle can be represented as a function of θ as $X_0(\theta) = \tilde{X}_0(t = \theta/\omega)$, which is a 2π -periodic function of θ , $X_0(\theta) = X_0(\theta + 2\pi)$. Similar to Eq. (2), in order to represent a time sequence on the limit cycle, we introduce a notation

$$X_0^{(\theta)}(\sigma) = X_0(\theta + \omega\sigma) \quad (-L \leqslant \sigma \leqslant 0) \tag{3}$$

and abbreviate this as $X_0^{(\theta)}$.

The linear response property of the oscillator phase to weak perturbations is characterized by the phase sensitivity function (PSF), defined by $\mathbf{Z}(\theta) = \nabla \Theta(\mathbf{X})|_{\mathbf{X} = \mathbf{X}_0(\theta)} : [0, 2\pi) \to \mathbb{R}^N$. That is, when the oscillator is weakly driven by a perturbation \mathbf{p} as $\dot{\mathbf{X}} = \mathbf{F}(\mathbf{X}) + \epsilon \mathbf{p}$, the phase θ of the oscillator approximately obeys a reduced phase equation, $\dot{\theta} = \omega + \epsilon \mathbf{Z}(\theta) \cdot \mathbf{p}$, which is correct up to $O(\epsilon)$. The PSF can be calculated as a 2π -periodic solution to an adjoint equation $\omega d\mathbf{Z}(\theta)/d\theta = -J^{\dagger}(\mathbf{X}_0(\theta))\mathbf{Z}(\theta)$ with a normalization condition $\mathbf{Z}(\theta) \cdot d\mathbf{X}_0(\theta)/d\theta = 1$, where $J(\mathbf{X}) : \mathbb{R}^N \to \mathbb{R}^{N \times N}$ is a Jacobian matrix of \mathbf{F} at \mathbf{X} , and \dagger denotes transpose. In the numerical analysis, $\mathbf{Z}(\theta)$ can be calculated easily using the backward time evolution of the adjoint equation as proposed by Ermentrout [10].

Let us now consider a pair of weakly coupled oscillators described by Eq. (1). We define the phase values of the oscillators 1, 2 as $\theta_{1,2} = \Theta(X_{1,2})$. When the perturbation applied to the oscillators is sufficiently weak and of $O(\epsilon)$, the state vector of each oscillator can be approximated as $X_{1,2}(t) \approx X_0(\theta_{1,2}(t))$ as a function of the phase $\theta_{1,2}(t)$ within the error of $O(\epsilon)$. More generally, we assume that the deviation of the oscillator state from the limit cycle is small and of $O(\epsilon)$ in the whole interval [t-L,t], i.e.,

$$X_{1,2}^{(t)}(\sigma) = X_0^{[\theta_{1,2}(t)]}(\sigma) + O(\epsilon) \quad (-L \leqslant \sigma \leqslant 0).$$
 (4)

It then follows from the smoothness of \hat{H} that

$$\hat{H}\left\{X_{1}^{(t)}, X_{2}^{(t)}\right\} = \hat{H}\left\{X_{0}^{[\theta_{1}(t)]}, X_{0}^{[\theta_{2}(t)]}\right\} + O(\epsilon L).$$
 (5)

We assume hereafter that the length L satisfies $L \ll 1/\epsilon$, i.e., $\hat{H}\{X_1^{(t)}, X_2^{(t)}\} \approx \hat{H}\{X_0^{[\theta_1(t)]}, X_0^{[\theta_2(t)]}\}$ within the error of $O(\epsilon)$.

By the phase reduction, we obtain the following pair of phase equations from Eq. (1):

$$\dot{\theta}_1(t) = \omega + \epsilon \mathbf{Z}(\theta_1(t)) \cdot \hat{\mathbf{H}} \left\{ \mathbf{X}_1^{(t)}, \mathbf{X}_2^{(t)} \right\},
\dot{\theta}_2(t) = \omega + \epsilon \mathbf{Z}(\theta_2(t)) \cdot \hat{\mathbf{H}} \left\{ \mathbf{X}_2^{(t)}, \mathbf{X}_1^{(t)} \right\},$$
(6)

which are correct up to $O(\epsilon)$. Substituting Eq. (5) into Eq. (6) and ignoring errors of $O(\epsilon^2)$, we obtain a pair of reduced phase equations,

$$\dot{\theta}_{1} = \omega + \epsilon \mathbf{Z}(\theta_{1}) \cdot \hat{\mathbf{H}} \left\{ X_{0}^{(\theta_{1})}, X_{0}^{(\theta_{2})} \right\},
\dot{\theta}_{2} = \omega + \epsilon \mathbf{Z}(\theta_{2}) \cdot \hat{\mathbf{H}} \left\{ X_{0}^{(\theta_{2})}, X_{0}^{(\theta_{1})} \right\},$$
(7)

which are also correct up to $O(\epsilon)$. Thus, we can neglect the deviations of the oscillator states from the limit cycle at the lowest order approximation.

The coupling term in Eq. (7), $\hat{H}\{X_0^{(\theta_1)}, X_0^{(\theta_2)}\}$, is formally a functional of the two time sequences $X_0^{(\theta_1)}$ and $X_0^{(\theta_2)}$. However,

these functions $X_0^{(\theta_1)}$ and $X_0^{(\theta_2)}$ are determined only by the two phase values θ_1 and θ_2 . Therefore, we can regard the coupling term $\hat{H}\{X_0^{(\theta_1)}, X_0^{(\theta_2)}\}$ as an ordinary function of θ_1 and θ_2 , defined by

$$H(\theta_1, \theta_2) = \hat{H}\{X_0^{(\theta_1)}, X_0^{(\theta_2)}\},$$
 (8)

and we rewrite the phase equations as

$$\dot{\theta}_1 = \omega + \epsilon \mathbf{Z}(\theta_1) \cdot \mathbf{H}(\theta_1, \theta_2),
\dot{\theta}_2 = \omega + \epsilon \mathbf{Z}(\theta_2) \cdot \mathbf{H}(\theta_2, \theta_1).$$
(9)

It should be stressed that, although we started from Eq. (1) with general coupling functionals that depend on the past time sequences of the oscillators, the coupled system reduces to a pair of simple ordinary differential equations that depend only on the present phase values θ_1 and θ_2 of the oscillators within the phase-reduction approximation.

Once we have obtained Eq. (9), we can follow the standard averaging procedure of the phase-reduction theory [8,9]. We introduce slow phase variables $\phi_{1,2} = \theta_{1,2} - \omega t$, rewrite the equations as

$$\dot{\phi}_1 = \epsilon \mathbf{Z}(\phi_1 + \omega t) \cdot \mathbf{H}(\phi_1 + \omega t, \phi_2 + \omega t),$$

$$\dot{\phi}_2 = \epsilon \mathbf{Z}(\phi_2 + \omega t) \cdot \mathbf{H}(\phi_2 + \omega t, \phi_1 + \omega t), \tag{10}$$

and average the small right-hand side of these equations over one period of oscillation. This yields the following averaged phase equations, which are correct up to $O(\epsilon)$:

$$\dot{\theta}_1 = \omega + \epsilon \Gamma(\theta_1 - \theta_2),
\dot{\theta}_2 = \omega + \epsilon \Gamma(\theta_2 - \theta_1),$$
(11)

where $\Gamma(\phi)$ is the phase coupling function defined as

$$\Gamma(\phi) = \frac{1}{2\pi} \int_0^{2\pi} \mathbf{Z}(\phi + \psi) \cdot \mathbf{H}(\phi + \psi, \psi) d\psi$$
$$= \langle \mathbf{Z}(\phi + \psi) \cdot \mathbf{H}(\phi + \psi, \psi) \rangle_{\psi}$$
$$= \langle \mathbf{Z}(\psi) \cdot \mathbf{H}(\psi, \psi - \phi) \rangle_{\psi}. \tag{12}$$

Here, we have defined an average of a function $f(\psi)$ over one period of oscillation as

$$\langle f(\psi) \rangle_{\psi} = \frac{1}{2\pi} \int_0^{2\pi} f(\psi) d\psi. \tag{13}$$

C. Linear stability of the in-phase synchronized state

From the coupled phase equations (11), the dynamics of the phase difference $\phi = \theta_1 - \theta_2$ obeys

$$\dot{\phi} = \epsilon [\Gamma(\phi) - \Gamma(-\phi)], \tag{14}$$

where the right-hand side is (twice) the antisymmetric part of the phase coupling function $\Gamma(\phi)$. This equation always has a fixed point at $\phi=0$ corresponding to the in-phase synchronized state. In a small vicinity of $\phi=0$, the above equation can be linearized as

$$\dot{\phi} \approx 2\epsilon \Gamma'(0)\phi.$$
 (15)

The derivative of the phase coupling function is given by

$$\Gamma'(\phi) = -\langle \mathbf{Z}(\psi) \cdot \mathbf{H}_2'(\psi, \psi - \phi) \rangle_{y_t}, \tag{16}$$

where

$$\mathbf{H}_{2}'(\psi_{1}, \psi_{2}) = \frac{\partial \mathbf{H}(\psi_{1}, \psi_{2})}{\partial \psi_{2}} \tag{17}$$

is the partial derivative of H with respect to the second argument. Thus, the linear stability of this state is characterized by the exponent $2\epsilon\Gamma'(0)$, and a larger $-\Gamma'(0)$ yields a higher linear stability of the in-phase synchronized state.

In this study, we consider optimization of either the parameters or functions included in the mutual coupling term H so that the linear stability

$$-\Gamma'(0) = \langle \mathbf{Z}(\psi) \cdot \mathbf{H}_2'(\psi, \psi) \rangle_{\psi} \tag{18}$$

of the in-phase synchronized state, $\phi=0$, is maximized under appropriate constraints on the intensity of the mutual coupling.

D. Examples of limit-cycle oscillators

The Stuart-Landau (SL) oscillator is used in the following numerical examples. It is a normal form of the supercritical Hopf bifurcation and is described by

$$X = \begin{pmatrix} x \\ y \end{pmatrix} \in \mathbb{R}^2, \tag{19}$$

$$F = \begin{pmatrix} F_x \\ F_y \end{pmatrix} = \begin{pmatrix} x - ay - (x^2 + y^2)(x - by) \\ ax + y - (x^2 + y^2)(bx + y) \end{pmatrix}, \tag{20}$$

where the parameters are fixed at a=2 and b=1. This oscillator has a stable limit cycle with a natural frequency $\omega = a - b = 1$ and period $T = 2\pi$, represented by

$$X_0(\theta) = \begin{pmatrix} x_0(\theta) \\ y_0(\theta) \end{pmatrix} = \begin{pmatrix} \cos \theta \\ \sin \theta \end{pmatrix} \quad (0 \leqslant \theta < 2\pi).$$
 (21)

The phase function can be explicitly represented by

$$\Theta(x, y) = \arctan \frac{y}{x} - \frac{b}{2} \ln(x^2 + y^2)$$
 (22)

on the whole xy-plane except (0, 0), and the PSF is given by

$$\mathbf{Z}(\theta) = \begin{pmatrix} Z_{x} \\ Z_{y} \end{pmatrix} = \begin{pmatrix} -\sin\theta - b\cos\theta \\ \cos\theta - b\sin\theta \end{pmatrix}. \tag{23}$$

As another example, we use the FitzHugh-Nagumo (FHN) oscillator, described by

$$X = \begin{pmatrix} x \\ y \end{pmatrix} \in \mathbb{R}^2, \tag{24}$$

$$F = \begin{pmatrix} F_x \\ F_y \end{pmatrix} = \begin{pmatrix} x(x-c)(1-x) - y \\ \mu^{-1}(x-dy) \end{pmatrix}, \tag{25}$$

where the parameters are fixed at c=-0.1, d=0.5, and $\mu=100$. As μ is large, this oscillator is a slow-fast system whose x variable evolves much faster than the y variable, leading to relaxation oscillations. With these parameters, the natural period of the oscillation is $T\approx 126.7$ and the natural frequency is $\omega\approx 0.0496$. The limit cycle $X_0(\theta)=(x_0(\theta),y_0(\theta))^{\dagger}$ and phase function $\Theta(X)$ cannot be obtained analytically for this model, but the PSF $Z(\theta)$ can be obtained by numerically solving the adjoint equation. Figure 1 shows the time sequences of the limit-cycle orbit $X_0(\theta)$ and PSF $Z(\theta)$ for one period of oscillation, $0\leqslant \theta<2\pi$.

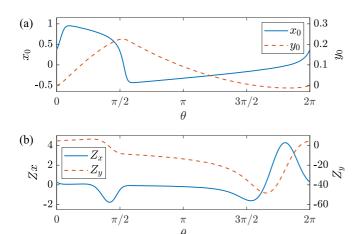


FIG. 1. Limit-cycle orbit and phase sensitivity function of the FitzHugh-Nagumo oscillator. Time sequences of the x and y components are plotted for one period of oscillation, $0 \le \theta < 2\pi$. (a) Limit cycle $(x_0(\theta), y_0(\theta))$. (b) Phase sensitivity function $(Z_x(\theta), Z_y(\theta))$.

III. LINEAR COUPLING

A. Linear coupling without self-coupling terms

In this section, we consider the case in which the oscillators are linearly coupled, i.e., the coupling functional can be expressed as $\hat{H}\{X_1^{(t)}, X_2^{(t)}\} = \hat{G}\{X_2^{(t)}\} + \hat{I}\{X_1^{(t)}\}$, where $\hat{G}\{X^{(t)}\}$ and $\hat{I}\{X^{(t)}\}$ are linear functionals of $X^{(t)}$. The oscillators obey

$$\dot{X}_{1}(t) = F(X_{1}(t)) + \epsilon \left[\hat{G} \left\{ X_{2}^{(t)} \right\} + \hat{I} \left\{ X_{1}^{(t)} \right\} \right],
\dot{X}_{2}(t) = F(X_{2}(t)) + \epsilon \left[\hat{G} \left\{ X_{1}^{(t)} \right\} + \hat{I} \left\{ X_{2}^{(t)} \right\} \right],$$
(26)

where $\hat{G}\{X_{2,1}^{(t)}\}$ and $\hat{I}\{X_{1,2}^{(t)}\}$ in each equation represent the effects of the coupling from the other oscillator and the self-coupling with itself, respectively. For example, when $\hat{I}\{X^{(t)}\} = -\hat{G}\{X^{(t)}\}$, we obtain diffusive coupling that depends on the state difference between the oscillators. In particular, when $\hat{I}\{X^{(t)}\} = 0$, we obtain coupled oscillators without the self-coupling terms,

$$\dot{X}_{1}(t) = F(X_{1}(t)) + \epsilon \hat{G}\{X_{2}^{(t)}\},
\dot{X}_{2}(t) = F(X_{2}(t)) + \epsilon \hat{G}\{X_{1}^{(t)}\}.$$
(27)

In Appendix A, it is shown that the above two models with and without the self-coupling term $\hat{I}\{X_{1,2}^{(t)}\}$ are equivalent within the phase-reduction approximation in the sense that both models have the in-phase synchronized state with the same linear stability as a solution. We thus analyze Eq. (27) in the following subsections.

As typical coupling schemes, we analyze a simple timedelayed coupling, where each oscillator is driven by the past state of the other oscillator with a fixed time delay, and a more general coupling via linear filtering, where each oscillator is driven by a linearly filtered signal of the past time sequences of the other oscillator.

B. Time-delayed coupling

First, we consider a simple time-delayed coupling, where each oscillator is driven by the past state of the other oscillator. The model is given by

$$\dot{X}_1 = F(X_1) + \epsilon \sqrt{P} K X_2(t - \tau),$$

$$\dot{X}_2 = F(X_2) + \epsilon \sqrt{P} K X_1(t - \tau),$$
(28)

where $0 < \epsilon \ll 1$ is a small parameter representing the strength of the interaction, P > 0 is a real constant that controls the norm of the coupling signal, $K \in \mathbb{R}^{N \times N}$ is a constant matrix specifying which components of the oscillator states $X_{1,2}(t)$ are coupled, and τ ($0 \leqslant \tau \leqslant L$) is a time delay. In our previous study [34], we considered optimization of the matrix K for the case in which the two oscillators are nearly identical and coupled without time delay. Here, we consider two oscillators with identical properties, fix the matrix K constant, and vary the time delay τ to improve the linear stability of the in-phase synchronized state.

In this case, the coupling functionals are given by

$$\hat{H}\left\{X_{1}^{(t)}, X_{2}^{(t)}\right\} = \sqrt{P}KX_{2}(t - \tau),$$

$$\hat{H}\left\{X_{2}^{(t)}, X_{1}^{(t)}\right\} = \sqrt{P}KX_{1}(t - \tau),$$
(29)

which can be expressed as functions of θ_1 and θ_2 as

$$H(\theta_1, \theta_2) = \sqrt{P} K X_0(\theta_2 - \omega \tau),$$

$$H(\theta_2, \theta_1) = \sqrt{P} K X_0(\theta_1 - \omega \tau),$$
(30)

after phase reduction. The phase coupling function is

$$\Gamma(\phi) = \langle \mathbf{Z}(\psi) \cdot \mathbf{H}(\psi, \psi - \phi) \rangle_{\psi}$$
$$= \langle \mathbf{Z}(\psi) \cdot \sqrt{P} \mathbf{K} X_0 (\psi - \phi - \omega \tau) \rangle_{\psi}, \qquad (31)$$

and the linear stability is characterized by

$$-\Gamma'(0) = \langle \mathbf{Z}(\psi) \cdot \mathbf{H}_2'(\psi, \psi) \rangle_{\psi}$$
$$= \langle \mathbf{Z}(\psi) \cdot \sqrt{P} \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau) \rangle_{\psi}, \tag{32}$$

where $X_0'(\theta) = dX_0(\theta)/d\theta$.

The maximum stability is attained only when τ satisfies

$$\frac{\partial}{\partial \tau} \{ -\Gamma'(0) \} = -\omega \langle \mathbf{Z}(\psi) \cdot \sqrt{P} \mathbf{K} \mathbf{X}_0''(\psi - \omega \tau) \rangle_{\psi} = 0, \quad (33)$$

where $X_0''(\theta) = d^2X_0(\theta)/d\theta^2$. We denote the value of τ satisfying the above equation as τ^* , i.e.,

$$\langle \mathbf{Z}(\psi) \cdot \sqrt{P} \mathbf{K} X_0''(\psi - \omega \tau^*) \rangle_{\psi} = 0. \tag{34}$$

By partial integration using the 2π -periodicity of $\mathbf{Z}(\theta)$ and $X_0(\theta)$, this can also be expressed as

$$\langle \mathbf{Z}'(\psi) \cdot \sqrt{P} \mathbf{K} \mathbf{X}'_0(\psi - \omega \tau^*) \rangle_{\psi} = 0, \tag{35}$$

which has the form of a cross-correlation function between $\mathbf{Z}'(\theta)$ and $\sqrt{P}\mathbf{K}\mathbf{X}'_0(\theta)$. Because both of these functions are 2π -periodic with zero-mean, the left-hand side of Eq. (35) is also 2π -periodic with zero mean. Thus, there are at least two values of τ satisfying the above equation, as long as $\mathbf{Z}'(\theta)$ and $\sqrt{P}\mathbf{K}\mathbf{X}'_0(\theta)$ are nonconstant functions (which holds generally for ordinary limit cycles). By choosing an appropriate value of τ^* , the maximum stability is given by

$$-\Gamma'(0) = \sqrt{P\langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau^*) \rangle_{\psi}^2}.$$
 (36)

C. Coupling via linear filtering

Generalizing the time-delayed coupling, we consider a case in which the past time sequences of both oscillator states are linearly filtered and used as driving signals for the other oscillators. The model is given by

$$\dot{X}_1 = F(X_1) + \epsilon \int_0^L h(\tau) \mathbf{K} X_2(t - \tau) d\tau,$$

$$\dot{X}_2 = F(X_2) + \epsilon \int_0^L h(\tau) \mathbf{K} X_1(t - \tau) d\tau,$$
(37)

where $h(\tau): [0, L] \to \mathbb{R}$ is a real scalar function representing a linear filter, which transforms the sequence of the oscillator state to a coupling signal, and $K \in \mathbb{R}^{N \times N}$ is a constant matrix specifying which components of X are coupled. We optimize the linear filter $h(\tau)$ for a given coupling matrix K under a constraint specified below.

The coupling functionals are given by

$$\hat{H}\left\{X_{1}^{(t)}, X_{2}^{(t)}\right\} = \int_{0}^{L} h(\tau) K X_{2}(t-\tau) d\tau,$$

$$\hat{H}\left\{X_{2}^{(t)}, X_{1}^{(t)}\right\} = \int_{0}^{L} h(\tau) K X_{1}(t-\tau) d\tau,$$
(38)

which simplify to ordinary functions

$$H(\theta_1, \theta_2) = \int_0^L h(\tau) K X_0(\theta_2 - \omega \tau) d\tau,$$

$$H(\theta_2, \theta_1) = \int_0^L h(\tau) K X_0(\theta_1 - \omega \tau) d\tau,$$
(39)

after phase reduction. The phase coupling function is given by

$$\Gamma(\phi) = \left\langle \mathbf{Z}(\phi + \psi) \cdot \int_0^L h(\tau) \mathbf{K} \mathbf{X}_0(\psi - \omega \tau) d\tau \right\rangle_{\psi}$$
$$= \left\langle \int_0^L \mathbf{Z}(\psi) \cdot h(\tau) \mathbf{K} \mathbf{X}_0(\psi - \omega \tau - \phi) d\tau \right\rangle_{\psi} \tag{40}$$

and the linear stability of the in-phase synchronized state is characterized by

$$-\Gamma'(0) = \left\langle \int_0^L \mathbf{Z}(\psi) \cdot h(\tau) \mathbf{K} X_0'(\psi - \omega \tau) d\tau \right\rangle_{\psi}.$$
 (41)

We constrain the L^2 -norm $||h(\tau)|| = \sqrt{\int_0^L h(\tau)^2 d\tau}$ of the linear filter, $h(\tau)$, as $||h(\tau)||^2 = Q$, where Q > 0 controls the overall coupling intensity, and we seek the optimal $h(\tau)$ that maximizes the linear stability, $-\Gamma'(0)$. That is, we consider an optimization problem:

maximize
$$-\Gamma'(0)$$
 subject to $||h(\tau)||^2 = Q$. (42)

To this end, we define an objective functional as

$$S\{h, \lambda\} = -\Gamma'(0) + \lambda [\|h(\psi)\|^2 - Q]$$

$$= \left\langle \int_0^L \mathbf{Z}(\psi) \cdot h(\tau) \mathbf{K} X_0'(-\omega \tau + \psi) d\tau \right\rangle_{\psi}$$

$$+ \lambda \left(\int_0^L h(\tau)^2 d\tau - Q \right), \tag{43}$$

where λ is a Lagrange multiplier. From the extremum condition of S, the functional derivative of S with respect to $h(\tau)$ should satisfy

$$\frac{\delta S}{\delta h(\tau)} = \langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(-\omega \tau + \psi) \rangle_{\psi} + 2\lambda h(\tau) = 0 \quad (44)$$

and the partial derivative of S by λ should satisfy

$$\frac{\partial S}{\partial \lambda} = \int_0^L h(\tau)^2 d\tau - Q = 0. \tag{45}$$

Thus, the optimal linear filter $h(\tau)$ is given by

$$h(\tau) = -\frac{1}{2\lambda} \langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau) \rangle_{\psi}. \tag{46}$$

The Lagrange multiplier λ is determined from the constraint $\|h(\tau)\|^2 = Q$, i.e.,

$$\frac{1}{4\lambda^2} \int_0^L \langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau) \rangle_{\psi}^2 d\tau = Q$$
 (47)

as

$$\lambda = -\sqrt{\frac{1}{4Q} \int_0^L \langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau) \rangle_{\psi}^2 d\tau}, \qquad (48)$$

where the negative sign should be chosen for the in-phase synchronized state to be linearly stable, $-\Gamma'(0) > 0$. The maximum linear stability with the optimized $h(\tau)$ is

$$-\Gamma'(0) = \sqrt{Q \int_0^L \langle \mathbf{Z}(\psi) \cdot \mathbf{K} \mathbf{X}_0'(\psi - \omega \tau) \rangle_{\psi}^2 d\tau}.$$
 (49)

D. Numerical examples

1. Setup

We use the SL and FHN oscillators in the following numerical illustrations. In the following examples, we assume that the length L of the time sequence is equal to the period T of the oscillation, L = T, that is, we use a whole period of the time sequence of the oscillation for the coupling. For both models, the coupling matrix is assumed to be

$$\mathbf{K} = \begin{pmatrix} 1 & 0 \\ 0 & 0 \end{pmatrix}. \tag{50}$$

We compare the optimized cases with the nonoptimized case, i.e.,

$$\dot{X}_1 = F(X_1) + \epsilon \sqrt{P} K X_2(t),$$

$$\dot{X}_2 = F(X_2) + \epsilon \sqrt{P} K X_1(t),$$
(51)

where ϵ is a small parameter that determines the coupling strength, and P controls the norm of the coupling signal.

2. Time-delayed coupling

In this case, the mean square of the coupling term over one period of oscillation is the same irrespective of the time delay, that is,

$$\langle |\sqrt{P}\mathbf{K}\boldsymbol{X}_{0}(\psi)|^{2}\rangle_{\psi} = \langle |\sqrt{P}\mathbf{K}\boldsymbol{X}_{0}(\psi - \omega\tau^{*})|^{2}\rangle_{\psi}. \tag{52}$$

First, for the SL oscillator, we can analytically calculate the optimal time delay. The linear stability of the in-phase

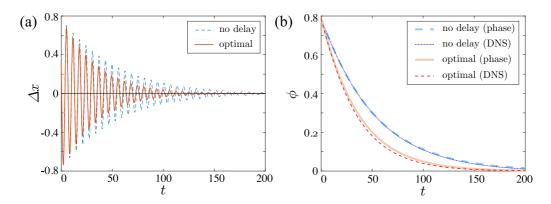


FIG. 2. Synchronization of two Stuart-Landau oscillators coupled with time delay. The results with the optimal time delay are compared with those without time delay. (a) Evolution of the difference Δx between x variables of the two oscillators. (b) Evolution of the phase difference ϕ between the oscillators, where results of direct numerical simulations (DNSs) of the original model are compared with those of the reduced phase model.

synchronized state, Eq. (32), is given by

$$-\Gamma'(0) = \sqrt{P} \langle Z_x(\psi) x_0'(\psi - \omega \tau) \rangle_{\psi}$$
$$= \frac{\sqrt{P}}{2} [\cos(\omega \tau) - b \sin(\omega \tau)]. \tag{53}$$

The optimal time delay $\tau = \tau^*$ is determined from Eq. (33), or equivalently from

$$\langle Z_x(\psi)x_0''(\psi-\omega\tau)\rangle_{\psi} = \frac{b\cos(\omega\tau) + \sin(\omega\tau)}{2} = 0. \quad (54)$$

For the parameter values b=1 and $\omega=1$, this equation is satisfied when $\tau=3\pi/4$ or $\tau=7\pi/4$ ($0\leqslant \tau < L=T$). Substituting this into Eq. (53), we find that $\tau^*=7\pi/4$ should be chosen, and the maximum linear stability is given by $-\Gamma'(0)=\sqrt{P}/\sqrt{2}$. For the case with no time delay, the linear stability is $-\Gamma'(0)=\sqrt{P}/2$. Thus, by appropriately choosing the time delay, the linear stability improves by a factor of $\sqrt{2}$ in this case.

Figure 2 shows synchronization of two SL oscillators for the cases with the optimal time delay and without time delay, where $\epsilon = 0.02$, P = 1, and the initial phase difference is $\phi(0) = \pi/4$. In Fig. 2(a), the difference Δx between the x variables of the two oscillators, obtained by direct numerical simulations of the coupled SL oscillators, is plotted as a function of t. It can be seen that the in-phase synchronized state is established faster in the optimized case because of the higher linear stability. Figure 2(b) shows the convergence of the phase difference ϕ to 0. It can be seen that the results of the reduced phase equation agree well with those of direct numerical simulations.

Figure 3 shows the results for two FHN oscillators, where $\epsilon = 0.003$, P = 1, and the initial phase difference is $\phi = \pi/4$. Figure 3(a) plots the linear stability $-\Gamma'(0)$ and its derivative $-\partial\Gamma'(0)/\partial\tau$ as functions of the time delay τ , where there are two extrema of $-\Gamma'(0)$. We choose the larger extremum, which is attained at the optimal time delay $\tau^* \approx 117.6$. The antisymmetric part of the phase coupling function, $\Gamma(\phi) - \Gamma(-\phi)$, is shown in Fig. 3(b) for the cases with the optimal delay and without delay.

It can be seen that the stability of the in-phase synchronized state $\phi = 0$ is improved, as indicated by the straight lines in

Fig. 3(b), where $-\Gamma'(0) \approx 0.654$ with the optimized time delay and $-\Gamma'(0) \approx 0.221$ without the time delay. The evolution of the difference Δx between the x variables of the oscillators is plotted as a function of t in Fig. 3(c). The phase differences ϕ converging toward 0, obtained from the phase equation and direct numerical simulations of the original model, are shown in Fig. 3(d). It can be seen that the convergence to the in-phase synchronization is faster with the optimized time delay, and the results of the reduced phase equation agree well with direct numerical simulations.

3. Coupling via linear filtering

We again assume that the coupling matrix K is given by Eq. (50), and we compare the results for the optimized case with linear filtering with those for the nonfiltered case given by Eq. (51). We choose the parameter Q that constrains the norm of the linear filter such that the squared average of the coupling term over one period of oscillation becomes equal to that in the nonfiltered case given by Eq. (51), i.e.,

$$\left\langle \left| \int_0^L h(\tau) \mathbf{K} X_0(\psi - \omega \tau) d\tau \right|^2 \right\rangle_{\psi} = \left\langle \left| \sqrt{P} \mathbf{K} X_0(\psi) \right|^2 \right\rangle_{\psi}. \tag{55}$$

For the SL oscillators, the optimal filter $h(\tau)$, Eq. (46), is explicitly calculated as

$$h(\tau) = \sqrt{\frac{Q\omega}{\pi (1 + b^2)}} [\cos(\omega \tau) - b \sin(\omega \tau)].$$
 (56)

The optimal phase coupling function, Eq. (40), and optimized linear stability, Eq. (49), are expressed as

$$\Gamma(\phi) = -\frac{1}{2} \sqrt{\frac{\pi (1 + b^2) Q}{\omega}} \sin \phi \tag{57}$$

and

$$-\Gamma'(0) = \frac{1}{2} \sqrt{\frac{\pi (1 + b^2)Q}{\omega}},$$
 (58)

respectively. We take $Q = \omega P/\pi$ so that Eq. (55) is satisfied. The linear stability is then $-\Gamma'(0) = \sqrt{(1+b^2)P}/2$ when the optimized linear filter is used and $-\Gamma'(0) = \sqrt{P}/2$ when no

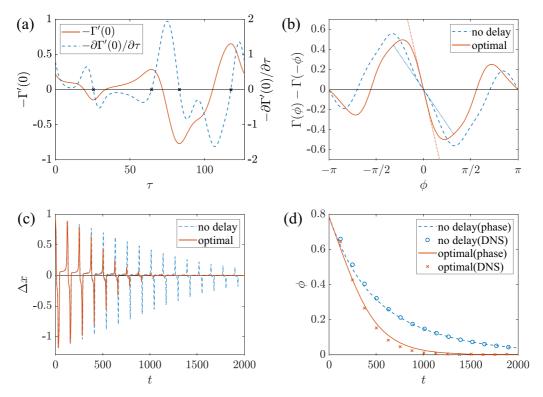


FIG. 3. Synchronization of two FitzHugh-Nagumo oscillators coupled with time delay. In (b)–(d), the results with the optimal time delay are compared with those without time delay. (a) Linear stability $-\Gamma'(0)$ and its derivative $-\partial\Gamma'(0)/\partial\tau$ vs time delay τ . The crosses indicate the values of τ where $-\partial\Gamma'(0)/\partial\tau=0$. (b) Antisymmetric part of the phase coupling function, $\Gamma(\phi)-\Gamma(-\phi)$. (c) Evolution of the difference Δx between x variables of the two oscillators. (d) Evolution of the phase difference ϕ between the oscillators. Results of DNSs of the original model are compared with those of the phase model.

filtering of the oscillator state is performed. Thus, the linear stability is improved by a factor of $\sqrt{2}$ when b = 1.

It is important to note that, in the SL oscillator case, $X_0(\psi)$, $Z(\psi)$, and hence the linear filter $h(\tau)$ contain only the fundamental frequency, i.e., they are purely sinusoidal. Thus, the linear filtering can only shift the phase of the coupling signal and gives the same result as the previous case with the simple time delay. It is also interesting to note that the stability cannot be improved (it is already optimal without filtering) when the parameter b, which characterizes nonisochronicity of the limit cycle, is zero.

Figure 4 shows the synchronization of two SL oscillators, with and without linear filtering, where $\epsilon=0.02$, P=1, and the initial phase difference is $\phi=\pi/4$. Figure 4(a) shows the evolution of the difference Δx between the x variables of the oscillators, and Fig. 4(b) shows the convergence of the phase difference ϕ to 0. We can see that the in-phase synchronized state is established faster in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well.

For the FHN oscillators, the optimal linear filter can be calculated from the time sequences of the limit-cycle solution

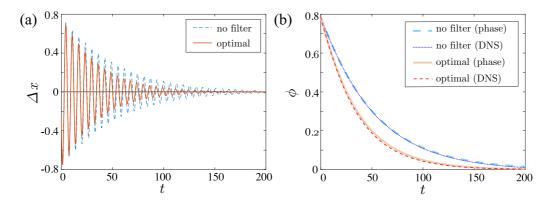


FIG. 4. Synchronization of two Stuart-Landau oscillators coupled with linear filtering. The results with the optimal filtering are compared with those without filtering. (a) Evolution of the difference Δx in the x variables between the oscillators. (b) Evolution of the phase difference ϕ between the oscillators. Results of DNSs of the original model are compared with those of the phase model.

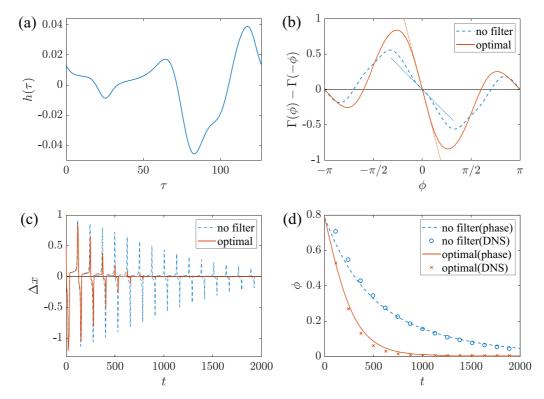


FIG. 5. Synchronization of two FitzHugh-Nagumo oscillators coupled with linear filtering. In (b)–(d), the results with the optimal filtering are compared with those without filtering. (a) Optimal linear filter $h(\tau)$. (b) Antisymmetric part of the phase coupling function $\Gamma(\phi) - \Gamma(-\phi)$. (c) Evolution of the difference Δx in the x variables between the oscillators. (d) Evolution of the phase difference ϕ between the oscillators. Results of DNSs of the original model are compared with those of the phase model.

and PSF obtained numerically. Figure 5 shows the synchronization of two coupled FHN oscillators, with and without linear filtering, where $\epsilon = 0.003$, P = 1, $Q \approx 0.0522$, and the initial phase difference is $\phi = \pi/4$. Figure 5(a) shows the optimal filter, (b) shows the antisymmetric part $\Gamma(\phi) - \Gamma(-\phi)$ of the phase coupling function $\Gamma(\phi)$, (c) shows the evolution of the difference Δx between the x variables of the oscillators, and (d) shows the convergence of the phase difference ϕ toward 0. The linear stability is given by $-\Gamma'(0) \approx 0.844$ for the case with the optimal filter and by $-\Gamma'(0) \approx 0.221$ for the case without filtering, as shown by the straight lines in Fig. 5(b). The in-phase synchronized state is established faster in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well. Because the FHN oscillator has the higher harmonic components in $X_0(\psi)$ and $Z(\psi)$, the optimal filter $h(\tau)$ can exploit these components, and hence the improvement in the linear stability is larger than that for the case with simple delay.

IV. NONLINEAR COUPLING

A. Mutual drive-response coupling

In this section, we consider the case in which the oscillators interact through nonlinear coupling functionals, and we seek the optimal forms of the coupling functional. It should be noted here that the functional form of $\hat{H}\{X_1^{(t)}, X_2^{(t)}\}$, which depends on both time sequences of the oscillators, cannot be determined in the present problem of optimizing the linear

stability $-\Gamma'(0)$ of the in-phase synchronized state. Because $-\Gamma'(0)$ is evaluated only in the completely in-phase synchronized state of the two oscillators, i.e., only when $X_1^{(t)} = X_2^{(t)} = X_0^{[\theta(t)]}$ [see Eqs. (8) and (18)], no information for the case with $X_1^{(t)} \neq X_2^{(t)}$ can be attained from $-\Gamma'(0)$.

We thus assume that the coupling is of a drive-response type, i.e., it can be written as a product of a response matrix of the driven oscillator and a driving function that transforms the signal from the other oscillator as

$$\hat{H}\{X_1^{(t)}, X_2^{(t)}\} = \hat{A}\{X_1^{(t)}\}\hat{G}\{X_2^{(t)}\},\tag{59}$$

where the matrix $\hat{\mathbf{A}}: C \to \mathbb{R}^{N \times N}$ is a functional of the time sequence of each oscillator representing its response properties, and $\hat{\mathbf{G}}: C \to \mathbb{R}^N$ is a functional that transforms the time sequence of the other oscillator to a driving signal. The model is given by

$$\dot{X}_{1}(t) = F(X_{1}(t)) + \epsilon \hat{A} \{X_{1}^{(t)}\} \hat{G} \{X_{2}^{(t)}\},
\dot{X}_{2}(t) = F(X_{2}(t)) + \epsilon \hat{A} \{X_{2}^{(t)}\} \hat{G} \{X_{1}^{(t)}\}.$$
(60)

Similar to the previous case of linear coupling, we may also include self-coupling terms of the form $\epsilon \hat{I}\{X_{1,2}^{(t)}\}$ to the model

$$\dot{X}_{1}(t) = F(X_{1}(t)) + \epsilon \left(\hat{A} \left\{ X_{1}^{(t)} \right\} \hat{G} \left\{ X_{2}^{(t)} \right\} + \hat{I} \left\{ X_{1}^{(t)} \right\} \right),
\dot{X}_{2}(t) = F(X_{2}(t)) + \epsilon \left(\hat{A} \left\{ X_{2}^{(t)} \right\} \hat{G} \left\{ X_{1}^{(t)} \right\} + \hat{I} \left\{ X_{2}^{(t)} \right\} \right).$$
(61)

For example, when $\hat{I}\{X^{(t)}\} = -\hat{A}\{X^{(t)}\}\hat{G}\{X^{(t)}\}$, we obtain (nonlinear) diffusive coupling. As explained in Appendix A,

inclusion of such self-coupling terms does not alter the results, and the linear stability remains the same in the framework of the phase-reduction theory. We thus analyze Eq. (60) in the following subsections.

The coupling functionals in this case are given by

$$\hat{\mathbf{H}} \{ X_1^{(t)}, X_2^{(t)} \} = \hat{\mathbf{A}} \{ X_1^{(t)} \} \hat{\mathbf{G}} \{ X_2^{(t)} \},
\hat{\mathbf{H}} \{ X_2^{(t)}, X_1^{(t)} \} = \hat{\mathbf{A}} \{ X_2^{(t)} \} \hat{\mathbf{G}} \{ X_1^{(t)} \},$$
(62)

and, as argued in Sec. II B, at the lowest-order phase reduction, these functionals can be expressed as ordinary functions of the phase θ_1 and θ_2 as

$$H(\theta_1, \theta_2) = A(\theta_1)G(\theta_2),$$

$$H(\theta_2, \theta_1) = A(\theta_2)G(\theta_1),$$
(63)

where we introduced ordinary 2π -periodic functions A and G of θ_1 and θ_2 . Using these functions, the phase coupling function is given by

$$\Gamma(\phi) = \langle \mathbf{Z}(\psi) \cdot \mathbf{A}(\psi)\mathbf{G}(\psi - \phi) \rangle_{\psi}, \tag{64}$$

and the linear stability is characterized by

$$-\Gamma'(0) = \langle \mathbf{Z}(\psi) \cdot \mathbf{A}(\psi) \mathbf{G}'(\psi) \rangle_{\psi}$$

$$= \langle \mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi) \cdot \mathbf{G}'(\psi) \rangle_{\psi}$$

$$= -\left\langle \frac{d}{d\psi} [\mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi)] \cdot \mathbf{G}(\psi) \right\rangle_{\psi}, \quad (65)$$

where the last expression is obtained by partial integration using 2π -periodicity of $A(\psi)$, $Z(\psi)$, and $G(\psi)$.

Therefore, although we started from Eq. (60) with a general drive-response coupling that depends on the past time sequences of the oscillators, the linear stability can be represented only by the present phase values of the oscillators at the lowest-order phase reduction. In the following subsections, we consider the optimization of the response matrix $A(\psi)$ or the driving function $G(\psi)$, represented as functions of the phase ψ .

B. Optimal response matrix

As for the first case, we optimize the response matrix $A(\psi)$ as a function of the phase ψ , assuming that the driving functional $\hat{\mathbf{G}}$ is given. We introduce a constraint that the squared Frobenius norm of $A(\psi)$ averaged over one period of oscillation is fixed as $\langle \|A(\psi)\|^2 \rangle_{\psi} = P$, and we consider an optimization problem:

maximize
$$-\Gamma'(0)$$
 subject to $(\|A(\psi)\|^2)_{\psi} = P$, (66)

where $||A|| = \sqrt{\sum_{i,j} A_{ij}^2}$ represents the Frobenius norm of the matrix $A = (A_{ij})$. By defining an objective functional,

$$S\{A, \lambda\} = -\Gamma'(0) + \lambda(\langle ||A(\psi)||^2)_{\psi} - P)$$

= $\langle \mathbf{Z}(\psi) \cdot A(\psi)\mathbf{G}'(\psi)\rangle_{\psi} + \lambda(\langle ||A(\psi)||^2)_{\psi} - P),$

where λ is a Lagrange multiplier, and by taking the functional derivative with respect to each component, A_{ij} , of A, we

obtain the extremum condition. In this case,

$$\frac{\delta S}{\delta A_{ii}(\psi)} = \frac{1}{2\pi} Z_i(\psi) G'_j(\psi) + \frac{\lambda}{\pi} A_{ij}(\psi) = 0, \qquad (68)$$

and we obtain

$$A_{ij}(\psi) = -\frac{1}{2\lambda} Z_i(\psi) G'_j(\psi), \tag{69}$$

i.e..

$$A(\psi) = -\frac{1}{2\lambda} \mathbf{Z}(\psi) \mathbf{G}'(\psi)^{\dagger}, \tag{70}$$

and the Lagrange multiplier is determined from the constraint,

$$\langle \|\mathbf{A}(\psi)\|^2 \rangle_{\psi} = \frac{1}{4\lambda^2} \langle \|\mathbf{Z}(\psi)\mathbf{G}'(\psi)^{\dagger}\|^2 \rangle_{\psi} = P \qquad (71)$$

as

$$\lambda = -\sqrt{\frac{1}{4P} \langle \|\mathbf{Z}(\psi)\mathbf{G}'(\psi)^{\dagger}\|^{2} \rangle_{\psi}}, \tag{72}$$

where the negative sign is chosen so that $-\Gamma'(0) > 0$. The maximum stability of the in-phase synchronized state is

$$-\Gamma'(0) = \sqrt{P(\|\mathbf{Z}(\psi)\mathbf{G}'(\psi)^{\dagger}\|^{2})_{\psi}}.$$
 (73)

C. Optimal driving function

We can also seek the function $G(\psi)$ that provides the optimal driving signal as a function of the phase ψ , assuming that the response matrix \hat{A} is given. We constrain the squared average of $G(\psi)$ over one period of oscillation as $\langle |G(\psi)|^2 \rangle_{\psi} = P$, and we maximize the linear stability of the in-phase state:

maximize
$$-\Gamma'(0)$$
 subject to $\langle |G(\psi)|^2 \rangle_{\psi} = P$. (74)

We define an objective functional,

$$S\{G, \lambda\} = -\Gamma'(0) + \lambda(\langle |G(\psi)|^2 \rangle_{\psi} - P)$$

$$= -\left\langle \frac{d}{d\psi} [A^{\dagger}(\psi) \mathbf{Z}(\psi)] \cdot G(\psi) \right\rangle_{\psi}$$

$$+ \lambda(\langle |G(\psi)|^2 \rangle_{\psi} - P), \tag{75}$$

where λ is a Lagrange multiplier. From the extremum condition for S, we obtain

$$\frac{\delta S}{\delta \mathbf{G}(\psi)} = -\frac{1}{2\pi} \frac{d}{d\psi} [\mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi)] + \frac{\lambda}{\pi} \mathbf{G}(\psi) = 0 \quad (76)$$

and the constraint on G. The optimal driving function is given by

$$G(\psi) = \frac{1}{2\lambda} \frac{d}{d\psi} [A^{\dagger}(\psi) \mathbf{Z}(\psi)], \tag{77}$$

where the Lagrange multiplier λ should be chosen to satisfy the norm constraint,

$$\frac{1}{4\lambda^2} \left\langle \left| \frac{d}{d\psi} [\mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi)] \right|^2 \right\rangle_{\psi} = P. \tag{78}$$

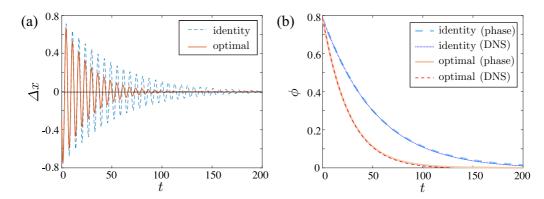


FIG. 6. Synchronization of Stuart-Landau oscillators with the optimal response matrix. (a) Evolution of the difference Δx between the x variables of the oscillators. (b) Evolution of the phase difference ϕ between the oscillators. DNSs of the original model vs the reduced model.

This yields

$$\lambda = -\sqrt{\frac{1}{4P} \left\langle \left| \frac{d}{d\psi} [\mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi)] \right|^2 \right\rangle_{\psi}}, \tag{79}$$

where the negative sign is taken to satisfy $\Gamma'(0) < 0$. The maximum stability is

$$-\Gamma'(0) = \sqrt{P\left\langle \left| \frac{d}{d\psi} [\mathbf{A}^{\dagger}(\psi) \mathbf{Z}(\psi)] \right|^2 \right\rangle_{\psi}}.$$
 (80)

D. Numerical examples

1. Optimal response matrix

As an example, we assume that the driving functional $\hat{G}\{X^{(t)}\}$ is simply given by $\hat{G}\{X^{(t)}\}=X(t)$, and we seek the optimal response matrix $A(\psi)$ satisfying $\langle \|A(\psi)\|^2 \rangle_{\psi} = P$. For comparison, we also consider an identity response matrix, $A_I = \operatorname{diag}(\sqrt{P/2}, \sqrt{P/2})$, normalized to satisfy $\langle \|A_I\|^2 \rangle_{\psi} = P$. Note that both the x and y components are coupled, in contrast to the previous section where only the x component is coupled.

For the SL oscillator, the optimal response matrix can be analytically expressed as

$$A(\psi) = \sqrt{\frac{P}{1+b^2}} \begin{pmatrix} \sin\psi(b\cos\psi + \sin\psi) & -\cos\psi(b\cos\psi + \sin\psi) \\ \sin\psi(b\sin\psi - \cos\psi) & \cos\psi(\cos\psi - b\sin\psi) \end{pmatrix}, \tag{81}$$

and the phase coupling function is given by $\Gamma(\phi) = -\sqrt{(1+b^2)P}\sin\phi$, which gives the optimal linear stability $-\Gamma'(0) = \sqrt{(1+b^2)P}$. In contrast, for the identity matrix A_I , the phase coupling function is $\Gamma(\phi) = -\sqrt{P/2}(b\cos\phi + \sin\phi)$ and the linear stability is $-\Gamma'(0) = \sqrt{P}/\sqrt{2}$. Thus, the linear stability is improved by a factor of $\sqrt{2}(1+b^2)$.

Figure 6 shows synchronization of two SL oscillators for the cases with the optimal response matrix $A(\psi)$ and with the identity matrix $A_I(\psi)$, where $b=1, \epsilon=0.01, P=2$, and the initial phase difference is $\phi=\pi/4$. Figure 6(a) shows the evolution of the difference Δx in the x variables between the two oscillators, and Fig. 6(b) shows the convergence of the phase difference ϕ to 0. The in-phase synchronized state is more quickly established in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well.

For the FHN oscillator, the optimal response matrix can be calculated numerically. Figure 7 compares the synchronization dynamics of two coupled FHN oscillators with the optimal and identity response matrices, where $\epsilon = 0.0002$, P = 2, and the initial phase difference is $\phi = \pi/4$. Figure 7(a) shows four components of the optimal response matrix $A(\psi)$

for $0 \le \psi < 2\pi$. It is notable that the magnitude of $A_{21}(\psi)$ is much larger than the other components, indicating that driving the y component of each oscillator by using the x component of the other oscillator is efficient in synchronizing the oscillators in this case. Figure 7(b) plots the antisymmetric part of the phase coupling functions for the optimal and identity response matrices, which shows that a much higher stability is attained in the optimized case $[-\Gamma'(0) \approx 10.1$ for the optimized response matrix and $-\Gamma'(0) \approx 0.999$ for the identity response matrix].

Figure 7(c) shows the time evolution of the difference Δx between the two oscillators, and Fig. 7(d) shows the convergence of the phase difference ϕ to 0. To use the optimal response matrix, instantaneous phase values of the oscillators are necessary. In the direct numerical simulations shown here, we approximately evaluated the phase value by linearly interpolating two consecutive crossings times of the oscillator state at an appropriate Poincaré section, and this value was used to generate the driving signal. It can be seen from the figures that the in-phase synchronized state is established much faster in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well.

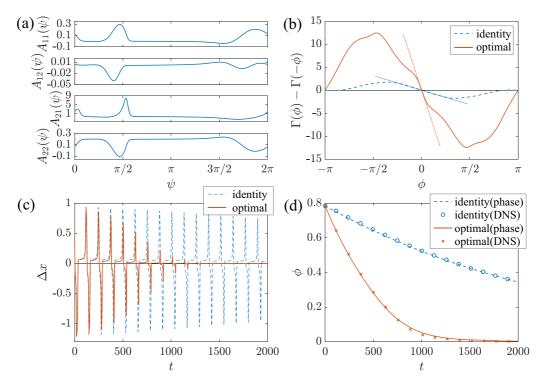


FIG. 7. Synchronization of FitzHugh-Nagumo oscillators with the optimal response matrix. In (b)–(d), the results with the optimal response matrix are compared with those with the identity response matrix. (a) Four components of the optimal response matrix $A(\psi)$. (b) Antisymmetric part of the phase coupling function, $\Gamma(\phi) - \Gamma(-\phi)$. (c) Evolution of the difference Δx between the x variables of the oscillators. (d) Evolution of the phase difference ϕ between the oscillators. DNSs of the original model vs the reduced model.

2. Optimal driving function

For the numerical simulations, we assume that $\hat{A}\{X^{(t)}\}$ is simply given by an identity matrix, diag(1, 1). The optimal driving function $G(\psi)$ is then simply given as $G(\psi) \propto \mathbf{Z}'(\psi)$ from Eq. (77), with the norm constraint $\langle |G(\psi)|^2 \rangle_{\psi} = P$. For the SL oscillator, the optimal driving function is explicitly given by

$$G(\psi) = \sqrt{\frac{P}{1+b^2}} \begin{pmatrix} \cos \psi - b \sin \psi \\ b \cos \psi + \sin \psi \end{pmatrix}. \tag{82}$$

Figure 8 shows synchronization of two SL oscillators coupled through the optimal driving function, and coupled without

transformation of the oscillator state, i.e., $\hat{G}\{X^{(t)}\} = X(t)$, where b = 1, $\epsilon = 0.01$, and P = 1, and the initial phase difference is $\phi = \pi/4$. Figure 8(a) shows the evolution of the difference Δx between the x variables of the two oscillators, and Fig. 8(b) shows the convergence of the phase difference ϕ to 0. It is confirmed that the linear stability of the in-phase synchronized state is improved in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well.

For the FHN oscillator, the norm of $X_0(\psi)$ is $\langle |X_0(\psi)|^2 \rangle_{\psi} \approx 0.221$, and we fix the norm P of $G(\psi)$ to this value. The optimal driving function can be calculated

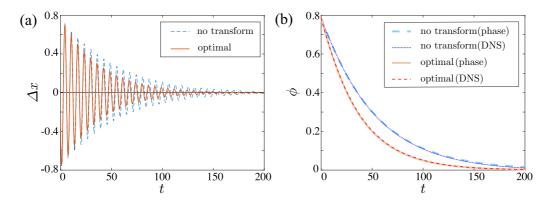


FIG. 8. Synchronization of Stuart-Landau oscillators coupled with the optimal driving function. (a) Evolution of the difference Δx between the x variables of the oscillators. (b) Evolution of the phase difference ϕ between the oscillators. DNSs of the original model vs the reduced model.

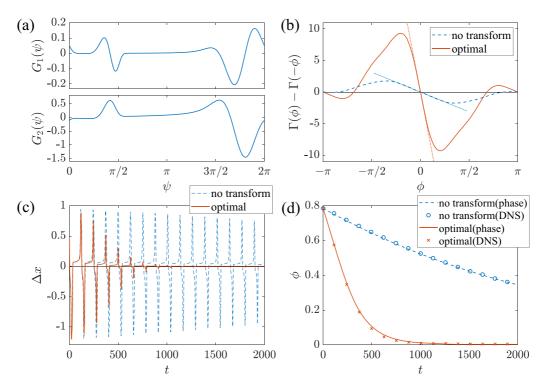


FIG. 9. Synchronization of FitzHugh-Nagumo oscillators coupled with the optimal driving function. In (b)–(d), the results with the optimal driving function are compared with those without transformation. (a) Optimal driving function $G(\psi) = (G_1(\psi), G_2(\psi))^{\dagger}$. (b) Antisymmetric part of the phase coupling function, $\Gamma(\phi) - \Gamma(-\phi)$. (c) Evolution of the difference Δx between the x variables of the oscillators. (d) Evolution of the phase difference ϕ . DNSs of the original model vs the reduced model.

from $X_0(\psi)$ and $Z(\psi)$ obtained numerically. Figure 9 shows synchronization of two FHN oscillators coupled with the optimal driving function, as well as a comparison with the nontransformed case, where $\epsilon=0.0002$, $P\approx0.221$, and the initial phase difference is $\phi=\pi/4$. Figure 9(a) shows the optimal driving function $G(\psi)$ for $0\leqslant\psi<2\pi$, which is proportional to the derivative $Z'(\psi)$. Figure 9(b) plots the antisymmetric part of the phase coupling function for the optimal driving function $G(\psi)$ with and without transformation, respectively, indicating a much higher linear stability in the optimized case $[-\Gamma'(0)\approx12.8$ with the optimized driving function and $-\Gamma'(0)\approx0.999$ without optimization].

Figure 9(c) shows a plot of the evolution of the difference Δx between x variables of the oscillators, and Fig. 9(d) shows the convergence of the phase difference ϕ to 0. Similar to the previous case with the optimal response matrix, instantaneous phase values of the oscillators are approximately evaluated by linear interpolation and used to generate the optimal driving signal in the direct numerical simulations. We can confirm that the in-phase synchronized state is established much faster in the optimized case, and the results of the reduced phase model and direct numerical simulations agree well.

V. DISCUSSION

We have shown that by optimizing the mutual coupling between coupled oscillators, the linear stability of the in-phase synchronized state can be improved, and faster convergence to the synchronization can be achieved. We have shown that, even if we start from a system of coupled oscillators with general coupling functionals that depend on the past time sequences of the oscillators, the system can be approximately reduced to a pair of simple ordinary differential equations that depend only on the present phase values of the oscillators within the phase-reduction theory, and the optimal coupling function between the oscillators can be obtained as a function of the phase values. Though we have considered only the simplest cases in which two oscillators with identical properties are symmetrically coupled without noise, the theory can also be extended to include heterogeneity of the oscillators or noise.

The linear coupling with time delay or linear filtering can be realized without measuring the phase values of the oscillator once the correlation functions of the PSF and the limit-cycle orbit (or their derivatives) are obtained. The nonlinear coupling requires the measurement of the phase values of the oscillators, but it can further improve the linear stability of the synchronized state. We have shown that a simple approximate evaluation of the phase values by a linear interpolation gives reasonable results even though it may yield a somewhat incorrect evaluation of the true phase values.

It is interesting to compare the present analysis for stable synchronization between the two oscillators with the optimization of driving signals for injection locking of a single oscillator, which has been analyzed by Zlotnik *et al.* [24] and others (briefly explained in Appendix B for a simple case). In Sec. IV C, we have obtained the optimal driving function. In

particular, when $A(\psi) = K$, where K is a constant matrix, the optimal driving signal is

$$G(\psi) = \frac{1}{2\lambda} \mathbf{K}^{\dagger} \mathbf{Z}'(\psi) \tag{83}$$

and the maximum stability is

$$-\Gamma'(0) = \sqrt{P\langle |\mathbf{K}^{\dagger} \mathbf{Z}'(\psi)|^2 \rangle_{\psi}}.$$
 (84)

This result coincides with the optimal injection signal for stable synchronization of a single oscillator, obtained by Zlotnik *et al.* [24]. Thus, the optimal coupling between the oscillators is realized by measuring the present phase ψ of the other oscillator and applying a driving signal that is proportional to $K^{\dagger}Z'(\psi)$ to the oscillator.

It is also interesting to note that we have obtained similar expressions for the maximum stability in all examples, $-\Gamma'(0) = \sqrt{P\langle \cdots \rangle_{\psi}^2}$, where \cdots depends on the quantity to be optimized. This is because we are essentially maximizing the inner product of the PSF with the derivative of the driving signal under a mean-square constraint on the parameters or functions included in the driving signal in all cases.

The linear coupling schemes in Sec. III would be easy to realize experimentally. The nonlinear coupling schemes in Sec. IV require an evaluation of the phase values from the oscillators, but they can yield an even higher linear stability. These methods may be useful when higher stability of the in-phase synchronized state between oscillators is desirable in technical applications. It would also be interesting to study interactions between rhythmic elements, e.g., in biological systems from the viewpoint of synchronization efficiency.

ACKNOWLEDGMENTS

This work is financially supported by Japan Society for the Promotion of Science KAKENHI Grants No. JP16K13847, No. JP17H03279, No. 18K03471, No. JP18H03287, and No. 18H06478, and JST CREST Grant No. JPMJCR1913.

APPENDIX A: MODELS WITH SELF COUPLING

We show here that the inclusion of self-coupling terms in the model does not alter the linear stability of the in-phase synchronized state under the phase-reduction approximation. Suppose that we have additional self-coupling terms in the model as

$$\dot{X}_{1} = F(X_{1}) + \epsilon \left[\hat{H} \left\{ X_{1}^{(t)}, X_{2}^{(t)} \right\} + \hat{I} \left\{ X_{1}^{(t)} \right\} \right],
\dot{X}_{2} = F(X_{2}) + \epsilon \left[\hat{H} \left\{ X_{2}^{(t)}, X_{1}^{(t)} \right\} + \hat{I} \left\{ X_{2}^{(t)} \right\} \right],$$
(A1)

where $\hat{I}\{X^{(t)}\}$ is a functional representing self-coupling. A typical example is coupled oscillators with linear diffusive coupling,

$$\dot{X}_1 = F(X_1) + \epsilon(X_2 - X_1),$$

$$\dot{X}_2 = F(X_2) + \epsilon(X_1 - X_2),$$
(A2)

where we may take $\hat{H}\{X_1^{(t)}, X_2^{(t)}\} = X_2(t)$ and $\hat{I}\{X_1^{(t)}\} = -X_1(t)$. By phase reduction, we obtain the phase equations

$$\dot{\theta}_1 = \omega + \epsilon \mathbf{Z}(\theta_1) \cdot [\mathbf{H}(\theta_1, \theta_2) + \mathbf{I}(\theta_1)],
\dot{\theta}_2 = \omega + \epsilon \mathbf{Z}(\theta_2) \cdot [\mathbf{H}(\theta_2, \theta_1) + \mathbf{I}(\theta_2)],$$
(A3)

and the phase coupling function

$$\tilde{\Gamma}(\phi) = \langle \mathbf{Z}(\psi) \cdot [\mathbf{H}(\psi, \psi - \phi) + \mathbf{I}(\psi)] \rangle_{\psi}$$

$$= \Gamma(\phi) + \langle \mathbf{Z}(\psi) \cdot \mathbf{I}(\psi) \rangle_{\psi}, \tag{A4}$$

where $\Gamma(\phi)$ is the phase coupling function for the case without the self-coupling term, and the second term is a constant. Thus, this model also has the in-phase synchronized state as a fixed point, and its linear stability is equal to the case without the self-coupling term,

$$-\epsilon \Gamma'(0) = -\epsilon \tilde{\Gamma}'(0). \tag{A5}$$

APPENDIX B: OPTIMAL SIGNAL FOR INJECTION LOCKING

In this Appendix, we briefly review the result of Zlotnik *et al.* [24] on the optimal driving signal for injection locking for a simple case. We consider a limit-cycle oscillator driven by a periodic driving signal whose period is the same as the natural period *T* of the oscillator,

$$\dot{X} = F(X) + \epsilon K f(t), \quad f(t) = f(t+T), \quad (B1)$$

where X is the oscillator state, F(X) represents its dynamics, and $\epsilon K f(t)$ is a weak periodic driving signal, where ϵ is a small positive parameter, and a constant matrix $K \in \mathbb{R}^{N \times N}$ represents which components of X are driven by f(t).

By phase reduction, we obtain a phase equation

$$\dot{\theta} = \omega + \mathbf{Z}(\theta) \cdot \mathbf{K} \mathbf{f}(t) \tag{B2}$$

for the oscillator phase θ , where $\mathbf{Z}(\theta)$ is the PSF. Defining $\theta - \omega t = \phi$ and averaging over one oscillation period yields

$$\dot{\phi} = \Gamma(\phi). \tag{B3}$$

The phase coupling function $\Gamma(\psi)$ is expressed as

$$\Gamma(\phi) = \frac{1}{2\pi} \int_0^{2\pi} \mathbf{Z}(\phi + \psi) \cdot \mathbf{K} f(\psi/\omega) d\psi$$
$$= \frac{1}{2\pi} \int_0^{2\pi} \mathbf{Z}(\phi + \psi) \cdot \mathbf{K} \tilde{f}(\psi) d\psi$$
$$= \langle \mathbf{Z}(\phi + \psi) \cdot \mathbf{K} \tilde{f}(\psi) \rangle_{\psi}, \tag{B4}$$

where we have defined $\tilde{f}(\psi) = f(\psi/\omega)$.

By choosing the origin of the phase of the periodic signal so that $\Gamma(0)=0$ holds, the oscillator synchronizes with the periodic signal at $\phi=0$, and the linear stability of this synchronized state is given by

$$\Gamma'(0) = \langle \mathbf{Z}'(\psi) \cdot \mathbf{K}\tilde{\mathbf{f}}(\psi) \rangle_{\psi}$$
$$= \langle \mathbf{K}^{\dagger}\mathbf{Z}'(\psi) \cdot \tilde{\mathbf{f}}(\psi) \rangle_{\psi}. \tag{B5}$$

We constrain the one-period average of $\hat{f}(\psi)$ as

$$\langle |\tilde{f}(\psi)|^2 \rangle_{\psi} = P,$$
 (B6)

and we consider an objective function

$$S\{\tilde{f},\lambda\} = -\Gamma'(0) + \lambda(\langle |\tilde{f}(\psi)|^2 \rangle_{\psi} - P)$$

= $-\langle K^{\dagger} \mathbf{Z}'(\psi) \cdot \tilde{f}(\psi) \rangle_{\psi} + \lambda(\langle |\tilde{f}(\psi)|^2 \rangle_{\psi} - P), \text{ (B7)}$

where λ is a Lagrange multiplier. Extremum conditions are given by

$$\frac{\delta S}{\delta \tilde{f}(\psi)} = -\frac{1}{2\pi} K^{\dagger} \mathbf{Z}'(\psi) + \frac{\lambda}{\pi} \tilde{f}(\psi) = 0, \quad (B8)$$

$$\frac{\partial S}{\partial \lambda} = \langle |\tilde{f}(\psi)|^2 \rangle_{\psi} - P = 0.$$
 (B9)

The optimal driving signal is given by

$$\tilde{f}(\psi) = \frac{1}{2\lambda} K^{\dagger} \mathbf{Z}'(\psi) \tag{B10}$$

and the constraint is

$$\frac{1}{4\lambda^2} \langle |\mathbf{K}^{\dagger} \mathbf{Z}'(\psi)|^2 \rangle_{\psi} = P, \tag{B11}$$

which yields

$$\lambda = -\sqrt{\frac{1}{4P} \langle |\mathbf{K}^{\dagger} \mathbf{Z}'(\psi)|^2 \rangle_{\psi}}, \tag{B12}$$

where the negative sign should be taken in order that $\Gamma'(0) < 0$. Thus, the optimal driving signal is given by

$$\tilde{f}(\psi) = -\sqrt{\frac{P}{\langle |\mathbf{K}^{\dagger} \mathbf{Z}'(\psi)|^2 \rangle_{\psi}}} \mathbf{K}^{\dagger} \mathbf{Z}'(\psi)$$
 (B13)

and the maximum linear stability is given by

$$-\Gamma'(0) = -\langle \mathbf{K}^{\dagger} \mathbf{Z}'(\psi) \cdot \tilde{\mathbf{f}}(\psi) \rangle_{\psi} = \sqrt{P \langle |\mathbf{K}^{\dagger} \mathbf{Z}'(\psi)|^{2} \rangle_{\psi}}.$$
(B14)

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