

Two Algorithms for Fast and Accurate Passivity-Preserving Model Order Reduction

Ngai Wong, *Member, IEEE*, Venkataramanan Balakrishnan, *Member, IEEE*,
Cheng-Kok Koh, *Member, IEEE*, and Tung-Sang Ng, *Fellow, IEEE*

Abstract—This paper presents two recently developed algorithms for efficient model order reduction. Both algorithms enable the fast solution of continuous-time algebraic Riccati equations (CAREs) that constitute the bottleneck in the passivity-preserving balanced stochastic truncation (BST). The first algorithm is a Smith-method-based Newton algorithm, called Newton/Smith CARE, that exploits low-rank matrices commonly found in physical system modeling. The second algorithm is a project-and-balance scheme that utilizes dominant eigenspace projection, followed by a simultaneous solution of a pair of dual CAREs through completely separating the stable and unstable invariant subspaces of a Hamiltonian matrix. The algorithms can be applied individually or together. Numerical examples show the proposed algorithms offer significant computational savings and better accuracy in reduced-order models over those from conventional schemes.

Index Terms—Algebraic Riccati equation, balanced stochastic truncation (BST), Newton method, Smith method, SR algorithm.

I. INTRODUCTION

IN BACKEND verification of very-large-scale integration (VLSI) design, initial state-space modeling of interconnect and pin packages easily involves thousands or millions of state variables, thereby prohibiting direct computer simulation and analysis. Model order reduction (MOR) (e.g., in [1]–[20]) has become an integral step wherein the original linear model is reduced to, and approximated by, a much smaller linear model. It is desirable that the reduced-order model has small error over the frequency and/or time domains. Important properties such as stability and passivity¹ must also be preserved along the reduction process in order for the reduced-order models to be useful [1], [2].

MOR techniques include transfer-function moment matching (e.g., asymptotic waveform evaluation (AWE) [3]), Krylov

subspace projection (e.g., Padé Approximation via Lanczos (PVL) [4], and matrix PVL (MPVL) [5]), and the passivity-preserving congruence transform (e.g., passive reduced-order interconnect macromodeling algorithm (PRIMA) [1]). These schemes can be implemented by computationally efficient Krylov iterations [21] and often work well; however, they are “feasible” designs and not “optimal” in terms of any approximation criterion. Another class of techniques stem from control theory. Examples include the optimal Hankel-norm approximation [6], standard balanced truncation (BT) [7]–[9], [13], and the passivity-preserving balanced stochastic truncation (BST) [9]–[12], [14], [20]. The merits of these control-theoretic approaches are their superior global accuracy and deterministic error bounds [6], [11]. These schemes are, however, expensive to deploy due to the need of solving large-size matrix equations and decompositions. For example, standard BT requires solving a pair of Lyapunov equations (linear matrix equations), while BST calls for the solution of a pair of continuous-time algebraic Riccati equations (CAREs), i.e., quadratic matrix equations [22]. To alleviate the cost of standard BT, a series of recent work took advantage of the low-rank input/output matrices arising in many physical systems and developed the Cholesky-factor standard BT variants (e.g., in [13], [15]–[17], and [23]). These schemes are mainly based on the alternating direction implicit (ADI) method of solving Lyapunov equations and have comparable speed to the popular projection-based methods. However, standard BT does not guarantee passivity. BST preserves both passivity and stability and poses no special structural requirements on the original state space [12] but suffers from the high computational cost of solving CAREs. In fact, solving even a moderately sized CARE can be computationally intensive [18]. The heuristics in [24] tackle large CAREs with low-rank and sparse matrices, but theoretical basis and convergence proof are unavailable.

Standard techniques of solving a CARE include forming a Hamiltonian matrix and identifying its stable invariant subspace [22], [25]–[27]. Another way, provided a stabilizing initial condition is known, is to use the Newton method, which solves a Lyapunov equation in each iteration [22], [28]. In this paper, we summarize and report our recent work on fast implementations of both the Newton and Hamiltonian approaches in the context of large-scale BST [18], [19]. The first contribution is a Smith-method-based Newton algorithm, called the Newton/Smith CARE (NSCARE) algorithm, for quickly solving a large-scale CARE containing low-rank input/output matrices [18]. The algorithm uses Krylov subspace iterations and is numerically stable. The second contribution is an effective

Manuscript received January 25, 2005; revised August 14, 2005. This work was supported in part by the Hong Kong Research Grants Council under Project HKU 7173/04E, by the University Research Committee of The University of Hong Kong, and by the National Science Foundation of the USA under Grants ECS-0200320 and CCR-0203362. This paper was recommended by Associate Editor J. R. Phillips.

N. Wong and T.-S. Ng are with the Department of Electrical and Electronic Engineering, The University of Hong Kong, Hong Kong (e-mail: nwong@eee.hku.hk; tsng@eee.hku.hk).

V. Balakrishnan and C.-K. Koh are with the School of Electrical and Computer Engineering, Purdue University, West Lafayette, IN 47907 USA (e-mail: ragu@ecn.purdue.edu; chengkok@ecn.purdue.edu).

Digital Object Identifier 10.1109/TCAD.2006.873893

¹A passive system is one that does not generate energy internally. A dissipative system, such as an *RLC* network, is passive.

two-stage project-and-balance reduction algorithm [19], which provides a framework for trading off computational cost against model approximation accuracy. The projection basis in the first stage is formed by the dominant eigenspaces of the controllability and observability Gramians [8], [15], [17]. The projected intermediate model is then further reduced by BST. A novel observation that relies on completely separating the stable and unstable invariant subspaces of a Hamiltonian matrix reveals that two dual CAREs in BST can be jointly solved at the cost of essentially one. Numerical examples show the proposed algorithms exhibit fast reduction and deliver excellent model accuracy.

The paper is organized as follows. Section II presents the problem setting and preliminaries. Section III introduces the NSCARE algorithm and Section IV presents the project-and-balance algorithm. Numerical examples in Section V demonstrate the effectiveness of the proposed algorithms over conventional approaches. Finally, Section VI draws the conclusion.

II. BACKGROUND AND PRELIMINARIES

A target application of the proposed algorithms is in the reduction of large-scale *RLC* (and therefore passive) circuits commonly encountered in VLSI interconnect and package simulations. Consider a minimal state-space model of

$$\dot{x} = Ax + Bu \quad (1a)$$

$$y = Cx + Du \quad (1b)$$

where $A \in \mathbb{R}^{n \times n}$, $B \in \mathbb{R}^{n \times m}$, $C \in \mathbb{R}^{m \times n}$, $D \in \mathbb{R}^{m \times m}$, B , and C are of low rank, i.e., $m \ll n$, and $u \in \mathbb{R}^m$ and $y \in \mathbb{R}^m$ are power conjugates.² The system matrix A is assumed stable or, equivalently, its spectrum is in the open left half-plane, denoted by $\text{spec}(A) \subset \mathbb{C}^-$. We assume that $D + D^T > 0$, where the notation $M > 0$ ($M \geq 0$) means that the symmetric matrix M is positive definite (positive semidefinite).

RLC state-space models from modified nodal analysis (MNA) [29] in VLSI interconnect modeling have the properties of $A + A^T \leq 0$, $B = C^T$, and $D = 0$ [2]. Such a system can be recast into an equivalent form with $D + D^T > 0$ [9], [30]. Moreover, an *RLC* system in the descriptor format [2] with a singular E ($\in \mathbb{R}^{n \times n}$) preceding \dot{x} can be transformed into an equivalent minimal form in (1) [13], so the settings in (1) are assumed without loss of generality.

The controllability Gramian W_c and observability Gramian W_o are solved through the following continuous-time Lyapunov equations:

$$AW_c + W_c A^T + BB^T = 0 \quad (2)$$

$$A^T W_o + W_o A + C^T C = 0. \quad (3)$$

²For every component of u that is a node voltage (branch current), the corresponding component of y is a branch current (node voltage) so that $u^T y$ represents the instantaneous power injected into the system.

The spans (ranges) of W_c and W_o denote the reachable and observable states, respectively. For many physical systems including *RLC* circuits, W_c and W_o are of low numerical rank or approximately so. The implication is that the state activities usually take place in, or through, projection can be well captured by some lower dimensional subspaces [23].

A. Smith Method

The Smith method (e.g., in [9], [13], and [16]) solves a continuous-time Lyapunov equation by transforming it into a discrete-time version having exactly the same solution. For instance, the following two equations solve the same W_c :

$$AW_c + W_c A^T + BB^T = 0 \quad (4a)$$

$$A_p W_c A_p^T + W_c + B_p B_p^T = 0 \quad (4b)$$

where $A_p = (A - pI)(A + pI)^{-1}$, $B_p = \sqrt{-2p}(A + pI)^{-1}B$, and $p \in \mathbb{C}^-$ is a shift parameter. It follows that $W_c = \sum_{i=0}^{\infty} A_p^i B_p B_p^T (A_p^T)^i$. In practice, we want to minimize the spectral radius of A_p so that the power terms decay quickly and the infinite summation can be well approximated by finite terms. A simple possible choice is $p = -\sqrt{|\lambda_{\max}(A)| |\lambda_{\min}(A)|}$ [9], where $\lambda_{\max}(\cdot)$ and $\lambda_{\min}(\cdot)$ denote the maximum- and minimum-modulus eigenvalues. An important observation is that W_c is naturally cast as a matrix factorization, namely, when the growth of the summation becomes negligible after k terms

$$W_c \approx \sum_{i=0}^{k-1} A_p^i B_p B_p^T (A_p^T)^i = \mathcal{K}_k(A_p, B_p) \mathcal{K}_k(A_p, B_p)^T \quad (5)$$

where $\mathcal{K}_k(A_p, B_p) = [B_p \ A_p B_p \ \cdots \ A_p^{k-1} B_p]$ is called the k th order Krylov matrix and serves as a Cholesky factor of W_c . Obviously, this Krylov matrix factorization is computationally advantageous when B_p is of low rank and when the rank information of the Krylov matrix can be revealed quickly. The application of the Smith method in standard BT of VLSI models can be found in [9] and [13]. It should be noted that the Smith method is mathematically equivalent to the ADI method with a single shift parameter [6], [17], [23]. In fact, in our algorithms, the Smith-method-based parts can readily be replaced with the ADI scheme with multiple shifts. The Smith method is chosen because of its ease of exposition and also because it requires only one large-scale matrix inversion (in finding A_p), which constitutes the most expensive step in both proposed algorithms.

B. Krylov Subspace Iteration

The structure of a Krylov matrix lends itself to iterative computation. Among others, Arnoldi and Lanczos algorithms [21] are numerically efficient procedures for obtaining the Krylov matrix. We present only the Arnoldi algorithm here, due to space limitation. The following MATLAB-style pseudocodes assume a rank-one B_p , but block versions of Arnoldi and

Lanczos algorithms are available for B_p of arbitrary ranks (e.g., in [1] and [2]).

SmithArnoldi: Input ($A_p, B_p, \max_itr, \text{tol}$)

```

j := 1;
q1 := B_p / ||B_p||_2; Q1 := [q1]; beta := 1;
H1 := []; R1 := [[||B_p||_2];
W1 := B_p B_p^T;
While j ≤ max_itr,
  for i := 1 to j
    h_ij := q_i^T A_p q_j;
  end for
  r_{j+1} := A_p q_j - sum_{i=1}^j h_ij q_i;

```

$$H_j := \begin{bmatrix} & & & & h_{1j} \\ & & & & \vdots \\ & H_{j-1} & & & h_{j-1,j} \\ [0 & \cdots & 0 & \beta] & h_{jj} \end{bmatrix}$$

if $j > 1$

$$R_j := \begin{bmatrix} R_{j-1} & \\ [0 & \cdots & 0] \end{bmatrix} \left| H_j \begin{bmatrix} R_{j-1}(:, j-1) \\ 0 \end{bmatrix} \right.$$

```

w_j := Q_j R_j(:, j);
W_j := W_{j-1} + w_j w_j^T;
if (||w_j||_2 < tol) break while loop;

```

end if

```

beta := ||r_{j+1}||_2;
if (beta < tol) break while loop;

```

```

q_{j+1} := r_{j+1} / beta;
Q_{j+1} := [Q_j q_{j+1}];
j := j + 1;

```

end while

k := number of columns in R_j ;

Return W_k, Q_k, R_k , and H_k .

In short, the Arnoldi algorithm iteratively computes the k orthogonal columns of $Q_k \in \mathbb{R}^{n \times k}$, an upper Hessenberg matrix $H_k \in \mathbb{R}^{k \times k}$, an upper-triangular matrix $R_k \in \mathbb{R}^{k \times k}$, and an accumulation matrix $W_k \in \mathbb{R}^{n \times n}$ such that

- 1) $Q_k^T Q_k = I_k$;
- 2) $H_k = Q_k^T A_p Q_k$;
- 3) $\mathcal{K}_k(A_p, B_p) = [B_p \ A_p B_p \ \cdots \ A_p^{k-1} B_p] = Q_k R_k$ is a QR factorization;
- 4) Q_k spans the range of $\mathcal{K}_k(A_p, B_p)$;
- 5) $W_k = \mathcal{K}_k(A_p, B_p) \mathcal{K}_k(A_p, B_p)^T = (Q_k R_k)(Q_k R_k)^T$.

C. BST

The positive real lemma [2] states that the system in (1) is passive if and only if there exists a $P (\in \mathbb{R}^{n \times n}) \geq 0$ satisfying the linear matrix inequality (LMI)

$$\begin{bmatrix} A^T P + P A & P B - C^T \\ B^T P - C & -(D + D^T) \end{bmatrix} \leq 0. \quad (6)$$

Using the Schur complement, (6) is equivalent to

$$A^T P + P A + (P B - C^T)(D + D^T)^{-1}(B^T P - C) \leq 0. \quad (7)$$

The solution of (7) being zero is a CARE. Taking the matrix root $LL^T = (D + D^T)^{-1}$ and defining $\hat{B} = BL$, $\hat{C} = L^T C$, and $\hat{A} = A - B(D + D^T)^{-1}C$, the CARE is expressible as

$$F(P) = \hat{A}^T P + P \hat{A} + P \hat{B} \hat{B}^T P + \hat{C}^T \hat{C} = 0. \quad (8)$$

The solution of (8), if it exists, is not unique. There is a unique stabilizing solution P_∞ in the sense that $\text{spec}(\hat{A} + \hat{B} \hat{B}^T P_\infty) \subset \mathbb{C}^-$. In the mechanics of BST, we first align (balance) the most reachable states with a given input energy, quantified by $\int_{-\infty}^0 u(t)^T y(t) dt$, with the states delivering the maximum energy to the output, quantified by $-\int_0^\infty u(t)^T y(t) dt$ [9], [12]. It starts by finding the stabilizing solutions, P_{\min} and Q_{\min} , to the following dual CAREs:

$$\hat{A}^T P_{\min} + P_{\min} \hat{A} + P_{\min} \hat{B} \hat{B}^T P_{\min} + \hat{C}^T \hat{C} = 0 \quad (9a)$$

$$\hat{A} Q_{\min} + Q_{\min} \hat{A}^T + Q_{\min} \hat{C}^T \hat{C} Q_{\min} + \hat{B} \hat{B}^T = 0. \quad (9b)$$

Let $Q_{\min} = X X^T$ and $P_{\min} = Y Y^T$ be any Cholesky factorizations; now, do the singular value decomposition (SVD)

$$X^T Y = U \Sigma V^T \quad (10)$$

where $\Sigma \geq 0$ is an ‘‘economy size’’ k -by- k ($k \leq n$) diagonal matrix with singular values in descending magnitudes. Suppose the singular values of Σ are

$$\sigma_1 \geq \sigma_2 \geq \cdots \geq \sigma_r \gg \sigma_{r+1} \geq \cdots \geq \sigma_k. \quad (11)$$

These values quantify the importance of the states in the input-to-output energy transfer. Accordingly, in MOR, the ‘‘tail’’ can be truncated so only the most significant states remain. To do this, define I_m to be the identity matrix of dimension m , $0_{m \times n}$ to be an $m \times n$ zero matrix, and

$$T_L = \begin{bmatrix} I_r & 0_{r \times (k-r)} \end{bmatrix} \Sigma^{-\frac{1}{2}} V^T Y^T$$

$$T_R = X U \Sigma^{-\frac{1}{2}} \begin{bmatrix} I_r \\ 0_{(k-r) \times r} \end{bmatrix}. \quad (12)$$

The system $(T_L A T_R, T_L B, C T_R, D)$ then represents the stochastically balanced (also referred to as positive-real balanced [12], [20]) and truncated model. The best bound to date for the frequency-domain approximation error can be found in [11]. BST is preferred to standard BT because it guarantees passivity, in addition to stability, in the reduced-order model (e.g., in [9]–[12]).

III. NSCARE ALGORITHM

We start with a brief recap of the Newton method in solving CAREs (more details can be found in [22] and [28]). The advantages of the Newton algorithm include its quadratic convergence (once attained) and high numerical accuracy. Letting P_j , $j = 0, 1, \dots$, be the progressive estimates of the stabilizing

solution, we define $P_{j+1} = P_j + \delta P_j$, where δP_j is the search direction or Newton step. Substituting P_{j+1} into (8)

$$F(P_j + \delta P_j) = F(P_j) + (\hat{A} + \hat{B}\hat{B}^T P_j)^T \delta P_j + \delta P_j (\hat{A} + \hat{B}\hat{B}^T P_j) + \delta P_j \hat{B}\hat{B}^T \delta P_j. \quad (13)$$

Every Newton iteration step is a first-order error correction such that the sum of the first three terms on the right of (13) goes to zero, i.e.,

$$(\hat{A} + \hat{B}\hat{B}^T P_j)^T \delta P_j + \delta P_j (\hat{A} + \hat{B}\hat{B}^T P_j) + F(P_j) = 0 \quad (14)$$

which is simply a Lyapunov equation. After each step, (13) is left with a quadratic residual term. Thus, from the second step, i.e., $j = 1$, onwards

$$F(P_j) = \delta P_{j-1} \hat{B}\hat{B}^T \delta P_{j-1} \quad (15)$$

which is a low-rank matrix due to \hat{B} . For compactness, define the Lyapunov operator $\mathcal{L}_A : \mathbb{R}^{n \times n} \rightarrow \mathbb{R}^{n \times n}$ as $\mathcal{L}_A(P) = A^T P + P A$, so that (14) is rewritten as

$$\mathcal{L}_{\hat{A} + \hat{B}\hat{B}^T P_j}(\delta P_j) + \delta P_{j-1} \hat{B}\hat{B}^T \delta P_{j-1} = 0 \quad (16)$$

for $j = 1, 2, \dots$. This reminds us of the Smith method [cf., (4)] and its associated Krylov matrix factorization. To take full advantage of the low-rank input/output matrices, we note that the Smith transformation involves computing the inverse $(\hat{A} + \hat{B}\hat{B}^T P_j + pI)^{-1}$. Letting $S_p = (\hat{A} + \hat{B}\hat{B}^T P_{j-1} + pI)^{-1}$ and using a matrix-inversion lemma, we get

$$\begin{aligned} & (\hat{A} + \hat{B}\hat{B}^T P_j + pI)^{-1} \\ &= (\hat{A} + \hat{B}\hat{B}^T P_{j-1} + pI + \hat{B}\hat{B}^T \delta P_{j-1})^{-1} \\ &= S_p - S_p \hat{B}(I + \hat{B}^T \delta P_{j-1} S_p \hat{B})^{-1} \hat{B}^T \delta P_{j-1} S_p. \end{aligned} \quad (17)$$

Therefore, if S_p is precomputed, the right-hand side of (17) requires only a $m \times m$ inversion in subsequent steps provided the same p is used. To sum up, the NSCARE algorithm that solves (8) is as follows:

NSCARE: Input $(\hat{A}, \hat{B}, \hat{C}, P_0, \max_itr, \text{tol})$
 Find the shift p corresponding to $\hat{A} + \hat{B}\hat{B}^T P_0$;
 $T_p := (\hat{A} + \hat{B}\hat{B}^T P_0 - pI)$;
 $S_p := (\hat{A} + \hat{B}\hat{B}^T P_0 + pI)^{-1}$;
 Solve for δP_0 in $\mathcal{L}_{\hat{A} + \hat{B}\hat{B}^T P_0}(\delta P_0) + F(P_0) = 0$ with standard solvers. In particular, when $P_0 = 0$, solve $\mathcal{L}_{\hat{A}}(\delta P_0) + \hat{C}^T \hat{C} = 0$ using the SmithArnoldi algorithm with input $[(T_p S_p)^T, \sqrt{-2p} S_p^T \hat{C}^T, \max_itr, \text{tol}]$;
 $j := 1$;
 While $j \leq \max_itr$,
 $P_j := P_{j-1} + \delta P_{j-1}$;
 $\Theta := \hat{B}^T \delta P_{j-1}$;
 If convergence is slow, update p by
 Find shift p corresponding to $\hat{A} + \hat{B}\hat{B}^T P_j$;
 $T_p := (\hat{A} + \hat{B}\hat{B}^T P_j - pI)$;
 $S_p := (\hat{A} + \hat{B}\hat{B}^T P_j + pI)^{-1}$;

else use the same shift p and

$$\begin{aligned} T_p &:= T_p + \hat{B}\Theta; \\ S_p &:= S_p - S_p \hat{B}(I + \Theta S_p \hat{B})^{-1} \Theta S_p; \end{aligned}$$

end if

Solve for δP_j in $\mathcal{L}_{\hat{A} + \hat{B}\hat{B}^T P_j}(\delta P_j) + \Theta^T \Theta = 0$

[i.e., (16)] using SmithArnoldi with input $[(T_p S_p)^T, \sqrt{-2p} S_p^T \Theta^T, \max_itr, \text{tol}]$;

If the Frobenius norm $\|\delta P_j\|_F < \text{tol}$

$$P_\infty := P_j + \delta P_j;$$

Break while loop and return P_∞ ;

end if

$j := j + 1$;

end while.

The convergence analysis of the NSCARE algorithm follows closely from those in [22] and [28]. To save space, the main results are given without elaboration: 1) for a stabilizing initial guess P_0 , the subsequent Lyapunov operators in each Newton step are nonsingular and P_j , $j = 1, 2, \dots$, are also stabilizing; 2) $0 \leq \dots \leq P_j \leq P_{j+1} \leq P_\infty$; and 3) $0 \leq \|P_\infty - P_{j+1}\| \leq \gamma \|P_\infty - P_j\|^2$, where γ is a positive constant, i.e., convergence is quadratic once P_j falls into the region of convergence.

In practice, the tolerance parameter tol is set to a small value such as the machine precision. In the first call to SmithArnoldi (i.e., finding δP_0 or δP_1 , depending on the initial condition P_0), the number of iterations is the highest and then decreases in subsequent runs once quadratic convergence is acquired. For a strictly dissipative system such as an *RLC* circuit modeled to high fidelity, it can be shown that there exists a representation such that strict inequality is satisfied in (6) with $P = I$ (see also [9] and [14]). It follows that \hat{A} is stable and the initial guess of $P_0 = 0$ is stabilizing. Moreover, since $0 \leq P_j \leq P_\infty < I$, we have $\|P_\infty - P_j\| < 1$, $j = 0, 1, \dots$. In other words, under the mild assumption of a strictly passive (dissipative) system, the quadratic convergence of the NSCARE algorithm is guaranteed.

Remarks:

- 1) In our NSCARE implementation, p is approximated by first applying a Lanczos algorithm for κ steps on $\hat{A} + \hat{B}\hat{B}^T P_j$ to obtain a tridiagonal matrix $T_\kappa \in \mathbb{R}^{\kappa \times \kappa}$, $\kappa \ll n$, whose eigenvalues closely approximate the extremal eigenvalues of $\hat{A} + \hat{B}\hat{B}^T P_j$. Then, a simple (inverse) power iteration [21] is used to estimate the magnitude of the maximum (minimum) eigenvalue of T_κ to form p . The initial Lanczos process has $O(\kappa n^2)$ work and the power iterations require $O(\kappa^3)$ work.
- 2) When a matrix-inversion lemma is used to update S_p , full-rank inversion is bypassed and the work reduces from $O(n^3)$ to $O(m^3)$. Subsequently, the only $O(n^3)$ steps in NSCARE are the explicit updates of S_p . In case of sparse matrices wherein the inversion in S_p can be done with $O(n^2)$ work, the whole algorithm will further reduce to an $O(n^2)$ one. In our examples later on, it is found that only one or two shift updates are needed. This is because when P_j is converging quadratically, the norm of δP_j is small and has little effect on p .
- 3) Suppose (9a) is solved with NSCARE in N Newton steps with $P_0 = 0$. Let the stabilizing solution be

$P_{\min} = P_{\infty}$ and the number of iterations in each call to SmithArnoldi be k_1, k_2, \dots, k_N , and $k_T = k_1 + k_2 + \dots + k_N$. Then, in terms of the outputs of SmithArnoldi, $P_{\min} = \sum_{i=1}^N (Q_{k_i} R_{k_i})(Q_{k_i} R_{k_i})^T = (Q_{k_T} R_{k_T})(Q_{k_T} R_{k_T})^T$, where $Q_{k_T} = [Q_{k_1} \dots Q_{k_N}]$ and $R_{k_T} = \text{diag}(R_{k_1}, \dots, R_{k_N})$. Thus, referring to (10), a factor of P_{\min} is given by $Y = Q_{k_T} R_{k_T}$. The factor X of Q_{\min} is obtained similarly. In our experiments, N is usually less than ten and k_T is in the order of tens or hundreds regardless of the CARE size; thus, only a medium-size SVD is needed even for high-order initial models. Consequently, the NSCARE algorithm also helps to elude two large-size matrix factorizations and one large-scale SVD required by the original BST implementation.

IV. PROJECT-AND-BALANCE ALGORITHM

The NSCARE algorithm is suitable for the BST of medium to large systems (say, orders from hundreds to thousands). For even higher orders (thousands to millions), it is advisable to adopt a two-stage project-and-balance approach. The idea of a stepwise reduction is not new (e.g., in [12] and [15]). The attention of our paper is on the efficient implementation of such a scheme. In particular, the first projection step is carried out by the fast Smith method as in NSCARE. More importantly, in the second BST step, we introduce an innovative way of simultaneously solving the dual CAREs at the cost of essentially one. The step adopts a Hamiltonian approach that does not rely on low-rank input/output matrices. As a result, it is by itself an attractive scheme for BST of systems with a large number of input/output ports.

A. Eigenspace Projection

The first stage of reduction is to select an appropriate subspace onto which the original high-order system is projected. It is therefore well justified to use the (approximate) spans of W_c and W_o as they capture (nearly) all state activities. This idea appeared as dominant subspaces projection in [8] and also as dominant Gramian eigenspaces method in [17]. The Smith method together with Krylov iteration are attractive candidates for the task. For example, the SmithArnoldi algorithm can be used for extracting the span of, say, W_c in (4). Suppose the algorithm converges in τ steps rendering $\mathcal{K}_{\tau}(A_p, B_p) = Q_{\tau} R_{\tau}$; it is obvious that Q_{τ} spans the column range of W_c . A counterpart Q_v corresponding to the column range of W_o is obtained similarly. A Gram-Schmidt (GS) orthogonalization of Q_v against Q_{τ} (columns in Q_{τ} are already orthogonal) produces an orthogonal $Q_k = \text{GS}([Q_{\tau} \ Q_v]) \in \mathbb{R}^{n \times k}$, $k \leq \tau + v$, which can be taken as the projection basis to generate an intermediate model of order k . Referring to (6), RLC models obtained from MNA have the properties $A + A^T \leq 0$, $B = C^T$, and $D = 0$ [14]. The passivity of the circuit is then borne out by the fact that $P = I$ is a solution satisfying (6). Performing a congruence transformation of compatible dimensions, we have

$$\begin{bmatrix} Q_k^T & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A + A^T & B - C^T \\ B^T - C & 0 \end{bmatrix} \begin{bmatrix} Q_k & 0 \\ 0 & I \end{bmatrix} \leq 0. \quad (18)$$

It is easily seen that the system $(Q_k^T A Q_k, Q_k^T B, C Q_k, 0)$ inherits its passivity from its parent. Using techniques in [9] and [30], this system with a zero D matrix can be transformed into an equivalent one with $D + D^T > 0$, from which the CAREs for BST can be derived.

B. Solving Dual CAREs

The intermediate passive model from projection is then subject to BST to achieve further reduction with guaranteed passivity. Many algorithms exist for solving a CARE via identifying the stable invariant subspace of an associated Hamiltonian matrix (e.g., in [22], [25], and [31]). While this is sufficient for the stabilizing solution, information about the unstable invariant subspace is just a few steps away and not utilized. We show that with slight extra effort, the stable and unstable invariant subspaces can be completely separated, which in turn enables the joint solution of the dual CAREs in (9). Consider the Hamiltonian matrices H and H' , corresponding to (9a) and (9b), respectively

$$H = \begin{bmatrix} \hat{A} & \hat{B}\hat{B}^T \\ -\hat{C}^T\hat{C} & -\hat{A}^T \end{bmatrix}, \quad H' = \begin{bmatrix} \hat{A}^T & \hat{C}^T\hat{C} \\ -\hat{B}\hat{B}^T & -\hat{A} \end{bmatrix}. \quad (19)$$

If λ is an eigenvalue of a Hamiltonian matrix, then so is $-\lambda$. Since H and H' are real, eigenvalues apart from the real and imaginary axes occur even in quadruple $(\lambda, -\lambda, \bar{\lambda}, -\bar{\lambda})$. By our assumption of a minimal passive system, H has no eigenvalues on the imaginary axis, and the stable and unstable invariant subspaces can be decoupled, namely

$$H \begin{bmatrix} X_{11} & X_{12} \\ X_{21} & X_{22} \end{bmatrix} = \begin{bmatrix} X_{11} & X_{12} \\ X_{21} & X_{22} \end{bmatrix} \begin{bmatrix} \Lambda_s & 0 \\ 0 & \Lambda_u \end{bmatrix} \quad (20)$$

where Λ_s contains the stable eigenvalues and Λ_u the unstable ones. A well-known fact is that X_{11} is invertible and $P_{\min} = X_{21}X_{11}^{-1} \geq 0$. A key observation is that from $H' = \begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix} (-H) \begin{bmatrix} 0 & I \\ I & 0 \end{bmatrix}$, we also get $Q_{\min} = X_{12}X_{22}^{-1} \geq 0$. In other words, all information about P_{\min} and Q_{\min} are contained in (20). The decoupling of the invariant subspaces can be achieved by standard ways (e.g., Grassmann manifolds [32]). Nonetheless, for completeness of this paper and interested practitioners, we propose a fast-converging quadruple-shift bulge-chasing SR algorithm for completely separating the stable and unstable invariant subspaces of a Hamiltonian matrix. The implementation details of this special SR algorithm are given in the Appendixes.

As in NSCARE, the most expensive step in this two-stage algorithm is the full matrix inversion ($O(n^3)$ work) in computing A_p in the first-stage Smith transformation. Apart from this, all other operations in the projection stage are at most $O(n^2)$. The intermediate model in the second-stage BST is much smaller ($k \ll n$) and imposes a minor burden on the overall cost. It should be stressed that the subspace separation technique here is independent of the projection applied in Section IV-A. As a

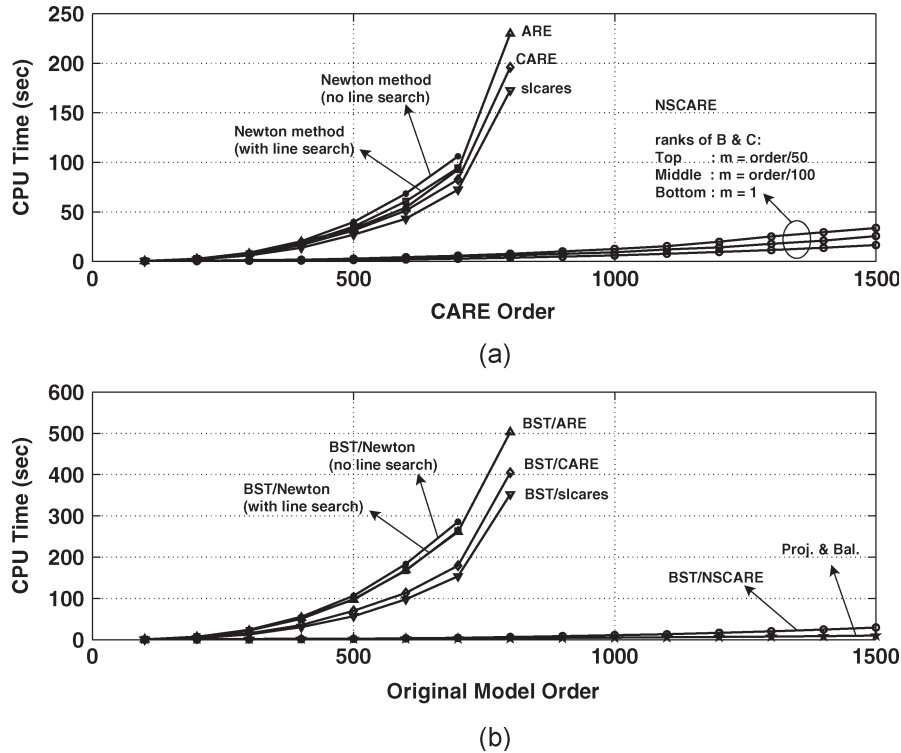


Fig. 1. (a) CPU time for solving single CARE. (b) CPU time for doing MOR.

result, it can also be employed in direct BST to approximately double the speed of solving dual CAREs.

V. NUMERICAL EXAMPLES

All numerical experiments in this section were carried out in the MATLAB R14 environment on a 3-GHz 3G-RAM PC with many applications running. The NSCARE algorithm, the project-and-balance algorithm together with the quadruple-shift SR procedure in the Appendixes, as well as the Newton method with and without line search [28] were coded and executed as MATLAB script (text) files. For comparison, we also employed the MATLAB CARE solver routines ARE and CARE that implement the Hamiltonian-based Schur vector and eigenvalue methods, respectively (see [22] and references therein). In addition, the prebuilt FORTRAN 77 routine slcares, a numerically reliable and efficient implementation of the Schur vector method, was invoked from the subroutine library in systems and control theory (SLICOT) library [27] via a MATLAB gateway.

First, we study randomly generated strictly passive systems, with $A + A^T < 0$ and rank-one B and C , satisfying the settings in (1). The CARE in (9a) was formed and solved by various solvers, as depicted in Fig. 1(a). For fairness, solutions from NSCARE were computed to the same or better accuracy than those by other solvers. Though all these solver algorithms are of $O(n^3)$ complexity, they behave quite differently. In particular, NSCARE easily handles orders as high as 1500, while others require intensive computation well below 1000. Perhaps more importantly, NSCARE scales favorably with increasing model order, e.g., at 500, it is at least 20 times faster and 50 times faster at 800. To investigate the effect of increasing ranks in B and C , Fig. 1(a) also includes the curves whereby the ranks of B and

C equal the CARE order divided by 100 and 50, respectively. Block Arnoldi algorithms were used in these cases. It is seen that the growth in computation is relatively mild and is more obvious at higher orders. In practical systems, however, the ranks of B and C (associated with number of input/output ports) usually remain constant and seldom grow with model order. This justifies the use of NSCARE whenever ranks of B and C are low. Fig. 2(b) shows the convergence of the NSCARE iterates at several CARE orders, wherein the relative residual $\|P_j - P_{\min}\|_F / \|P_{\min}\|_F$ ($\|\cdot\|_F$ being the Frobenius norm of a matrix) is plotted. Expectedly, a quadratic rate is observed since NSCARE is simply an efficient implementation of the Newton method. Fig. 2(a) plots the convergence of the iterates in the SmithArnoldi algorithm when solving for δP_0 in the first step of NSCARE (which for δP_j s are similar). The rate is superlinear, which is again expected because the Smith method, being a special case of ADI, inherits the superlinear convergence of the latter.

Fig. 1(b) plots the CPU time for realizing BST and the project-and-balance algorithm. Rank-one B and C are used. BST requires solving two CAREs plus matrix factorizations, so the time curves corresponding to different solvers, compared with their counterparts in Fig. 1(a), are generally more than double. As stated in the remarks in Section III, for models with low-rank input/output matrices, BST/NSCARE has the additional advantage of eluding large-scale matrix factorizations and SVDs. On the other hand, the project-and-balance algorithm employed an intermediate model of order about 50 at all CARE orders. Its curve corresponds to the sum of projection time and BST time, in contrast to the direct BST in other curves. As the intermediate model order remains almost constant, the time for BST (using the proposed SR algorithm) is

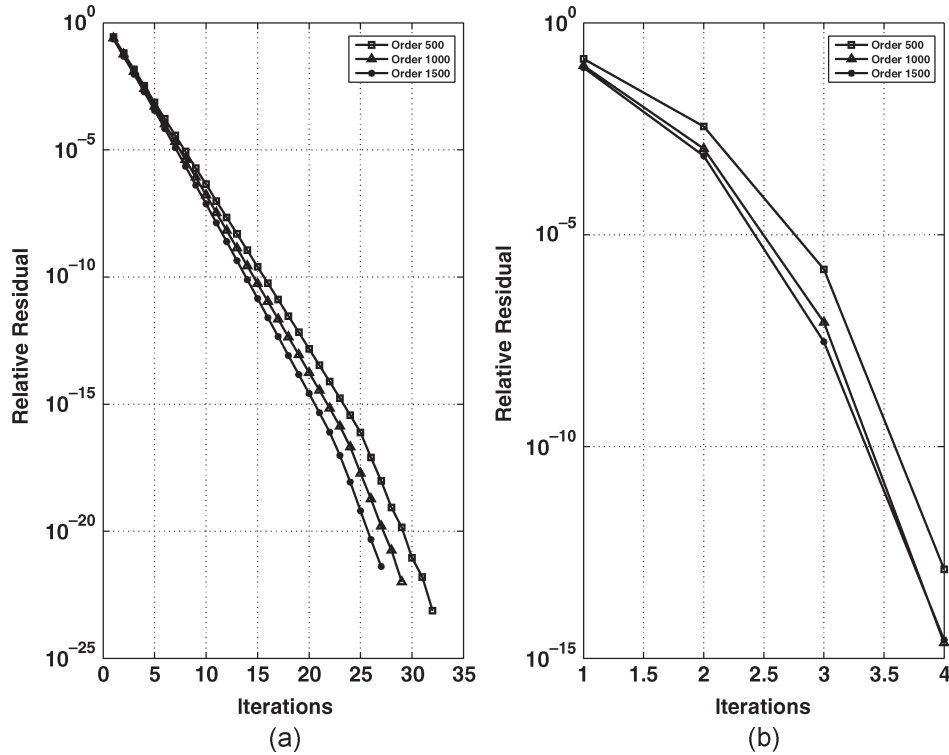


Fig. 2. (a) Superlinear convergence of iterates in SmithArnoldi algorithm. (b) Quadratic convergence of iterates in NSCARE, at several CARE orders.

independent of the original model order. From the figure, it can be seen that NSCARE and the project-and-balance algorithm consume similar resources and are superior to conventional BST implementations.

Next, we apply NSCARE and the project-and-balance scheme towards some large-scale MOR problems. The first example results from the extraction of an on-chip planar square spiral inductor suspended over a copper plane [23]. The initial model of order 500 is reduced using different schemes including PRIMA [1], low-rank square root (LRSR) [13], [23], the proposed project-and-balance algorithm [19], BST with NSCARE [18], and BST with conventional solvers. The PRIMA model is set to the same order as that from BST/NSCARE. The frequency response and error plots are shown in Fig. 3, and the CPU times and final model orders are tabulated in Table I. It should be noted that if the Hamiltonian solver routines (in this case, *scars* and CARE) could be modified to implement the complete subspace separation of Section IV-B, the time in solving the dual CAREs would have been approximately halved, but even so, BST/NSCARE is still more efficient due to the exploitation of low-rank B and C . From the figure and table, with a comparable model order, PRIMA exhibits a bigger mismatch over the frequency axis. Models from BST, unsurprisingly, tend to have better global accuracy [12], [19]. In our two-stage project-and-balance implementation, the initial model was first reduced to an intermediate model of order 100 (the same in the next two examples), followed by BST using the proposed SR algorithm. The excellent accuracy can be attributed to the effectiveness of dominant eigenspace projection in capturing most state activities, as has been observed in [17]. Another possible reason is the better numerical conditioning in a stepwise reduction.

The second example is the simulation of a wire model with 500 repeated RLC sections in Fig. 4(a), producing a model of order 1000 [20]. For simplicity, identical sections are used. The input and output are taken as the voltage and current into the first section, respectively. The reduction results are shown in Fig. 5 and Table I. As before, PRIMA is less accurate in approximating the original response, while models from other schemes have almost indistinguishable responses from the original. The third example in Fig. 4(b) depicts another wire modeling wherein the center loop is repeatedly inserted to generate a model of order 1000. Again, similar observations can be drawn from Fig. 6 and Table I. To this end, a few technical details are worth mentioning.

- 1) Runtimes of BST/NSCARE and the project-and-balance algorithm are comparable to “fast” algorithms such as PRIMA and LRSR. In fact, BST/NSCARE is about an order faster than conventional BST realizations.
- 2) LRSR does not guarantee passivity of the reduced state-space model. A passivity test and enforcement may be needed before the reduced model is connected for global simulation.
- 3) Projection-type algorithms like PRIMA and the first stage of the project-and-balance algorithm require the initial (passive) state space to be in a certain form [2], [12] (essentially for (18) to hold) and this is not always convenient or feasible. BST, on the other hand, poses no constraints on the internal structure of the state-space model.
- 4) BST avoids the selection of expansion points and final model order as in PRIMA, which would involve *a priori* knowledge of the original response.

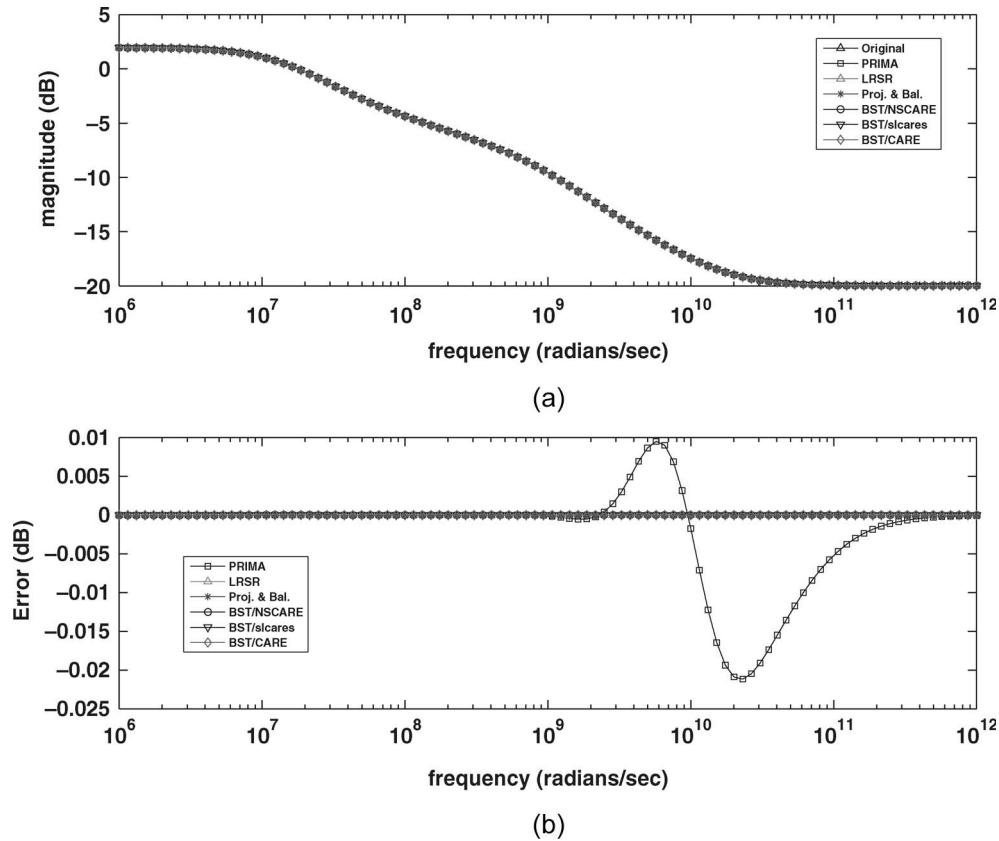


Fig. 3. (a) Frequency responses of spiral inductor and reduced-order models. (b) Deviation from original response.

TABLE I
CPU TIMES AND REDUCED MODEL ORDERS (TIME/ORDER) FOR VARIOUS SCHEMES WITH TIME OF PRIMA NORMALIZED TO ONE

	PRIMA	LRSR	Proj. & Bal.	BST/NSCARE	BST/slcares	BST/CARE
Inductor		(1.76)	(1.58)	(7.65)	(66.67)	(79.94)
Order=500		(0.02)	(1.00)	(0.73)	(6.04)	(5.84)
	1 / 10	1.78 / 9	2.58 / 10	8.38 / 10	72.71 / 11	85.78 / 11
Wire 1		(1.19)	(5.05)	(7.31)	(107.75)	(117.33)
Order=1000		(0.003)	(0.15)	(0.003)	(6.87)	(7.23)
	1 / 5	1.19 / 6	5.20 / 8	7.31 / 5	114.62 / 5	124.56 / 8
Wire 2		(1.26)	(4.83)	(7.41)	(103.53)	(115.45)
Order=1000		(0.003)	(0.12)	(0.006)	(6.82)	(6.17)
	1 / 6	1.26 / 8	4.95 / 9	7.42 / 6	110.35 / 6	121.62 / 9

Note:

1. The upper bracket in Proj. & Bal. is the time for projection and the lower bracket is the time for BST.

2. The upper bracket in the direct BST (LRSR) is the time for solving dual CAREs (Lyapunov equations) and the lower bracket is the time for matrix factorizations.

- 5) From our experiments, it is seen that reduced-order models from BST/NSCARE generally have a lower order for the same accuracy.

To summarize, both NSCARE and the project-and-balance algorithm are important candidates in large-scale reduction problems. These control-theoretic approaches generally produce reduced-order models of high global accuracy. The algorithms can be applied individually or together to enable the deployment of previously impractical large-scale passivity-preserving MOR. As a general guideline, for very high model order (millions), the projection-type method is used to bring the order down to thousands or hundreds. It is then followed by BST to further compress the order down to hundreds or tens.

When the ranks of input/output matrices are low, NSCARE provides a fast means for implementing BST. When there are a large number of input/output ports, NSCARE may not be advantageous and the solution of dual CAREs through subspace separation may be considered due to its independence on the number of ports. Preservation of passivity, in addition to stability, in the reduced-order models is guaranteed throughout the process.

VI. CONCLUSION

This paper has described two algorithms for fast and accurate passivity-preserving MOR. The first algorithm is a Newton-method variant based on Smith and Krylov methods, called

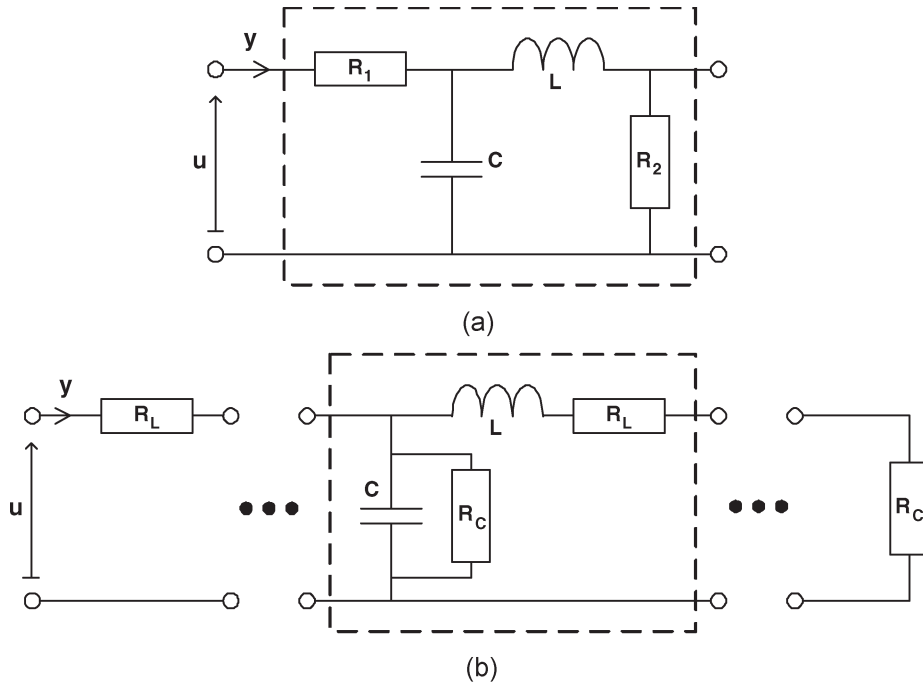


Fig. 4. (a) One RLC section of wire modeling. (b) Another RLC modeling. For simplicity, values are used in all sections: $R_1 = R_L = 0.1$, $R_2 = R_C = 1.0$, $C = 0.1$, and $L = 0.1$.

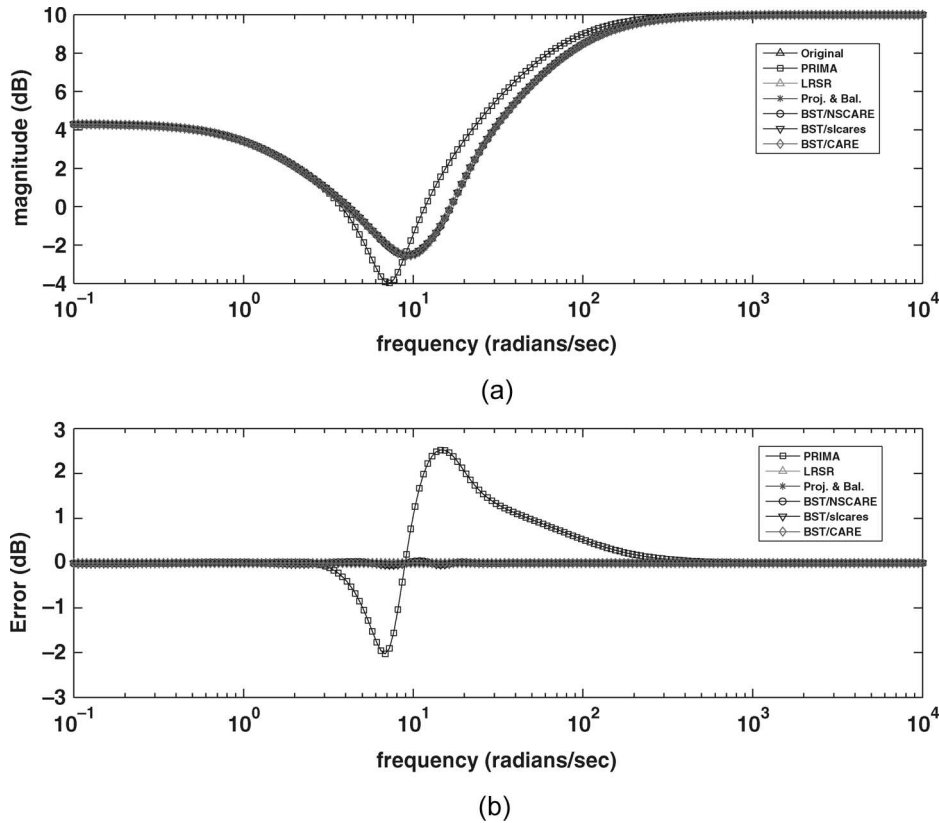


Fig. 5. (a) Frequency responses of first wire model and its reduced-order models. (b) Deviation from original response.

NSCARE, which exploits low-rank input/output matrices to quickly solve a CARE. It can help avoid two large-size matrix factorizations and one large-size SVD in traditional BST. The second algorithm is an efficient implementation of a two-stage project-and-balance reduction procedure. The first stage

consists of dominant eigenspace projection, again using the fast Smith method. Moreover, applying the novel idea of separating the stable and unstable invariant subspaces of a Hamiltonian matrix, the second stage solves two dual CAREs in BST at the cost of slightly more than one. An effective quadruple-shift

Here, $\Theta = I_k + [(1+v^2)^{-1/4} - 1](e_{i-1}e_{i-1}^T + e_i e_i^T)$ and $\Phi = (v(1+v^2)^{-1/4})(e_{i-1}e_i^T + e_i e_{i-1}^T)$. Algorithm G zeroes a single entry in the upper half of a column of the Hamiltonian matrix when $y_{k+i} = 0$ (algorithm J does not work) and $y_{k+i-1} \neq 0$. Given i , $2 \leq i \leq k$, $x \in \mathbb{R}^{2k}$, we have $G(i, v)x = y$, where $y_i = 0$.

Algorithms J and H use orthogonal symplectic matrices, while algorithm G uses a nonorthogonal symplectic matrix of condition number $\text{cond}_2[G(i, v)] = (1 + v^2)^{1/2} + |v|$.

• Implicit Quadruple-Shift SR Algorithm

As in modern implementations of the QR algorithm [21], the SR counterpart utilizes Implicit S bulge chasing such that all computations are in the real domain. Single- and double-shift strategies are investigated in the technical report version of [31], in which the shifts are chosen from the real and imaginary axes only. Our implementation waives this constraint and complies better with the quadruple occurrence of eigenvalues away from the axes. A proven heuristic to speed up convergence is to choose the four shifts as eigenvalues of the 4×4 subblock [cf., (21)]

$$N_j = \begin{bmatrix} a_j & 0 & c_j & b_j \\ 0 & a_{j+1} & b_j & c_{j+1} \\ q_j & 0 & -a_j & 0 \\ 0 & q_{j+1} & 0 & -a_{j+1} \end{bmatrix} \quad (25)$$

where $j = k - 1$ in the first iteration and gradually decreases when the J -tridiagonal matrix deflates [21], [33]. Defining the expression $\alpha_j = a_j^2 + c_j q_j$, the characteristic polynomial of (25) is found to be

$$s^4 - (\alpha_j + \alpha_{j+1})s^2 + \alpha_j \alpha_{j+1} - b_j^2 q_j q_{j+1} = 0. \quad (26)$$

The roots of (26) are used as shifts. Analogous to the Implicit Q theorem in the QR algorithm, the first column of the following matrix product is required for Implicit S similarity transform:

$$\begin{aligned} p(\lambda) &= (H - \lambda I)(H + \lambda I)(H - \bar{\lambda} I)(H + \bar{\lambda} I) \\ &= H^4 - 2\text{Re}(\lambda^2)H^2 + |\lambda|^4 I \\ &= H^4 - (\alpha_j + \alpha_{j+1})H^2 + (\alpha_j \alpha_{j+1} - b_j^2 q_j q_{j+1})I. \end{aligned} \quad (27)$$

Reusing the definition of α_j and a MATLAB-style representation, the first columns of H^2 and H^4 are

$$H^2(:, 1) = \begin{bmatrix} \alpha_1 \\ b_1 q_1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}, \quad H^4(:, 1) = \begin{bmatrix} \alpha_1^2 + b_1^2 q_1 q_2 \\ b_1 q_1 (\alpha_1 + \alpha_2) \\ b_1 q_1 b_2 q_2 \\ 0 \\ \vdots \\ 0 \end{bmatrix}. \quad (28)$$

Setting $H_1 := H$ and using algorithm H to find an $H(1, 3, w)$ such that $H(1, 3, w)p(\lambda)e_1$ is a multiple

of e_1 , the bulge chasing begins by forming $H_2 := H(1, 3, w)H_1H(1, 3, w)^T$ and $\Pi := H(1, 3, w)^T$. An example $H_2 \in \mathbb{R}^{12 \times 12}$ looks like

$$\left[\begin{array}{ccc|cccc} \times & \times & \times & \times & \times & \times & \times \\ \times & \times & \times & \times & \times & \times & \times \\ \times & \times & \times & \times & \times & \times & \times \\ & & & \times & \times & \times & \times \\ & & & & \times & \times & \times \\ & & & & & \times & \times \\ \hline \times & \times & \times & \times & \times & \times & \times \\ \times & \times & \times & \times & \times & \times & \times \\ \times & \times & \times & \times & \times & \times & \times \\ & & & \times & \times & \times & \times \\ & & & & \times & \times & \times \\ & & & & & \times & \times \end{array} \right].$$

To restore the J -tridiagonal structure, we refer to the following matrix. Circles represent zeroing of entries and asterisks stand for newly generated entries. The (9, 1) entry is zeroed using $H_3 := J(3, c, s)H_2J(3, c, s)^T$ and the update $\Pi := \Pi J(3, c, s)^T$. The entry at (7, 3) is automatically zeroed due to the (Hamiltonian) structure-preserving symplectic transform. Similarly, algorithm J is used to zero (8, 1). Then, (3, 1) is zeroed by algorithm H with $H(2, 2, w)$, followed by algorithm G for (2, 1). Next, on the right half, (9, 7) and (8, 7) are zeroed by two times of algorithm J , and on the upper right quadrant, (4, 7) and (3, 7) are zeroed with algorithm H with $H(2, 3, w)$. Consequently, the bulge is pushed to the lower right and the process is continued until it is completely driven out

$$\left[\begin{array}{cccc|ccccc} \times & \otimes & \otimes & & \times & \times & \otimes & \otimes \\ \otimes & \times & \times & * & \times & \times & \times & \times & * \\ \otimes & \times & \times & * & \otimes & \times & \times & \times & * \\ & & * & * & \times & \times & \times & \times & \times \\ & & & & & \times & & & \times \\ & & & & & & * & * & \times & \times & \times \\ \hline \times & \otimes & \otimes & & \times & \otimes & \otimes & & & & \\ \otimes & \times & \times & * & \otimes & \times & \times & * & & & \\ \otimes & \times & \times & * & \otimes & \times & \times & * & & & \\ & & * & * & \times & * & * & \times & & & \\ & & & & & \times & & & & \times & \\ & & & & & & \times & & & & \times \end{array} \right].$$

As the iteration proceeds, some of the b_j s become negligibly small and the problem size deflates as in the QR algorithm. Ultimately, the SR algorithm reduces the J -tridiagonal matrix into decoupled 2×2 and 4×4 subblocks. The stability and passivity of the intermediate model (Section IV-A) implies the absence of purely imaginary eigenvalues. Using the procedures in [33], the 2×2 (4×4) subblock can then be transformed into an upper (block) triangular form with the upper left (block) entry containing the eigenvalue(s) with negative real part(s). The example that follows show an upper block

triangular form with interleaved 2×2 and 4×4 subblocks (lifted into appropriate planes). The subblocks in the upper left (lower right) quadrant contain all the stable (unstable) eigenvalues. Appendix II studies the zeroing of the upper right quadrant to eventually arrive at (20)

$$\left[\begin{array}{ccc|ccc} \times & & & \times & & \\ & \times & \times & & \times & \times \\ & \times & \times & & \times & \times \\ & & & \times & & \\ \hline & & & & \times & \\ & & & & \times & \times \\ & & & & \times & \times \\ & & & & & \times \\ & & & & & \times & \times \\ & & & & & \times & \times \end{array} \right].$$

Remarks:

- 1) An alternative to perform the projection in the project-and-balance algorithm is by an implicitly restarted Lanczos algorithm [31]. In that case, H is readily in J -tridiagonal form, but the projection basis may not be as good as the dominant eigenspace in capturing state transitions.
- 2) JHESS algorithm in [33] can be used to transform a Hamiltonian matrix into J -tridiagonal form. The existence of this transformation is strongly dependent on the first column of the similarity transform matrix [31]. The set of these breakdown-free vectors is dense in \mathbb{R}^{2k} . Should breakdown (or near breakdown) occur due to high condition numbers in algorithm G , a different projection basis Q_k in Section IV-A is chosen by varying the order and/or number of columns in Q_τ and Q_v . If the implicitly restarted Lanczos algorithm is used [31], then it is a simple matter of invoking an implicit restart.
- 3) Convergence of the quadruple-shift SR algorithm is excellent (usually within ten iterations) under mild conditions. In the few cases where algorithm G produces a very large condition number (only during early iterates), an exceptional shift is performed and the process is continued [33]. In the BST of the intermediate model, the transformation to J -tridiagonal form requires $O(k^3)$ work (not required in implicitly restarted Lanczos), while that of the SR algorithm is $O(k^2)$. As mentioned, $k \ll n$ and the cost of the second stage of the project-and-balance algorithm is insignificant.

APPENDIX II

SEPARATION OF INVARIANT SUBSPACES

We introduce additional symplectic transforms for each type of subblock, at a small cost, to completely separate the stable and unstable invariant subspaces. This brings about the solution of the dual CAREs in (9) at the complexity of practically one.

1) 2×2 Subblock

Let N_j be an ordered subblock taken from the $j, k + j$ plane of the $2k \times 2k$ matrix

$$N_j = \begin{bmatrix} -\lambda_j & x_j \\ 0 & \lambda_j \end{bmatrix} \quad (29)$$

where $-\lambda_j < 0$ and x_j is nonzero (otherwise, no processing is required). Defining the 2×2 symplectic matrix

$$T_j = \begin{bmatrix} 1/2\lambda_j & x_j \\ 0 & 2\lambda_j \end{bmatrix} \quad (30)$$

it is easy to verify that $T_j^{-1}N_jT_j$ gives the diagonal matrix $\text{diag}(-\lambda_j, \lambda_j)$. Lifting T_j into the $j, k + j$ plane and updating Π completes the subspace separation in this subblock.

2) 4×4 Subblock

Let N_j be an ordered subblock taken from the $j, j + 1, k + j, k + j + 1$ plane of the $2k \times 2k$ matrix

$$N_j = \begin{bmatrix} \Delta_j & \Omega_j \\ 0 & -\Delta_j^T \end{bmatrix} \quad (31)$$

where $\Delta_j, \Omega_j (= \Omega_j^T)$ are 2×2 matrices. Assume Δ_j contains the stable eigenvalues $-\lambda_j, -\bar{\lambda}_j$ whose real parts are negative. The key to separating the subspaces is to realize that the column range of $U_j = (N_j + \lambda_j I)(N_j + \bar{\lambda}_j I)$ spans the unstable invariant subspace. A simple manipulation shows

$$\text{span}(U_j) = \text{span} \left(\left[\begin{array}{c} \Delta_j \Omega_j - \Omega_j \Delta_j^T + 2\text{Re}(\lambda) \Omega_j \\ -4\text{Re}(\lambda) \Delta_j^T \end{array} \right] \right). \quad (32)$$

On the right-hand side of (32), denoting the upper partition by Z_1 and the lower partition by Z_2 , we define

$$F_j = \begin{bmatrix} Z_2^{-T} & Z_1 \\ 0 & Z_2 \end{bmatrix}. \quad (33)$$

It is easy to see that F_j is well defined (Z_2 invertible) and symplectic. Moreover, $F_j^{-1}N_jF_j$ gives $\text{diag}(\Delta_j, -\Delta_j^T)$. Lifting F_j into the $j, j + 1, k + j, k + j + 1$ plane and updating Π completes the subspace separation in this subblock. Finally, we have $H\Pi = \Pi \text{diag}(\Lambda_s, -\Lambda_s^T)$, and solutions to the dual CAREs can be extracted from Π as in (20).

ACKNOWLEDGMENT

The authors are grateful to the Associate Editor and anonymous reviewers for their helpful and constructive comments, particularly with the experimental part of a previous manuscript. Any opinions, findings, and conclusions or recommendations expressed in this material are those of the author(s) and do not necessarily reflect the views of the funding agencies.

REFERENCES

- [1] A. Odabasioglu, M. Celik, and L. T. Pileggi, "PRIMA: Passive reduced-order interconnect macromodeling algorithm," *IEEE Trans. Comput.-Aided Des. Integr. Circuits Syst.*, vol. 17, no. 8, pp. 645–654, Aug. 1998.

- [2] Z. Bai, P. M. Dewilde, and R. W. Freund, *Reduced-Order Modeling*, Mar. 2002, Murray Hill, NJ: Bell Laboratories. Numerical Analysis Manuscript 02-4-13.
- [3] L. T. Pillage and R. A. Rohrer, "Asymptotic waveform evaluation for timing analysis," *IEEE Trans. Comput.-Aided Des. Integr. Circuits Syst.*, vol. 9, no. 4, pp. 352–366, Apr. 1990.
- [4] P. Feldmann and R. W. Freund, "Efficient linear circuit analysis by Padé approximation via the Lanczos process," *IEEE Trans. Comput.-Aided Des. Integr. Circuits Syst.*, vol. 14, no. 5, pp. 639–649, May 1995.
- [5] —, "Reduced-order modeling of large linear subcircuits via a block Lanczos algorithm," in *Proc. ACM/IEEE Des. Autom. Conf.*, Jun. 1995, pp. 474–479.
- [6] K. Glover, "All optimal Hankel-norm approximation of linear multivariable systems and their L^∞ -error bounds," *Int. J. Contr.*, vol. 39, no. 6, pp. 1115–1193, Jun. 1984.
- [7] B. Moore, "Principal component analysis in linear systems: Controllability, observability, and model reduction," *IEEE Trans. Automat. Contr.*, vol. AC-26, no. 1, pp. 17–32, Feb. 1981.
- [8] T. Penzl, "Algorithms for model reduction of large dynamical systems," FRG, Chemnitz, Germany, Tech. Rep. SFB393/99-40, 1999. Sonderforschungsbereich 393 Numerische Simulation auf massiv parallelen Rechnern, TU Chemnitz.
- [9] Q. Su, "Algorithms for model reduction of large scale RLC systems," Ph.D. dissertation, School ECE, Purdue Univ., Lafayette, IN, Aug. 2002.
- [10] M. Green, "Balanced stochastic realizations," *Linear Algebra Appl.*, vol. 98, pp. 211–247, Jan. 1988.
- [11] X. Chen and J. T. Wen, "Positive realness preserving model reduction with H_∞ norm error bounds," *IEEE Trans. Circuits Syst. I, Fundam. Theory Appl.*, vol. 42, no. 1, pp. 23–29, Jan. 1995.
- [12] J. R. Phillips, L. Daniel, and L. M. Silveira, "Guaranteed passive balancing transformations for model order reduction," *IEEE Trans. Comput.-Aided Des. Integr. Circuits Syst.*, vol. 22, no. 8, pp. 1027–1041, Aug. 2003.
- [13] Q. Su, V. Balakrishnan, and C.-K. Koh, "Efficient approximate balanced truncation of general large-scale RLC systems via Krylov methods," in *Proc. ASPDAC/Int. Conf. VLSI Des.*, Jan. 2002, pp. 311–316.
- [14] —, "A factorization-based framework for passivity-preserving model reduction of RLC systems," in *Proc. IEEE Design Automation Conf.*, Jun. 2002, pp. 40–45.
- [15] J. Li and J. White, "Efficient model reduction of interconnect via approximate system Gramians," in *Proc. IEEE Int. Conf. Comput.-Aided Des.*, Nov. 1999, pp. 380–383.
- [16] T. Penzl, "A cyclic low rank Smith method for large sparse Lyapunov equations with applications in model reduction and optimal control," *SIAM J. Sci. Comput.*, vol. 21, no. 4, pp. 1401–1418, 2000.
- [17] J. Li, "Model reduction of large linear systems via low rank system Gramians," Ph.D. dissertation, Dept. Math., MIT, Cambridge, MA, Sep. 2000.
- [18] N. Wong, V. Balakrishnan, C.-K. Koh, and T. S. Ng, "A fast Newton/Smith algorithm for solving algebraic Riccati equations and its application in model order reduction," in *Proc. IEEE Conf. Acoust., Speech, Signal Process.*, May 2004, pp. 53–56.
- [19] N. Wong, V. Balakrishnan, and C.-K. Koh, "Passivity-preserving model reduction via a computationally efficient project-and-balance scheme," in *Proc. IEEE Design Automation Conf.*, Jun. 2004, pp. 369–374.
- [20] S. Gugercin and A. C. Antoulas, "A survey of model reduction by balanced truncation and some new results," *Int. J. Contr.*, vol. 77, no. 8, pp. 748–766, 2004.
- [21] G. Golub and C. V. Loan, *Matrix Computations*, 2nd ed. Baltimore, MD: Johns Hopkins Univ. Press, 1989.
- [22] V. Mehrmann, *The Autonomous Linear Quadratic Control Problem: Theory and Numerical Solution*, ser. 163 Lecture Notes in Control and Information Sciences. Berlin, Germany: Springer-Verlag, Nov. 1991.
- [23] J. Li and J. White, "Low-rank solution of Lyapunov equations," *SIAM Rev.*, vol. 46, no. 4, pp. 693–713, 2004.
- [24] A. S. Hodel and K. R. Poolla, "Heuristic approaches to the solution of very large sparse Lyapunov and algebraic Riccati equations," in *Proc. Conf. Decision Control*, Dec. 1988, pp. 2217–2221.
- [25] G. S. Ammar, P. Benner, and V. Mehrmann, "A multishift algorithm for the numerical solution of algebraic Riccati equations," *Electron. Trans. Numer. Anal.*, vol. 1, pp. 33–48, Sep. 1993.
- [26] P. Benner, V. Mehrmann, and H. Xu, "A new method for computing the stable invariant subspace of a real Hamiltonian matrix," *J. Comput. Appl. Math.*, vol. 86, no. 1, pp. 17–43, Nov. 1997.
- [27] P. Benner, V. Mehrmann, V. Sima, S. Van-Huffel, and A. Varga, "SLICOT—A subroutine library in systems and control theory," in *Applied and Computational Control, Signals and Circuits*, vol. 1, B. N. Datta, Ed. Boston, MA: Birkhauser, 1999, ch. 10, pp. 499–539.
- [28] P. Benner and R. Byers, "An exact line search method for solving generalized continuous-time algebraic Riccati equations," *IEEE Trans. Automat. Contr.*, vol. 43, no. 1, pp. 101–107, Jan. 1998.
- [29] J. Vlach and K. Singhal, *Computer Methods for Circuit Analysis and Design*. Norwell, MA: Kluwer, Jul. 1993.
- [30] H. Weiss, Q. Wang, and J. L. Speyer, "System characterization of positive real conditions," *IEEE Trans. Automat. Contr.*, vol. 39, no. 3, pp. 540–544, Mar. 1994.
- [31] P. Benner and H. Faßbender, "An implicitly restarted symplectic Lanczos method for the Hamiltonian eigenvalue problem," *Linear Algebra Appl.*, vol. 263, no. 1–3, pp. 75–111, Sep. 1997.
- [32] P.-A. Absil, R. Mahony, and R. Sepulchre, "Riemannian geometry of Grassmann manifolds with a view on algorithmic computation," *Acta Appl. Math.*, vol. 80, no. 2, pp. 199–220, Jan. 2004.
- [33] A. Bunse-Gerstner and V. Mehrmann, "A symplectic QR like algorithm for the solution of the real algebraic Riccati equation," *IEEE Trans. Automat. Contr.*, vol. AC-31, no. 12, pp. 1104–1113, Dec. 1986.



Ngai Wong (S'98–M'02) received the B.Eng. degree (first class honors) and Ph.D. degree, both in electrical and electronic engineering, from The University of Hong Kong, Hong Kong, in 1999 and 2003, respectively.

He was an Intern at Motorola Inc., Hong Kong, from 1997 to 1998, specializing in product testing. He was a Visiting Scholar at Purdue University, West Lafayette, IN, in 2003. Currently, he is an Assistant Professor at the University of Hong Kong. His research interests include very-large-scale integration

(VLSI) model order reduction and simulation, digital filter design, sigma-delta modulators, and optimization problems in communication and VLSI applications.

Dr. Wong received the P. K. Yu Memorial Scholarship, in 2000, the Sir Edward Youde Memorial Fellowship, and the Leung Wai Sun Fellowship, in 2002.



Venkataramanan Balakrishnan (M'94) received the B.Tech degree in electronics and communication from the Indian Institute of Technology, Madras, India, in 1985, the M.S. degree in statistics in 1992, and the M.S. and Ph.D. degrees in electrical engineering from Stanford University, Stanford, CA, in 1989 and 1992, respectively.

In 1994, he joined Purdue University, West Lafayette, IN, as an Assistant Professor, following Post-Doctoral stints at Stanford University, California Institute of Technology (Caltech), Pasadena, and the Institute for Systems Research, University of Maryland, College Park. Currently, he is a Professor and the Associate Head of Electrical and Computer Engineering at Purdue University. He is the coauthor of the monograph *Linear Matrix Inequalities in System and Control Theory* (Philadelphia: SIAM, 1994). His primary research interests include applying convex optimization to problems in systems and control.

Dr. Balakrishnan received the President of India Gold medal from the Indian Institute of Technology, Madras, in 1985, the Young Investigator Award from the Office of Naval Research, in 1997, the Ruth and Joel Spira Outstanding Teacher Award from the School of Electrical and Computer Engineering at Purdue University, in 1998, and the Honeywell Award for excellence in teaching from the School of Electrical and Computer Engineering at Purdue University, in 2001.



Cheng-Kok Koh (S'92–M'98) received the B.S. degree (first class honors) and M.S. degree, both in computer science, from the National University of Singapore, Singapore, in 1998 and 1992, respectively, and the Ph.D. degree in computer science from the University of California, Los Angeles, in 1996.

Currently, he is an Associate Professor of Electrical and Computer Engineering at Purdue University, West Lafayette, IN. His research interests include physical design of very-large-scale integra-

tion circuits and modeling and analysis of large-scale systems.

Dr. Koh received the Lim Soo Peng Book Prize for Best Computer Science Student from the National University of Singapore, in 1990, and the Tan Kah Kee Foundation Postgraduate Scholarship, in 1993 and 1994. He received the GTE Fellowship and the Chorafas Foundation Prize from the University of California, Los Angeles, in 1995 and 1996, respectively. He received the Association for Computing Machinery (ACM) Special Interest Group on Design Automation (SIGDA) Meritorious Service Award and Distinguished Service Award, in 1998, the Chicago Alumni Award from Purdue University, in 1999, the National Science Foundation CAREER Award, in 2000, and the ACM/SIGDA Distinguished Service Award, in 2002.



Tung-Sang Ng (S'74–M'78–SM'90–F'03) received the B.Sc.(Eng.) degree from The University of Hong Kong, Hong Kong, in 1972, and the M.Eng.Sc. and Ph.D. degrees from the University of Newcastle, Australia, in electrical engineering, in 1974 and 1977, respectively.

He worked for BHP Steel International and The University of Wollongong, Australia, after graduation for 14 years before returning to The University of Hong Kong, in 1991, taking up the position of Professor and Chair of Electronic Engineering. He was

the Head of the Department of Electrical and Electronic Engineering from 2000 to 2003 and is currently Dean of Engineering. His research interests include wireless communication systems, spread-spectrum techniques, code division multiple access (CDMA), and digital signal processing. He has published over 250 international journal and conference papers.

Dr. Ng was the General Chair of ISCAS'97 and the VP-Region 10 of IEEE CAS Society in 1999 and 2000. He was an Executive Committee Member and a Board Member of the IEE Informatics Divisional Board (1999–2001) and was an Ordinary Member of the IEE Council (1999–2001). He was awarded the Honorary Doctor of Engineering Degree by the University of Newcastle, Australia, in 1997, the Senior Croucher Foundation Fellowship, in 1999, the IEEE Third Millennium medal, in 2000, and the Outstanding Researcher Award by The University of Hong Kong, in 2003. He is a Fellow of the IEE and HKIE.