

Radiation patterns of circular apertures with prescribed sidelobe levels

Citation for published version (APA):

Worm, S. C. J. (1979). Radiation pátterns of circular apertures with prescribed sidelobe levels. (EUT report. E. Fac. of Electrical Engineering; Vol. 79-E-097). Technische Hogeschool Eindhoven.

Document status and date: Published: 01/01/1979

Document Version:

Publisher's PDF, also known as Version of Record (includes final page, issue and volume numbers)

Please check the document version of this publication:

• A submitted manuscript is the version of the article upon submission and before peer-review. There can be important differences between the submitted version and the official published version of record. People interested in the research are advised to contact the author for the final version of the publication, or visit the DOI to the publisher's website.

• The final author version and the galley proof are versions of the publication after peer review.

• The final published version features the final layout of the paper including the volume, issue and page numbers.

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by

S, C, J. Worm

August 1979

RADIATION PATTERNS OF CIRCULAR APERTURES WITH PRESCRIBED SIDELOBE LEVELS by S.C.J. Worm

> TH-Report 79-E-97 ISBN 90-6144-097-1

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Abstract

Some possibilities of controlling the sidelobes of circular apertures are investigated in this report. Special attention is paid to modifications of a method published by Ishimaru and Held. Source functions with different behaviour at the aperture edge are used to synthesize radiation patterns. The differences concern the value of the edge field and the value of the first edge-field derivative. Aperture distributions comprise series of Bessel functions, namely $\sum_{n=1}^{\infty} a_n J_0(u_n r)$ and $\sum_{n=1}^{\infty} a_n \{J_0(u_n r) - J_0(u_n)\}$ with u_n solution of $J_1(u) = 0$, and $\sum_{n=1}^{\infty} a_n J_0(u_n r)$ with u_n solution of $J_{(u)} = 0$. The synthesis methods which are described, compared and modified allow a number of equal sidelobes, or a number of different sidelobe extrema over a certain region of space beyond the main beam to be prescribed. The computed patterns are compared with copolar specifications for satellite transmitting antennas and earth-station antennas. The patterns are computed to show the influence of the types of series expansions for the aperture distributions, and to show the possibilities of the synthesis methods.

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RADIATION PATTERNS OF CIRCULAR APERTURES WITH PRESCRIBED SIDELOBE LEVELS. TH-Report 79-E-97. Eindhoven University of Technology, Department of Electrical Engineering, Eindhoven, The Netherlands. August 1979.

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Acknowledgement

The author wishes to express his thanks to dr. V. Vokurka and dr. M. Jeuken for suggesting the problem.

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Introduction

Controlling the sidelobes of radiating apertures is acquiring ever greater importance in view of the growth of satellite communications. The reduction of sidelobe levels or sidelobe power reduces the interference between systems and allows more efficient use of the RF spectrum and the geostationary orbit.

The radiation patterns which can be realized theoretically depend on the type of source functions used in the expansion of the aperture illumination and the relative strength of the source functions.

There are synthesis methods which maximize the fraction of power radiated in a prescribed solid angle. As a result the fraction of power contained in the region outside this solid angle is minimized. Examples of these methods are given by Borgiotti [1] who uses hyperspheroidal functions and by Mironenko [8] who uses Legendre polynomials. Both types of source functions yield aperture fields with a finite value at the aperture edge.

Methods to synthesize a radiation pattern with a prescribed sidelobe level are given for instance by Taylor [10], Ishimaru and Held [5], and Kritskiy [7]. They are concerned merely with patterns having a number of equal sidelobes, so as to achieve minimum beam width for a given sidelobe level. The aperture illuminations consist of a series of Bessel functions of the first kind and the zeroth order. The aperture fields have a finite value at the aperture edge, except for one type of illumination used by Kritskiy, which has a zero value there.

A method developed by Kouznetsov [6] who uses power series in r, the normalized radial variable in the aperture, and reduces the level of the first sidelobe to a prescribed value is worth mentioning. This can be achieved with aperture fields having a finite or a zero edge value. For the same level of the first sidelobe the finite value edge fields result in a higher aperture efficiency and a lower decay rate of the remote sidelobes.

-I.1-

In this report, patterns are computed by a number of methods and a comparison is made with certain antenna gain specifications. In section 1 the Taylor distribution for a circular aperture [10] is investigated in connection with satellite antenna gain specifications [3]. In section 2 a synthesis method described by Ishimaru and Held [5] is briefly treated. Modified versions of this method are presented in section 3. The aperture fields consist of series of Bessel functions of the first kind and zeroth order. By judiciously choosing these functions one can generate aperture illuminations with a zero value and a nonzero first derivative of the edge field, or aperture illuminations with at the edge both the value and the first derivative of the field equal to zero. With the modified methods it is possible to prescribe individual sidelobe levels, allowing several different types of envelopes to be realized. The computed radiation patterns are compared with existing [2] and new specifications for the gain of earth-station antennas. The computations assume a nonblocked circular aperture throughout.

1. The Taylor distribution for a circular aperture

The formulas derived in [4], [9], and [10] are used to compare the radiation patterns of the Taylor distribution with the satellite transmitting antenna reference pattern which, for the copolar component, reads [3]

$$-12\left(\frac{\phi}{\phi_{0}}\right)^{2} \text{ for } 0 \leqslant \phi \leqslant 1.58 \phi_{0}$$

$$-30 \text{ for } 1.58 \phi_{0} < \phi \leqslant 3.16 \phi_{0} \qquad (1.1)$$

$$-[17+25^{10}\log(\frac{\phi}{\phi_{0}})] \text{ for } 3.16 \phi_{0} < \phi$$

after intersection with the minus on-axis gain, as the minus on-axis gain.

The angle measured from the main beam axis is ϕ and ϕ_0 is the 3 dB beam width. The reference pattern is shown in fig. 1.



Fig. 1. Copolar component of satellite transmitting antenna reference pattern.

For easy reference the formulas for the aperture distributions, the far field [4], [10] and the efficiency [9] are summarized below. The aperture distribution g(p) is

$$g(p) = \frac{2}{\pi^2} \sum_{m=0}^{\bar{n}-1} \frac{F(u_m, \bar{A}, \bar{n})}{J_0^2(\pi u_m)} J_0(u_m p)$$
(1.2)

where $F(u_0, A, \overline{n}) = 1$ if m = 0,

$$F(u_{m}, A, \bar{n}) = -J_{o}(\pi u_{m}) \xrightarrow[n=1]{n=1} \left\{ \begin{array}{c} 1 - \frac{u_{m}^{2}}{\sigma^{2} \left[A^{2} + \left(n - \frac{1}{2}\right)^{2}\right]} \\ \frac{n=1}{n-1} \left\{ 1 - \frac{u_{m}^{2}}{u_{n}^{2}} \right\} \\ n=1 \\ n \neq m \end{array} \right\} \text{ if } m > 0,$$

 $p = \pi \rho/a$ is a normalized radial variable if a is the aperture radius, A is related to the design sidelobe level η by $\eta = \cosh(\pi A)$, σ is the beam-broadening factor which equals $u_{\overline{n}}/\sqrt{A^2+(\overline{n}-\frac{1}{2})^2}$,

 \bar{n} is a parameter for $\bar{n} - l$ equal sidelobes, πu_{1} is a zero of $J_{1}(\pi u) = 0$ so that the aperture field has a finite value at the aperture edge.

The point $\pi u_{\overline{n}}$ separates the region of equal sidelobes from that of decaying sidelobes. As \overline{n} is increased more sidelobes are forced towards the level n and the beam width is decreased.

The radiation pattern $F(z,A,\bar{n})$ is

$$F(z,A,\bar{n}) = \frac{2J_{1}(\pi z)}{\pi z} \prod_{n=1}^{\bar{n}-1} \frac{\left\{1 - \frac{z^{2}}{\sigma^{2} \left[A^{2} + (n-\frac{1}{2})^{2}\right]}\right\}}{\left\{1 - \frac{z^{2}}{u_{n}^{2}}\right\}}$$
(1.3)

where z is related to the wavelength λ , the aperture radius a, and the angle θ by $z = \frac{2a}{\lambda} \sin \theta$.

With aperture distributions the efficiency $\boldsymbol{\eta}_{T}$ is

$$n_{\rm T} = \left[\mathbf{i} + \sum_{n=1}^{\bar{n}-1} \left\{ \frac{F(u_n, A, \bar{n})}{J_o(\pi u_n)} \right\}^2 \right]^{-1}.$$
 (1.4)

The efficiency is defined by the ratio of the maximum gain from the Taylor pattern to that from the uniformly illuminated aperture.

1.1. Some observations

a. The normalized radiation pattern of the ideal Taylor distribution for a circular aperture is [9]

$$F(z,A) = \frac{\cosh\{\pi(A^2-z^2)^{\frac{1}{2}}\}}{\eta} \quad \text{for the main beam,} \quad (1.5)$$

and

$$F(z,A) = \frac{\cos\{\pi(z^2 - A^2)^{\frac{1}{2}}\}}{\eta} \quad \text{for the sidelobes.} \quad (1.6)$$

This optimum pattern is shown in fig. 2, it is optimum in its beam width sidelobe relationship.



Fig. 2. An optimum pattern.

An approximation to the ideal pattern is realized by setting the first \bar{n} -1 sidelobes equal to each other while the remaining sidelobes decay as $(\sin\theta)^{-3/2}$. For higher values of \bar{n} the approximation is closer and the beam-broadening factor tends to unity. b. The zeros of the approximating pattern are $\sigma \sqrt{A^2 + (n-\frac{1}{2})^2}$ with $n = 1, 2, \ldots, \bar{n}-1$, \bar{n} if $z \leq u_{\bar{n}}$. These are the zeros of $\cos\{\pi\sqrt{(\frac{z}{\sigma})^2 - A^2}\}$. If $z \geq u_{\bar{n}}$ the zeros of the pattern are equal to the zeros of $J_1(\pi z)$. The beam-broadening factor σ coordinates the \bar{n} th zero of $\cos\{\pi\sqrt{(\frac{z}{\sigma})^2 - A^2}\}$ and the \bar{n} th zero of $\frac{J_1(\pi z)}{\pi z}$, so that it can be computed from

$$\sigma = \frac{u_{\bar{n}}}{\sqrt{A^2 + (\bar{n} - \frac{1}{2})^2}}$$
 (1.7)

c. Taylor [11] uses a series of J_o -functions with a finite value at the aperture edge and adjusts the first zeros of the radiation pattern to the zeros of $\cos\{\pi\sqrt{(\frac{z}{\sigma})^2 - A^2}\}$. Perhaps this kind of zeros can be employed if a series of J_o -functions is used with zero-edge values. In that case the radiation pattern is

$$F(z,A,\bar{n}) = \frac{J_{0}(\pi z)}{1 - \frac{z^{2}}{u_{0}^{2}}} \prod_{n=1}^{\bar{n}-1} \left(\frac{1 - \frac{z^{2}}{\sigma^{2} [A^{2} + (n - \frac{1}{2})^{2}]}}{1 - \frac{z^{2}}{u_{n}^{2}}} \right)$$
(1.8)

The aperture field is

$$g(p) = \sum_{m=0}^{\overline{n}-1} a_{m}^{J} o(u_{m}^{p}) , \qquad (1.9)$$

where πu_{m} is a zero of $J_{o}(\pi u) = 0$ and a_{m} is an excitation coefficient which equals

$$2 \frac{F(z,A,\bar{n})}{J_{1}^{2}(\pi u_{m})} z = u_{m} .$$
 (1.10)

Similar reasoning applies if the aperture field consists of the series \overline{n}_{-1}

$$g(p) = \sum_{m=1}^{n-1} a_{m} \{J_{o}(u_{m}p) - J_{o}(\pi u_{m})\}$$
(1.11)

where πu_m is a zero of $J_1(\pi u) = 0$. In this case the zero $u_0 = 0$ can be dropped. The radiation pattern is 2^2

$$F(z,A,\bar{n}) = \frac{2J_{1}(\pi z)}{\pi z (1 - \frac{z^{2}}{u_{1}^{2}})} \prod_{n=2}^{\bar{n}-1} \left\{ \begin{array}{c} 1 - \frac{z}{\sigma^{2} [A + (n - \frac{1}{2})^{2}]} \\ \sigma^{2} [A + (n - \frac{1}{2})^{2}] \\ 1 - \frac{z^{2}}{u_{n}^{2}} \\ u_{n} \end{array} \right\}$$
(1.12)

and the excitation coefficient a equals

$$2 \frac{F(z,A,\overline{n})}{J_{o}^{2}(\pi u_{m})} z=u_{m}$$
(1.13)

The equations (1.8) and (1.12) have not been investigated further, because they don't allow individual sidelobes to be prescribed.

1.2. Results and conclusions

Equations (1.2), (1.3) and (1.4) have been programmed with $\bar{n} = 3,4,5$ and 6, a design sidelobe level of -30 dB, and a design 3 dB beam width of 3 degrees.

In figures 6 - 9 the radiation patterns are shown together with the specifications (1.1). The computed sidelobe extrema, 3 dB beam width, aperture size and efficiency are also indicated. In figures 10 and 11 the accompanying aperture fields are shown. The first \overline{n} -1 computed sidelobes are not exactly equal to each other. They decrease slightly with increasing distance from the main beam axis. The difference between the level of the first sidelobe and the design level is at least 0.4 dB.

One can notice that the decay rate of the far-out sidelobes is low. The efficiencies are high and range from 0.8377 to 0.8735. With a prescribed sidelobe level of -30 dB and with \bar{n} = 3,4,5 the computed copolar patterns fulfil the requirements imposed by eq. (1.1). If \bar{n} = 6 the requirements are not fulfilled.

2. The method of Ishimaru and Held for the synthesis of radiation patterns from circular apertures

In order to be able to show the analogy between the method of Ishimaru and Held [5] and the further-developed methods and because we believe that the Ishimaru and Held paper has not been widely available we will deal with the relevant part of it briefly now.

Ishimaru and Held [5, part I] describe a method to synthesize a radiation pattern having a number of equal sidelobes. They want to realize a pattern approximating the optimum beam width sidelobe-level relationship. The optimum pattern has a minimum beam width for a given sidelobe level employing a given number of terms in the series expansion of the source field. After determining the level and the number N of the equal sidelobes, the first N zeros of the radiation pattern are computed.

The radiation pattern g(u) from a circular aperture with radius a is given by

$$g(u) = \int_{0}^{1} f(r)J_{0}(ur)rdr , \qquad (2.1)$$

where

r is the normalized radial variable,

f(r) is the aperture distribution which is r-dependent only, u equals $\frac{2\pi}{\lambda} a \sin\theta$ with θ the angle from broadside.

With N+1 terms in the aperture-field series expansion one can control N sidelobe levels. Assume that the field can be expanded in a series of Bessel functions,

$$f(\mathbf{r}) = \sum_{n=0}^{N} a_n J_0(\mathbf{u}_n \mathbf{r}) \quad \text{if } 0 \leq \mathbf{r} \leq 1$$

$$= 0 \qquad \qquad \text{if } \mathbf{r} > 1 . \qquad (2.2)$$

The possible values of u are derived from the homogeneous boundary condition at r = 1

$$u_n J'_o(u_n) + h J_o(u_n) = 0$$
 (2.3)

where h is a constant.

An acceptable approximation of the optimum pattern can be obtained if the following requirements are fulfilled.

a. The zeros of the radiation pattern are real. Nonreal zeros tend to produce higher sidelobe levels and broader beam width.

b. The constant h is zero. Then the first zero of the radiation pattern is as near to the origin as possible.

c. The number N is finite. The lower limit follows from the desired level and the number of sidelobes higher than that level. The upper limit is mainly bounded by the supergain consideration.

With h=0 eq. (2.3) reduces to

$$J'_{0}(u_{n}) = -J_{1}(u_{n}) = 0$$
(2.4)

so $u_0 = 0$, $u_1 = 3.8317$, $u_2 = 7.0156$ etc. up to and including u_N . At r = 1 the value of the field is nonzero and that of the first derivative of the field is zero. A few source functions are shown in fig. 3.



Substituting eq. (2.2) in the integral for the far field and employing eq. (2.4) yields

$$g(u) = \sum_{n=0}^{N} \left\{ -a_n u \frac{J_o(u_n)}{u_n^2 - u^2} J_1(u) \right\} \quad \text{if } u \neq u_n \quad (2.5)$$

$$= \frac{1}{2}a_{n}J_{0}^{2}(u_{n}) \qquad \text{if } u = u_{n} . \qquad (2.6)$$

After rewriting and normalizing eq. (2.5) the following form is obtained for the far field (see appendix A),

$$g(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{2}} \right\}$$
(2.7)

The first N zeros of the radiation pattern $(u < u_{N+1})$ are $u_n + t_n$ with n = 1, 2, ..., N. If $u \ge u_{N+1}$ the zeros of g(u) are the zeros of $2J_1(u)$

In the region $u_1 < u < u_{N+1}$ there are N sidelobes. Assume that the extrema occur in

$$u = M_{n}$$
 with $n = 1, 2, ..., N.$ (2.8)

The conditions for the optimum pattern are

$$|g(M_1)| = |g(M_2)| = \dots = |g(M_N)| = b$$
 (2.9)

where b is the desired sidelobe level and g is given by eq. (2.7).

A relationship between the unknowns t_n , n = 1, 2, ..., N and the desired sidelobe level is determined as follows.

If the variables t_n are small compared with u_n an expression for $g(M_n)$ can be deduced from eq. (2.7) which, together with eq. (2.9), yields N linear equations for N unknowns. Rewrite eq. (2.7) as

$$g(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \begin{array}{c} 1 + \frac{2u_{n}}{u_{n}^{2} - u^{2}} t_{n} + \frac{1}{u_{n}^{2} - u^{2}} t_{n}^{2} \\ \frac{u_{n}^{2} - u^{2}}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u^{2}} t_{n}^{2} \\ \frac{1 + \frac{2}{u_{n}} t_{n} + \frac{1}{u_{n}^{2}} t_{n}^{2} \\ \frac{1 + \frac{2}{u_{n}} t_{n} + \frac{1}{u_{n}^{2}} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2}} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n} + \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\ \frac{1}{u_{n}^{2} - u} t_{n}^{2} \\$$

Assume that the terms with t_n^2 can be neglected and that the points M_n are located at about the middle of u_n and u_{n+1} . The remaining products in nominator and denominator can be approximated by series, so that N 2u

$$g(M_{i}) \approx \frac{2J_{1}(M_{i})}{M_{i}} \cdot \frac{1 + \sum_{n=1}^{N} \frac{1 - n}{u_{n}^{2} - M_{i}^{2}} t_{n}}{1 + \sum_{n=1}^{N} \frac{2}{u_{n}} t_{n}}$$
(2.11)

with i = 1, 2, ..., N.

Assume that the points M are independent of t,

$$M_{n} = \frac{u_{n}^{+}u_{n+1}}{2} . \qquad (2.12)$$

Then eqs. (2.11) and (2.9) yield N linear equations for N unknowns t_n . If it is found that the variables t_n are not small enough to make eq. (2.11) valid and the optimum pattern is not approximated sufficiently closely, another set of variables t'_n can be computed. Assume that the first N zeros of the pattern closer to the optimum are $u_n + t_n + t'_n$. The pattern representation is

$$g_{2}(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}^{*})^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}$$

$$= \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}^{*})^{2}}}{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}} \right\} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}} \right\}$$

$$= g_{1}(u) \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}^{*})^{2}}}{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}^{*})^{2}}} \right\}, \quad (2.13)$$

if the first approximation to the optimum pattern is

$$g_{1}(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{2}} \right\}.$$
 (2.14)

The eqs. (2.13) and (2.7) are similar. The procedure for computing t_n can also be used for t'_n . In eqs. (2.7) - (2.12) replace $\frac{2J_1(u)}{u}$ by $g_1(u)$, $u_n(+t_n)$ by $u_n+t_n(+t'_n)$, t_n by t'_n , $M_n = \frac{u_n+u_n+1}{2}$ by $M'_n = \frac{u_n+t_n+u_n+1+t_n+1}{2}$ and $\frac{2J_1(M_1)}{M_1}$ by $g_1(M'_1)$.

The adjusted equation for $g(M_i)$ and the optimum condition (N equal sidelobes) yield N linear equations for N unknowns t'_n . As an example, a pattern is synthesized with three -25 dB sidelobes. The zeros of

J₁(u) are assumed to be u₀ = 0, u₁ = 3.83, u₂ = 7.016 and u₃ = 10.173. The sidelobe extrema of $\frac{2J_1(u)}{u}$ are taken to be in M₁ = 5.33, M₂ = 8.53 and M₃ = 11.70. The -25 dB level means b = 17.78⁻¹. Solving for t_n, the first approximation is

$$g_{1}(u) = \frac{2J_{1}(u)}{u} \frac{\left[1 - \left(\frac{u}{4.1625}\right)^{2}\right]\left[1 - \left(\frac{u}{6.652}\right)^{2}\right]\left[1 - \left(\frac{u}{9.620}\right)^{2}\right]}{\left[1 - \left(\frac{u}{3.83}\right)^{2}\right]\left[1 - \left(\frac{u}{7.016}\right)^{2}\right]\left[1 - \left(\frac{u}{10.173}\right)^{2}\right]} .$$
 (2.15)

The first sidelobe of $g_1(u)$ is still higher than -25 dB so that it is necessary to compute t_n . The improved pattern is

$$g_{2}(u) = \frac{2J_{1}(u)}{u} \frac{\left[1 - \left(\frac{u}{4.2885}\right)^{2}\right]\left[1 - \left(\frac{u}{6.561}\right)^{2}\right]\left[1 - \left(\frac{u}{9.576}\right)^{2}\right]}{\left[1 - \left(\frac{u}{3.83}\right)^{2}\right]\left[1 - \left(\frac{u}{7.016}\right)^{2}\right]\left[1 - \left(\frac{u}{10.173}\right)^{2}\right]} .$$
 (2.16)

The coefficients a are computed with

$$a_n = 2 \frac{g_2(u_n)}{J_o^2(u_n)}$$
 $n = 1, 2, ..., N.$ (2.17)

The aperture field which generates $g_2(u)$ is

$$f(r) = 2 + 0.935J_o(u_1r) + 0.702J_o(u_2r) - 1.10J_o(u_3r).$$
 (2.18)

The coefficients we have computed differ from those given by Ishimaru and Held [5].

If the method of Taylor is used in synthesizing a radiation pattern with three -25 dB sidelobes one gets

$$g_{T}(u) = \frac{2J_{1}(u)}{u} \frac{\left[1 - \left(\frac{u}{4.34}\right)^{2}\right] \left[1 - \left(\frac{u}{6.57}\right)^{2}\right] \left[1 - \left(\frac{u}{9.58}\right)^{2}\right]}{\left[1 - \left(\frac{u}{3.83}\right)^{2}\right] \left[1 - \left(\frac{u}{7.016}\right)^{2}\right] \left[1 - \left(\frac{u}{10.173}\right)^{2}\right]}$$
(2.19)

The patterns $g_1(u)$, $g_2(u)$ and $g_T(u)$ are shown in figs. 12 and 13. The first zero of $g_2(u)$ is nearer to the origin than the first zero of $g_T(u)$ so that $g_2(u)$ has the minimum beam width. For $g_2(u)$ and $g_T(u)$ the deviations from the required levels are of the same magnitude. The maximum gain G is given by

$$G = \frac{8\pi^{2}a^{2}}{\lambda^{2}} \frac{\int_{0}^{1} f(r)rdr|^{2}}{\int_{0}^{1} |f(r)|^{2}rdr} .$$
 (2.20)

$$G = \frac{8\pi^{2}a^{2}}{\lambda^{2}} \frac{g^{2}(0)}{\sum_{n=0}^{N} \frac{1}{2}a_{n}J_{0}^{2}(u_{n})}.$$
(2.21)
Using 2a = D and $a_{n} = \frac{2g(u_{n})}{J_{0}^{2}(u_{n})}$ results in

$$G = (\frac{\pi D}{\lambda})^{2} \frac{g^{2}(0)}{\sum_{n=0}^{N} \frac{g^{2}(u_{n})}{J_{0}^{2}(u_{n})}}.$$
(2.22)

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3. The modified methods with other source functions

3.1. Source functions with a zero edge field and a nonzero first derivative of the edge field

In this section we will employ the source functions which arise on the assumption $h = \infty$ in eq. (2.3). The aperture field f(r) consists of a series of Bessel functions

$$f(r) = \sum_{n=0}^{N} a_{n} J_{0}(u_{n}r) \qquad 0 \leqslant r \leqslant 1 ,$$

$$= 0 \qquad r > 1 . \qquad (3.1)$$

If $h = \infty$ the homogeneous boundary condition, eq. (2.3), reduces to

$$J_{o}(u_{n}) = 0$$
 (3.2)

The possible values of u_n are now $u_0 = 2.4048$, $u_1 = 5.5201$ etc. up to and including u_N . A few source functions are shown in fig. 4.



Fig. 4. Source functions J₀(u,r).

N sidelobe levels can be controlled with N+1 source functions. The far field is

$$g(u) = \int_{0}^{1} f(r) J_{0}(ur) r dr \qquad (2.1)$$

and with eq. (3.1) and eq. (3.2) g(u) is computed as

$$g(u) = \sum_{n=0}^{N} \{a_n \frac{u_n}{u_n^2 - u^2} J_1(u_n) J_0(u)\} \text{ if } u \neq u_n$$
(3.3)

$$g(u) = \frac{1}{2}a_n J_1^2(u_n)$$
 if $u = u_n$. (3.4)

Rewriting and normalizing eq. (3.3) in the same way as eq. (2.5) gives N $\begin{pmatrix} 2 \\ 2 \end{pmatrix}$

$$g(u) = J_{0}(u) \frac{\prod_{n=1}^{n-1} \left\{ 1 - \frac{u}{(u_{n} + t_{n})^{2}} \right\}}{\prod_{n=0}^{n-1} \left\{ 1 - \frac{u^{2}}{u_{n}^{2}} \right\}}.$$
 (3.5)

If $u \ge u_{N+1}$ the zeros of g(u) are the zeros of $J_o(u) = 0$. If $u < u_{N+1}$ the zeros of g(u) are $u_n + t_n$, n = 1, 2, ..., N. To show the analogy with the far field where h = 0, eq. (3.5) is written as

$$g(u) = \frac{J_{o}(u)}{1 - \frac{u^{2}}{u_{o}^{2}}} \prod_{n=1}^{\Pi} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}$$
(3.6)

The factor $\frac{J_0(u)}{\frac{2}{u}}$ is proportional to the field of $J_0(u_0r)$ shown in $I - \frac{u}{\frac{2}{u}}$ fig. 14. The first sidelobe is -27.5 dB and the o second is -36.4 dB down. The zeros, the efficiency

and the 3 dB beam width are also indicated. The changes in the aperture field can be small in order to lower the first sidelobe to -30 dB, for instance.

Where h = 0 the expression for the far field is

$$g(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}$$
(2.7)

The factor $\frac{2J_1(u)}{u}$ is proportional to the field of the uniformly illuminated aperture. The first sidelobes are -17.6 dB and -23.8 dB down. Write eq. (3.6) as

$$g(u) = \frac{J_{o}(u)}{1 - \frac{u^{2}}{u^{2}_{o}}} \prod_{n=1}^{n} \left\{ \frac{1 + \frac{2u_{n}}{2u_{-u}} t_{n} + \frac{1}{u^{2}_{-u}} t_{n}^{2}}{1 + \frac{2}{u_{n}} t_{n} + \frac{1}{u^{2}_{-u}} t_{n}^{2}}{1 + \frac{2}{u_{n}} t_{n} + \frac{1}{u^{2}_{n}} t_{n}^{2}} \right\}.$$
(3.7)

-3.2.~

In order to determine a relationship between the variables t_n and the prescribed sidelobe extrema we proceed in the same way as in section 2. Assume that the extrema occur at the points M_i at about the middle of u_i and u_{i+1} . Assume t_n is small compared to u_n so that terms with t_n^2 can be dropped. The products in nominator and denominator can be replaced by series and

$$g(M_{i}) \approx \frac{J_{o}(M_{i})}{1 - \frac{M_{i}^{2}}{u_{o}^{2}}} \frac{1 + \sum_{n=1}^{N} \frac{2u_{n}}{u_{n}^{2} - M_{i}^{2}} t_{n}}{1 - \frac{M_{i}^{2}}{u_{o}^{2}}} (3.8)$$

with i = 1, 2, ..., N.

We introduce the following conditions for the sidelobe levels

$$|g(M_{1})| = b_{1}$$

.
.
 $|g(M_{N})| = b_{N}$
(3.9)

N equal sidelobes is a special case of eq. (3.9). The points M are taken to be independent of t

$$M_n = \frac{u_n + u_{n+1}}{2} . (3.10)$$

Then the equations (3.8) and (3.9) provide N linear equations for N unknowns t_n . The first approximation $g_1(u)$ to the desired pattern is

$$g_{1}(u) = \frac{J_{0}(u)}{1 - \frac{u^{2}}{u_{0}^{2}}} \prod_{n=1}^{\Pi} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}.$$
 (3.11)

If this approximation deviates too much one can introduce the variables t'_n to describe the better pattern $g_2(u)$

$$g_{2}(u) = \frac{J_{0}(u) N}{1 - \frac{u^{2}}{u^{2}} \prod_{n=1}^{n=1} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t'_{n})^{2}}}{1 - \frac{u^{2}}{u^{2}} \right\}}$$

-3.4.-

$$= g_{1}(u) \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}')^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}.$$
 (3.12)

The computation of t' follows the same lines as the computation of t_n due to the similarity of eq. (3.12) and eq. (3.6).

In eqs. (3.6) - (3.11) replace $J_0(u)/(1-\frac{u^2}{u_0^2})$ by $g_1(u)$, $u_n(+t_n)$ by $u_n+t_n(+t_n')$

$$t_n$$
 by t'_n , $M_n = \frac{u_n + u_{n+1}}{2}$ by $M'_n = \frac{u_n + t_n + u_{n+1} + t_{n+1}}{2}$ and $J_o(M_i) / (1 - \frac{M_i^2}{u_o^2})$ by $g_1(M'_i)$

Again N linear equations are obtained for N variables and the t'_n are easily solved.

Instead of eq. (3.10) one can also use

$$M_{n} = \frac{\frac{u + d u}{n + 1}}{2}$$
(3.13)

where $d_n \lesssim 1$. Then the points M_n can be chosen in better agreement with the real positions of the sidelobe extrema. The weight factors d_n can be estimated from the pattern of $J_0(u_0r)$. The excitation coefficients a_n are computed with

$$a_n = 2 \frac{g(u_n)}{J_1^2(u_n)}$$
 $n = 0, 1, 2, ..., N.$ (3.14)

The maximum gain G is derived from

$$G = \frac{8\pi^{2}a^{2}}{\lambda^{2}} \frac{\int_{0}^{1} f(r)rdr|^{2}}{\int_{0}^{1} |f(r)|^{2}rdr}$$
(2.20)

Substituting the source functions we are concerned with, yields

$$G = \frac{8\pi^2 a^2}{\lambda^2} \frac{g^2(0)}{\sum_{n=0}^{N} \frac{1}{2}a_n^2 J_1^2(u_n)}$$
(3.15)

and with 2a = D and eq. (3.14),

$$G = \left(\frac{\pi D}{\lambda}\right)^2 \frac{g^2(0)}{\sum_{n=0}^{N} \frac{g^2(u_n)}{J_1^2(u_n)}} .$$
(3.16)

The difference $\Delta g(dB)$ between the normalized power patterns of $J_o(u_o r)$ and $\sum_{n=0}^{N} a_n J_o(u_n r)$ for large values of u is estimated as

$$\Delta g(dB) \approx 40 \sum_{n=1}^{N} \frac{10}{\log(1 + \frac{t_n}{u_n})}$$
 (3.17)

With suitable values of the variables b_i in eq. (3.9) the pattern envelopes can be made to fulfil, for instance, the specifications for earth-station antennas [2]. The following requirements for G_i , the gain relative to isotropic, are represented in fig. 15:

$$I \quad G_{i} = 32 - 25 * \log(\theta^{0}) \quad dB \qquad (3.18a)$$

II
$$G_{i} = 32 - 30 * \log(\theta^{0})$$
 dB (3.18b)

III
$$G_1 = 32 - 35 * \log(\theta^{\circ})$$
 dB (3.18c)

IV
$$G_1 = 25 - 25 * \log(\theta^{\circ}) dB$$
 (3.18d)

These formulas are assumed to apply to the region beyond the first sidelobe peak, that is at and beyond $\theta(\text{degr}) \approx 100 \frac{\lambda}{D}$, until -10 dB relative to isotropic is reached.

3.2. Results and conclusions

The method described in section 3.1. is used to compute a number of radiation patterns. In figs. 16-20 we represent computed patterns and aperture distributions if only the first sidelobe is prescribed at a certain level. Furthermore, some zeros, the efficiency, the gain reduction relative to the uniformly illuminated aperture, the 3 dB beam width and the excitation coefficients are indicated. If we compare these 5 patterns with each other and with the 'reference' pattern of fig. 14 (no sidelobes prescribed) we note that:

- The deviation between the prescribed and computed level increases if the difference between the required level and the level of the first sidelobe of the 'reference' pattern is increased. The deviations in the first approximation patterns are less than 0.9 dB.
- Increasing the first sidelobe, relative to the 'reference' pattern,

-3.5.-

results in a shift of the first zero $u_1 + t_1$ towards the origin, an increase of the efficiency and a decrease of the 3 dB beam width. The shift of the first zero results in opposite signs for the excitation coefficients a_0 and a_1 .

- Lowering the first sidelobe, relative to the 'reference' pattern, results in a shift of the first zero $u_1 + t_1$ away from the origin, a decrease of the efficiency and an increase of the 3 dB beam width. The shift of the first zero now results in a_0 and a_1 having the same sign.
- For greater values of the difference between the prescribed level and the level of the first sidelobe of the 'reference' pattern, the shift of the first zero as well as the ratio of the excitation coefficients $\left|\frac{a_1}{a_0}\right|$ are greater.
- The difference between the first and the second sidelobe level of a pattern decreases if the level of the first sidelobe is decreased.
- The estimated differences Δg , eq. (3.17), range from -4.05 dB to 2.11 dB. These values are computed for the -17.0 dB and the -35.1 dB sidelobe patterns. The differences between the computed and the 'reference' pattern range from -4.4 dB to 2.2 dB for the fifth sidelobe. Hence the value of Δg is almost reached at u \approx 20.
- The decay of the sidelobe extrema is greater than the decay of G_i, eq. (3.18a-d). These specifications are therefore fulfilled if the peak of the first sidelobe is below the curves I IV of fig. 15 when the gain reduction is taken into account.
- If the first sidelobe is prescribed to have a higher (lower) value than the first lobe of the 'reference' pattern, the normalized aperture field for 0 < r < 1 is higher (lower) than the aperture field generating the 'reference' pattern.

In figs. 21-25 we represent computed patterns and aperture distributions if two or more sidelobes are prescribed to have equal levels. In most cases it is necessary to compute a second approximation because the levels in the first approximation deviate more than 1 dB (otherwise an arbitrarily chosen value) from the required levels. The results of each approximation are indicated: the computed levels, the zeros, the excitation coefficients, the efficiency, the gain reduction relative to a uniformly illuminated aperture and the 3 dB beam width. If we compare the pattern with three -25 dB sidelobes of fig. 21 with those of figs. 12 and 13 we observe that the sidelobe levels not prescribed are lower and decay faster, and that the 3 dB beam width has increased (the first zero of the pattern is further away from the origin). These effects are due to the fact that we are now dealing with zero-edge aperture fields. If we compare the patterns of figs. 22-25, having two or more sidelobes prescribed to have the -30 dB level, with the Taylor patterns of figs. 6-9 we note that the required levels are now reached more accurately, that the 3 dB beam widths are greater and that the sidelobes not prescribed are lower and decay faster. The efficiencies are lower than in the case of the Taylor patterns. If we look at the accompanying aperture distributions we see that the Taylor distributions (figs. 10 and 11) are smoother, which is imputed to the different source functions and to the fact that the sidelobes of the Taylor patterns are actually decaying instead of being equal. If we compare the patterns in figs. 22-25 with each other we see that with an increasing number of equal sidelobes the zeros have shifted more and more towards the origin, the 3 dB beam width is decreased and efficiency is increased. With N-1 prescribed sidelobes the width of the lobes 1 up to and including N-1 increases. For instance in fig. 25 these widths are successively 2.22, 2.80, 3.34, 3.57 and 4.26. The width of sidelobes not prescribed is always equal to the distance between the zeros of J (u), which is approximately 3.14.

The requirements set by eq. (3.18a-d) are not all fulfilled. The patterns of figs. 22-24 meet a-d, a-c and a-b, respectively, while the pattern of fig. 25 meets none of these requirements.

In figs. 26-33 we show computed patterns, aperture distributions and other relevant data when, starting at different levels for the first sidelobe, the decay rate of the first 3 or 4 extrema is prescribed. Once again, if there is a deviation of more than 1 dB between a required and a computed level a second approximation is made. Decay rates of 4,5,6,7 and 8 dB are required and the starting levels are -30, -33, -36 and -48 dB. However, not all the combinations of rates and levels are investigated. At the cost of the efficiency and with increasing 3 dB beam width the starting level can be decreased and the decay rate can be increased. All the diagrams shown meet the requirements of eq. (3.18a-d).

The computed differences Δg according to eq. (3.17) for the patterns of fig. 26 and fig. 33 are, respectively, 0.2 dB and 7.8 dB. These final values are almost reached at the fifth sidelobe extrema (u \approx 20) because the actual differences (compare with fig. 14) are 0.3 dB and 9.4 dB.

The normalized aperture distributions for the described patterns are smooth functions which, for 0 < r < 1, are lower than the distributions for the pattern with none of the sidelobes prescribed (fig. 14).

Finally, in figs. 34-36 some cases are illustrated in which the first sidelobe is required to be lower than the second. We observe that the required levels are reached within the I dB limit in the second approximation, except the -50 dB level in fig. 36. The 3 dB beam widths are increased and the efficiencies are decreased compared with the pattern of fig. 14. The differences Δg according to eq. (3.17) and the actual differences to the pattern of fig. 14 at u \approx 20 are equal for these three patterns. They are respectively -0.3, 0.2 and 1.6 dB.

The patterns meet the specifications imposed by the equations (3.18a-d).

The normalized aperture distributions for these patterns are smooth functions having lower values for 0 < r < 1 than the distribution $J_0(u_0 r)$ shown in fig. 14.

In conclusion we can say that the method described in section 3.1. is flexible, accurate and capable of handling several types of sidelobe envelopes. The source functions having zero-edge values ensure a more rapid decay of the sidelobes not prescribed than do the distributions with non-zero edge values.

3.3. Source functions with both the field and the first derivative of the field equal to zero at r = 1

The aperture field is taken to be a series of the form

$$f(r) = \sum_{n=1}^{N} a_n \{J_o(u_n r) - J_o(u_n)\} \quad 0 \le r \le 1,$$

$$= 0 \qquad r > 1,$$
(3.19)

where the values of u_n are determined by

$$J_1(u_n r) = 0$$
 at $r = 1$. (3.20)

The normalized source functions with $u_1 = 3.8317$ and $u_2 = 7.0156$ are shown in fig. 5.



Fig. 5. Source functions.

With N source functions, N-1 sidelobe levels can be controlled.

Now the radiation pattern of a circular aperture is

$$g(u) = \sum_{n=1}^{N} \left\{ \frac{-a_n u_n^2}{(u_n^2 - u^2)u} \right\} J_0(u_n) J_1(u) \quad \text{if } u \neq u_n , \quad (3.21)$$
$$= \frac{1}{2} a_n J_0^2(u_n) \quad \text{if } u = u_n . \quad (3.22)$$

.

Eq. (3.21) can be written in a form similar to eq. (2.7) and eq. (3.6), namely

$$g(u) = \frac{2J_{1}(u)}{u(1-\frac{u^{2}}{u_{1}^{2}})} \prod_{n=2}^{N} \left\{ \frac{1-\frac{u^{2}}{(u_{n}+t_{n})^{2}}}{1-\frac{u^{2}}{u_{n}^{2}}} \right\}$$
(3.23)

The factor $2J_1(u)/u(1-\frac{u^2}{u_1^2})$ is proportional to the field of the aperture distribution $J_0(u_1r) - J_0(u_1)$, see fig. 37. The first sidelobe is -35.1 dB down.

Assume that the variables t_n are small compared to u_n and the sidelobe extrema occur in the points M_i at about the middle of u_i and u_{i+1} . Suppose further that the points M_i are independent of the variables t_n . It will be evident from the foregoing treatment in sections 2 and 3.1. that N-1 linear equations for N-1 unknowns t_n are deduced from the approximation

$$g(M_{i}) \approx \frac{2J_{1}(M_{i})}{M_{i}(1 - \frac{M_{i}^{2}}{u_{1}^{2}})} \cdot \frac{1 + \sum_{n=2}^{N} \frac{2u_{n}}{u_{n}^{2} - M_{i}^{2}} t_{n}}{1 + \sum_{N=2}^{N} \frac{2u_{n}}{u_{n}} t_{n}}$$
(3.24)

with $i = 1, 2, \dots, N-1$ and the condition

$$|g(M_1)| = b_1$$

...
 $|g(M_{N-1})| = b_{N-1}$
(3.25)

The variables $b_1 - b_{N-1}$ are the prescribed sidelobe levels. The first approximation g,(u) to the desired pattern is

$$g_{1}(u) = \frac{2J_{1}(u)}{u(1-\frac{u^{2}}{u_{1}^{2}})} \prod_{n=2}^{N} \left\{ \frac{1-\frac{u^{2}}{(u_{n}+t_{n})^{2}}}{1-\frac{u^{2}}{u_{n}^{2}}} \right\} .$$
(3.26)

If the sidelobe levels of $g_1(u)$ differ too much from the desired levels, the variables t'_n are introduced to make the computation of the better approximation $g_2(u)$ possible,

$$g_{2}(u) = g_{1}(u) \prod_{n=2}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n} + t_{n}^{*})^{2}}}{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}} \right\} .$$
(3.27)

Eq. (3.27) and eq. (3.23) are similar so that the scheme used to relate linearly the sidelobe levels and the variables t_n can be employed to do the same for the levels and the variables t'_n .

After the desired pattern has been obtained the excitation coefficients a are computed from

$$a_{n} = \frac{2g(u)}{J_{o}^{2}(u_{n})} u = u_{n} \quad \text{with } n = 1, 2, \dots, N. \quad (3.28)$$

The maximum gain G for the aperture with radius a is

$$G = \frac{8\pi^{2}a^{2}}{\lambda^{2}} \frac{\int_{0}^{1} f(r)rdr|^{2}}{\int_{0}^{1} |f(r)|^{2}rdr} \qquad (2.20)$$

With the applied source functions, 2a = D and eq. (3.28) G becomes

$$G = \left(\frac{\pi D}{\lambda}\right)^{2} \frac{g^{2}(0)}{\left\{\sum_{n=1}^{N} \frac{g(u_{n})}{J_{o}(u_{n})}\right\}^{2} + \sum_{n=1}^{N} \frac{g^{2}(u_{n})}{J_{o}^{2}(u_{n})}} \qquad (3.29)$$

An estimate for the difference $\Delta g(dB)$ between the extrema of the normalized power patterns of $\{J_0(u_1r)-J_0(u_1)\}$ and $\sum_{n=1}^{N} a_n \{J_0(u_nr)-J_0(u_n)\}$ is for great values of u

$$\Delta g(dB) \approx 40 \sum_{n=2}^{N} \frac{10}{\log(1 + \frac{t_n}{u_n})}$$
 (3.30)

3.4. Results and conclusions

The method of section 3.3. is used in the numerical computation of several radiation patterns.

In figs. 38 and 39 we show the patterns having the first sidelobe prescribed at -30 dB and -33 dB, respectively. These levels are higher than the -35.1 dB level of the pattern of $J_0(u_1r) - J_0(u_1)$, fig. 37, which for the moment, is the 'reference' pattern. The first zero of each of the patterns shifts towards the origin, the 3 dB beam width

decreases and the efficiency increases relative to the 'reference' pattern. Because of the shift towards the origin, the excitation coefficients a, and a₂ have opposite signs.

The differences Δg according to eq. (3.30) are for the patterns of fig. 38 and fig. 39, -1.6 dB and -0.6 dB, respectively, while the actual differences for the fifth sidelobes are respectively -1.8 dB and -0.7 dB.

The patterns meet the requirements imposed by the equation (3.18a-d).

The normalized aperture fields for 0 < r < 1 are somewhat higher than the aperture field for the 'reference' pattern.

For equal sidelobes we have given some examples in figs. 40-44. In some of these cases even a second approximation does not allow us to obtain differences between required and computed levels less than 1 dB. The diagrams shown have two or more equally prescribed sidelobes. The efficiencies are higher and the 3 dB beam widths are lower than in the case of the 'reference' pattern. With two -30 dB or -40 dB lobes the required levels are adequately approximated and the aperture fields remain rather smooth. With three or more -30 dB sidelobes the required levels are not reached in two approximations and the aperture fields are not very smooth. Apparently the changes in the levels relative to the 'reference' pattern are too great, and although the method is capable of handling more than two approximations, no more than two have been carried out.

In figs. 45-48 we show some computed patterns, aperture distributions and relevant data, starting at different levels for the first sidelobe. The decay rate of the first three extrema is prescribed. Now the deviations between required and computed levels are less than 1 dB with one or two approximations. The efficiencies and 3 dB beam widths differ little from the 'reference' values (fig. 37) 0.5 and 4.8 due to the chosen sidelobe levels.

The patterns meet the requirements of eq. (3.18a-d).

The differences Δg according to eq. (3.30), computed for instance for the patterns of figs. 47 and 48, are respectively 2.3 dB and 1.2 dB. The actual differences at the fifth sidelobe extrema are 2.4 dB and 1.2 dB.

The normalized aperture fields of figs. 45-48 are smooth and have changed little compared to the aperture distribution of fig. 37.

In fact many more cases could be handled but the results and observations would be similar to those of section 3.2. The differences are found in the magnitude of some characteristics. If patterns with the same prescribed levels are synthesized with the source functions of section 3.1. and section 3.3., then with the last-named we find the following:

- the zeros of the radiation pattern are further away from the origin, so that the 3 dB beam width is greater;
- the efficiency is lower;
- the sidelobes not prescribed are lower and decay faster.

This is clearly demonstrated in the patterns of figs. 45 and 31 for sidelobes prescribed at -36, -41 and -46 dB and in the patterns of figs. 47 and 33 for sidelobes prescribed at -48, -52 and -56 dB.

In conclusion, we can say also that the method described in section 3.3., using source functions with both the field and the first derivative of the aperture field equal to zero at r = 1 enables us to prescribe several types of sidelobe envelopes. These source functions ensure an even more rapid decay of the unprescribed sidelobes than do the source functions with only the field values equal to zero at r=1.

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Appendix A

Derivation of eq. (2.7).

Start with eq. (2.5) for the far field g(u). Use $J_0(u_0)=J_0(0)=1$, then g(u) becomes

$$g(u) = J_{1}(u) \sum_{n=0}^{N} \frac{-a_{n}^{u} J_{0}(u_{n})}{u_{n}^{2} - u^{2}} =$$

$$= \frac{J_{1}(u)}{u} \left[a_{0}^{-} \frac{a_{1}^{u} u_{0}^{2} J_{0}(u_{1})}{u_{1}^{2} - u^{2}} - \dots - \frac{a_{N}^{u} u_{0}^{2} J_{0}(u_{N})}{u_{N}^{2} - u^{2}} \right]$$

$$= \frac{J_{1}(u)}{u} \left[a_{0}^{-} \frac{a_{1}^{\frac{1}{2}} J_{0}(u_{1})}{1 - \frac{u_{1}^{2}}{u_{1}^{2}} - \dots - \frac{a_{N}^{\frac{1}{2}} J_{0}(u_{N})}{1 - \frac{u_{2}^{2}}{u_{N}^{2}}} \right]$$

$$= \frac{J_{1}(u)}{u} \left[a_{0}^{-} \frac{a_{0}^{-} \frac{a_{1}^{\frac{1}{2}} J_{0}(u_{1})}{1 - \frac{u_{2}^{2}}{u_{1}^{2}} - \dots - a_{N}^{\frac{1}{2}} J_{0}(u_{N}) (1 - \frac{u^{2}}{u_{N}^{2}}) \dots (1 - \frac{u^{2}}{u_{N}^{2}}) \right]$$

$$= \frac{J_{1}(u)}{u} \left[\frac{a_{0}(1 - \frac{u^{2}}{u_{1}^{2}}) \dots (1 - \frac{u^{2}}{u_{N}^{2}}) - \dots - a_{N}^{\frac{1}{2}} J_{0}(u_{N}) (1 - \frac{u^{2}}{u_{N}^{2}}) \dots (1 - \frac{u^{2}}{u_{N}^{2}}) \right]$$

The nominator of the form between brackets is a polynomial in u^2 of order N, the zeros of which can be expressed by $(u_n + t_n)^2$ with n=1,2,...,N, so that g(u) can be written as

$$g(u) = \frac{2J_{1}(u)}{u} \prod_{n=1}^{N} \left\{ \frac{1 - \frac{u^{2}}{(u_{n} + t_{n})^{2}}}{1 - \frac{u^{2}}{u_{n}^{2}}} \right\}$$
(2.7)

if at the same time g(0) is normalized to 1.



WITH

77 = 3.




-A.3.-





WITH TT = 5.

-A.4.-





WITH # = 6.

• • TAYLOR CIRCULAR APERTURE PATTERN FOR -30 dB SIDELOBES





A-1 SIDELOBE EXTREMA OF -30 dB.



FIG. 11: CIRCULAR APERTURE DISTRIBUTIONS FOR TAYLOR PATTERNS WITH

\$ -1 SIDELOBE EXTREMA OF -30 dB.



-A.7.-





FIG. 15 : EARTH STATION ANTENNA RADIATION PATTERN ENVELOPES



6 8 4 10 12 14 16 18 2 20 ٥ + u 0 -/0 -20 -22.2 -30 -33.8 RELATIVE POWER (db) - 40.3 -40 -45.2 -49.2 -50 -60 -70 PRESCRIBED -22.5 dB SIDELOBE LEVEL Complited -22.2 dB 1 $u_1 + t_1 = 4.9553$ $u_2 = 8.6537$ ZEROS NORMALIZED APERTURE DISTRIBUTION EXCITATION COEFFICIENTS $a_0 = 4.37308$ $a_1 = -0.91311$ EFFICIENCY = 0.7623 0.5 GAIN REDUCTION-1.18 dB 43dB = 3.94



1

as Radius L

° | ,

NORMALIZED

-A.11.-

-A.12.-



-A.13.-



FIG. 19: APERTURE DISTRIBUTION $\sum_{m=0}^{1} a_m \int_{0} (u_m x)$ AND RADIATION PATTERN.



FIG. 20: APERTURE DISTRIBUTION $\sum_{m=0}^{1} a_m \int_0 (u_m r)$ and Radiation Pattern.

-A.15.-



-A.16.-



-A.17.-



.



FIRST APPROXIMATION

--- SECOND APPROXIMATION

10

16

8

4



-A.19.-

-A.20.-



FIG. 26: APERTURE DISTRIBUTION $\sum_{m=0}^{4} a_m \int_0 (u_m x)$ AND RADIATION PATTERN

-A.21.-





-A.22.-



-A.23.-



FIG. 29: APERTURE DISTRIBUTION $\sum_{m=0}^{4} a_m \int_{0} (u_m r)$ AND RADIATION PATTERN.



-A.25.-

- ---







-A.27.-



FIRST APPROXIMATION

SECOND APPROXIMATION.







Q 6 EFFICIENCY

GAIN REDUCTION

0.6575

1.82 dB

0.6455

-A.29.-

a second and a second

`~



-A.30.-



RADIATION PATTERN

÷

-A.31.-

I

-A.32.-



-A.33.-





-A.35.-



-



-A.37.-




-A.39.-



k.



-A.41.-



-A.42.-



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