

Wavelength-flattened directional couplers: a geometrical approach

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A new approach to design a wavelength-insensitive optical power splitter is presented. First, a coupled-mode theory is cast in operatorial form. This allows us to solve the equivalent of coupled differential equations as simple limits. The operators are then represented on a generalized Poincaré sphere, and the resulting graphical tool is applied to different structures, giving a clear interpretation of previous results in literature as well as hints on how to find improved solutions. © 2003 Optical Society of America

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1. Introduction

The need for wideband optical communication demands that optical devices able to cover the entire band of interest. A very basic element of a typical device is, of course, the power splitter.

Power exchange between optical waveguides is usually obtained via directional couplers.^{1–3} The simplest splitting device is the synchronous coupler, obtained by keeping two identical waveguides close together over a given length. This coupling mechanism is based on the difference in the propagation constants of the coupler supermodes. On the other hand, this means that it can depend strongly on the wavelength.

Many proposals have been presented to overcome this problem, for example, coupling different waveguides (asynchronous coupler),^{4–6} tapering the coupled waveguides (tapered coupler),⁷ or implementing interferometric structures (Mach–Zehnder coupler).^{8–11} Takagi *et al.*⁴ fabricated an asynchronous coupler with a coupling ratio of $(50 \pm 5)\%$ over a wavelength range of 400 nm, as well as a tapered coupler⁷ achieving $(50 \pm 5)\%$ over 500 nm. Jinguji *et al.*¹¹ built a Mach–Zehnder coupler that achieves a splitting ratio of $(50 \pm 1.9)\%$ over 400 nm, and with a similar structure Gonthier *et al.*⁹ obtained $(50 \pm 2.5)\%$ over 300 nm.

In this paper we will focus on wavelength sensitivity of 50/50 splitters, but a similar approach applies to all splitting ratios as well as to changes of parameters other than wavelength.

2. Unitary Transformations

We will assume a power splitter as a lossless device made of two waveguides, one beside the other over a length L along the propagation axis z (see Fig. 1). In general their cross sections, distances, and refractive index profiles may vary along z . We will also assume that, for any fixed z_0 , the two eigenmodes of the single waveguides cross section, $E_j[z_0] \equiv E_j(x, y, z_0)$, ($j = 1, 2$) can be considered as a basis¹² for the field at that point, so that it is always possible to write

$$E[z] \equiv a_1[z]E_1[z] + a_2[z]E_2[z],$$

where $a_j[z]$ are complex numbers, so that $|a_1[z]|^2 + |a_2[z]|^2 = 1$ and $E_j[z]$ are normalized so that $|a_j|^2$ represents the power fractions in each of the waveguides. We will regard $E[z]$ as a scalar quantity, since we will consider only singly copolarized modes of the structure. With these assumptions we can always represent the generic state of the system through the complex vector

$$\mathbf{u}[z] \equiv \begin{pmatrix} a_1[z] \\ a_2[z] \end{pmatrix}. \quad (1)$$

The evolution of the system can be described by an unitary 2×2 matrix or, up to an overall phase,¹³ by an element \mathbf{U} of $SU(2)$ (i.e., an unitary matrix so that

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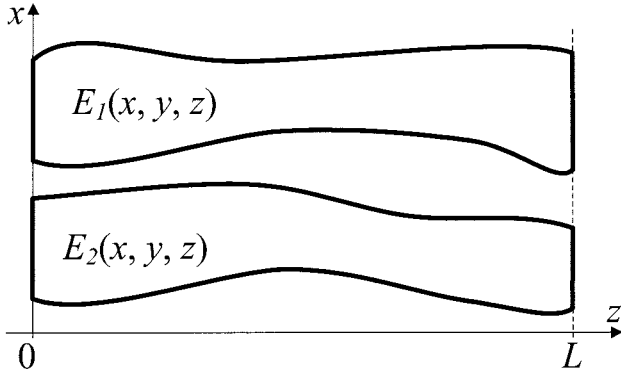


Fig. 1. Schematic of a splitter.

$\det \mathbf{U} = 1$). From its definition, the most general form for \mathbf{U} is¹⁴:

$$\begin{pmatrix} \exp(-i\theta/2)\cos\phi & \exp(i\xi)\exp(-i\theta/2)\sin\phi \\ -\exp(-i\xi)\exp(i\theta/2)\sin\phi & \exp(i\theta/2)\cos\phi \end{pmatrix}. \quad (2)$$

If optical power is launched in just one of the two waveguides, the splitting ratio (defined as the power fraction transferred to the other waveguide) does not depend on which waveguide is chosen, and it is clearly given by $\sin^2\phi$.

A. Global View

Equation (2) describes a power splitter as a black box that propagates the input state from $z = 0$ to $z = L$, i.e., so that $\mathbf{u}[L] = \mathbf{U}[L, 0]\mathbf{u}[0]$. It can be cast in the form $\mathbf{U} = \mathbf{V}(\theta) \mathbf{W}(\phi, \xi)$, where

$$\mathbf{V}(\theta) \equiv \begin{pmatrix} \exp(-i\theta/2) & 0 \\ 0 & \exp(i\theta/2) \end{pmatrix}, \quad (3)$$

$$\mathbf{W}(\phi, \xi) \equiv \begin{pmatrix} \cos\phi & \exp(i\xi)\sin\phi \\ -\exp(-i\xi)\sin\phi & \cos\phi \end{pmatrix}, \quad (4)$$

which physically means that any unitary transformation can be decomposed in the product of a θ -phase shift between the two branches and a coupling in which the field transferred from the upper (or lower) branch acquires a ξ (or $\pi - \xi$)-phase shift.

B. Local View

If we look locally at the whole transformation $\mathbf{U}[L, 0]$, we can decompose it in an infinite number of infinitesimal unitary transformations $\mathbf{U}_n \equiv \mathbf{U}[z_{n+1}, z_n]$:

$$\mathbf{U}[L, 0] = \lim_{N \rightarrow \infty} \overset{N-1}{\leftarrow} \prod_{n=0} \mathbf{U}_n, \quad (5)$$

where $\overset{\leftarrow}{\prod}$ means matrix product on the left, and we have defined $\Delta z \equiv L/N$, $z_n \equiv n\Delta z$ [from now on $f_n \equiv f(z_n) \forall f(z)$].

From the physical interpretation of Eqs. (3) and (4), we can think of every small section of the splitter as a composition of a local phase shifter $\mathbf{V}(\theta_n)$ with $\theta_n \equiv \Delta\beta_n\Delta z$ (where $\Delta\beta_n \equiv \beta_2[z_n] - \beta_1[z_n]$ is the local dif-

ference in propagation constants of the two branches) and a local coupler $\mathbf{W}(\phi_n, \xi_n)$ with $\phi_n = \kappa_n\Delta z$ (where κ_n is the local coupling coefficient between the two waveguides and ξ_n is the phase acquired locally, passing from the first to the second branch). To first order in Δz can rewrite

$$\mathbf{U}_n \equiv \mathbf{I} + \boldsymbol{\tau}_n\Delta z,$$

where \mathbf{I} is the identity matrix and

$$\boldsymbol{\tau}_n \equiv \begin{bmatrix} -i/2\Delta\beta_n & \kappa_n \exp(i\xi_n) \\ -\kappa_n \exp(-i\xi_n) & i/2\Delta\beta_n \end{bmatrix}. \quad (6)$$

\mathbf{U}_n is the infinitesimal propagator from z_n to z_{n+1} , i.e., the matrix so that

$$\mathbf{u}_{n+1} = \mathbf{U}_n\mathbf{u}_n. \quad (7)$$

Notice that Eq. (7) is nothing but a compact form for writing the standard¹⁻³ system of coupled differential equations in the coefficients $a_1[z]$, $a_2[z]$ of Eq. (1).

C. Special Cases

In some special cases Eq. (5) can be cast in closed form:

1) Null coupling.

Setting $\kappa = 0$ gives $\mathbf{U}[L, 0] = \mathbf{V}(\Theta)$ with $\Theta \equiv \int_0^L \Delta\beta dz$. This result describes a “pure” phase shifter and is easily shown by noting that $\mathbf{V}(\theta_1)\mathbf{V}(\theta_2) = \mathbf{V}(\theta_1 + \theta_2)$.

If we define $\langle\beta[z]\rangle \equiv (\beta_1[z] + \beta_2[z])/2$, the neglected overall phase¹³ is $\exp(i\eta)$, with $\eta \equiv \int_0^L \langle\beta[z]\rangle dz$.

2) Null phase shift.

Imposing $\beta_1 = \beta_2 = \beta$ and $d\xi/dz = 0$ gives $\mathbf{U}[L, 0] = \mathbf{W}(\Phi, \xi)$ with $\Phi \equiv \int_0^L \kappa dz$. This describes a “pure” coupler and, as before, it is a consequence of the property $\mathbf{W}(\phi_1, \xi)\mathbf{W}(\phi_2, \xi) = \mathbf{W}(\phi_1 + \phi_2, \xi)$.

In this case the neglected overall phase amounts to $\exp(i\eta)$, with $\eta \equiv \int_0^L \beta dz$.

Actually the only coupling mechanism known to the author is reciprocal, i.e., time reversal. It can be shown² that the time reversal combined to the unitarity condition implies $\mathbf{U}^\dagger = \mathbf{U}^*$; that in Eq. (2) requires $\xi = \pi/2$. Physically this means that the phase acquired during coupling from the first to the second branch must be equal to the phase acquired in the reverse process. Nevertheless we will allow for generic ξ values, so that our formalism holds for any unitary system (e.g., for polarization states, where the Faraday rotation is nonreciprocal).

3) Constant coupling and phase shift (asynchronous coupler).

Suppose $d\Delta\beta/dz = 0$, $d\kappa/dz = 0$ and $d\xi/dz = 0$. Eq. (5) then becomes

$$\begin{aligned} \mathbf{U}[L, 0] &\equiv \bar{\mathbf{U}}[L, 0] = \lim_{N \rightarrow \infty} \left(\mathbf{I} + \frac{L}{N} \boldsymbol{\tau} \right)^N \\ &= \lim_{N \rightarrow \infty} \sum_{k=0}^N \binom{N}{k} \left(\frac{L}{N} \boldsymbol{\tau} \right)^k = \lim_{N \rightarrow \infty} \sum_{k=0}^N \frac{(L\boldsymbol{\tau})^k}{k!} \\ &= \exp(L\boldsymbol{\tau}), \end{aligned}$$

where, from now on, the overbar will denote the solution with constant coupling and phase shift. Let us now introduce the hermitian matrix $\mathbf{v} \equiv \boldsymbol{\tau}/(i\mu)$, where $\mu^2 \equiv \det \boldsymbol{\tau} = (\Delta\beta/2)^2 + \kappa^2$. It is easily verified that $\mathbf{v}^{2n} \equiv (-\mathbf{I})^n$. So the sum splits in even and odd terms, yielding

$$\overline{\mathbf{U}}[L] = \cos(\mu L)\mathbf{I} + i \sin(\mu L)\mathbf{v}. \quad (8)$$

The matrix \mathbf{v} is the so-called infinitesimal generator^{14,15} of $\overline{\mathbf{U}}$.

In this case the neglected overall phase is clearly $\exp(i\langle\beta\rangle L)$, where

$$\langle\beta\rangle \equiv (\beta_1 + \beta_2)/2. \quad (9)$$

If we define $\cos \gamma \equiv -\Delta\beta/(2\mu)$ (which implies $\sin \gamma \equiv \kappa/\mu$), Eq. (8) becomes

$$\overline{\mathbf{U}}(L) \equiv \overline{\mathbf{U}}(\mu L, \xi, \gamma) = \begin{pmatrix} \cos(\mu L) + i \sin(\mu L)\cos \gamma & \exp(i\xi)\sin(\mu L)\sin \gamma \\ -\exp(-i\xi)\sin(\mu L)\sin \gamma & \cos(\mu L) - i \sin(\mu L)\cos \gamma \end{pmatrix}. \quad (10)$$

Since we are neglecting overall phases

$$\overline{\mathbf{U}}(\mu L + \pi, \xi, \gamma) = -\overline{\mathbf{U}}(\mu L, \xi, \gamma) \equiv \overline{\mathbf{U}}(\mu L, \xi, \gamma), \quad (11)$$

i.e., $\overline{\mathbf{U}}(\mu L, \xi, \gamma)$ is π -periodic in μL .

Notice that in solving for the propagator matrix $\mathbf{U}[L, 0]$ [Eq. (5)], we do not need to impose any boundary conditions, as for usual differential equations, because they are implicit in the input state $\mathbf{u}[0]$ on which $\mathbf{U}[L, 0]$ operates.

D. Eigenstates

To interpret the last result, it is useful to determine the eigenstates of $\overline{\mathbf{U}}$. From the determinantal equation the eigenvalues are found to be

$$\lambda_{\pm} = \exp(\pm i\mu L), \quad (12)$$

and, correspondingly, the eigenvectors can be cast in the form

$$\mathbf{u}_{+} = \begin{Bmatrix} \exp[-i/2(\pi/2 - \xi)]\cos(\gamma/2) \\ \exp[i/2(\pi/2 - \xi)]\sin(\gamma/2) \end{Bmatrix} \\ \mathbf{u}_{-} = \begin{Bmatrix} -\exp[-i/2(\pi/2 - \xi)]\sin(\gamma/2) \\ \exp[i/2(\pi/2 - \xi)]\cos(\gamma/2) \end{Bmatrix}. \quad (13)$$

From Eqs. (3) and (4), defining $\mathbf{R}(\gamma, \xi) \equiv \mathbf{V}(\pi/2 - \xi)\mathbf{W}(\gamma/2, 0)$, we can also rewrite

$$\mathbf{u}_{+,-} = \mathbf{R}(\gamma, \xi)\mathbf{u}_{1,2} \\ \overline{\mathbf{U}}(\mu L, \xi, \gamma) = \mathbf{R}(\gamma, \xi)\mathbf{V}(-2\mu L)\mathbf{R}^{-1}(\gamma, \xi). \quad (14)$$

Physically, Eq. (12) means that the difference between the propagation constants of the two eigenstates is $\Delta\beta_{\pm} \equiv \beta_{-} - \beta_{+} = -2\mu$. On the other hand Eq. (14) means that $\overline{\mathbf{U}}(\mu L, \xi, \gamma)$ is nothing but a $\mathbf{V}(\Delta\beta_{\pm}L)$ diagonal transformation (i.e., a transformation repre-

sented in the basis of its eigenstates) represented in the rotated basis $[\mathbf{R}^{-1}(\gamma, \xi)\mathbf{u}_{+}, \mathbf{R}^{-1}(\gamma, \xi)\mathbf{u}_{-}]$.

Looking at $\overline{\mathbf{U}}$ globally, we can also rewrite the neglected overall phase as¹³ $\exp(i\langle\beta_{\mp}\rangle L)$, where $\langle\beta_{\mp}\rangle \equiv (\beta_{+} + \beta_{-})/2$. From Eq. (9) (which was obtained looking at the local \mathbf{U}_n) we get $\langle\beta_{\mp}\rangle = \langle\beta\rangle$, which implies $\beta_{\mp} = \langle\beta\rangle \mp \mu$.

Notice that, for $\kappa = 0$, the eigenstates of $\overline{\mathbf{U}}$ do not depend on $\Delta\beta$, so they must also be the eigenstates of $\mathbf{V}(\Theta)$ (the first special case discussed earlier). Similarly, for $\Delta\beta = 0$ they do not depend on κ , so they must also be the eigenstates of $\mathbf{W}(\Phi, \xi)$.

3. Generalized Poincaré Sphere

All the results obtained so far are better understood introducing a convenient geometrical representation.

It is well known^{15,16} that SU(2) transformations may be mapped into SO(3) transformations through a homomorphism. This means that all the transformations we have analyzed before can be represented as rotations on a spherical surface, analogous of the Poincaré sphere¹⁷ for polarization states. In Fig. 2 are displayed all the intersection of the S_1, S_2, S_3 axes within the sphere. They represent the single waveguides modes E_1, E_2 and their linear combinations $E_{S,A} \equiv 1/\sqrt{2}(E_1 \pm E_2)$ and $E_{R,L} \equiv 1/\sqrt{2}(E_1 \pm iE_2)$. Also plotted is the generic normalized mode $P \equiv a_1E_1 + a_2E_2 \equiv \cos \alpha E_1 + \exp(i\theta) \sin \alpha E_2$, with Stokes parameters¹⁷

$$S_0 \equiv |a_1|^2 + |a_2|^2 = 1, \\ S_1 \equiv |a_1|^2 - |a_2|^2 = \cos 2\alpha, \\ S_2 \equiv a_1a_2^* + a_1^*a_2 = \sin 2\alpha \cos \theta, \\ S_3 \equiv i(a_1a_2^* - a_1^*a_2) = \sin 2\alpha \sin \theta.$$

In Fig. 3 we have plotted the eigenmodes $E_{+,-}$ [corresponding to the eigenstates $\mathbf{u}_{+,-}$ of Eq. (13)]. It is clear that the generic transformation $\overline{\mathbf{U}}$ [Eq. (8)], corresponds, on the sphere, to a rotation about the axis of its eigenstates $\overline{E}_{+}E_{-} \equiv S(\gamma, \pi/2 - \xi)$ (which is the rotated of the axis S_1 by an angle γ about S_3 and then by an angle $\pi/2 - \xi$ about S_1). This means that a generic input state P will be rotated by an angle $-2\mu L$ [see Eq. (14)] on the circle of revolution about $S(\gamma, \pi/2 - \xi)$ passing through P . The three special cases discussed earlier may be represented as rotations on the sphere:

1) $\kappa = 0$ implies $E_{+,-} = E_{1,2}$ and $\mathbf{U} = \mathbf{V}(\Theta)$.

Physically it is clear that $E_{1,2}$ are the eigenmodes of the phase shifter. On the sphere a generic input state P will be rotated by an angle Θ on the circle of revolution about S_1 (Fig. 4) passing through P . In

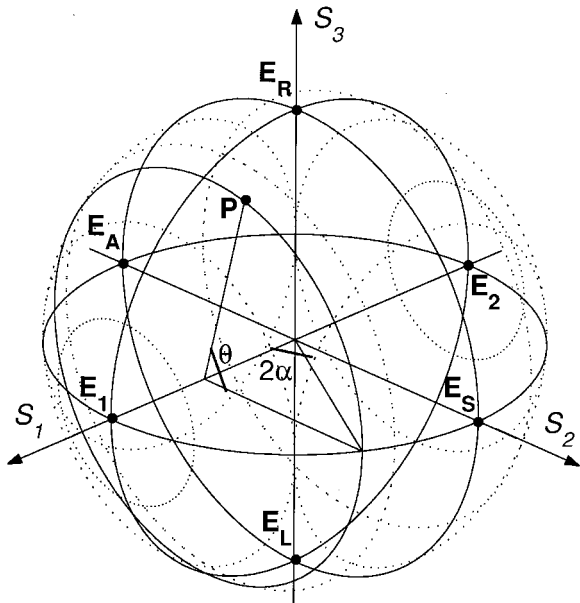


Fig. 2. Generalized Poincaré sphere.

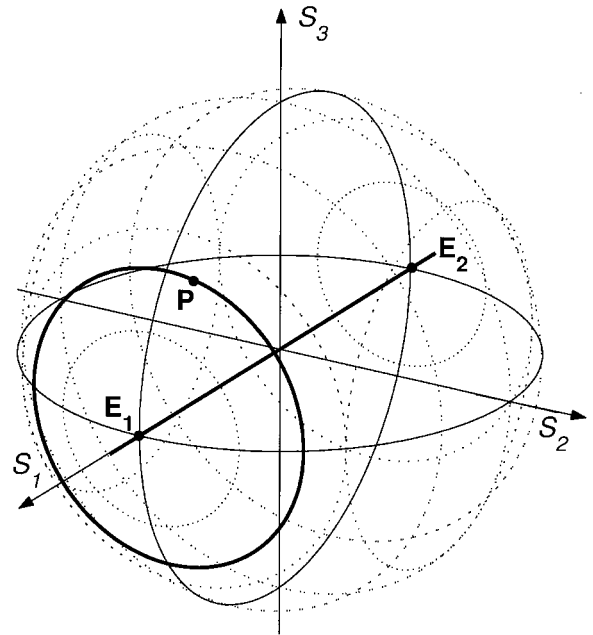


Fig. 4. Action of a phase shifter.

the special case of constant phase shift will be $\Theta = \Delta\beta L$.

Notice that the circles of revolution about S_1 represent the loci of constant power splitting, which we will call isodias ($\omega o = \text{equal}$, $\delta\alpha = \text{split in two parts}$).

2) $\Delta\beta = 0$ and $\xi = \pi/2$ means $E_{+,-} = E_{S,A}$ and $\mathbf{U} = \mathbf{W}(\Phi, \pi/2)$.

Physically it is clear that the symmetric and anti-symmetric superpositions of the single waveguide modes are the eigenmodes of the synchronous coupler. On the sphere a generic input state P will be rotated by an angle 2Φ on the circle of revolution about S_2 (Fig. 5) passing through P . In the special

case of constant coupling will be $-2\Phi = -2\kappa L = \Delta\beta_{AS}L$, where we have defined $\Delta\beta_{AS} \equiv \beta_A - \beta_S$.

If one allows for generic $\xi[\mathbf{U} = \mathbf{W}(\Phi, \xi)]$, it is easily seen that the loci of constant $(\pi/2 - \xi)$ phase shift are the semicircles with diameter on S_1 , which we will call isophases (notice that isodias and isophases can be regarded as a parallels and meridians for the sphere, with poles on E_1 and E_2).

3) $\Delta\beta \neq 0$ and $\xi = \pi/2$ means $\mathbf{u}_{+,-} = \mathbf{W}(\gamma/2, 0)\mathbf{u}_{1,2}$ and $\mathbf{U} \equiv \bar{\mathbf{U}}(\mu L, \pi/2, \gamma)$.

In the limit of validity of our approximation,¹² the eigenmodes of an asynchronous coupler are similar to $E_{S,A}$, but with unbalanced power in the waveguides.

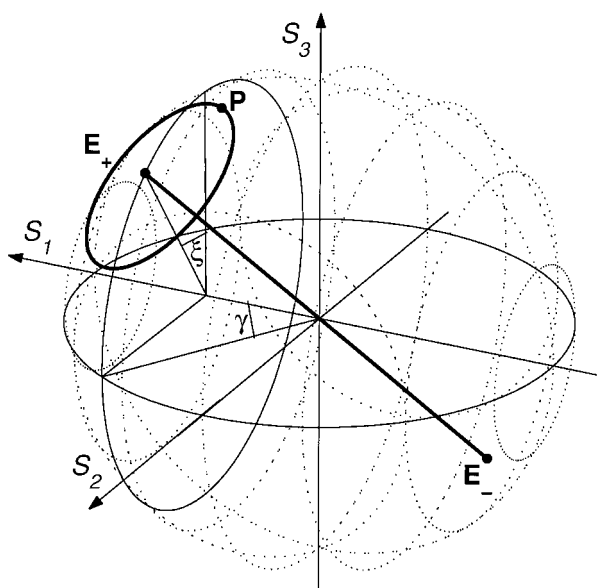


Fig. 3. Eigenstates of $\mathbf{U}(\mu L, \xi, \gamma)$.

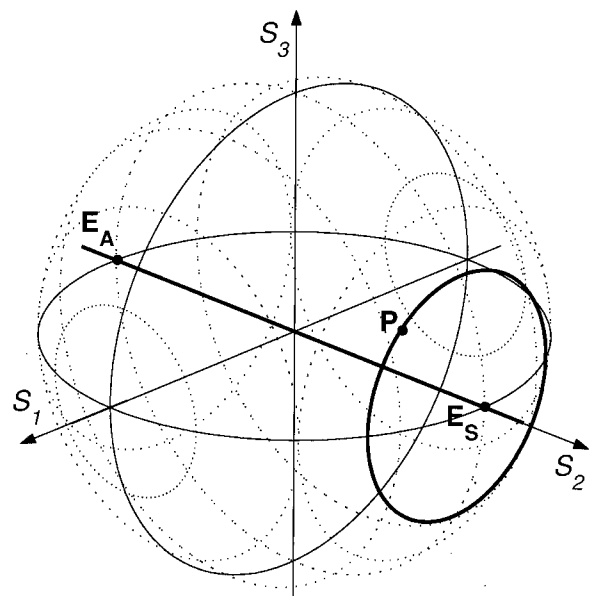


Fig. 5. Action of a synchronous coupler.

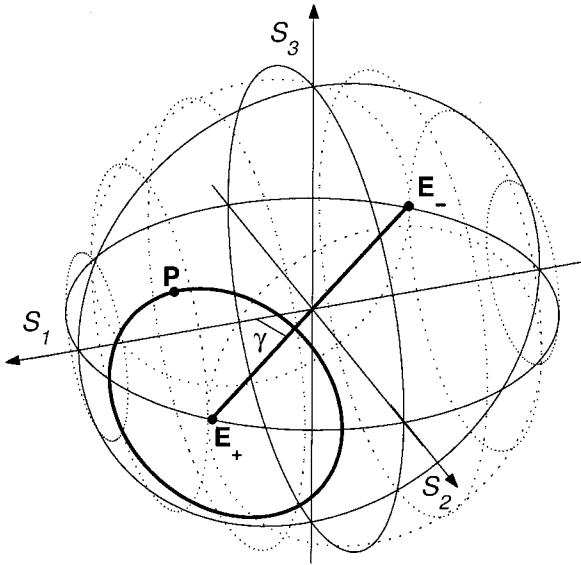


Fig. 6. Action of an asynchronous coupler.

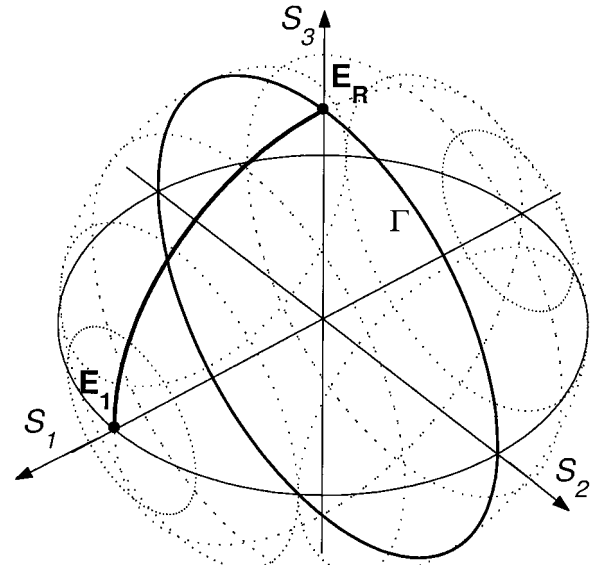


Fig. 7. 50/50 synchronous coupler.

On the sphere a generic input state P will be rotated by an angle $-2\mu L = \Delta\beta_{\mp}L$ on the circle of revolution about $S(\gamma, 0)$ (Fig. 6) passing through P .

4. 50/50 Splitters

We will now analyze the wavelength dependence of some 50/50 splitters. Under a change $d\lambda$ in the wavelength λ , a relative change df/f will correspond to any $f(\lambda)$ (relative changes are easier to calculate when products and divisions resembling the rules of relative error propagation are used).

On the sphere a 50/50 splitter is any transformation that, starting from E_1 (or, equivalently, from E_2), reaches any point on the isodia Γ with diameter on S_2 . The straightest way to do so is by a synchronous coupler, i.e., by a $\mathbf{W}(\pi/4, \pi/2)$ transformation. Power transfer is given by $P_2 = 1/2[1 - \cos(\Delta\beta_{AS}L)]$, and varies as

$$dP_2 = \frac{1}{2} \sin(\Delta\beta_{AS}L) L d\Delta\beta_{AS} = \frac{1}{2} \Delta\beta_{AS} L \sin(\Delta\beta_{AS}L) \left(\frac{d\Delta n_{AS}}{\Delta n_{AS}} - \frac{d\lambda}{\lambda} \right), \quad (15)$$

where $\Delta n_{AS} \equiv n_A - n_S$ is the difference between the effective indexes of the symmetric and antisymmetric modes. In our case

$$dP_2 = \frac{1}{2} L d\Delta\beta_{AS} = \frac{\pi}{4} \left(\frac{d\Delta n_{AS}}{\Delta n_{AS}} - \frac{d\lambda}{\lambda} \right).$$

Therefore the only way to minimize the wavelength sensitivity is through a convenient choice of the waveguides and their distance.¹⁸

On the sphere (Fig. 7) we observe that power splitting is measured by the S_1 parameter (by definition, the difference between the optical power in the two waveguides). Since a synchronous coupler ap-

proaches the isodia Γ parallel to S_1 , a change in the angle $L\Delta\beta_{AS}$ translates immediately into a power change.

An alternative approach could be using an asynchronous coupler.⁴⁻⁶ Power transfer in this case is given by $P_2 = P_0 F$, where $P_0 \equiv \sin^2 \gamma = 1 - (\Delta\beta / \Delta\beta_{\mp})^2$ is the maximum transferable power and $F \equiv 1/2[1 - \cos(\Delta\beta_{\mp}L)]$ is the power oscillation along propagation. A λ change will now give $dP_2/P_2 = dP_0/P_0 + dF/F$, where

$$dP_0 = 2(P_0 - 1) \left(\frac{d\Delta n}{\Delta n} - \frac{d\Delta n_{\mp}}{\Delta n_{\mp}} \right) \equiv 2(P_0 - 1)\delta, \quad (16)$$

$$dF = \frac{1}{2} \sin(\Delta\beta_{\mp}L) L d\Delta\beta_{\mp} = \frac{1}{2} \Delta\beta_{\mp} L \sin(\Delta\beta_{\mp}L) \left(\frac{d\Delta n_{\mp}}{\Delta n_{\mp}} - \frac{d\lambda}{\lambda} \right), \quad (17)$$

where $\Delta n \equiv n_2 - n_1$ is the difference in effective index of the single-waveguide modes.

In particular when $\Delta\beta_{\mp}L = k\pi$, the oscillating contribution vanishes and we can get a 50/50 splitter if $\gamma = \pi/4$. Regarding the dP_0 contribution we notice that it is null for $\gamma = m\pi/2$ and does not feature $d\lambda$ term. Therefore, to get an insensitive coupler, a convenient choice of the coupler parameters must be made for δ to become negligible.

On the sphere (Fig. 8) dP_0 is due to a γ change, i.e., a change of the rotation axis, while dF is related to a $\Delta\beta_{\mp}$ change, i.e., a change of the rotation angle. It is clear that this asynchronous coupler approaches the isodia Γ perpendicular to S_1 direction so that, if the rotation axis does not change, power splitting it is invariant under $\Delta\beta_{\mp}$ changes.

This approach can be generalized by cascading N asynchronous couplers so that $\Delta\beta_{\mp}^{(i)}L^{(i)} = k\pi \forall i$ and $\sum_{i=1}^N (-)^{i+N} \gamma_i = \pi/4$. In Fig. 9 it is shown the case

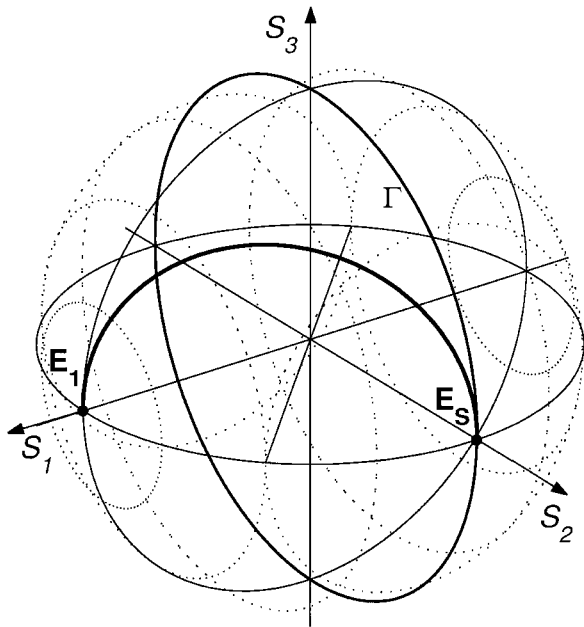


Fig. 8. 50/50 asynchronous coupler.

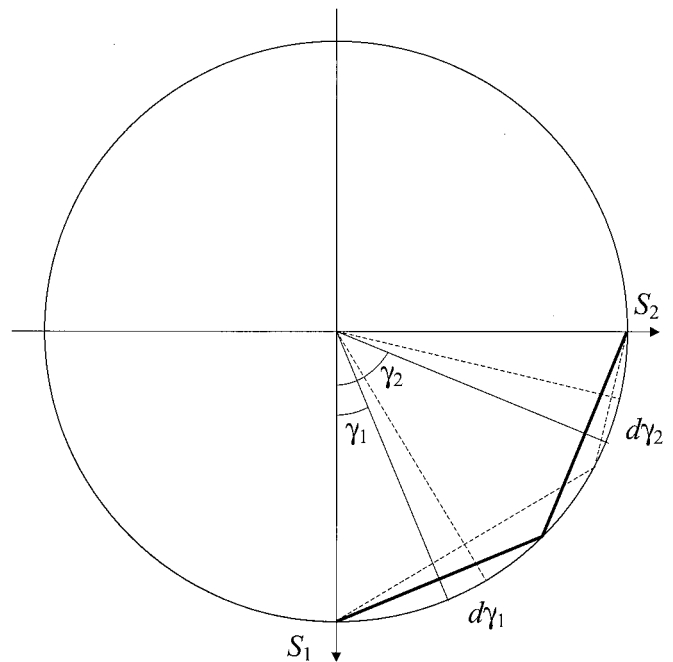


Fig. 10. Insensitivity of the configuration of Fig. 9.

$\gamma_1 = \pi/8$ and $\gamma_2 = 3\pi/8$. Under the hypothesis that $d\gamma_1 = d\gamma_2$, we expect this configuration to solve the problem of rotation axis changes, as shown in Fig. 10 [in general it will be true for any choice of $\{\gamma_i\}$ so that $\sum_{i=1}^N (-)^{i+N} d\gamma_i = 0$].

Another approach may be to find a combination of γ and $\Delta\beta_{\pm}L$ reaching a point on Γ that make dP_0/P_0 and dF/F cancel each other out, instead of setting each single contribution to zero.

These simple examples show how use of a pictorial view can help not only to interpret well known results but also to find better solutions.

Now that we understand the working principle of

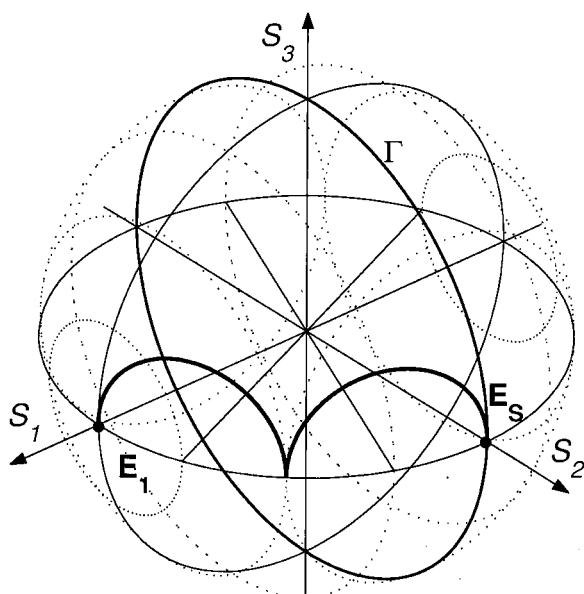


Fig. 9. 50/50 double-asynchronous coupler.

the asynchronous coupler, it seems that even better insensitivity could be obtained if we approached Γ descending from points closer to E_R . This could be done with a structure that starts as a pure synchronous coupler and, along propagation, becomes an almost-pure phase shifter. This is the tapered coupler,⁷ schematically shown in Fig. 11. The trajectory on the sphere (Fig. 12) can be seen as a composition of small rotations about different axes $S[\gamma(z), 0]$, with $\gamma(0) = \pi/2$ and $\gamma(L) \approx 0$. Notice that this structure is tolerant at the beginning [see Eq. (15)] and at the end, but a dilatation (contraction) in the middle part of the trajectory [see Eqs. (16) and (17)] could shift the ending part of the trajectory on an isodia different by Γ . So, from a pictorial point of view, it is not clear whether a tapered coupler may be better than an asynchronous coupler. The answer can come only from a numerical study and will depend on the parameters of the specific structure under investigation. A completely different approach

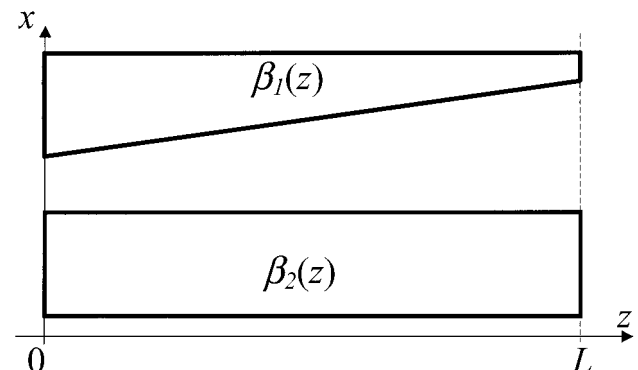


Fig. 11. Schematic of a tapered coupler.

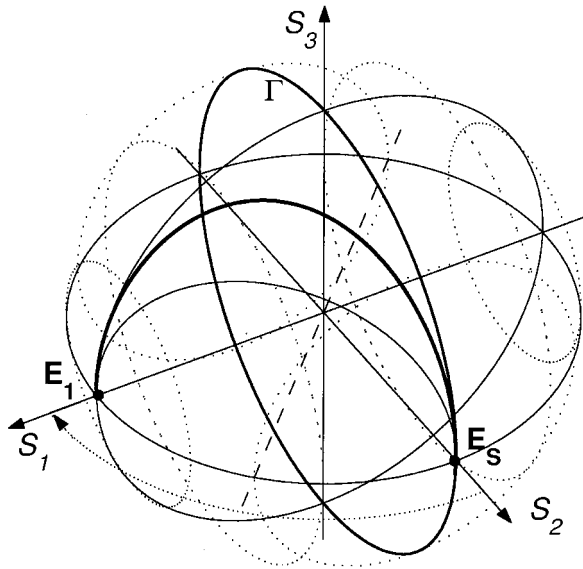


Fig. 12. 50/50 tapered coupler.

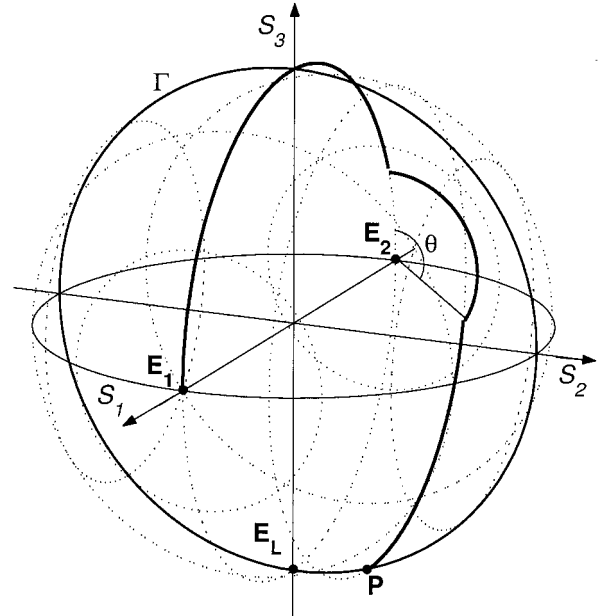


Fig. 13. 50/50 Mach-Zehnder coupler.

is based on an interferometric scheme.⁸⁻¹¹ Consider a synchronous directional coupler so that (at a certain working wavelength λ_0) $\phi_0 = \kappa_0 L = \pi/2$ (100% power transfer), cascaded with a θ -phase shifter and another directional coupler, identical to the first one, but with half the length (50% power transfer). At a generic λ will be $\phi = \phi_0 + \Delta\phi$, and the system will be described by the matrix

$$\mathbf{M} = \mathbf{W}[1/2(\pi/2 + \Delta\phi), \pi/2]\mathbf{V}(\theta)\mathbf{W}(\pi/2 + \Delta\phi, \pi/2).$$

Of course when $\Delta\phi = 0$, this is perfectly equivalent to a $3\pi/4$, 50% splitter disregarding the θ -value.

When a (noninfinitesimal) fixed value is assigned to $\Delta\phi$, our aim is to determine a corresponding θ value that still gives 50/50 power splitting. This means requiring $|M_{11}|^2 = 1/2$, which gives

$$\cos \theta = \frac{t(t-1) - 1/2}{1-t^2}, \quad (18)$$

where $t \equiv \sin \Delta\phi$ and the condition $|\cos \theta| \leq 1$ implies $t \leq 1/2$. For small $\Delta\phi$ we have $\cos \theta \approx -1/2 - t \rightarrow -1/2$ or $\theta \rightarrow \pm 2\pi/3 + 2k\pi$.

This result is easily understood on the sphere (Fig. 13). Since we have a $\Delta\phi$ angular shift in the first coupler and a $\Delta\phi/2$ angular shift in the second coupler, a $2\pi/3$ rotation about S_1 (which goes from $\phi_0 + \Delta\phi$ to $\phi_0 - \Delta\phi/2$) will compensate, at once, the shifts of both couplers (being the change of the circle representing the second coupler a second-order effect).

Notice that in general the phase shifter will also be wavelength dependent. In the case of a concentrated phase shifter, made of two identical waveguides of different length and effective index n , it will be $d\theta/\theta = d\Delta n/\Delta n - d\lambda/\lambda$. In the case of a distributed phase shifter, made of two different

waveguides of the same length and a difference Δn in their effective indexes, it will be $d\theta/\theta = d\Delta n/\Delta n - d\lambda/\lambda$. So it may be convenient to set the working point of the couplers and the working point of the phase shifter at different wavelengths in the desired band.

This example shows the power of the geometrical representation, which becomes apparent especially when dealing with interferometric schemes.

5. Conclusions

We have cast coupled-mode theory in an operatorial form. This formalism allows us to solve the equivalent of the usual differential equations as simple limits. Furthermore, the homomorphism between the SU(2) group and the SO(3) group allows us to represent all these transformations on a generalized Poincaré sphere, which is found to be a powerful tool in the design and understanding of tolerant 2×2 devices, once all parameter dependences are determined.

References and Notes

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 12. In the present work we will regard $B \equiv \{E_1, E_2\}$ as an orthonormal basis, but this is not a rigorous treatment. First, the expansions of the single-waveguide modes on the orthogonal basis of the coupler supermodes are not, in general, orthogonal.^{1,3} This problem can be solved by working, for example, in the orthonormal basis $B' \equiv \{E_1, E_1^\perp\}$ and expanding E_2 on B' . Furthermore, none of the bases above can be regarded, in general, as complete, but only as truncated bases of the system.
 13. Working with $SU(2)$ transformations physically means keeping track of just the phase difference between the two branches. However the neglected overall phase can be easily recovered. In fact, an unitary matrix \mathbf{X} [in general, $\mathbf{X} \notin SU(2)$] can be always diagonalized, in the basis of its eigenvectors, as $\mathbf{X} = \mathbf{RDR}^{-1}$. Furthermore, a diagonal matrix $\mathbf{D} = \text{diag}[\exp(i\theta_1), \exp(i\theta_2)]$ can be always rewritten as $\exp(i\langle\theta\rangle)\mathbf{V}(\Delta\theta)$, where $\langle\theta\rangle \equiv (\theta_1 + \theta_2)/2$ and $\Delta\theta \equiv \theta_2 - \theta_1$. So we can always rewrite $\mathbf{X} = \exp(i\langle\theta\rangle)\mathbf{RV}(\Delta\theta)\mathbf{R}^{-1} \equiv \exp(i\langle\theta\rangle)\mathbf{U}$, where, from its definition, $\mathbf{U} \in SU(2)$.
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