Performance Analysis of High Speed and Area Efficient Finite Impulse Response Filters

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Abstract

This study represents the designing and implementation of a bandpass finite impulse response (FIR) filter of order 31, using windowing techniques. The frequency parameters used are of a typical GSM receiver,¹⁹ which is one of the applications of software-defined radio (SDR). To minimize filter area, various multiplication techniques, like a canonical signed digit, *Vedic* multiplier, booth multiplier, and modified booth multiplier are used. Adders, like ripple carry adder, carry save adder, carry look ahead adder, and Kogge-Stone adder are used to add the product from the multiplier unit. A comparison between three different windows has been made. The FIR is designed in MATLAB using a windowing technique. Then, it is synthesized on Xilinx 14.7, Virtex 6 XC6VLX760, whose results are included in this paper.

Keywords: Digital filters, Finite impulse response (FIR), FPGA.

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INTRODUCTION

n today's world, we witness radios in numerous applications such as cell phones, door openers, televisions, computers, etc. Radio is a device that is capable of transmitting and receiving signals in the radio frequency. When some part of the radio's physical layers function is software-defined, then they are known to be SDR. The most common application of SDR is found in communication systems, which are cost-efficient and are flexible. Their application can also be found in intensive care units where a large number of wired and wireless electronic devices must coexist.¹ The basic block diagram of a digital communication system is shown in Figure 1. SDR is used to implement the digital part.² Many applications work at a high sampling rate; this rapidly changing sampling rates require fast multipliers and efficient filters. The performance of the system depends on filters, mixers, amplifiers modulators, and demodulators, etc. The goal must be to reduce area and increase speed performance. A higher-order FIR Filter used for various SDR applications is suggested³ using canonical signed digit (CSD) and common subexpression elimination method (CSE). A re-configurable FIR filter has been suggested using CSD.⁴ Various multipliers, and adders⁵ were used, and a comparison was drawn to reduce hardware cost, i.e., area utilization. We extensively studied the performance of the FIR filter for various computational (multipliers and adders) elements for the transposed FIR filter structure. Windowing techniques, like a rectangular, Hamming window, Hanning window were found to fit the given filter specifications. The filter was designed for a word length of 32 bits.

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This paper is structured as follows. Section 2 is about FIR filters. The architecture used is discussed in section 3, followed by multipliers used in the proposed method

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FIR FILTER WITH WINDOWING TECHNIQUE

Filters are classified into two types, digital filters, and analog filters. Digital filters contain FIR and infinite impulse





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response (IIR). As FIR is defined for N number of values, a finite value, they are called finite impulse response. IIR filters cannot guarantee stability, and they are not linear phased following these reasons, they are undesirable.⁶

$$y[n] = \sum_{k=0}^{M-1} b[k]x[n-k]$$

The output here depends only upon current input and previous input, unlike IIR, which also depends on previous output values. FIR filter is shown below in Figure 2. M is the order of the filter. x[n] is the input sequence, b[k] is filter coefficients, which were calculated using the windowing technique.

The transposed direct form has been used to implement the filter. It is constructed from the direct form by following the steps given below.^{7,8}

- Exchanging the input and output.
- Inverting the direction of signal flow.

The main blocks of the filter are multiplier, adder, and a delay block.¹ The multiplier multiplies input with the coefficients; adder adds the outputs from delay block and multipliers. The linear phase filters are positive or negative symmetry depending on M, as shown below equation.

$$h(n) = \pm h(M - n - 1)$$

When M is even the number of multiplications can be reduced from M to M/2 and for odd value (M-1)/2, thereby reducing hardware utilization to half.^{9,10}

The filter length is 32, which is M + 1. The filter order can be calculated using Kaiser or Hermann, Rabiner developed formula.⁵ The following equations are used to calculate filter coefficients using the windowing technique.

$$h_{a}(n) = \frac{1}{2 \prod} \int_{-\Pi}^{\Pi} H_{d}(w) e^{jwn} dw$$
$$h(n) = h_{d}(n) w(n)$$
$$H(w) = \sum_{n=0}^{M-1} h(n) e^{-jwn}$$

Direct truncation of $h_d(n)$ to M terms to obtain h(n) leads to Gibbs effect due to which overshoot and ripples, before



Figure 2: Transposed direct form FIR filter structure

Table 1: FIR filter design parameter			
Filter parameter	Set		
Sampling frequency	4,800 MHz		
Passband frequency 1	925 MHz		
Passband frequency 2	960 MHz		
Order	31		

and after approximated discontinuity, occur.^{5,12} Rectangular, Hamming, and Hanning window are found most suitable for the given filter specifications. A comparison is drawn in these three windows. The generalized cosine windows W(N) is given below.

$$W(n) = A - B\cos(2\pi(n+1)(N+1) + C\cos(4\pi(n+1)(N+1)))$$

n = 0, 1, 2,..., n - 1

The filter designed using Table 1 has symmetric coefficients. The filter coefficients were represented in fixed-point representation (Figure 3).

ARCHITECTURE

The filter is implemented using a transposed form structure. The multiplication of filter coefficients is performed using multipliers, like CSD, booth multiplier, modified booth multiplier, *Vedic* multiplier. Adders, like carry save adder, carry look-ahead adder are used to add delayed outputs of multipliers.

Vedic Multiplier

Veda is a Sanskrit word meaning "storehouse of knowledge." They are divided into four parts. *Vedic* mathematics contains 16 principles called "sutra." "Urdhva Tiryabham" (UT) is the most popular multiplication technique. It is the fast and easiest technique to perform a multiplication operation. UT means vertically and clockwise.¹¹ The hardware architecture of the 32-bit *Vedic* multiplier is shown in Figure 4 and Table 2.

Modified Booth

This algorithm promises fast multiplication by reducing partial products. Here, multipliers LSB's three bits are encoded to -2, -1, 0, 1, and 2.¹² A zero is appended to the LSB, and grouping of three bits is done as shown in Figure 5.

After grouping, the respective operation is performed, as given below in Table 3.



Figure 3: Magnitude response of finite impulse response (FIR) filter with given set of frequencies using different windows

Table 2: Value of coefficier	nts for A, B, and C
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Window	A	В	С	
Rectangular	1	0	0	
Hanning	0.5	0.5	0	
Hamming	0.54	0.46	0	

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Figure 5: Grouping of numbers by overlapping technique

Booth Multiplier

The booth algorithm performs multiplication.¹³ It examines adjacent bits in a pair of two bits. N/2 cycles are examined, and respective operation is performed as given in the Table 4. It uses a two's complement notation of a signed binary number for multiplication.

The steps followed are:

- Choose multiplier and multiplicand.
- Only in the first step initialize accumulator with zero.
- Number of multiplicand bits is initialized in count register.
- Current LSB and previous LSB are used to decide the operation to be performed.
- After every arithmetic operation, a right shift is performed.
- Until count registers become "0," continue the above steps in a cycle.

Canonical Signed Digit (CSD)

The CSD is used to encode a number in signed digit representation. Digits 1, 0, and -1 are used for coding.¹⁴⁻¹⁶ The following rules are needed to be taken care of while coding in CSD format.¹⁷

Two consecutive numbers must not be non-zero.

Minimum non-zero digits must be used to represent the value.

RESULTS AND **D**ISCUSSION

Tables-5 and 6 compare the area consumed and delay taken by adders and multipliers. From the table, it is clear that the *Vedic* multiplier achieves minimum delay and also a minimum area. Among the adders, the ripple carry adder has minimum slice utilization and carry-save adder minimum delay. Here, MATLAB FDA toolbox is used to generate coefficients of the filter using windowing technique. The device package used in Xilinx ISE 14.7 is Virtex 6 XC6VLX760.¹⁸

Figures 6 to 8 are simulation results of the proposed work. Table 7 shows the comparison of synthesis results; from

Modified booth recording table multiplier bits		Encoded multiplier	Partial		
Qi + 1	Qi	Qi - 1	value∑	products	Operation
0	0	0	0	0 M	No action
0	0	1	1	1 M	Add
0	1	0	1	1 M	Add
0	1	1	2	2 M	Shift left and add
1	0	0	-2	-2 M	Shift left and subtract
1	0	1	-1	-1 M	Subtract
1	1	0	-1	-1 M	Subtract
1	1	1	0	0 M	No action

Table 3. Modified booth recoding table

Table 4: Operations to be performed

bi bi-1 Arithmetic operation 0 0 No arithmetic operation 1 No arithmetic operation 0 1 Add multiplicand to the current value of the accumulator product 1 0 Subtract multiplicand from the current value of the accumulator product

Table 5: Comparison of 32-bit multipliers

		•		
Attributes	Vedic multiplier	Booth multiplier	Modified booth	Canonical signed digit multiplier
No. of slices	1,364	2,001	2,203	2,122
No. of LUT's	2,486	4,001	4,139	3,896
Delay (ns)	32.146	128.091	89.884	63.897

Table 6: Comparison of 64-bit adders

		•		
Attributes	Kogge- Stone	Ripple carry adder	Carry save adder	Carry look- ahead adder
No. of slices	329	74	188	107
No. of LUT's	599	128	334	176
Delay (ns)	22.889	72.616	20.248	58.372







Figure 7: Ripple carry adder with booth multiplier for Hanning window



Computational elements		Rectangular w	Rectangular window		Hanning window		, Hamming window	
Adders	Multipliers	No. of slice register	Maximum frequency (MHz)	No. of slice register	Maximum frequency (MHz)	No. of slice register	Maximum frequency (MHz)	
CSA	Booth	61,532	266.38	57,689	289.6	61,642	266.386	
CLA	CSD	95,113	284.79	94,555	284.79	99,624	284.799	
Kogge	Modified booth	66,348	137.88	62,331	137.88	66,380	137.885	
CSA	Vedic	19,215	289.6	18,816	289.6	19,884	266.386	
RCA	Booth	60,108	288.08	56,273	280.48	59,948	280.485	
CLA	Modified booth	54,850	284.79	51,460	284.79	54,950	277.4	
Kogge	CSD	106,777	140.18	106,076	140.18	111,548	140.187	
RCA	Vedic	17,830	279.39	17,408	279.39	18,325	280.485	

 Table 7: Comparison of all proposed filters based on maximum frequency



Figure 8: Ripple carry adder with booth multiplier for Hamming window

this table, it is observed that RCA with *Vedic* multiplier and CSA with *Vedic* multiplier, consumes less slices for hanning window. 111,548 slices are the maximum slice consumption found by Hamming window.¹⁹

17,408 is the minimum slice consumption, which is achieved by the combinations of RCA and *Vedic* multiplier. A 60% increase in slice utilization is observed from Hanning to a rectangular window. The designed filters have a wide transition width. This can be reduced by increasing the filter order. 137.885 MHz is the minimum, maximum frequency for the combination of Kogge-Stone adder and modified booth. 289.607 MHz is the maximum frequency that is less delay. These results can be considered while designing a receiver, etc.²⁰

CONCLUSION

This work demonstrates important decisions to make while opting for the filter computational elements for an area-efficient filter or minimal delay filter. From the above results, which are shown in Table 7, it is evident that *Vedic* multiplier with CSA and RCA offers minimum slice utilization for the Hanning window. Among all the three windows, *viz.*, rectangular window, Hanning window, and Hamming window. Hanning window utilizes fewer slices for almost all the combinations, and the rectangular window utilizes high. From Table 7 and Figure 8, *Vedic* multiplier with RCA has a minimum delay, whereas Kogge-Stone adder with modified booth multiplier or CSD multiplier has a maximum delay. For area constraints, filter with RCA and CSA combinations can be utilized similarly for delay filter designed using RCA, and *Vedic* multiplier can be used.

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